

2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

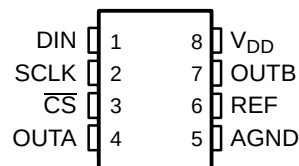
FEATURES

- Dual 12-Bit Voltage Output DAC
- Programmable Internal Reference
- Programmable Settling Time:
 - 1 μ s in Fast Mode,
 - 3.5 μ s in Slow Mode
- Compatible With TMS320 and SPI™ Serial Ports
- Differential Nonlinearity <0.5 LSB Typ
- Monotonic Over Temperature

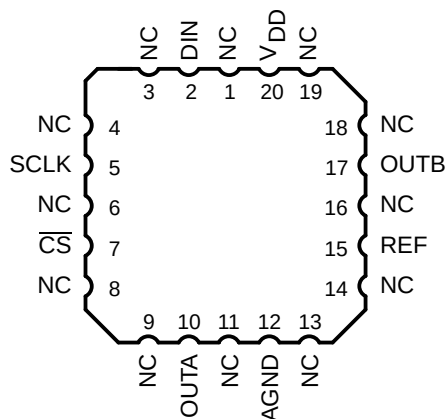
APPLICATIONS

- Digital Servo Control Loops
- Digital Offset and Gain Adjustment
- Industrial Process Control
- Machine and Motion Control Devices
- Mass Storage Devices

D, JG PACKAGE
(TOP VIEW)



FK PACKAGE
(TOP VIEW)



DESCRIPTION

The TLV5638 is a dual 12-bit voltage output DAC with a flexible 3-wire serial interface. The serial interface allows glueless interface to TMS320, SPI™, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing 4 control and 12 data bits.

The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed vs power dissipation. With its on-chip programmable precision voltage reference, the TLV5638 simplifies overall system design.

Because of its ability to source up to 1 mA, the reference can also be used as a system reference. Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package to reduce board space in standard commercial, industrial, and automotive temperature ranges. It is also available in JG and FK packages in the military temperature range.



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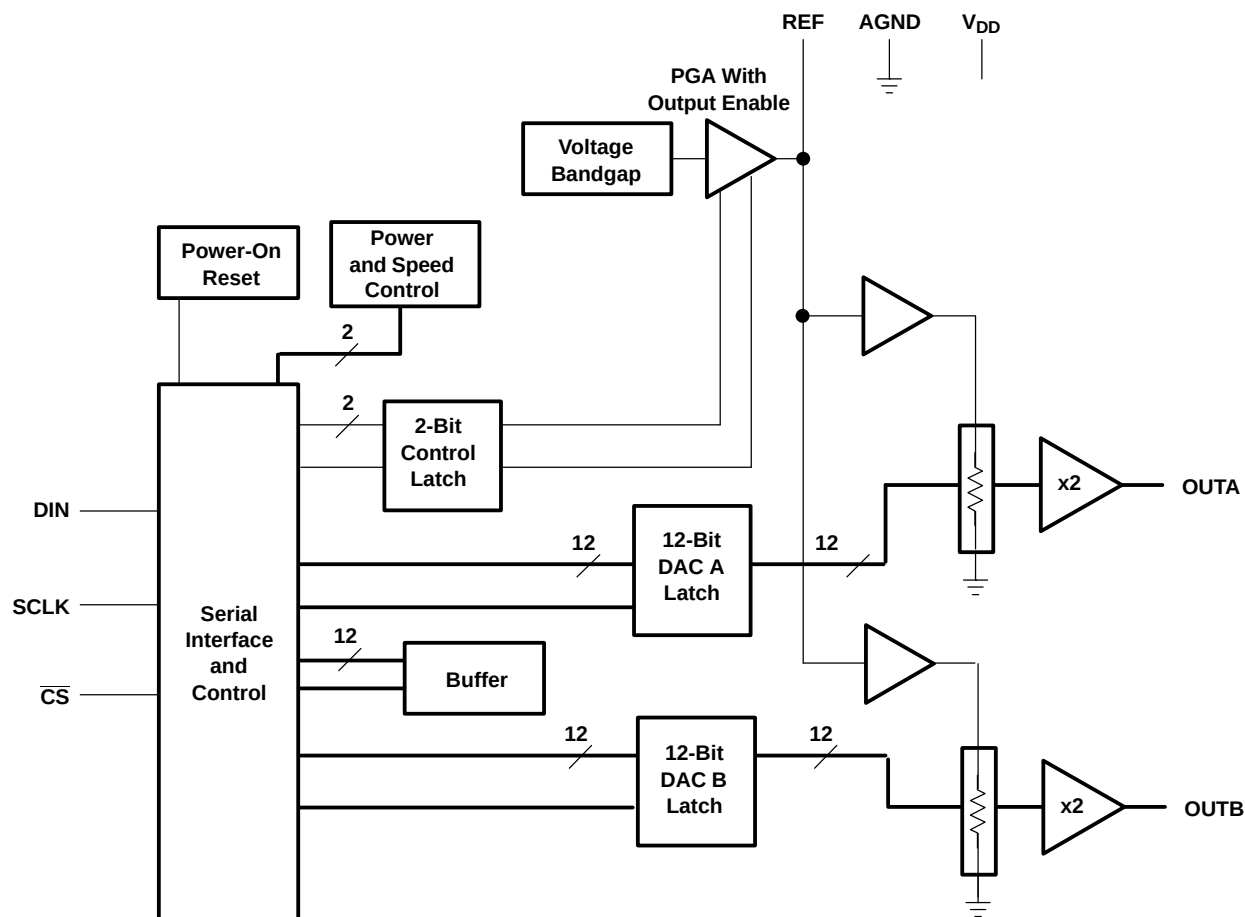
This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

AVAILABLE OPTIONS

T _A	PACKAGE		
	SOIC (D)	CERAMIC DIP (JG)	20 PAD LCCC (FK)
0°C to 70°C	TLV5638CD	—	—
40°C to 85°C	TLV5638ID	—	—
40°C to 125°C	TLV5638QD TLV5638QDR	—	—
55°C to 125°C	—	TLV5638MJG	TLV5638MFK

FUNCTIONAL BLOCK DIAGRAM



Terminal Functions

TERMINAL NAME	NO.	I/O/P	DESCRIPTION
AGND	5	P	Ground
$\overline{\text{CS}}$	3	I	Chip select. Digital input active low, used to enable/disable inputs
DIN	1	I	Digital serial data input
OUT A	4	O	DAC A analog voltage output
OUT B	7	O	DAC B analog voltage output
REF	6	I/O	Analog reference voltage input/output
SCLK	2	I	Digital serial clock input
V _{DD}	8	P	Positive power supply

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		UNIT
Supply voltage (V _{DD} to AGND)		7 V
Reference input voltage range		-0.3 V to V _{DD} + 0.3 V
Digital input voltage range		-0.3 V to V _{DD} + 0.3 V
Operating free-air temperature range, T _A	TLV5638C	0°C to 70°C
	TLV5638I	-40°C to 85°C
	TLV5638Q	-40°C to 125°C
	TLV5638M	-55°C to 125°C
Storage temperature range, T _{stg}		-65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds		260°C

- (1) Stresses beyond those listed under „absolute maximum ratings“ may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under „recommended operating conditions“ is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

DISSIPATION RATING TABLE

PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C ⁽¹⁾	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
D	635 mW	5.08 mW/°C	407 mW	330 mW	127 mW
FK	1375 mW	11.00 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.40 mW/°C	672 mW	546 mW	210 mW

- (1) This is the inverse of the traditional Junction-to-Ambient thermal Resistance (R_{θJA}). Thermal Resistances are not production tested and are for informational purposes only.

RECOMMENDED OPERATING CONDITIONS

			MIN	NOM	MAX	UNIT
Supply voltage, V_{DD}	$V_{DD} = 5\text{ V}$		4.5	5	5.5	V
	$V_{DD} = 3\text{ V}$		2.7	3	3.3	V
Power on reset, POR			0.55 ⁽¹⁾		2 ⁽¹⁾	V
High-level digital input voltage, V_{IH}	$V_{DD} = 2.7\text{ V}$		2			V
	$V_{DD} = 5.5\text{ V}$		2.4			
Low-level digital input voltage, V_{IL}	$V_{DD} = 2.7\text{ V}$				0.6	V
	$V_{DD} = 5.5\text{ V}$	TLV5638C and TLV5638I			1	V
		TLV5638Q and TLV5638M			0.8	
	$V_{DD} = 5.5\text{ V}$					
Reference voltage, V_{ref} to REF terminal	$V_{DD} = 5\text{ V}$ ⁽²⁾		AGND	2.048	$V_{DD}-1.5$	V
Reference voltage, V_{ref} to REF terminal	$V_{DD} = 3\text{ V}$ ⁽²⁾		AGND	1.024	$V_{DD}-1.5$	V
Load resistance, R_L			2			k Ω
Load capacitance, C_L					100	pF
Clock frequency, f_{CLK}					20	MHz
Operating free-air temperature, T_A	TLV5638C		0		70	°C
	TLV5638I		40		85	
	TLV5638Q		40		125	
	TLV5638M		55		125	

(1) This parameter is not tested for Q and M suffix devices.

(2) Due to the x2 output buffer, a reference input voltage $\geq (V_{DD}-0.4\text{ V})/2$ causes clipping of the transfer function. The output buffer of the internal reference must be disabled, if an external reference is used.

ELECTRICAL CHARACTERISTICS

over recommended operating conditions, $V_{ref} = 2.048\text{ V}$, $V_{ref} = 1.024\text{ V}$ (unless otherwise noted)

POWER SUPPLY							
PARAMETER		TEST CONDITIONS			TLV5638C, I TLV5638M		UNIT
					MIN	TYP	
I _{DD}	Power supply current	No load, All inputs = AGND or V _{DD} , DAC latch = 0x800	V _{DD} = 5 V, Int. ref.	Fast	4.3	7	mA
				Slow	2.2	3.6	
			V _{DD} = 3 V, Int. ref.	Fast	3.8	6.3	mA
				Slow	1.8	3.0	
			V _{DD} = 5 V, Ext. ref.	Fast	3.9	6.3	mA
				Slow	1.8	3.0	
			V _{DD} = 3 V, Ext. ref.	Fast	3.5	5.7	mA
				Slow	1.5	2.6	
Power-down supply current					0.01	10	μA
PSRR	Power supply rejection ratio	Zero scale, ⁽¹⁾			65	dB	
		Full scale, ⁽²⁾			65		

(1) Power supply rejection ratio at zero scale is measured by varying V_{DD} and is given by: $PSRR = 20 \log [(E_{ZS}(V_{DDmax}) - E_{ZS}(V_{DDmin}))/V_{DDmax}]$

(2) Power supply rejection ratio at full scale is measured by varying V_{DD} and is given by: $PSRR = 20 \log [(E_G(V_{DDmax}) - E_G(V_{DDmin}))/V_{DDmax}]$

ELECTRICAL CHARACTERISTICS (Continued)

over recommended operating conditions, $V_{ref} = 2.048\text{ V}$, $V_{ref} = 1.024\text{ V}$ (unless otherwise noted)

STATIC DAC SPECIFICATIONS							
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
Resolution				12			bits
INL	Integral nonlinearity, end point adjusted	See (1)	C and I suffixes	±1.7		±4	LSB
			Q and M suffixes	±1.7		±6	LSB
DNL	Differential nonlinearity	See (2)		±0.4		±1	LSB
E _{ZS}	Zero-scale error (offset error at zero scale)	See (3)				±24	mV
E _{ZS} TC	Zero-scale-error temperature coefficient	See (4)		10			ppm/°C
E _G	Gain error	See(5)				±0.6	% full scale V
E _G T _C	Gain error temperature coefficient	See (6)		10			ppm/°C
OUTPUT SPECIFICATIONS							
V _O	Output voltage	R _L = 10 kΩ		0	V _{DD} -0.4		V
Output load regulation accuracy		V _O = 4.096 V, 2.048 V, R _L = 2 kΩ				±0.25	% full scale V
REFERENCE PIN CONFIGURED AS OUTPUT (REF)							
V _{ref} (OUTL)	Low reference voltage			1.003	1.024	1.045	V
V _{ref} (OUTH)	High reference voltage	V _{DD} > 4.75 V		2.027	2.048	2.069	V
I _{ref} (source)	Output source current			1			mA
I _{ref} (sink)	Output sink current			-1			mA
Load capacitance				100			pF
PSRR	Power supply rejection ratio			-65			dB
REFERENCE PIN CONFIGURED AS INPUT (REF)							
V _I	Input voltage			0	V _{DD} -1.5		V
R _I	Input resistance			10			MΩ
C _I	Input capacitance			5			pF
Reference input bandwidth		REF = 0.2 V _{pp} + 1.024 V dc	Fast	1.3			MHz
			Slow	525			kHz
Reference feedthrough		REF = 1 V _{pp} at 1.024 V dc ⁽⁷⁾		-80			dB
DIGITAL INPUTS							
I _{IH}	High-level digital input current	V _I = V _{DD}		1			μA
I _{IL}	Low-level digital input current	V _I = 0 V		-1			μA
C _i	Input capacitance			8			pF

- (1) The relative accuracy or integral nonlinearity (INL) sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors. Tested from code 32 to 4095.
- (2) The differential nonlinearity (DNL) sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.
- (3) Zero-scale error is the deviation from zero voltage output when the digital input code is zero.
- (4) Zero-scale-error temperature coefficient is given by: $E_{ZS} TC = [E_{ZS}(T_{max}) - E_{ZS}(T_{min})]/V_{ref} \times 10^6/(T_{max} - T_{min})$.
- (5) Gain error is the deviation from the ideal output ($2V_{ref} - 1\text{ LSB}$) with an output load of $10\text{ k}\Omega$ excluding the effects of the zero-error.
- (6) Gain temperature coefficient is given by: $E_G TC = [E_G(T_{max}) - E_G(T_{min})]/V_{ref} \times 10^6/(T_{max} - T_{min})$.
- (7) Reference feedthrough is measured at the DAC output with an input code = 0x000.

ELECTRICAL CHARACTERISTICS (Continued)over recommended operating conditions, $V_{ref} = 2.048\text{ V}$, $V_{ref} = 1.024\text{ V}$ (unless otherwise noted)

ANALOG OUTPUT DYNAMIC PERFORMANCE								
PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT	
t _{s(FS)}	Output settling time, full scale	R _L = 10 kΩ, C _L = 100 pF, See ⁽¹⁾	Fast		1	3	μs	
			Slow		3.5	7		
t _{s(CC)}	Output settling time, code to code	R _L = 10 kΩ, C _L = 100 pF, See ⁽²⁾	Fast		0.5	1.5	μs	
			Slow		1	2		
SR	Slew rate	R _L = 10 kΩ, C _L = 100 pF, See ⁽³⁾	Fast		12		V/μs	
			Slow		1.8			
Glitch energy		DIN = 0 to 1, FCLK = 100 kHz, $\overline{CS}$ = V _{DD}			5		nV-s	
SNR	Signal-to-noise ratio	f _s = 480 kSPS, f _{out} = 1 kHz, R _L = 10 kΩ, C _L = 100 pF			69	74	dB	
S/(N+D)	Signal-to-noise + distortion				58	67		
THD	Total harmonic distortion					69		57
Spurious free dynamic range					57	72		

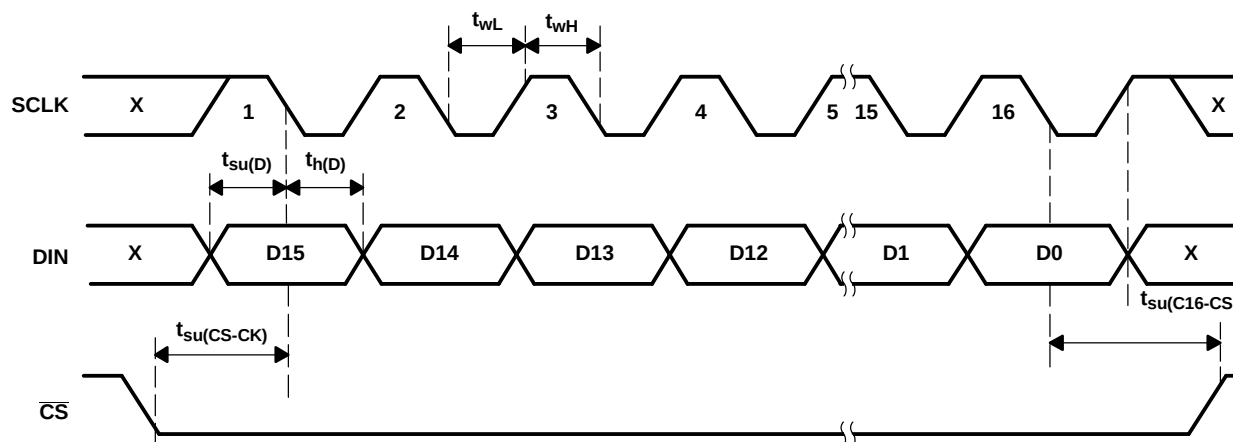
(1) Settling time is the time for the output signal to remain within $\pm 0.5\text{ LSB}$ of the final measured value for a digital input code change of 0x020 to 0xFDF and 0xFDF to 0x020 respectively. Not tested, assured by design.

(2) Settling time is the time for the output signal to remain within $\pm 0.5\text{ LSB}$ of the final measured value for a digital input code change of one count. Not tested, assured by design.

(3) Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.

DIGITAL INPUT TIMING REQUIREMENTS

	MIN	NOM	MAX	UNIT
$t_{su(CS-CK)}$	Setup time, $\overline{CS}$ low before first negative SCLK edge			ns
$t_{su(C16-CS)}$	Setup time, 16 th negative SCLK edge (when D0 is sampled) before $\overline{CS}$ rising edge			ns
t_{WH}	SCLK pulse width high			ns
t_{WL}	SCLK pulse width low			ns
$t_{su(D)}$	Setup time, data ready before SCLK falling edge			ns
$t_{h(D)}$	Hold time, data held valid after SCLK falling edge			ns

PARAMETER MEASUREMENT INFORMATION**Figure 1. Timing Diagram**

TYPICAL CHARACTERISTICS

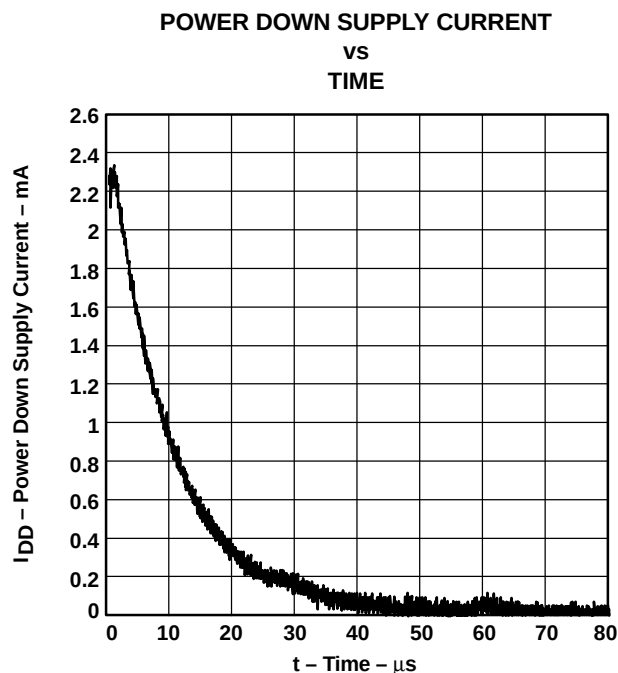


Figure 2.

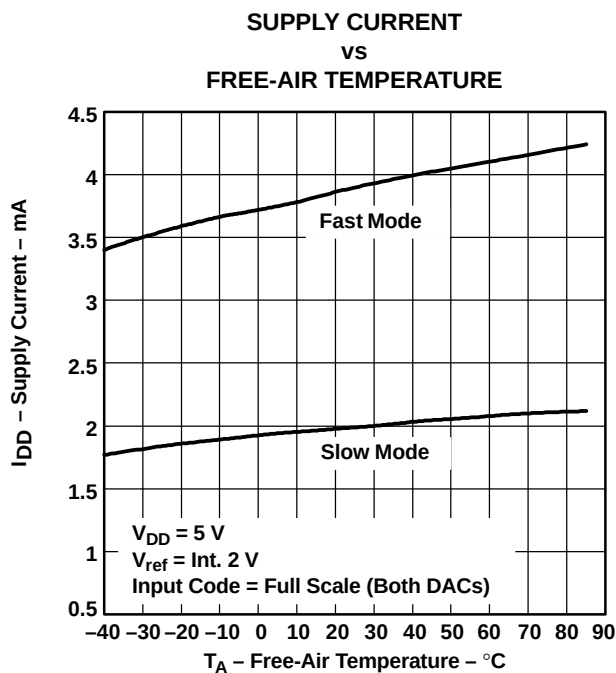


Figure 3.

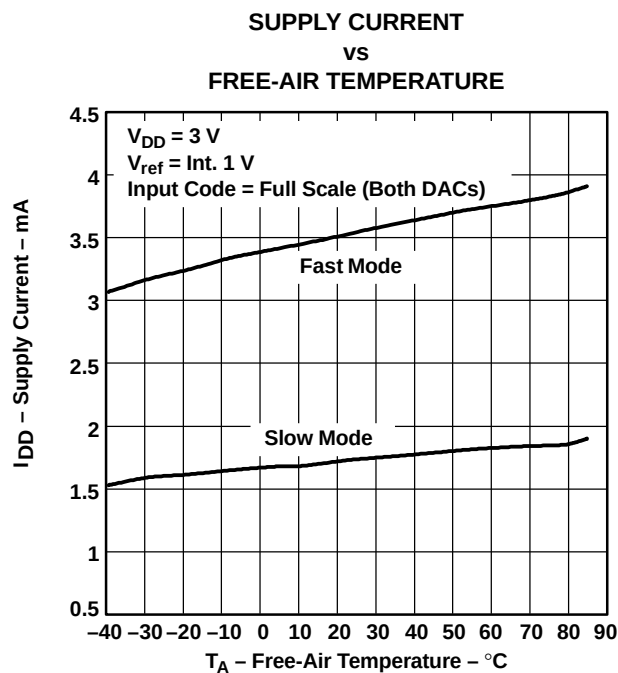


Figure 4.

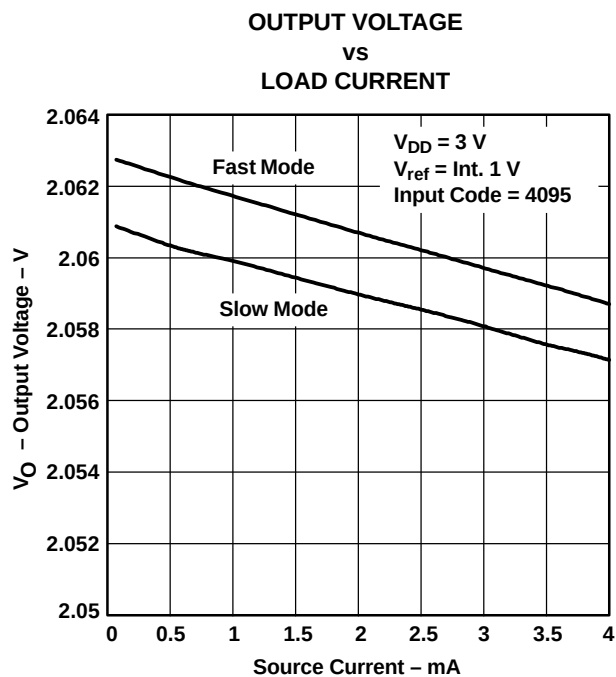


Figure 5.

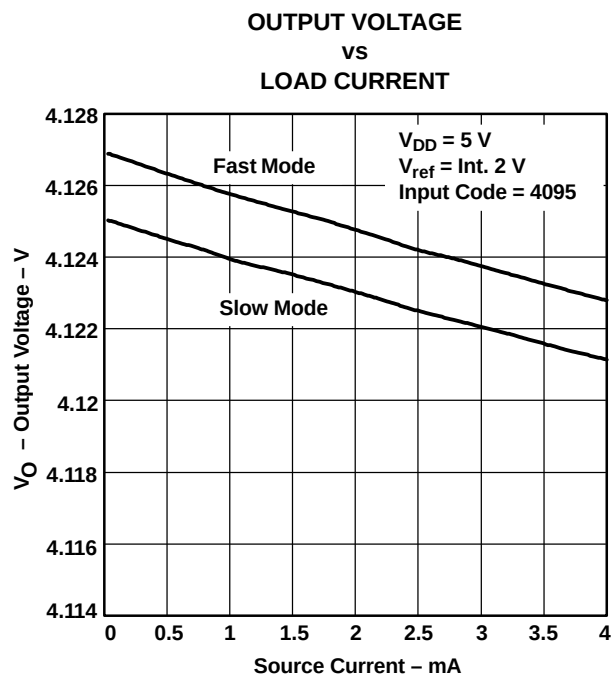
TYPICAL CHARACTERISTICS (continued)

Figure 6.

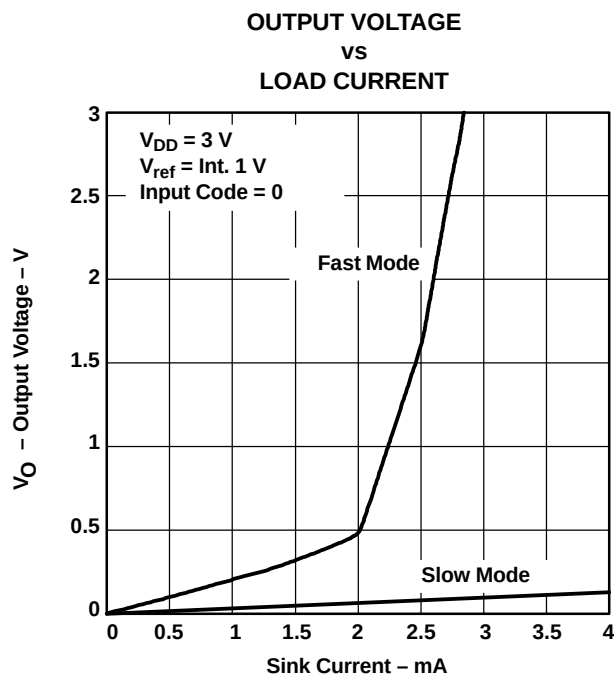


Figure 7.

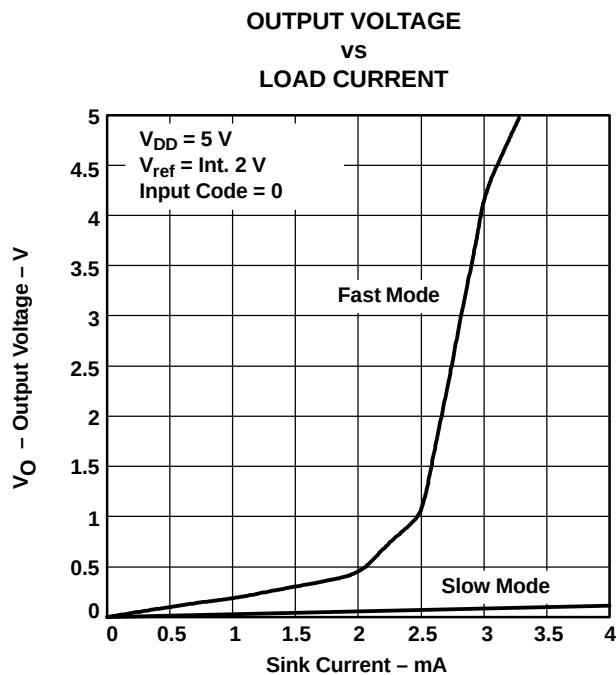


Figure 8.

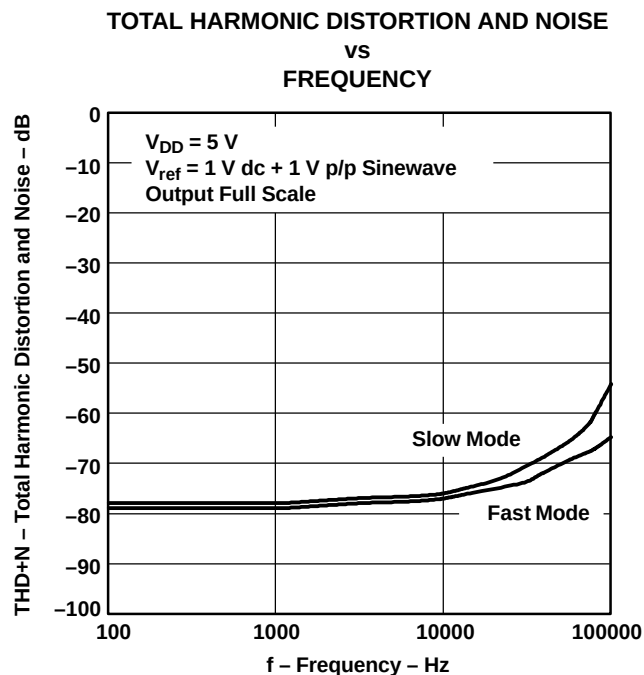


Figure 9.

TYPICAL CHARACTERISTICS (continued)

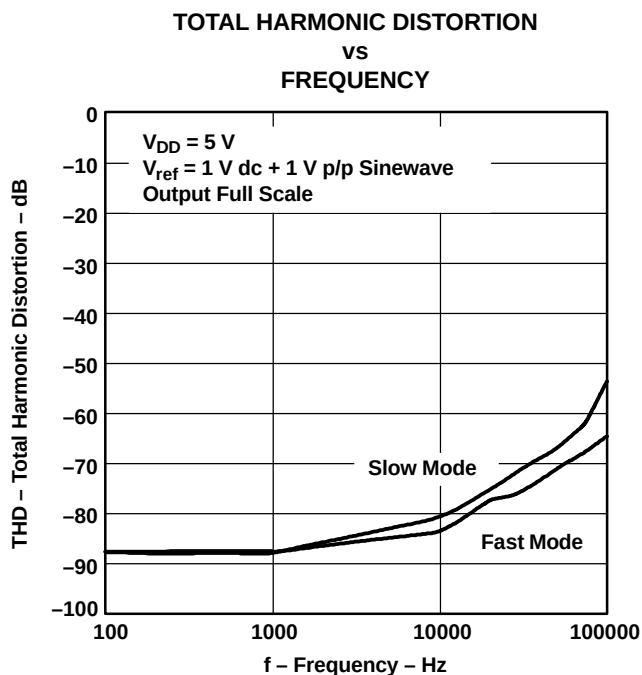
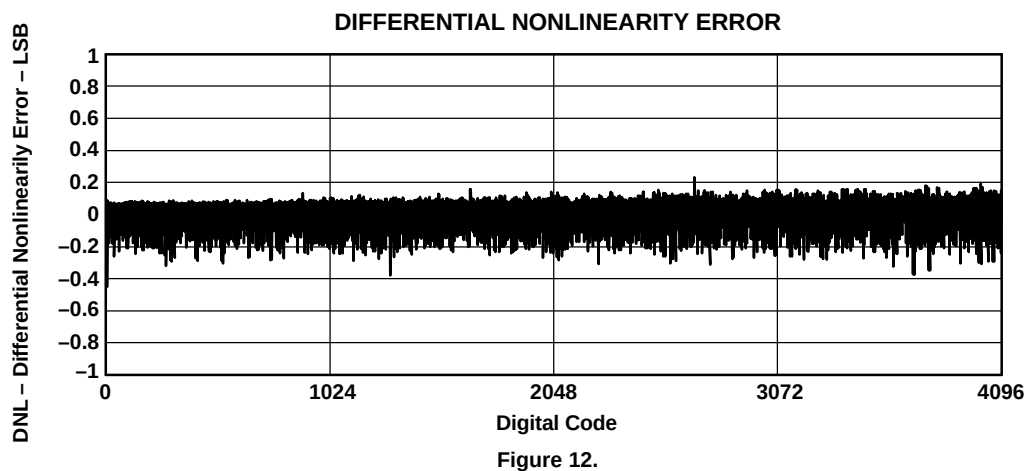
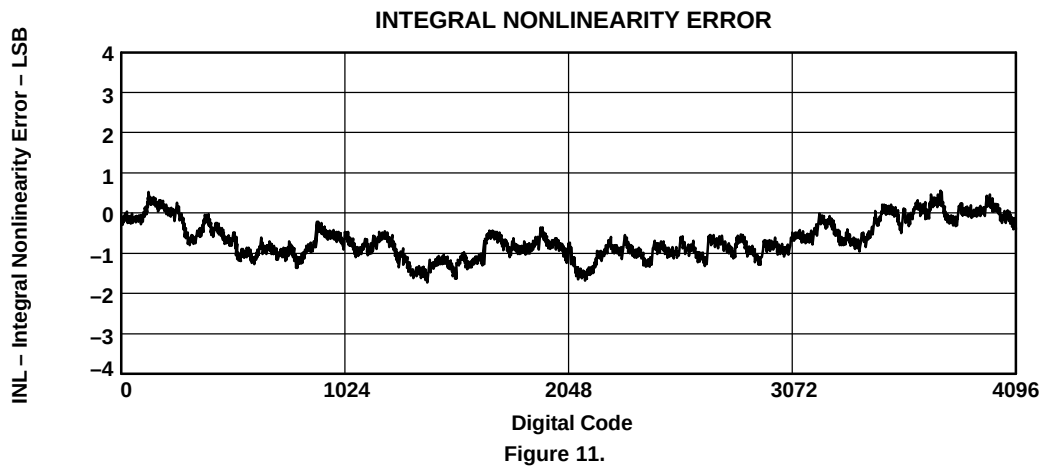


Figure 10.



APPLICATION INFORMATION

GENERAL FUNCTION

The TLV5638 is a dual 12-bit, single supply DAC, based on a resistor string architecture. It consists of a serial interface, a speed and power-down control logic, a programmable internal reference, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by reference) is given by:

$$2 \text{ REF } \frac{\text{CODE}}{0x1000} [\text{V}]$$

Where REF is the reference voltage and CODE is the digital input value in the range 0x000 to 0xFFFF. A power on reset initially puts the internal latches to a defined state (all bits zero).

SERIAL INTERFACE

A falling edge of $\overline{\text{CS}}$ starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or $\overline{\text{CS}}$ rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 13 shows examples of how to connect the TLV5638 to TMS320, SPI™, and Microwire™.

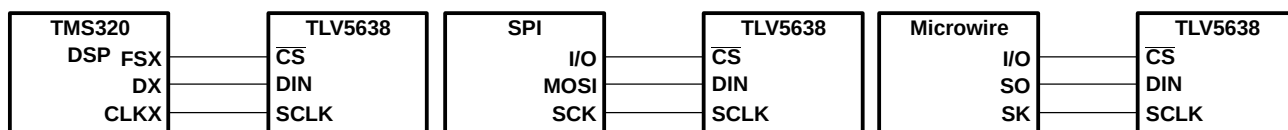


Figure 13. Three-Wire Interface

Notes on SPI™ and Microwire™: Before the controller starts the data transfer, the software has to generate a falling edge on the pin connected to $\overline{\text{CS}}$. If the word width is 8 bits (SPI™ and Microwire™), two write operations must be performed to program the TLV5638. After the write operation(s), the holding registers or the control register are updated automatically on the 16th positive clock edge.

SERIAL CLOCK FREQUENCY AND UPDATE RATE

The maximum serial clock frequency is given by:

$$f_{\text{sclmax}} = \frac{1}{t_{\text{whmin}} + t_{\text{wlmin}}} = 20 \text{ MHz}$$

The maximum update rate is:

$$f_{\text{updatemax}} = \frac{1}{16 (t_{\text{whmin}} + t_{\text{wlmin}})} = 1.25 \text{ MHz}$$

Note, that the maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5638 has to be considered, too.

APPLICATION INFORMATION (continued)

DATA FORMAT

The 16-bit data word for the TLV5638 consists of two parts:

- Program bits (D15..D12)
- New data (D11..D0)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R1	SPD	PWR	R0	12 Data bits											

SPD: Speed control bit	1 → fast mode	0 → slow mode
PWR: Power control bit	1 → power down	0 → normal operation

The following table lists the possible combination of the register select bits:

REGISTERED SELECT BITS

R1	R0	REGISTER
0	0	Write data to DAC B and BUFFER
0	1	Write data to BUFFER
1	0	Write data to DAC A and update DAC B with BUFFER content
1	1	Write data to control register

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

DATA BITS: DAC A, DAC B and BUFFER

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
New DAC Value											

If control is selected, then D1, D0 of the 12 data bits are used to program the reference voltage:

DATA BITS: CONTROL

[illegible]

REF1 and REF0 determine the reference source and, if internal reference is selected, the reference voltage.

REFERENCE BITS

REF1	REF0	REFERENCE
0	0	External
0	1	1.024 V
1	0	2.048 V
1	1	External

CAUTION:

If external reference voltage is applied to the REF pin, external reference **MUST** be selected.

EXAMPLES OF OPERATION:

- Set DAC A output, select fast mode, select internal reference at 2.048 V:

- Set reference voltage to 2.048 V (CONTROL register)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	1	0

- Write new DAC A value and update DAC A output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	0	New DAC A output value											

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set DAC B output, select fast mode, select external reference:

- Select external reference (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0

- Write new DAC B value to BUFFER and update DAC B output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	1	0	0	New BUFFER content and DAC B output value											

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set DAC A value, set DAC B value, update both simultaneously, select slow mode, select internal reference at 1.024 V:

- Set reference voltage to 1.024 V (CONTROL register)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	1	0	0	0	0	0	0	0	0	0	0	0	1

- Write data for DAC B to BUFFER:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1	New DAC B value											

- Write new DAC A value and update DAC A and B simultaneously:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	0	New DAC A value											

Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- Set power-down mode:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
X	X	1	X	X	X	X	X	X	X	X	X	X	X	X	X

X = Don't care

LINEARITY, OFFSET, AND GAIN ERROR USING SINGLE ENDED SUPPLIES

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in [Figure 14](#).

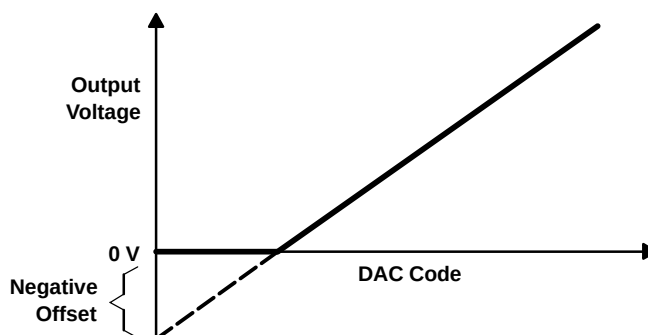


Figure 14. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.

DEFINITIONS OF SPECIFICATIONS AND TERMINOLOGY

Integral Nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

Differential Nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

Zero-Scale Error (E_{zs})

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

Gain Error (E_g)

Gain error is the error in slope of the DAC transfer function.

Total Harmonic Distortion (THD)

THD is the ratio of the rms value of the first six harmonic components to the value of the fundamental signal. The value for THD is expressed in decibels.

Signal-to-Noise Ratio + Distortion (S/N+D)

S/N+D is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

Spurious Free Dynamic Range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the largest spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
5962-9957601Q2A	Active	Production	LCCC (FK) 20	55 TUBE	ROHS Exempt	SNPB	N/A for Pkg Type	-55 to 125	5962- 9957601Q2A TLV5638 MFKB
5962-9957601QPA	Active	Production	CDIP (JG) 8	50 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	9957601QPA TLV5638M
TLV5638CD	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C
TLV5638CDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C
TLV5638ID	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I
TLV5638IDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I
TLV5638MFKB	Active	Production	LCCC (FK) 20	55 TUBE	ROHS Exempt	SNPB	N/A for Pkg Type	-55 to 125	5962- 9957601Q2A TLV5638 MFKB
TLV5638MJGB	Active	Production	CDIP (JG) 8	50 TUBE	No	SNPB	N/A for Pkg Type	-55 to 125	9957601QPA TLV5638M
TLV5638QD	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	V5638
TLV5638QDG4	Active	Production	SOIC (D) 8	75 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-	V5638
TLV5638QDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	V5638
TLV5638QDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-	V5638

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF TLV5638, TLV5638M :

- Catalog : [TLV5638](#)
- Enhanced Product : [TLV5638-EP](#), [TLV5638-EP](#)
- Military : [TLV5638M](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product
- Enhanced Product - Supports Defense, Aerospace and Medical Applications
- Military - QML certified for Military and Defense Applications

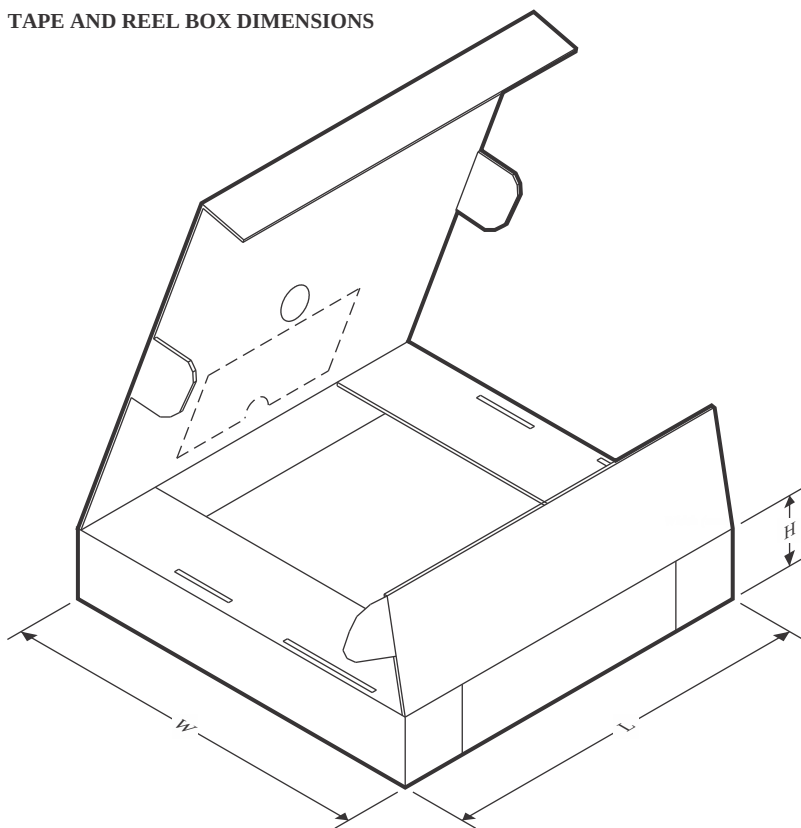
TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV5638CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5638IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5638QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV5638CDR	SOIC	D	8	2500	350.0	350.0	43.0
TLV5638IDR	SOIC	D	8	2500	356.0	356.0	35.0
TLV5638QDR	SOIC	D	8	2500	350.0	350.0	43.0

TUBE


*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
5962-9957601Q2A	FK	LCCC	20	55	506.98	12.06	2030	NA
TLV5638CD	D	SOIC	8	75	505.46	6.76	3810	4
TLV5638CDG4	D	SOIC	8	75	505.46	6.76	3810	4
TLV5638ID	D	SOIC	8	75	506.6	8	3940	4.32
TLV5638IDG4	D	SOIC	8	75	506.6	8	3940	4.32
TLV5638MFKB	FK	LCCC	20	55	506.98	12.06	2030	NA
TLV5638QD	D	SOIC	8	75	505.46	6.76	3810	4
TLV5638QDG4	D	SOIC	8	75	505.46	6.76	3810	4

GENERIC PACKAGE VIEW

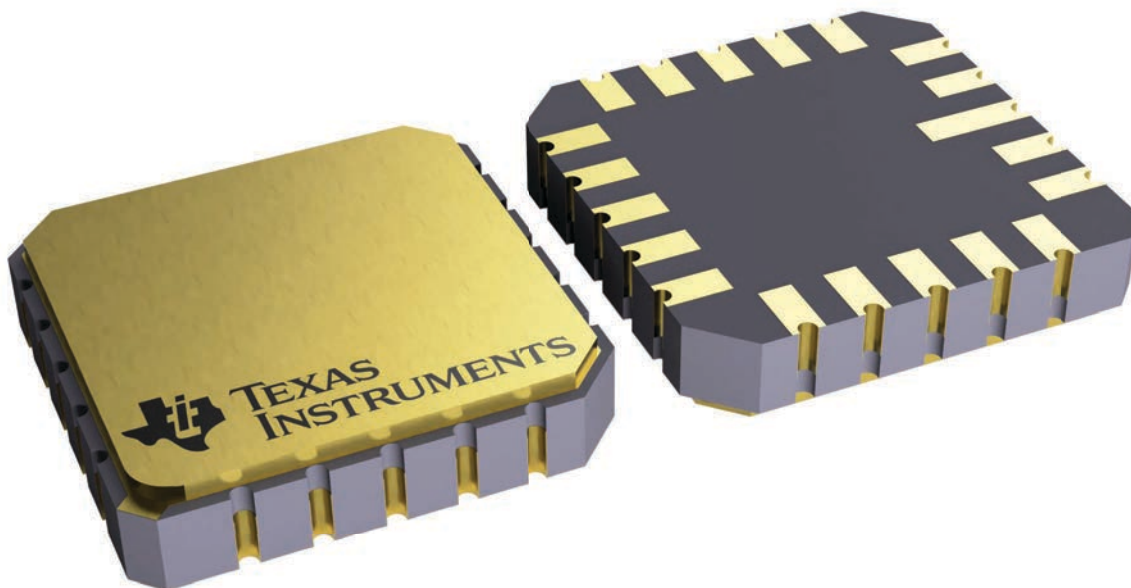
FK 20

LCCC - 2.03 mm max height

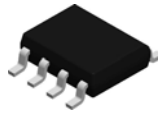
8.89 x 8.89, 1.27 mm pitch

LEADLESS CERAMIC CHIP CARRIER

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4229370VA\

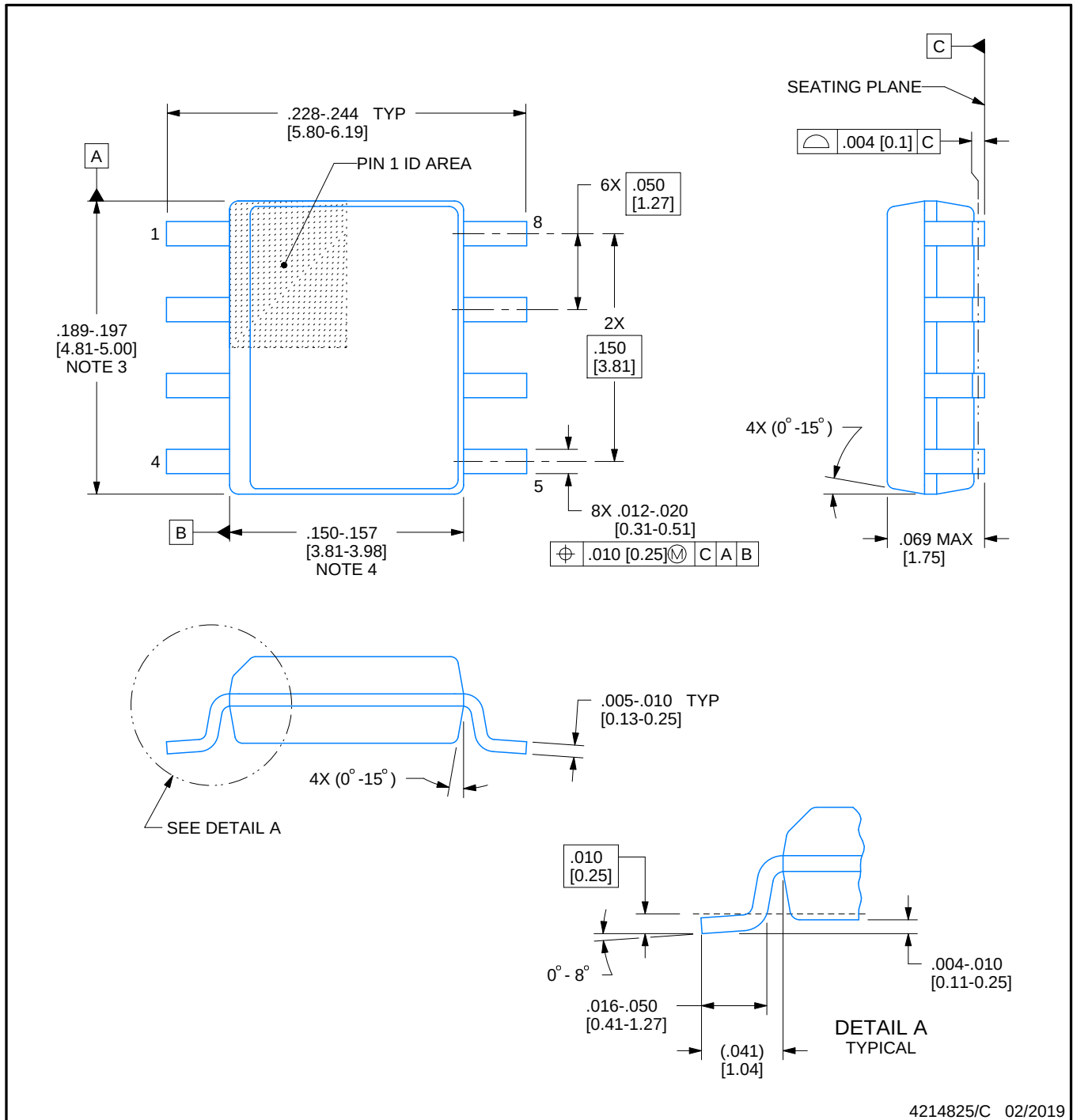


D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

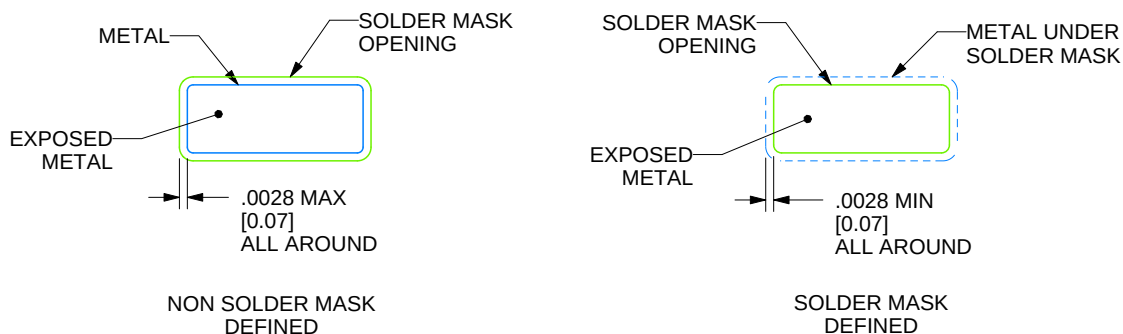
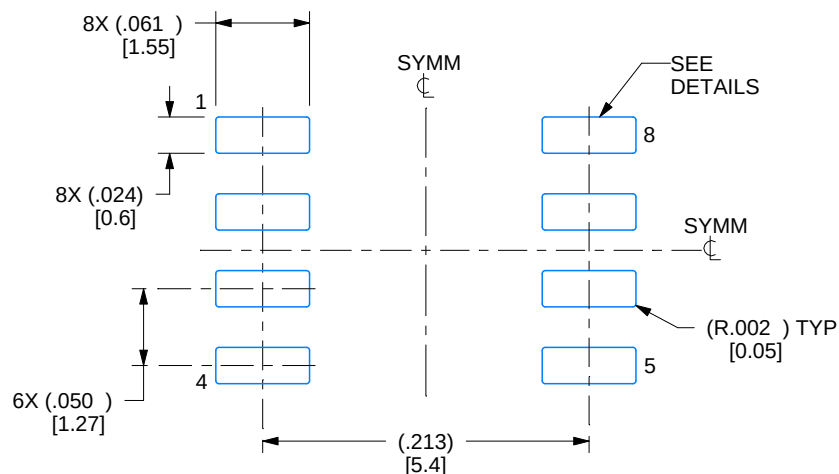
1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER MASK DETAILS

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NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

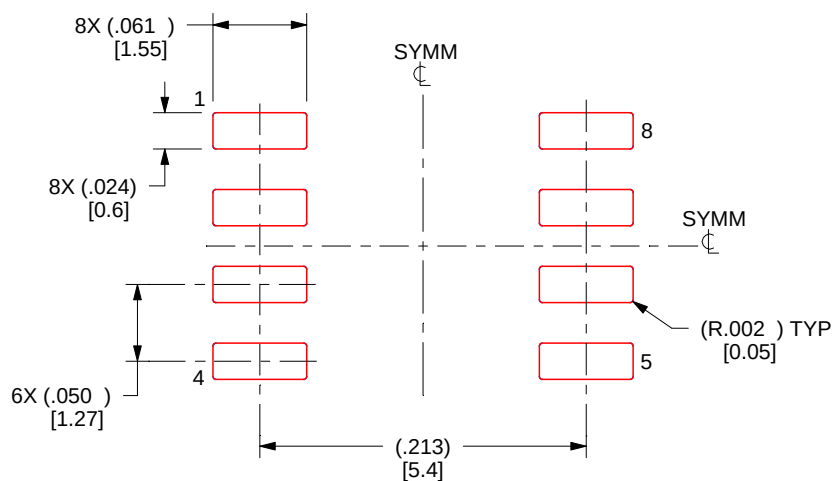
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

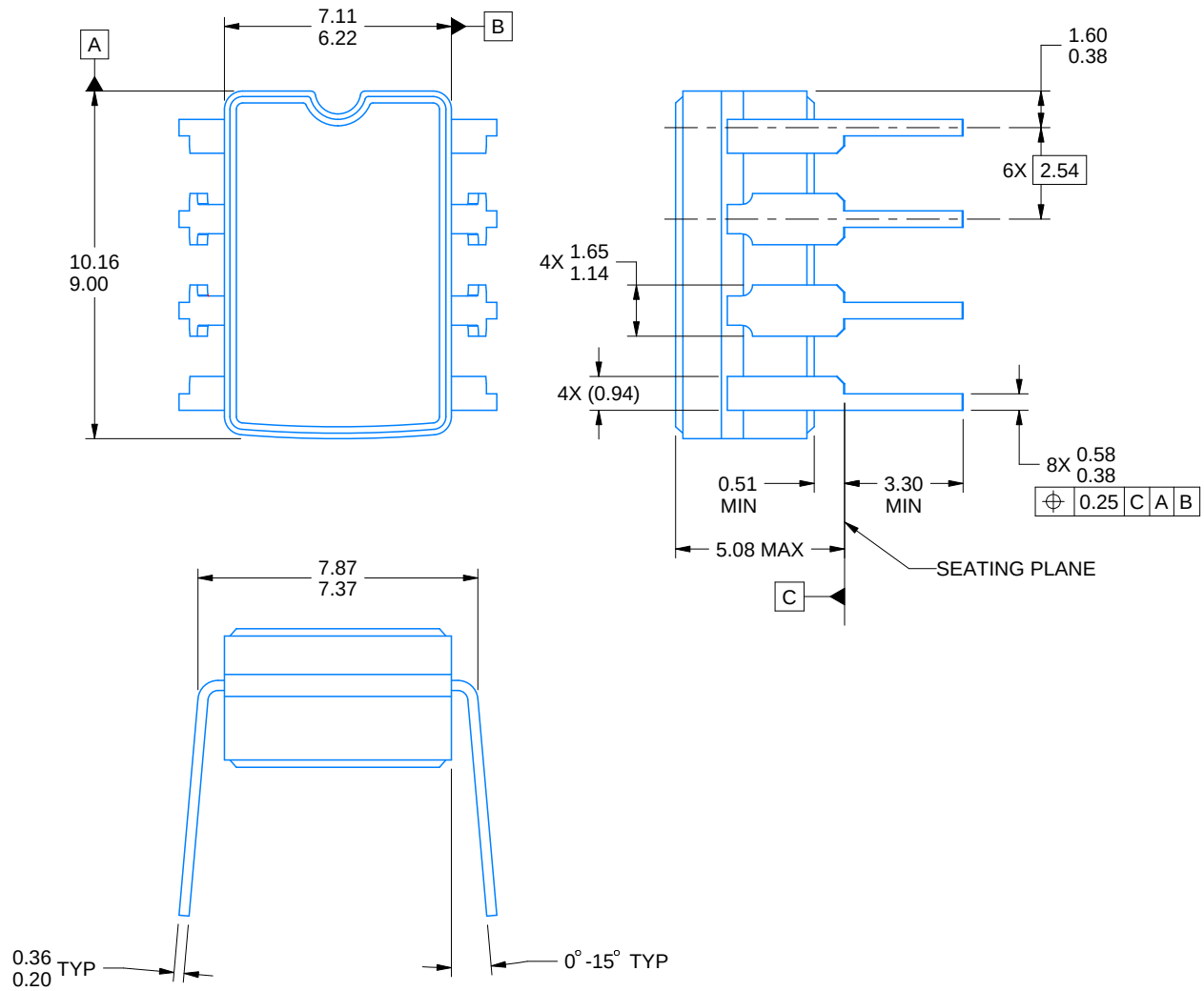
NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

JG0008A

CDIP - 5.08 mm max height

CERAMIC DUAL IN-LINE PACKAGE



4230036/A 09/2023

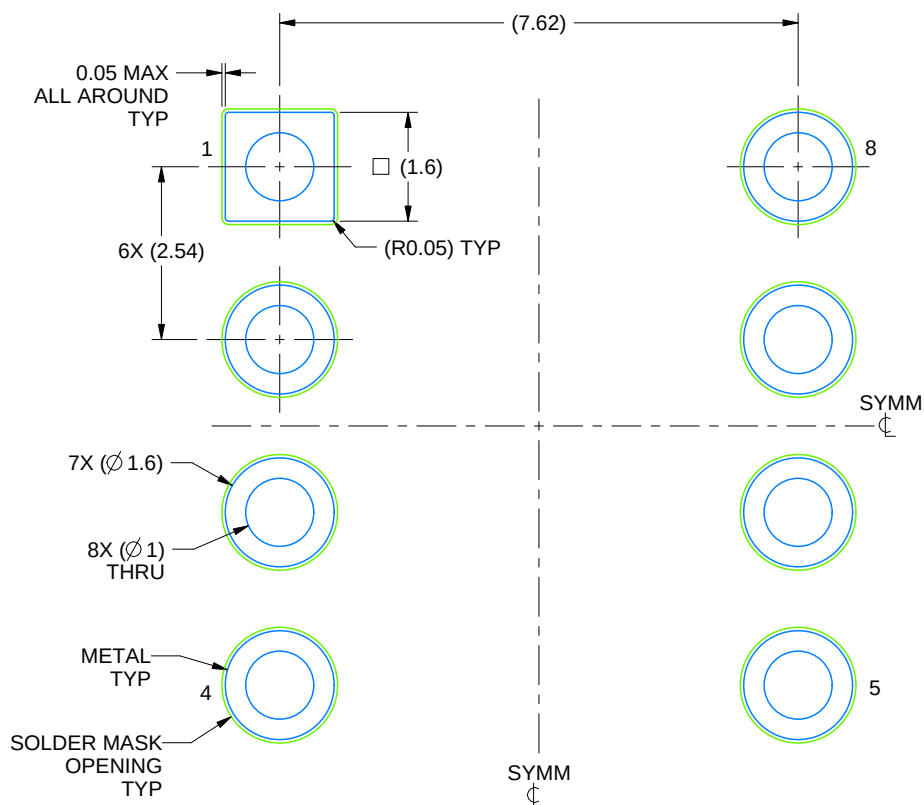
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This package can be hermetically sealed with a ceramic lid using glass frit.
4. Index point is provided on cap for terminal identification.
5. Falls within MIL STD 1835 GDIP1-T8

JG0008A

CDIP - 5.08 mm max height

CERAMIC DUAL IN-LINE PACKAGE



LAND PATTERN EXAMPLE
NON SOLDER MASK DEFINED
SCALE: 9X

4230036/A 09/2023

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