High-Side Current-Sensing Circuit Design with MSP430™ Smart Analog Combo



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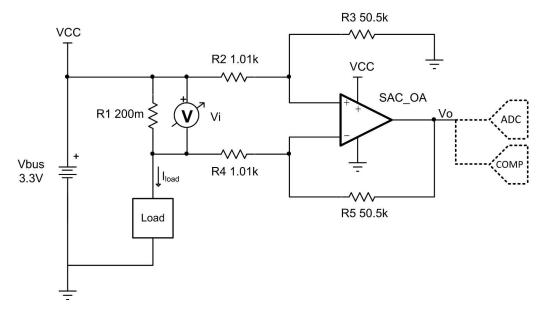
Design Goals

Input		Output		Supply	
l _{iMin}	I _{iMax}	V _{oMin}	V _{oMax}	V _{cc}	V _{ee}
25mA	300mA	0.25V	3V	3.3V	0V

Design Description

Some MSP430™ microcontrollers (MCUs) contain configurable integrated signal chain elements such as opamps, DACs, and programmable gain stages. These elements make up a peripheral called the smart analog combo (SAC). For information on the different types of SACs and how to leverage configurable analog signal chain capabilities, visit *MSP430 MCUs Smart Analog Combo Training*. To get started with your design, download the *High-Side Current Sensing Circuit Design Files*.

This single-supply, high-side, low-cost current sensing option detects load current between 25mA and 300mA and converts it to an output voltage from 0.25V to 3V. High-side sensing allows for the system to identify ground shorts and does not create a ground disturbance on the load. The circuit uses the MSP430FR2311 SAC_L1 op-amp in general-purpose (GP) mode with OAx+ and OAx- dedicated as noninverting and inverting inputs. The same approach can be implemented with the MSP430FR2355, featuring four SAC_L3 peripherals with additional built-in DAC and PGA capabilities. The output of the integrated SAC op-amp can be sampled directly by the on-board ADC or monitored by the on-board comparator for further processing inside the MCU.



Design Notes

- DC common-mode rejection ratio (CMRR) performance is dependent on the matching of the gain setting resistors, R₂-R₅.
- Increasing the shunt resistor increases power dissipation.
- Verify that the common-mode voltage is within the linear input operating region of the amplifier. The common-mode voltage is set by the resistor divider formed by R₂, R₃, and the bus voltage. Depending on the common-mode voltage determined by the resistor divider a rail-to-rail input (RRI) amplifier may not be required for this application.
- An op amp that does not have a common-mode voltage range that extends to V_{cc} can be used in low-gain or an attenuating configuration.
- A capacitor placed in parallel with the feedback resistor limits bandwidth, improves stability, and reduces noise.
- Use the op amp in a linear output operating region. Linear output swing is usually specified under the A_{OL} test conditions.
- If the process is implemented with the MSP430FR2311 SAC_L1 or with the MSP430FR2355 SAC_L3, the op-amp is configured in general-purpose mode.
- If the process is implemented using the MSP430FR2311 TIA, the input voltage range is limited to V_{CC}/2, so
 the gain or range must be adjusted accordingly.
- The High-Side Current Sensing Circuit Design Files include code examples showing how to properly initialize the SAC peripherals.

Design Steps

1. The full transfer function of the circuit is provided below.

$$\begin{aligned} &V_0 = I_{in} \times R_1 \times \frac{R_5}{R_4} \\ &\text{Given} \quad R_2 = R_4 \quad \text{and} \quad R_3 = R_5 \end{aligned}$$

2. Calculate the maximum shunt resistance. Set the maximum voltage across the shunt to 60mV.

$$R_1 = \frac{V_{iMax}}{I_{iMax}} = \frac{60 \text{mV}}{300 \text{mA}} = 200 \text{m}\Omega$$

3. Calculate the gain to set the maximum output swing range.

$$Gain = \frac{V_{0Max} - V_{0Min}}{(I_{iMax} - I_{iMin}) \times R_1} = \frac{3V - 0.25V}{(0.3A - 0.025A) \times 200m\Omega} = 50\frac{V}{V}$$

4. Calculate the gain setting resistors to set the gain calculated in step 3.

Choose
$$R_2=R_4=1.01 k\,\Omega$$
 (Standard value)
$$R_3=R_5=R_2\times Gain=1.01 k\,\Omega\,\times 50 \frac{V}{V}=50.5 k\,\Omega \, \mbox{(Standard value)}$$

5. Calculate the common-mode voltage of the amplifier to verify linear operation.

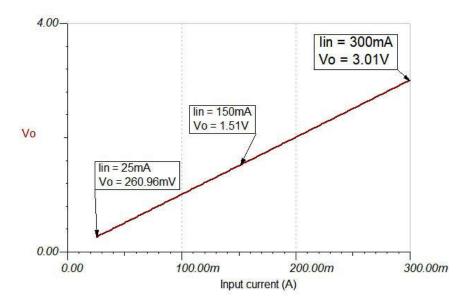
$$V_{cm} = V_{CC} \times \frac{R_3}{R_2 + R_3} = 3.3V \times \frac{50.5k}{1.01k + 50.5k} = 3.235V$$

6. The upper cutoff frequency (f_H) is set by the non-inverting gain (noise gain) of the circuit and the gain bandwidth (GBW) of the op amp.

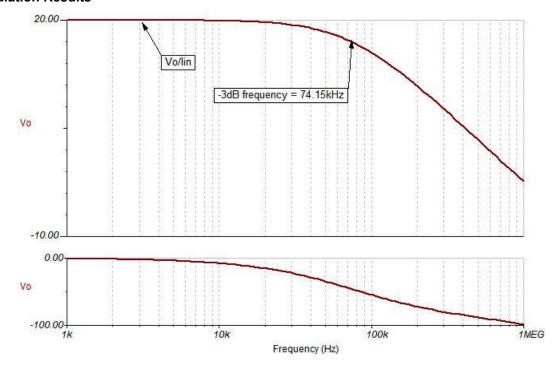
$$f_H = \frac{GBW}{Noise Gain} = \frac{4MHz}{51\frac{V}{V}} = 78.43 \text{ kHz}$$

Design Simulations

DC Simulation Results



AC Simulation Results



Target Applications

- Battery pack: cordless power tool
- HEV/EV battery-management system (BMS)
- Motor drives
- Lighting
- Energy infrastructure

References

- 1. Texas Instruments, High-Side Current Sensing Circuit, design files
- 2. Texas Instruments, MSP430FR2311 TINA-TI Spice Model, file download
- 3. Texas Instruments, MSP430 MCUs Smart Analog Combo, training video

Design Featured Op Amp

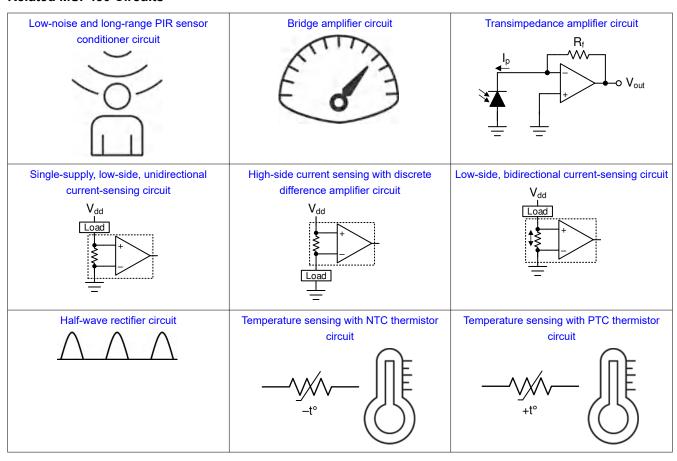
MSP430FRxx Smart Analog Combo						
	MSP430FR2311 SAC_L1	MSP430FR2355 SAC_L3				
V _{cc}	2.0V to 3.6V					
V _{CM}	-0.1V to V _{CC} + 0.1V					
V _{out}	Rail-to-rail					
V _{os}	±5mV					
A _{OL}	100dB					
	350μA (high-speed mode)					
Iq	120μA (low-power mode)					
l _b	50pA					
UGBW	4MHz (high-speed mode)	2.8MHz (high-speed mode)				
OGDII	1.4MHz (low-power mode)	1MHz (low-power mode)				
SR	3V/µs (high-speed mode)					
JK	1V/μs (low-power mode)					
Number of channels	1	4				
	MSP430FR2311	MSP430FR2355				

Design Alternate Op Amp

MSP430FR2311 Transimpedance Amplifier				
V _{cc}	2.0V to 3.6V			
V _{CM}	-0.1V to V _{CC} /2V			
V _{out}	Rail-to-rail			
V _{os}	±5mV			
A _{OL}	100dB			
	350μA (high-speed mode)			
Iq	120μA (low-power mode)			
	5pA (TSSOP-16 with OA-dedicated pin input)			
l _b	50pA (TSSOP-20 and VQFN-16)			
UGBW	5MHz (high-speed mode)			
UGBW	1.8MHz (low-power mode)			
SR	4V/µs (high-speed mode)			
SK .	1V/μs (low-power mode)			
Number of channels	1			
	MSP430FR2311			

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Related MSP430 Circuits



Trademarks

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Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (March 2020) to Revision C (October 2024)				
Updated the format for tables, figures, and cross-references throughout the document				
Changes from Revision A (November 2019) to Revision B (March 2020)	Page			
Added Related MSP430 Circuits section	1			
Changes from Revision * (November 2019) to Revision A (November 2019)	Page			

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