

AN-1956 LM5001 Boost Evaluation Board

1 Introduction

The LM5001 boost evaluation board is designed to provide the design engineer with a fully functional power converter based on the boost topology to evaluate the LM5001 high voltage switch mode regulator.

The performance of the evaluation board is as follows:

Input Operating Range: 16 to 36V

Output Voltage: 48V

Output Current: 0 to 150 mA

Measured Efficiency: 91% @ 150 mA, 86% @ 75 mA

Frequency of Operation: 240 kHz

Board Size: 1.75 X 1.75 inches

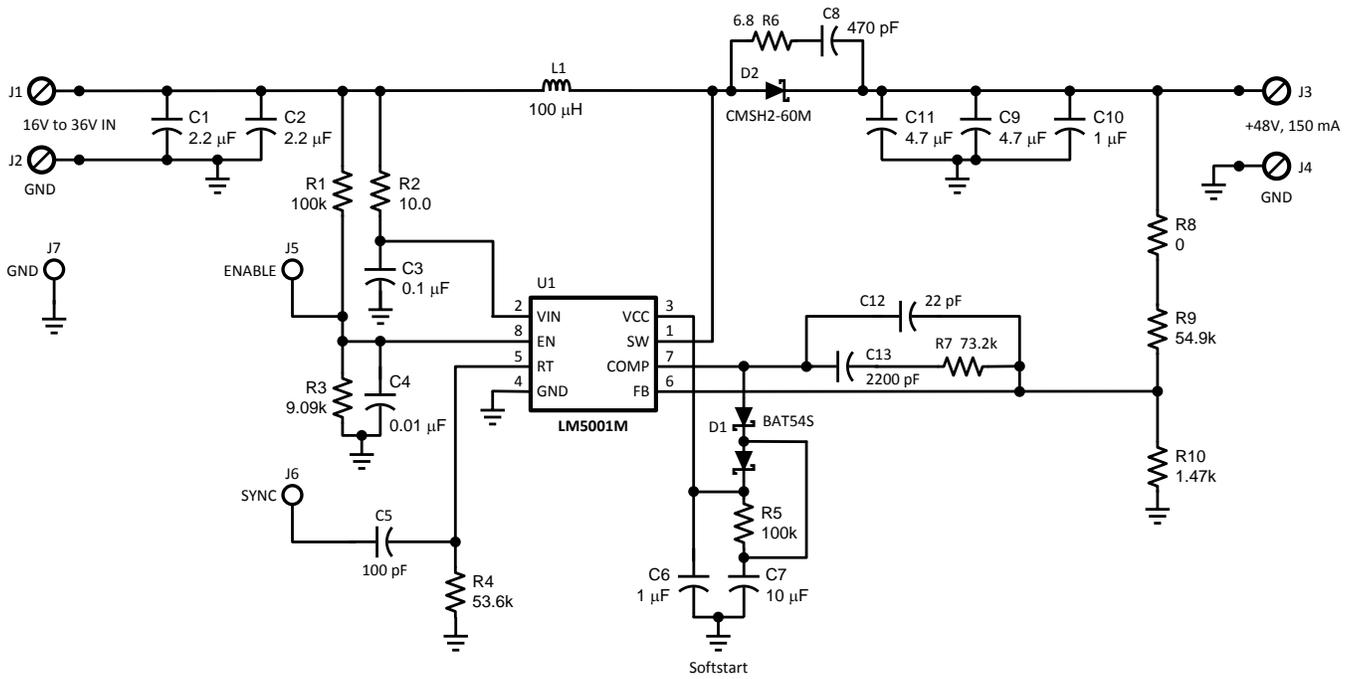
Load Regulation: 1%

Line Regulation: 0.1%

The printed circuit board consists of 2 layers; 1 ounce copper layers FR4 material with a total thickness of 0.062 inches.

When laying out the PCB note the proximity of the ground pin (pin 4) to the output capacitors (see the schematic in [Section 2](#)). Placing the ground pin near the output capacitor will minimize the ripple in the output by forcing a constant current to flow across the board for both the switch on and switch off portions of the cycle. If the board is laid out with the ground pin near the input capacitor then a high di/dt condition will occur due to the small conduction loop area during the switch on time and large loop conduction area during the switch off time. The output ripple and noise will be minimized if the conduction loop area and current both remain constant. Placing the ground pin near the output capacitor accomplishes this goal.

2 Schematic



3 Powering and Loading Considerations

When applying power to the LM5001 Boost evaluation board certain precautions need to be followed. A misconnection can damage the board.

3.1 Proper Connections

When operated at low input voltages the evaluation board can draw up to 500mA of current at full load. The maximum rated output current is 150mA. Be sure to choose the correct connector and wire size when attaching the source supply and the load. Monitor the current into and out of the evaluation board. Monitor the voltage directly at the output terminals of the evaluation board. The voltage drop across the load connecting wires will give inaccurate measurements. This is especially true for accurate efficiency measurements. When measuring output ripple with an oscilloscope. Do not use the wire ground lead for the ground connection. The loop formed by the wire lead will pick up noise from the switching circuits and make the ripple voltage look larger than it actually is. Instead use a spring ground clip on the exposed ground ring on the scope probe to minimize the loop area of the ground lead. An alternative is to remove the shroud covering the scope probe. Then touch the exposed scope probe ground connection to the output ground terminal while simultaneously connecting the probe tip to the output terminal.

3.2 Source Power

The power supply and cabling must look like a low impedance voltage source to the evaluation board. High inductance power supply leads like the type typically used for bench power supplies, could cause the LM5001 to become unstable or have poor response to load transients. This is due to the inductance of the power supply wiring interacting with the evaluation board input capacitor and causing a series resonant LC oscillation at a frequency defined by the inductance of the input wiring and the value of the input capacitor. In some cases it may be necessary to add an additional capacitor in parallel with input capacitor to move the resonate frequency away from the unity gain crossover frequency of the LM5001. Twisting the input supply lines together will reduce the inductance and potential for problems. Powering up at max rated voltage or close to this voltage can cause damage due to the inductance of the supply lines. Over shoot and ringing can be several volts under a sudden application of power. When operating near maximum input voltage slowly ramp up the voltage to avoid overshoot.

3.3 Loading

An appropriate electronic load, with specified operation up to 48V maximum or more, is desirable. Monitor both current and voltage at all times. Ensure there is sufficient cooling provided for the load.

3.4 Over Current Protection

The LM5001 monitors the peak current through the inductor on a cycle by cycle basis. If the inductor is sized large enough to not saturate when operating at peak current limit. Then the short circuit can be left on indefinitely with out damaging the device or causing it to go into thermal shutdown.

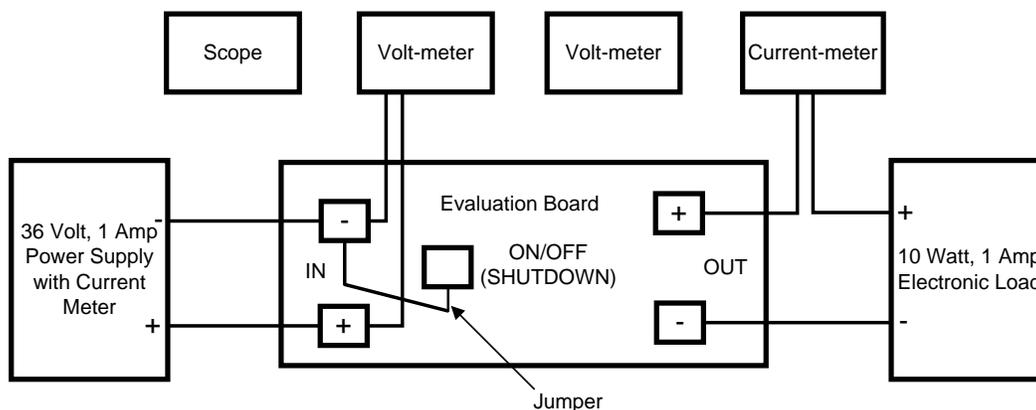


Figure 1. Typical Evaluation Setup

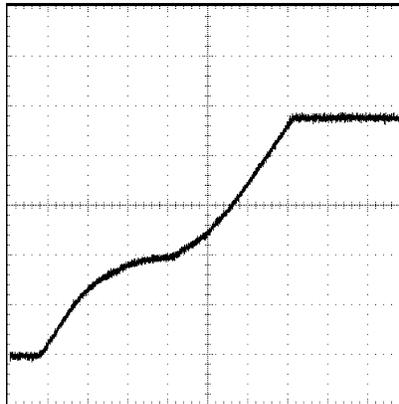
4 Performance Characteristics

Turn-on Waveforms

Figure 2 shows the output voltage during a typical start-up with a 20V input and a load of 150 mA. There is no overshoot during startup.

Output Ripple Waveforms

Figure 3 shows the transient response for a load of change from 15 mA to 150 mA. The upper trace shows minimal output voltage droop and overshoot during the sudden change in output current shown by the lower trace.



Conditions:

Input Voltage = 20VDC

Output Current = 150 mA

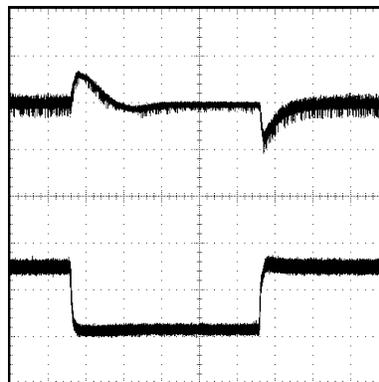
Trace 1:

Output Voltage

Volts/div = 10V

Horizontal Resolution = 4.0 ms/div

Figure 2. Output Voltage During a Typical Start-Up With a 20V Input and a Load of 150 mA



Conditions:

Input Voltage = 20VDC

Output Current = 15 mA to 150 mA

Upper Trace:

Output Voltage

Volts/div = 500 mV

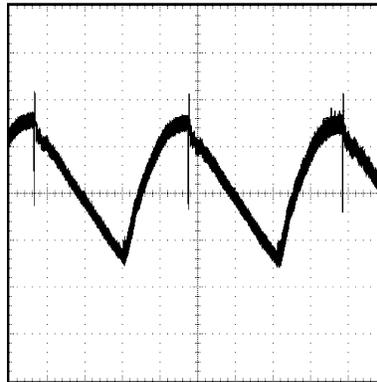
Lower Trace:

Output Current

150 mA to 15 mA to 150 mA

Horizontal Resolution = 0.4 ms/div

Figure 3. Transient Response for a Load of Change From 15 mA to 150 mA



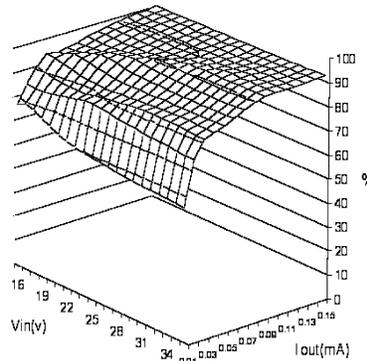
Conditions:
 Input Voltage = 20VDC
 Output Current = 150 mA
 Bandwidth Limit = 20 MHz
Trace 1:
 Output Voltage
 Volts/div = 20 mV
 Horizontal Resolution = 1 μs/div

Figure 4. Typical Output Ripple for an Input Voltage of 20V and a Load of 150 mA

Figure 4 shows typical output ripple, seen directly across the output capacitor, for an input voltage of 20V and a load of 150 mA. This waveform is typical of most loads and input voltages.

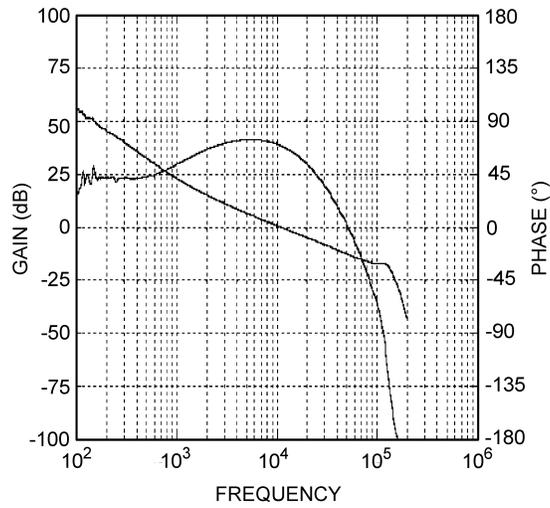
Figure 5 shows power efficiency over full input voltage and output current range. Peak efficiency is at full rated load and is greater than 90% across the input voltage range.

Figure 6 shows the small signal closed loop response with 20V input and 150 mA load current into a resistive load. The gain curve starts at around 60dB the phase curve starts at around 45°. 0dB of crossover frequency is at 11 kHz with a phase margin of 70°.



Conditions:
 Input Voltage = 16 - 36VDC
 Output Current = 10 mA - 150 mA

Figure 5. Power Efficiency Over Full Input Voltage and Output Current Range



Conditions:

Input Voltage = 20VDC

Output Current = 150 mA

Figure 6. Small Signal Closed Loop Response With 20V Input and 150 mA Load

5 Bill of Materials

Table 1. Bill of Materials

| Designator | Qty | Part Number | Description | Value |
|----------------|--------|--------------------|--|-------------------|
| C1, C2 | 2 | GRM31CR71H225KA88L | CAPACITOR, 1206 X7R CER, Murata | 2.2 μ F, 50V |
| C3 | 1 | C2012X7R1H104M | CAPACITOR, 0805 X7R CER, TDK | 0.1 μ F, 50V |
| C4 | 1 | C2012X7R1H103M | CAPACITOR, 0805 X7R CER, TDK | 0.01 μ F, 50V |
| C5 | 1 | C2012COG1H101J | CAPACITOR, 0805 COG CER, TDK | 100pF, 50V |
| C6 | 1 | C3216X7R1C105K | CAPACITOR, 0805 X7R CER, TDK | 1 μ F, 16V |
| C7 | 1 | GRM21BR61C106KE15L | CAPACITOR, 0805 X7R CER, Murata | 10 μ F, 16V |
| C8 | 1 | C2012COG1H471J | CAPACITOR, 0805 COG CER, TDK | 470pF, 100V |
| C9, C11 | 2 | C5750X7R2A475M | CAPACITOR, 2220 X7R CER, TDK | 4.7 μ F, 100V |
| C10 | 1 | C3225X7R2A105K | CAPACITOR, 1210 X7R CER, TDK | 1 μ F, 100V |
| C12 | 1 | C2012COG1H220J | CAPACITOR, 0805 COG CER, TDK | 22pF, 50V |
| C13 | 1 | C2012COG1H222J | CAPACITOR, 0805 COG CER, TDK | 2200pF, 50V |
| D1 | 1 | BAT54S | DIODE, SOT-23, DUAL, SCHOTTKY, Fairchild Semiconductor | 200mA, 30V |
| D2 | | CMSH2-60M | DIODE, SMA, SCHOTTKY, Central Semiconductor Corp. | 2A, 60V |
| L1 | 1 | MSS1260 | INDUCTOR, COILCRAFT | 100 μ H, 1.8A |
| R1, R5 | 2 | CRCW08051003F | RESISTOR, 0805, VISHAY | 100K |
| R2 | 1 | CRCW080510R0F | RESISTOR, 0805, VISHAY | 10 |
| R3 | 1 | CRCW08059091F | RESISTOR, 0805, VISHAY | 9.09K |
| R4 | 1 | CRCW08055362F | RESISTOR, 0805, VISHAY | 53.6K |
| R6 | 1 | CRCW080568R1F | RESISTOR, 0805, VISHAY | 6.8 |
| R7 | 1 | CRCW08057322F | RESISTOR, 0805, VISHAY | 73.2K |
| R8 | 1 | CRCW08050000F | RESISTOR, 0805, VISHAY | 0 |
| R9 | 1 | CRCW08055492F | RESISTOR, 0805, VISHAY | 54.9K |
| R10 | | CRCW08051471F | RESISTOR, 0805, VISHAY | 1.47K |
| J1, J2, J3, J4 | 4 | 7693 | Keystone Screw Terminal | |
| J5, J6, J7 | Mar-36 | PTC36SAAN | 0.025" Sq post, 36 position, Sullins | 3 posts used |
| U1 | 1 | LM5001 | High Voltage Switch Mode Regulator, Texas Instruments | |

6 Printed Circuit Layout

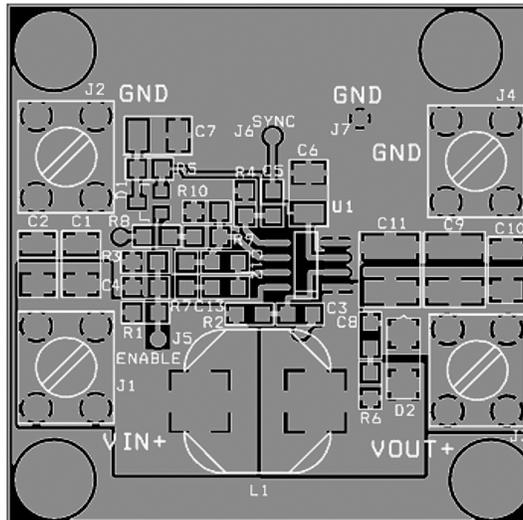


Figure 7. Silkscreen Layer

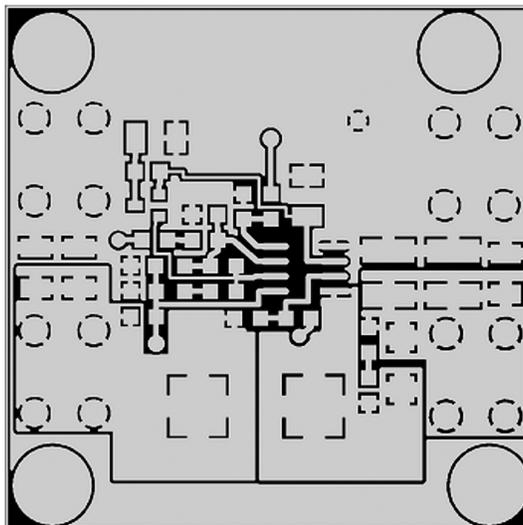


Figure 8. Top Layer

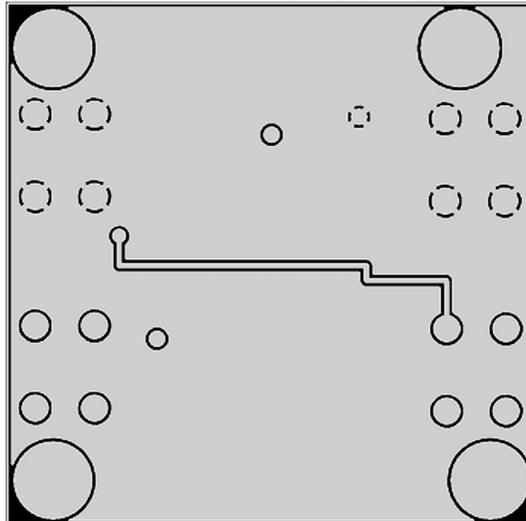


Figure 9. Bottom Layer

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