

Continuous Monitor of the PLL Frequency With the DCC

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ABSTRACT

The Dual Clock Compare (DCC) may be configured to provide autonomous, real-time monitoring of the average frequency of a signal. This application report illustrates a configuration in which the DCC monitors the average PLL frequency. This comparison triggers an error when the average PLL falls out of a specified range. The accuracy window and the duration over which the frequency is averaged must be determined. These parameters are tightly linked so that they are not independent, and trade-offs must be made based upon the application requirements.

Project collateral and source code discussed in this document can be downloaded from the following URL: http://www.ti.com/lit/zip/spna211.

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1 DCC Theory of Operation

The DCC accepts two clock inputs: Clock₀ and Clock₁. These clocks decrement counters in order to compare their relative frequencies. Three different time periods are created by the DCC:

- T₀ is created by decrementing Counter₀ with Clock₀
- T_{Valid} is created by decrementing Valid with Clock₀, referred to as the valid window
- T₁ is created by decrementing Counter₁ with Clock₁

 T_0 and T_{Valid} create the reference time and window for measuring T_1 . The counters are configured so that T_1 expires within the valid window; if T_1 does not expire in the Valid window, an error is generated. This behavior is the purpose of the DCC.

Figure 1 shows the behavior of the DCC.

- Initially, all counters are loaded with their user-defined preload value.
- Counter₀ and Counter₁ decrement at rates determined by the frequencies of Clock₀ and Clock₁, respectively.
- When Counter₀ equals 0 (expires), the Valid counter begins decrementing at a rate determined by Clock₀.
- If Counter₁ decrements to 0 in the valid window, then:
 - No error is generated
 - All counters (Counter₀, Valid, Counter₁) are reloaded when Valid decrements to 0 ⁽¹⁾.
- This circuit description also shows that the DCC does not monitor Clock₁ in the period from expiration of Counter₁ to expiration of Valid.



Counter, is shown decrementing at different rates in Figure 1.

- The rates that expire outside of the Valid window generate an error and are shown in red.
- The rates that expire within the Valid window do not generate an error; these rates are shown in black.

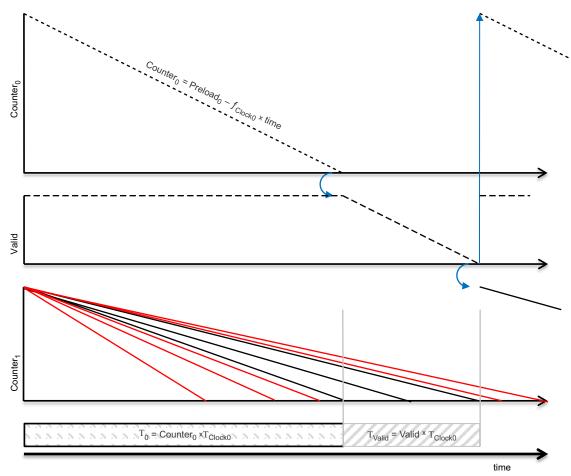


Figure 1. DCC Measurement

The expected operation of the DCC (with no errors) is shown in Figure 2. In this waveform, Counter₁ expires in the middle of the Valid count-down. After Valid has expired, all counters are again pre-loaded and Counter₀ and Counter₁ begins to decrement immediately.



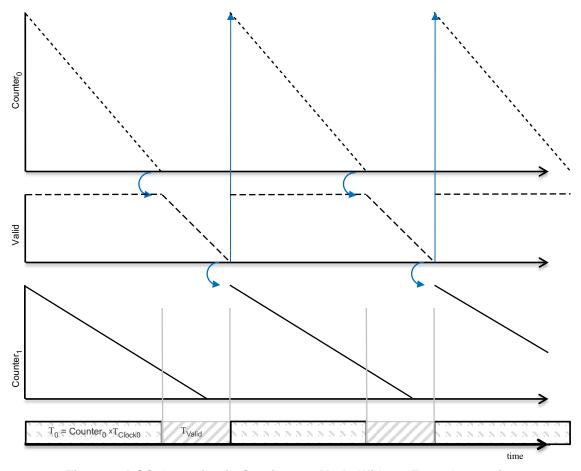


Figure 2. DCC Operating in Continuous Mode Without Error Generation

2 Calculating DCC Parameters

The counters are normally configured so that Counter, decrements to 0 in the middle of the Valid counter.

$$T_{Clock0}$$
 $\left(Counter_0 + \frac{1}{2}Valid\right) = T_{Clock1} \times Counter_1$ (1)

The steps to setup this relation are:

- 1. Set $T_{Clock0} \times Counter_0 = T_{Clock1} \times Counter_1$
- 2. Compute Valid.
- 3. Adjust Counter₀ so that the new $Counter_0 = Counter_0 \frac{1}{2}Valid$.

Given, a fixed period Clock₀ and Clock₁, the DCC Counters – Counter₀, Valid, and Counter₁ must be configured for optimal performance. The configuration of the Valid counter has some constraints, and these configuration guidelines for Valid form the subject of this application report.

3 Configuring DCC Valid

The Valid counter's configuration is determined by considering two different sources of errors:

- Errors due to the asynchronous timing of Clock₀ and Clock₁
- Digitization error



www.ti.com Configuring DCC Valid

3.1 Errors Due to Asynchronous Timing of Clock, and Clock,

Since Clock₀ and Clock₁ are asynchronous, the mechanism for loading the counters does not occur synchronously; this circuitry is shown in Figure 3 The diagram shows that Down Counter 0 and Down Counter 1 cannot start at the same time because Reload is synchronized to Clock₀ and Clock₁. There are two conditions to consider:

- $T_{Clock0} > T_{Clock1}$ (for example, $f_{Clock1} > f_{Clock0}$), the offset requires 2 cycle offset in Valid (2).
- $T_{Clock1} > T_{Clock0}$ (for example, $f_{Clock0} > f_{Clock1}$), the offset requires $2 \times \frac{I \ Clock1}{T \ Clock0}$ cycle offset in Valid (3).

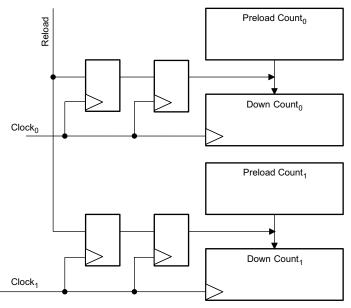


Figure 3. Asynchronous Timing Between Clock, and Clock, Starts Counters at Different Times

The minimum number of Valid counts (excluding digitizing error) when $f_{Clock1} > f_{Clock0}$ is derived:

$$\begin{split} Valid_{Min} \times T_{Clock0} &= 2T_{Clock0} - T_{Clock1} \\ Valid_{Min} \times T_{Clock0} &= 2T_{Clock0} - \frac{T_{Clock1}}{T_{Clock0}} T_{Clock0} \\ Valid_{MIN} &= 2 - \frac{T_{Clock1}}{T_{Clock0}} \rightarrow 2 \end{split}$$

The minimum number of Valid counts (excluding digitizing error) when $f_{Clock0} > f_{Clock1}$ is derived:

$$\begin{split} Valid_{Min} \times T_{Clock0} &= 2T_{Clock1} - T_{Clock0} \\ Valid_{Min} \times T_{Clock0} &= 2\frac{T_{Clock1}}{T_{Clock0}} T_{Clock0} - T_{Clock0} \\ Valid_{Min} &= 2\frac{T_{Clock1}}{T_{Clock0}} - 1 \rightarrow 2\frac{T_{Clock1}}{T_{Clock0}} \end{split}$$

Configuring DCC Valid www.ti.com

3.2 Digitization Error

Additionally, the counters have digitizing error. The code assigns 3 Clock 0 cycles for digitizing error.

3.3 Minimum Valid Count

The minimum count for Valid is expressed in terms of the synchronization error and the digitization error. The valid count is configured to be symmetric to the error sources (which gives a multiplication factor of 2).

$$Valid_{Min} = 2 \left(Synchronization + Digitization \right) = \begin{cases} 2(2+3) = 10 & T_{Clock0} > T_{Clock1} \\ 2\left(2 \times \frac{T_{Clock1}}{T_{Clock0}} + 3\right) & T_{Clock1} > T_{Clock0} \end{cases}$$
(2)

4 Relationship Between Resolution and Duration

It is obvious that, independent of error terms, there is a relationship between the accuracy of the frequency measurement and the duration of the count. Since the frequency is correct if Counter₁ expires within the Valid window, the valid window represents an uncertainty in the timing.

$$\frac{\frac{1}{2}^{T}Valid}{T_{0}} = \frac{1}{Re \, solution} \tag{3}$$

This resolution of the measurement can be maximized (as a percentage) by extending the duration of the accumulation. The duration can be express as ⁽⁴⁾:

$$Duration = \frac{Valid \times Re solution}{2^f Clock0}$$
(4)

in which the Resolution is expressed as 1 part in Resolution (1% is 1/100 and Resolution is 100; 0.2% is 1/500 and Resolution is 500).

⁴⁾ The relationship between Valid, Resolution and Duration is derived:

$$\begin{split} & \frac{\pm \Delta t}{T} = \frac{1}{Re \, solution} \\ & \frac{\pm \Delta t}{T} = \pm \frac{\frac{Valid}{2} \times T_{Clock0}}{Counter_0 \times T_{Clock0}} = \frac{1}{Re \, solution} \\ & \pm \frac{Valid \times Re \, solution}{2f_{Clock0}} = Counter_0 \times T_{Clock0} = Duration \end{split}$$

4.1 Setting Counter, Valid and Counter,

There is a trade-off between Duration and Resolution (see Equation 4). As the resolution is increased, so is the duration of the sample. For a given resolution, the duration of the sample can be minimized by:

- Minimizing Valid
- Using the fastest (accurate) Clock₀

Since the oscillator is usually determined by other system criteria, the minimum duration (for a given resolution) is achieved by selecting the minimum Valid.



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5 Example Calculations

In this example, the PLL frequency will be monitored with the main oscillator (OSCIN). The OSCIN will be selected as $Clock_0$ and a PLL will be selected as $Clock_1$.

1. Calculate the minimum Valid Counter using Equation 2. Since $T_{Clock1} > T_{Clock1}$

When
$$T_{Clock0} > T_{Clock1}$$
, $Valid_{Min} = 2 \times (2 + 3) = 10$.

2. Use the frequency accuracy (see Equation 3) to find the minimum duration for the sample.

With a frequency accuracy of 0.1%, this translates to $\frac{0.1}{100} = \frac{1}{1000} = \frac{1}{Re \ solution}$. Thus, the minimum duration of the sample is:

Duration_{Min} =
$$\frac{Valid_{Min} \times Re \, solution}{2^f Clock0}$$

$$Duration_{Min} = \frac{10 \times 1000}{2 \times 16 \, [MHz]}$$

$$Duration_{Min} = 312.5 \, \mu s$$
(5)

With the duration, it is easy to compute the Counter₀ and Counter₁ values.

$$T_{Clock1} \times Counter_{1} = \frac{Counter_{1}}{f_{Clock1}} = Duration$$

$$\frac{Counter_{1}}{160[MHz]} = 312.5 [\mu s]$$

$$Counter_{1} = 5 \times 10^{4}$$
(6)

and Counter₀ = $5 \times 10^3 - 5 = 4995$.

Summary of the Results:

With 16 MHz Clock₀ and 160 MHz Clock₁ and accuracy requirements of 0.1%:

- Valid_{Min} = 10
- Resolution = 1000
- DurationMin = 312.5 μs
- $(Counter_0)_{Min} = 4995$
- (Counter₁)_{Min} =50000



Code Configuration www.ti.com

6 Code Configuration

The configuration of the DCC is carried out in the *dccInit_demo* function. The function contains error checking that is not discussed in this application report.

Table 1. Code Configuration

The function takes five arguments:	1	unsigned int dccInit_demo(unsigned int CLK0_src, unsigned int CLK1_src, unsigned int CLK1_freq, unsigned int resolution)
 CLK0_src – Clock₀ signal; this example code is written for use on DCC1 with 	2	{
the LAUNCHXL2-RM46 (which uses	3	unsigned int valid0seed_val;
RM46L852) or LAUNCHXL2-	4	double long cnt0seed_val, cnt1seed_val, duration;
TMS57012 (which uses TMS570LS1224). Clock Source 0 can	5	
accept:	6	if(CLK1_freq >= CLK0_freq)
- OSCIN (0)	7	valid0seed_val = 10; // derived in accompanying application note
- HF LPO (5)	8	else
- TCK (0xA)	9	valid0seed_val = 2*((2*(CLK0_freg/CLK1_freg) + 1) +3);
For use with other devices, see	10	
the device-specific datasheet.	11	// duration is computed in accompanying application note.
This clock source is programmed in line 36.	12	duration = (double long) valid0seed_val*(double long)resolution/2;
• CLK1_src – Clock₁ signal; DCC1 for	13	
RM46L852 or TMS570LS1224 accepts	14	
Clock Source 0 can accept:	15	// CLK0_freq provides the number of CLK0 cycles to count.
– N2HET1[31]	16	// (CLK1_freq/CLK0_freq)*duration provides the number of CLK1 cycles to count.
– PLL1 (0)	17	
- PLL2	18	// in order to center the expiration of the CLK1 counter within the CLK0 counter window,
– LF HPO	19	// cnt0seed is programmed with the number of edges in one-half the valid counts.
- HF LPO	20	cnt0seed_val = duration - valid0seed_val/2;
- EXTCLKIN1	21	cnt1seed_val = (double long)CLK1_freq*duration/(double long)CLK0_freq - 1;
- EXTCLKIN2	22	
- VCLK	23	// Call HalCoGen function that sets count 0, count 1, and valid
For use with other devices, see the device-specific data sheet.	24	dccSetSeed(dccREG1, (unsigned int) cnt0seed_val, valid0seed_val, (unsigned int) cnt1seed_val);
This clock source is programmed in line 34.	25	
CLK0_freq – frequency of Clock ₀ in	26	// Generate error if PLL count expires outside of the valid window
Hertz. • CLK1_freq – frequency of Clock1 in	27	dccREG1->GCTRL = ((dccREG1->GCTRL & 0xFFFF0F0FU) dccNOTIFICATION_ERROR 0x5000);
Hertz.	28	
The frequency of Clock ₀ and Clock ₁ are	29	
compared in Line 6. Valid is computed on lines 7 and 9 as in Equation 2. The	30	// Configure clock sources
frequency of Clock ₀ and Clock ₁ are used	31	if(CLK1_src == 100)
to compute Counter ₀ and Counter ₁ in	32	dccREG1->CNT1CLKSRC = 0;
Lines 20 and 21. Resolution – frequency accuracy of the	33	else
measurement. The accuracy is passed	34	dccREG1->CNT1CLKSRC = (0xA000 CLK1_src);
1	35	
as Resolution. That is, a 1/2 %	36	dccREG1->CNT0CLKSRC = CLK0_src;
accuracy has a resolution of 1 part in 200, and Resolution is passed as 200.	37	
Resolution is required to compute the	38	// Call HalCoGen function to enable DCC1
duration in Line 12 following Equation 4.	39	dccEnable(dccREG1);
Counter, Valid, and Counter, are passed	40	
to dccSetSeed() in Line 24.	41	// returns the time duration over which the frequencies are compared.
The error response of the DCC is configured in Line 27.	42	return duration;
Line 38 enables the DCC.	43	}



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7 Example Code

The key purpose of the application report and example code is to demonstrate the DCC configuration. In order to demonstrate the DCC's monitoring capabilities, the code makes small adjustments to the PLL's frequency.

7.1 Hardware

The source code is written for use on either LAUNCHXL2-RM46 or LAUNCHXL2-TMS57012. This hardware provides two user controlled buttons: User Switch A and User Switch B.

- User Switch A increases the frequency of the PLL.
 - If the PLL is at its baseline frequency, then the increased frequency generates a DCC failure based on running too fast. (In terms of Figure 1, Counter₁ expires before Counter₀ and outside of the Valid window.)
 - If the PLL is running too slow, then the increased frequency returns the PLL to its baseline frequency. At its baseline frequency, the PLL expires within the Valid window and does not generate a DCC error.
 - If the PLL is running too fast, then User Switch A does not affect the PLL frequency since it is already too fast.
- User Switch B decreases the frequency of the PLL (similar to User Switch A)
 - If the PLL is at its baseline frequency, then the decreased frequency generates a DCC failure based on running too slow. (In terms of Figure 1, Counter₁ does not expire until after the Valid window.)
 - If the PLL is running too fast, then the decreased frequency returns the PLL to its baseline frequency. At its baseline frequency, the PLL expires within the Valid window and does not generate a DCC error.
 - If the PLL is running too slow, then User Switch A does not affect the PLL frequency since it is already too slow.

The code is written to be executed from flash, be sure to load the correct code into the device. The same source code is compiled for either RM46 or TMS570 and the output code is not interchangeable between the hardware.



Example Code www.ti.com

7.2 Software

The software is built from HalCoGen. Rather than using the dccInit function from Halcogen, the code develops the dccInit_demo function in order to define the relationship between Counter₀, Valid, and Counter₁. Outside of the DCC function, the software.

- Sits in an infinite loop, monitoring User Switch A and User Switch B
- Periodic interrupts are generated from the Real Time Interrupt Module (RTI) in order to toggle the LED

The PLL frequency is changed (based on User Switches A and B) by an amount that scales to the userdefined resolution. The frequency offset is computed as:

$$\Delta f = \frac{f}{\left(1 + \frac{Duration[inCLK0]}{Valid}\right)}$$
(7)

Table 2. Frequency Offset Computation

Resolution	Frequency Accuracy [%]	Frequency Offset [MHz]
10	10%	26.71
20	5%	14.56
50	2%	6.16
100	1%	3.14
200	0.5%	1.58
500	0.2%	0.64
1000	0.1%	0.32
2000	0.05%	0.16
5000	0.02%	0.06
10000	0.01%	0.03

NOTE: Code is written to clarify the relation to the DCC Concepts, not to optimize code execution.

While this application report has dealt with monitoring the PLL frequency with the DCC, the DCC can monitor other clock sources. If the application generates a 100KHz output on N2HET1[31], the frequency can be monitored by the DCC. In this case, the function dccInit_demo is called as dccInit_demo(0, 100, 16000000, 100000, resolution).

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