# TPS54302 Step-Down Converter Evaluation Module User's Guide



#### **ABSTRACT**

This user's guide contains background information for the TPS54302 as well as support documentation for the TPS54302EVM-716 evaluation module (PWR716-003). Also included are the performance specifications, the schematic, and the bill of materials for the TPS54302EVM-716.

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#### 1 Introduction

This user's guide contains background information for the TPS54302 as well as support documentation for the TPS54302EVM-716 evaluation module (PWR716-003). Also included are the performance specifications, the schematic, and the bill of materials for the TPS54302EVM-716.

#### 1.1 Background

The TPS54302 dc/dc converter is designed to provide up to a 3-A output from an input voltage source of 8 V to 28 V. Rated input voltage and output current range for the evaluation module are given in Table 1-1. In order to reduce EMI, the TPS54302 introduces frequency spread spectrum. The jittering span is ±6% of the switching frequency with 1/512 swing frequency. This evaluation module is designed to demonstrate the small printed-circuit-board areas that may be achieved when designing with the TPS54302 regulator. The switching frequency is internally set at a nominal 400 kHz. The high-side and low-side MOSFETs are incorporated inside the TPS54302 package along with the gate-drive circuitry. The low drain-to-source on resistance of the MOSFETs allow the TPS54302 to achieve high efficiencies and helps keep the junction temperature low at high output currents. The compensation components are integrated to the integrated circuit (IC), and an external divider allows for an adjustable output voltage. Additionally, the TPS54302 provides an adjustable undervoltage lockout input. The absolute maximum input voltage is 30 V for the TPS54302EVM-716.

Table 1-1. Input Voltage and Output Current Summary

EVM	Input Voltage Range	Output Current Range
TPS54302EVM-716	V <sub>IN</sub> = 8 V to 28 V	0 A to 3 A

## 1.2 Performance Specification Summary

A summary of the TPS54302EVM-716 performance specifications is provided in Table 1-2. Specifications are given for an input voltage of  $V_{\text{IN}}$  = 24 V and an output voltage of 5.0 V, unless otherwise specified. The TPS54302EVM-716 is designed and tested for  $V_{\text{IN}}$  = 8 V to 28 V. The ambient temperature is 25°C for all measurements, unless otherwise noted.

Table 1-2. TPS54302EVM-716 Performance Specification Summary

Specification	Test C	onditions	MIN	TYP	MAX	Unit
V <sub>IN</sub> operating voltage range			8	24	28	V
V <sub>IN</sub> start voltage				6.74		V
V <sub>IN</sub> stop voltage				5.83		V
Output voltage set point				5		V
Output current range	V <sub>IN</sub> = 8 V to 28 V		0		3	Α
Line regulation	I <sub>O</sub> = 1.5 A, V <sub>IN</sub> = 8 V to	I <sub>O</sub> = 1.5 A, V <sub>IN</sub> = 8 V to 28 V		±0.5%		
Load regulation	V <sub>IN</sub> = 12 V, I <sub>O</sub> = 0 A to 3	3 A		±0.5%		
	I <sub>O</sub> = 0.75 A to 2.25 A	Voltage change		-150		mV
Load transient recognic		Recovery time		150		μs
Load transient response	I <sub>O</sub> = 2.25 A to 0.75 A	Voltage change		150		mV
	10 - 2.25 A 10 0.75 A	Recovery time		150		μs
Input ripple voltage	I <sub>O</sub> = 3 A			400		$mV_{PP}$
Output ripple voltage	I <sub>O</sub> = 3 A	I <sub>O</sub> = 3 A		<30		$mV_{PP}$
Output rise time				5		ms
Center operating frequency				400		kHz
Maximum Efficiency	TPS54302EVM-716, V	TPS54302EVM-716, V <sub>IN</sub> = 12 V, I <sub>O</sub> = 1 A				

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#### 1.3 Modifications

These evaluation modules are designed to provide access to the features of the TPS54302. Some modifications can be made to this module.

#### 1.3.1 Output Voltage Set Point

The voltage divider, R2 and R3, is used to set the output voltage. To change the output voltage of the EVM, it is necessary to change the value of resistor R3. Changing the value of R3 can change the output voltage above 0.596 V. The value of R3 for a specific output voltage can be calculated using Equation 1. Use 100 k $\Omega$  for R2.

$$R3 = \frac{R2 \times 0.596 \text{ V}}{V_{\text{OUT}} - 0.596 \text{ V}}$$
(1)

Table 1-3 lists the R2 and R3 values for some common output voltages. Note that  $V_{IN}$  must be in a range so that the minimum on-time is greater than 150 ns. The values in Table 1-3 are standard values, not the exact value calculated using Equation 1.

Table 1-5. Recommended Component Values					
V <sub>OUT</sub> (V)	L (µH)	C <sub>OUT</sub> (µF)	R2 (kΩ)	R3 (kΩ)	C8 (pF)
1.8	4.7	66	100	49.9	33
2.5	5.6	66	100	31.6	47
3.3	6.8	44	100	22.1	47
5	10	44	100	13.3	75
12	15	44	100	5.23	100

**Table 1-3. Recommended Component Values** 

# 1.3.2 Output Capacitor and Feed-Forward Capacitor

Considering the loop stability and the effect of the internal parasitic parameters, choose a crossover frequency less than 40 kHz, without considering the feed-forward capacitor. A simple estimation for the crossover frequency without feed-forward capacitor C8 is shown in Equation 2, assuming C<sub>OUT</sub> has small ESR.

$$f_{O} = \frac{5.1}{V_{OUT} \times C_{OUT}}$$
 (2)

Depending on  $V_{OUT}$ , if the output capacitor,  $C_{OUT}$ , is dominated by low-ESR (ceramic types) capacitors, a low phase margin could result. To improve the phase boost, an external feed-forward capacitor, C8, can be added in parallel with R2. C8 is chosen such that phase margin is boosted at the crossover frequency.

C8 is calculated in Equation 3:

$$C8 = \frac{1}{2\pi f_o} \times \frac{1}{R_2} \tag{3}$$

For this design, C8 = 75 pF. C8 is not needed when  $C_{OUT}$  has high ESR, and C8 calculated from Equation 3 should be reduced with medium ESR. Use Table 1-3 as a starting point.

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#### 1.3.3 Adjustable UVLO

The under voltage lock out (UVLO) can be adjusted externally using R4 and R5. The EVM is set for a start voltage of 6.74 V and a stop voltage of 5.83 V using R4 = 510 k $\Omega$  and R5 = 105 k $\Omega$ . Use Equation 4 and Equation 5 to calculate required resistor values for different start and stop voltages. For higher light-load efficiency, consider choosing a larger R4 and R5. Make adjustments to V<sub>START</sub> or V<sub>STOP</sub> for a proper R4.

$$R4 = \frac{V_{START} \left( \frac{V_{ENFALLING}}{V_{ENRISING}} \right) - V_{STOP}}{I_{p} \left( 1 - \frac{V_{ENFALLING}}{V_{ENRISING}} \right) + I_{h}}$$
(4)

$$R5 = \frac{R4 \times V_{ENFALLING}}{V_{STOP} - V_{ENFALLING} + R4 (I_p + I_h)}$$
(5)

 $I_{\text{p}}$  = 0.7  $\mu\text{A},~I_{\text{h}}$  = 1.55  $\mu\text{A},~V_{\text{ENFALLING}}$  = 1.19 V and  $V_{\text{ENRISING}}$  = 1.22 V

## 2 Test Setup and Results

This section describes how to properly connect, set up, and use the TPS54302EVM-716 evaluation module. The section also includes test results typical for the evaluation module and covers efficiency, output voltage regulation, load transients, loop response, output ripple, input ripple, and start-up.

## 2.1 Input/Output Connections

The TPS54302EVM-716 is provided with input/output connectors and test points as shown in Table 2-1. A power supply capable of supplying 3 A must be connected to J1 through a pair of 20-AWG wires. The load must be connected to J2 through a pair of 20-AWG wires. The maximum load current capability must be at least 4 A to use the full capability of this EVM. Wire lengths must be minimized to reduce losses in the wires. Test-point TP1 provides a place to monitor the  $V_{IN}$  input voltages with TP2 providing a convenient ground reference. TP6 is used to monitor the output voltage with TP7 as the ground reference.

Reference Designator **Function** VIN (see Table 1-1 for V<sub>IN</sub> range) J1 J2 VOUT, 5 V at 3 A maximum JP1 2-pin header for enable. Connect EN to ground to disable, open to enable. TP1 VIN test point at VIN connector GND test point at VIN TP2 TP3 GND test point TP4 SW test point TP5 Test point between voltage divider network and output. Used for loop response measurements. TP6 Output voltage test point at OUT connector TP7 GND test point at VOUT connector

**Table 2-1. EVM Connectors and Test Points** 

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## 2.2 Efficiency

The efficiency of this EVM peaks at a load current of about 0.5 A to 1 A, and then decreases as the load current increases towards full load. Figure 2-1 shows the efficiency for the TPS54302EVM-716 at an ambient temperature of 25°C.

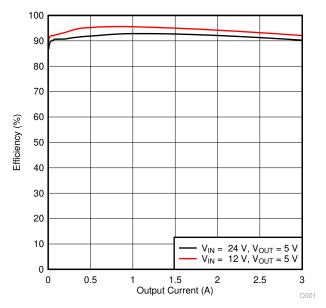


Figure 2-1. TPS54302EVM-716 Efficiency

Figure 2-2 shows the efficiency for the TPS54302EVM-716 on a semi-log scale to better show light load efficiency. The ambient temperature is 25°C.

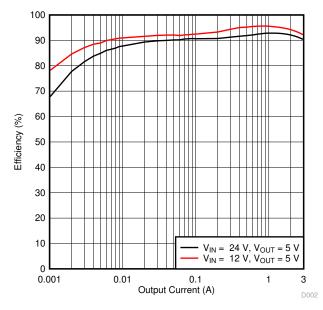


Figure 2-2. TPS54302EVM-716 Low Current Efficiency

The efficiency may be lower at higher ambient temperatures, due to temperature variation in the drain-to-source resistance of the internal MOSFET.

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## 2.3 Output Voltage Load Regulation

Figure 2-3 shows the load regulation for the TPS54302EVM-716.

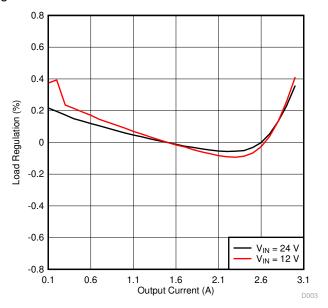


Figure 2-3. TPS54302EVM-716 Load Regulation

Measurements are given for an ambient temperature of 25°C.

## 2.4 Output Voltage Line Regulation

Figure 2-4 shows the line regulation for the TPS54302EVM-716.

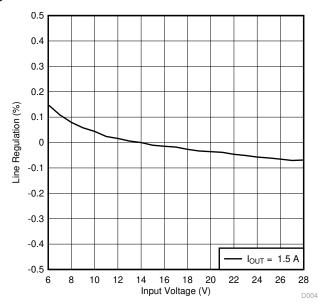


Figure 2-4. TPS54302EVM-716 Line Regulation

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## 2.5 Load Transients

Figure 2-5 shows the TPS54302EVM-716 response to load transients. The current step is from 25% to 75% of maximum rated load at 24-V input. Total peak-to-peak voltage variation is as shown, including ripple and noise on the output.

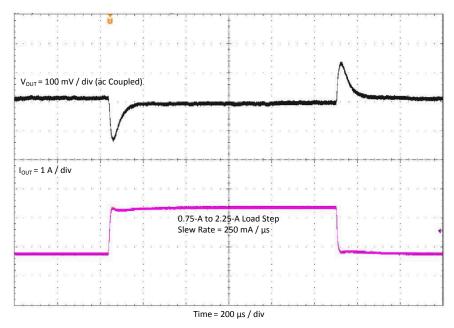


Figure 2-5. TPS54302EVM-716 Transient Response

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## 2.6 Output Voltage Ripple

Figure 2-6, Figure 2-7, Figure 2-8, and Figure 2-9 show the TPS54302EVM-716 output voltage ripple for full-load, skip-mode, light-load and no-load operation.  $V_{IN}$  = 24 V. The output The ripple voltage is measured directly across the output capacitors.

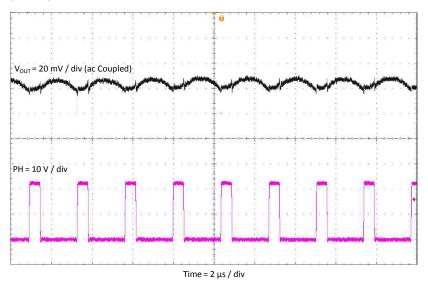


Figure 2-6. TPS54302EVM-716 Output Ripple, I<sub>OUT</sub> = 3 A

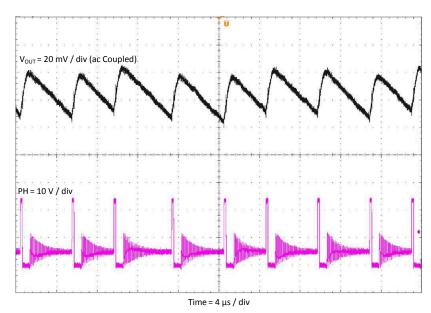


Figure 2-7. TPS54302EVM-716 Output Ripple, I<sub>OUT</sub> = 100 mA



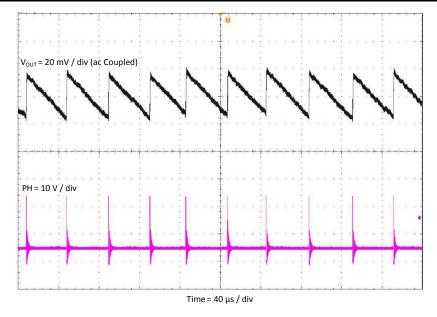


Figure 2-8. TPS54302EVM-716 Output Ripple, I<sub>OUT</sub> = 10 mA

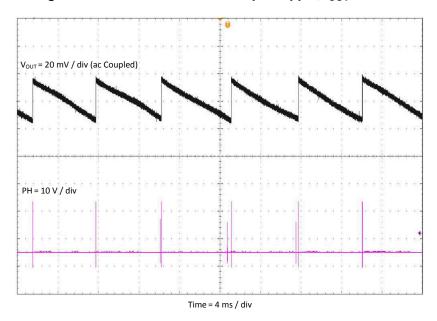


Figure 2-9. TPS54302EVM-716 Output Ripple, I<sub>OUT</sub> = 0 A

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# 2.7 Input Voltage Ripple

Figure 2-10 shows the TPS54302EVM-716 input voltage ripple. The output current is the rated full load of 3 Å and  $V_{IN}$  = 24 V. The ripple voltage is measured directly across the input capacitors.

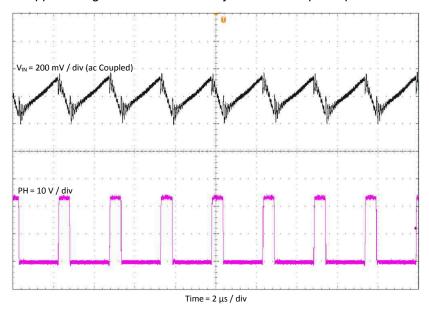


Figure 2-10. TPS54302EVM-716 Input Ripple

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## 2.8 Powering Up

Figure 2-11 and Figure 2-12 show the start-up waveforms for the TPS54302EVM-716. In Figure 2-11, the output voltage ramps up as soon as the input voltage reaches the UVLO threshold as set by the R4 and R5 resistor divider network. In Figure 2-12, the input voltage is initially applied and the output is inhibited by using a 3.3-V logic signal between EN and GND. When the EN voltage reaches the enable-threshold voltage, the start-up sequence begins and the output voltage ramps up to the externally set value of 5 V. The input voltage for these plots is 24 V and the load is  $3.3~\Omega$ .

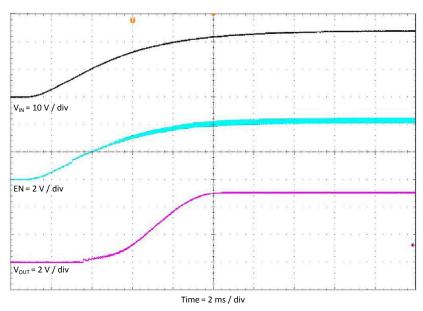


Figure 2-11. TPS54302EVM-716 Startup Relative to VIN

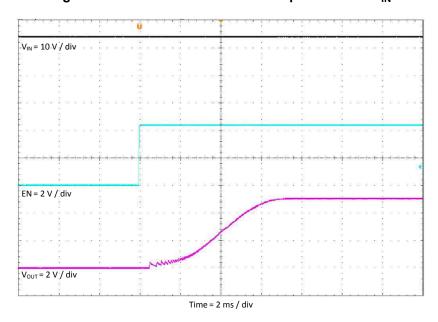


Figure 2-12. TPS54302EVM-716 Startup Relative to Enable

Test Setup and Results

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## 2.9 Powering Down

Figure 2-13 and Figure 2-14 show the start-up waveforms for the TPS54302EVM-716. In Figure 2-13, the output voltage ramps down as soon as the input voltage falls below the UVLO stop threshold as set by the R4 and R5 resistor divider network. In Figure 2-14, the output is inhibited by using a 3.3-V logic signal between EN and GND. The input voltage for these plots is 24 V and the load is  $3.3~\Omega$ .

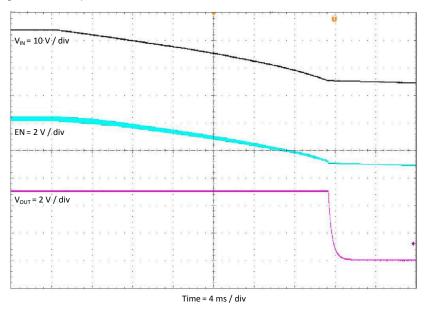


Figure 2-13. TPS54302EVM-716 Shutdown Relative to  $V_{\text{IN}}$ 

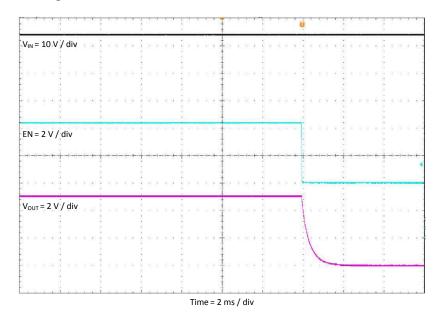


Figure 2-14. TPS54302EVM-716 Shutdown Relative to EN

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## 3 Board Layout

This section provides a description of the TPS54302EVM-716, board layout, and layer illustrations.

#### 3.1 Layout

Figure 3-1 and Figure 3-2 show the board layout for the TPS54302EVM-716. The topside layer of the EVM is laid out in a manner typical of a user application. The top and bottom layers are 2-oz. copper.

The top layer contains the main power traces for  $V_{IN}$ ,  $V_{OUT}$ , and SW. Also on the top layer are connections for the remaining pins of the TPS54302 and a large area filled with ground. To facilitate the placement of the main input bypass capacitor as close to the  $V_{IN}$  and GND pins as possible, the trace for SW is routed to the bottom layer immediately at the pin 3 connection. It is routed back to the top layer at the L1 inductor and C4 BOOT capacitor. The bottom layer contains a ground plane plus a copper fill area for SW, an etch run to connect the upper resistor of the voltage set point divider to the regulation point at the J2 output connector, and a trace to connect the upper resistor of the UVLO set point divider network to  $V_{IN}$ . The top-side ground areas are connected to the bottom and internal ground planes with multiple vias placed around the board to provide a thermal path from the top-side ground area to the bottom-side and internal ground planes.

The input decoupling capacitors (C2, and C1) and bootstrap capacitor (C4) are all located as close to the IC as possible. In addition, the voltage set-point resistor divider components are also kept close to the IC. For the TPS54302, an additional input bulk capacitor may be required, depending on the EVM connection to the input supply.

Board Layout www.ti.com

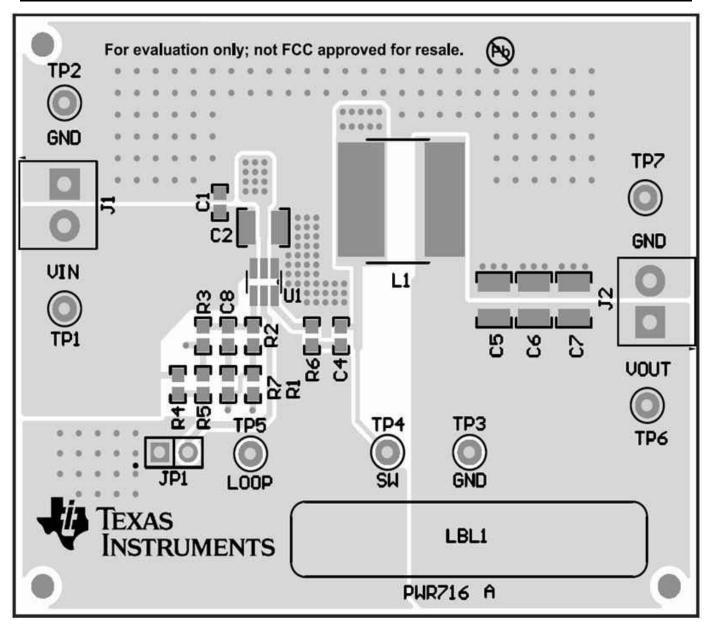


Figure 3-1. TPS54302EVM-716 Top-Side Assembly



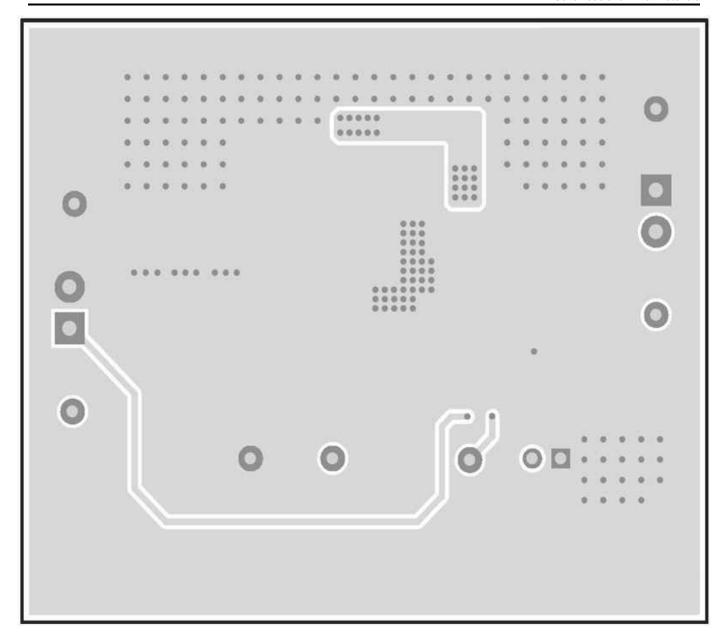


Figure 3-2. TPS54302EVM-716 Bottom-Side Layout

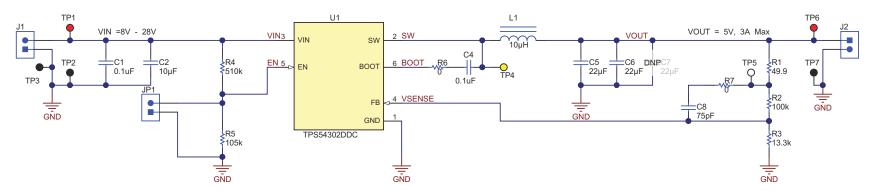
# 4 Schematic and Bill of Materials

This section presents the TPS54302EVM-716 schematic and bill of materials.



#### 4.1 Schematic

Figure 4-1 is the schematic for the TPS54302EVM-716.



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Figure 4-1. TPS54302EVM-716 Schematic

## 4.2 Bill of Materials

Table 4-1 presents the bill of materials for the TPS54302EVM-716.

Table 4-1. TPS54302EVM-716 Bill of Materials

Designator	Qty	Value	Description	PackageReference	PartNumber	Manufacturer
C1, C4	2	0.1uF	CAP, CERM, 0.1uF, 25V, +/-10%, X5R, 0603	0603	GRM188R61E104KA01D	Murata
C2	1	10uF	CAP, CERM, 10 µF, 35 V, +/- 10%, X7R, 1210	1210	GRM32ER7YA106KA12L	Murata
C5, C6	2	22uF	CAP, CERM, 22 µF, 25 V, +/- 10%, X7R, 1210	1210	GRM32ER71E226KE15L	Murata
C8	1	75pF	CAP, CERM, 75 pF, 50 V, +/- 5%, C0G/NP0, 0603	0603	GRM1885C1H750JA01D	Murata
J1, J2	2		Terminal Block, 6A, 3.5mm Pitch, 2-Pos, TH	7.0x8.2x6.5mm	ED555/2DS	On-Shore Technology
JP1	1		Header, 100mil, 2x1, Gold, TH	2x1 Header	TSW-102-07-G-S	Samtec
L1	1	10uH	Inductor, Shielded Drum Core, Ferrite, 10 µH, 4.3 A, 0.023 ohm, SMD	10x5x10mm	7447714100	Wurth Elektronik
R1	1	49.9	RES, 49.9, 1%, 0.1 W, 0603	0603	CRCW060349R9FKEA	Vishay-Dale
R2	1	100k	RES, 100 k, 1%, 0.1 W, 0603	0603	CRCW0603100KFKEA	Vishay-Dale
R3	1	13.3k	RES, 13.3 k, 1%, 0.1 W, 0603	0603	CRCW060313K3FKEA	Vishay-Dale
R4	1	510k	RES, 510 k, 5%, 0.1 W, 0603	0603	CRCW0603510KJNEA	Vishay-Dale
R5	1	105k	RES, 105 k, 1%, 0.1 W, 0603	0603	CRCW0603105KFKEA	Vishay-Dale
R6, R7	2	0	RES, 0 ohm, 5%, 0.1W, 0603	0603	ERJ-3GEY0R00V	Panasonic
TP1, TP6	2	Red	Test Point, Miniature, Red, TH	Red Miniature Testpoint	5000	Keystone
TP2, TP3, TP7	3	Black	Test Point, Miniature, Black, TH	Black Miniature Testpoint	5001	Keystone
TP4	1	Yellow	Test Point, Miniature, Yellow, TH	Yellow Miniature Testpoint	5004	Keystone



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# Table 4-1. TPS54302EVM-716 Bill of Materials (continued)

Designator	Qty	Value	Description	PackageReference	PartNumber	Manufacturer
TP5	1	White	Test Point, Miniature, White, TH	White Miniature Testpoint	5002	Keystone
U1	1		4.5-V TO 28-V INPUT, 3-A OUTPUT, EMI FRIENDLY SYNCHRONUOUS STEP-DOWN CONVERTOR, DDC0006A	DDC0006A	TPS54302DDC	Texas Instruments
C7	0	22uF	CAP, CERM, 22 μF, 25 V, +/- 10%, X7R, 1210	1210	GRM32ER71E226KE15L	Murata

Revision History www.ti.com

# **5 Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision A (September 2016) to Revision B (October 2021)	Page
• Updated the numbering format for tables, figures, and cross-references throughout the document	<mark>2</mark>
Updated the user's guide title	2
Changes from Revision * (May 2016) to Revision A (September 2016)	Page
Switched 'BOOT' and 'SW' pin labels in the schematic	16

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