

SN74AC126 Quadruple Non-Inverting Buffers with 3-State Outputs

1 Features

- Wide operating range of 1.5V to 6V
- Inputs accept voltages up to 6V
- Continuous ±24mA output drive at 5V
- Supports up to ±75mA output drive at 5V in short bursts
- Drives 50Ω transmission lines
- Maximum t_{pd} of 6.6ns at 5V, 50pF load

2 Applications

- Drive an indicator LED
- Redrive a digital signal
- Drive a transmission line
- Hold a signal during controller reset

3 Description

The SN74AC126 contains four independent buffers with 3-state outputs.

Device Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE(2)	BODY SIZE(3)
SN74AC126	PW (TSSOP, 14)	5mm × 6.4mm	5mm × 4.4mm
3N74AC120	BQA (WQFN, 14)	3mm x 2.5mm	3mm x 2.5mm

- For more information, see Mechanical, Packaging, and Orderable Information.
- The package size (length × width) is a nominal value and includes pins, where applicable.
- The body size (length × width) is a nominal value and does not include pins.

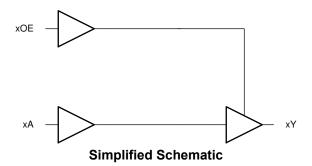


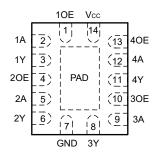


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4 Pin Configuration and Functions



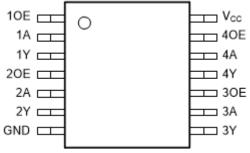


Figure 4-1. SN74AC126 BQA Package (Top View)

Figure 4-2. SN74AC126, PW Package (Top View)

Pin Functions

ı	PIN	TYPE ⁽¹⁾	DESCRIPTION
NAME	NO.	ITPE	DESCRIPTION
1 OE	1	I	Channel 1, output enable
1A	2	I	Channel 1, input A
1Y	3	0	Channel 1, output Y
2 OE	4	I	Channel 2, output enable
2A	5	I	Channel 2, input B
2Y	6	0	Channel 2, output Y
GND	7	G	Ground
3Y	8	0	Channel 3, output Y
3A	9	I	Channel 3, input A
3 OE	10	I	Channel 3, output enable
4Y	11	0	Channel 4, output Y
4A	12	I	Channel 4, input A
4 OE	13	I	Channel 4, output enable
V _{CC}	14	Р	Positive supply
Therm	nal Pad ⁽²⁾	_	The thermal pad can be connected to GND or left floating. Do not connect to any other signal or supply.

- (1) Signal Types: I = Input, O = Output, G = Ground, P = Power.
- (2) BQA package only.



5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

			MIN	MAX	UNIT
V _{CC}	Supply voltage range		-0.5	7	V
VI	Input voltage range ⁽²⁾		-0.5	V _{CC} + 0.5	V
Vo	Output voltage range ⁽²⁾		-0.5	V _{CC} + 0.5	V
I _{IK}	Input clamp current	$V_{I} < -0.5V \text{ or } V_{I} > V_{CC} + 0.5V$		±20	mA
I _{OK}	Output clamp current	V_{O} < -0.5V or V_{O} > V_{CC} + 0.5V		±50	mA
Io	Continuous output current	$V_O = 0$ to V_{CC}		±50	mA
	Continuous output current through V _{CC} or GND			±200	mA
TJ	Junction temperature			150	°C
T _{stg}	Storage temperature		-65	150	°C

⁽¹⁾ Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute maximum ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If briefly operating outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not sustain damage, but it may not be fully functional. Operating the device in this manner may affect device reliability, functionality, performance, and shorten the device lifetime.

(2) The input and output voltage ratings may be exceeded if the input and output current ratings are observed.

5.2 ESD Ratings

			VALUE	UNIT
V	Electrostatic	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	\/
V _(ESD)	discharge	Charged-device model (CDM), per ANSI/ESDA/JEDEC JS-002 ⁽²⁾	±1000	V

- (1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

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5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V _{CC}	Supply voltage		1.5	6	V
		V _{CC} = 1.5V	1.2		
		V _{CC} = 1.8V	1.26		
.,	Lligh level input valtage	V _{CC} = 2.5V	1.75		V
V _{IH}	High-level input voltage	V _{CC} = 3V	2.1		V
		V _{CC} = 4.5V	3.15		
		V _{CC} = 6V	4.2		
		V _{CC} = 1.5V		0.3	
		V _{CC} = 1.8V		0.54	
.,	Low Lovel input veltage	V _{CC} = 2.5V		0.75	V
V _{IL} L	Low-Level input voltage	V _{CC} = 3V		0.9	V
		V _{CC} = 4.5V		1.35	
		V _{CC} = 6V		1.8	
Vı	Input Voltage		0	V _{CC}	V
Vo	Output Voltage		0	V _{CC}	V
		V _{CC} = 1.8V		-1	
1	High level output ourrent	V _{CC} = 2.5V		-2	m A
ОН	High-level output current	V _{CC} = 3V		-12	mA
		V _{CC} = 4.5V to 6V		-24	
		V _{CC} = 1.8V		1	
1	Low lovel output ourrent	V _{CC} = 2.5V		2	m A
I _{OL}	Low-level output current	V _{CC} = 3V		12	mA
		V _{CC} = 4.5V to 6V		24	
Δt/Δν	Input transition rise or fall rate	V _{CC} = 1.5V to 3V		50	ns/V
Δι/Δν	Imput transition rise of fall rate	V _{CC} = 3.6V to 6V		20	115/ V
T _A	Operating free-air temperature		-40	125	°C

5.4 Thermal Information

PACKAGE	PACKAGE PINS		THERMAL METRIC(1)						
FACRAGE	FINO	R _{0JA}	R _{0JC(top)}	$R_{\theta JB}$	$\Psi_{ m JT}$	Ψ_{JB}	R _{0JC(bot)}	UNIT	
PW (TSSOP)	14	149.8	81.6	106.0	23.5	104.7	N/A	°C/W	
BQA (WQFN)	14	95.2	106.6	64.8	19.8	64.6	40.4	°C/W	

⁽¹⁾ For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application note.



5.5 Electrical Characteristics

PARAMETER	TEST CONDITIONS	V _{CC}	MIN	TYP	MAX	UNIT
		1.5V	1.4	1.49		
V		1.8V	1.7	1.79		
	I = 50uA	2.5V	2.4	2.49		
	I _{OH} = -50μA	3V	2.9	2.99		
		4.5V	4.4	4.49		
		6V	5.4	5.99		V
V _{OH}	I _{OH} = -1mA	1.8V	1.44			V
	I _{OH} = -2mA	2.5V	2			
	I _{OH} = -12mA	3V	2.4			
	I _{OH} = -24mA	4.5V	3.7			
	I _{OH} = -24mA	6V	4.7			
	I _{OH} = -75mA	6V	3.85			
		1.5V		0.01	0.1	
		1.8V		0.01	0.1	
	I = 50A	2.5V		0.01	0.1	
	$I_{OL} = 50\mu A$	3V		0.01	0.1	
		4.5V		0.01	0.1	
		6V		0.01	0.1	
V_{OL}	I _{OL} = 1mA	1.8V			0.36	V
	I _{OL} = 2mA	2.5V			0.5	
	I _{OL} = 12mA	3V			0.5	
	I _{OL} = 24mA	4.5V			0.5	
	I _{OL} = 24mA	6V			0.5	
	I _{OL} = 75mA	6V			1.65	
I _I	V _I = 6V or GND	0V to 6V			±1	μA
l _{oz}	V _O = V _{CC} or GND	6V			±10	μA
I _{cc}	V _I = V _{CC} or GND, I _O = 0	6V			20	μA
C _I	V _I = V _{CC} or GND	5V		2		pF
Co	Output disabled, V _O = V _{CC} or GND	5V		2		pF



5.6 Switching Characteristics

PARAMETER	FROM	то	V	-40°	C to 125°C		LINUT
PARAMETER	(INPUT)	(OUTPUT)	V _{CC}	MIN	TYP	MAX	UNIT
C _L = 15 pF	'			-			
t _{PLH}		Υ	1.5V		13.4	20.8	20
t _{PHL}	A	Y	1.5V		12.7	19.5	ns
t _{PLZ}	05	Υ	1.5V		9.4	13.5	
t _{PHZ}	OE	Y	1.5V		15.2	21.6	ns
t _{PZL}	05	\ <u></u>	1.5V		12.5	19.7	
t _{PZH}	OE	Υ	1.5V		14.9	23.1	ns
t _r		Υ	1.5V		3.7	5.8	
t _f		Y	1.5V		4.1	6.5	ns
t _{sk(o)}		Υ	1.5V		0.5	1.1	ns
t _{PLH}		\ <u></u>	1.8V		9.5	14.5	
t _{PHL}	A	Υ	1.8V		9.3	14	ns
t _{PLZ}	05	\ <u>\</u>	1.8V		6.9	9.8	
t _{PHZ}	OE	Y	1.8V		11.6	15.8	ns
t _{PZL}	05		1.8V		9.1	13.9	
t _{PZH}	OE	Υ	1.8V		10.7	16.2	ns
t _r		.,	1.8V		2.9	4.4	
t _f		Υ	1.8V		3.1	4.7	ns
t _{sk(o)}		Υ	1.8V		0.3	0.7	ns
t _{PLH}			2.5V		6.1	9	
t _{PHL}	─ A	Y	2.5V		5.9	8.6	ns
t _{PLZ}			2.5V		3.9	5.5	
t _{PHZ}	OE	Y	2.5V		6.3	8.9	ns
t _{PZL}			2.5V		6	9	
t _{PZH}	OE	Y	2.5V		6.9	10.2	ns
t _r			2.5V		2.2	3.5	
t _f		Y	2.5V		2.1	3.2	ns
t _{sk(o)}		Υ	2.5V		0.2	0.4	ns
t _{PLH}			3.3V		5.1	7.4	
t _{PHL}	— A	Υ	3.3V		5	7.1	ns
t _{PLZ}			3.3V		3.3	4.5	
t _{PHZ}	OE	Y	3.3V		5.2	7.2	ns
t _{PZL}			3.3V		5	7.3	
t _{PZH}	OE	Y	3.3V		5.7	8.3	ns
t _r			3.3V		1.8	2.9	
t _f		Υ	3.3V		1.8	2.7	ns
		Υ	3.3V		0.1	0.4	ns
t _{sk(o)}			5V		3.7	5.2	110
	— A	Υ	5V		3.8	5.1	ns
t _{PHL}			5V		2.4	3.1	
t _{PLZ}	— OE	Υ	5V		3.9	5.1	ns
t _{PHZ}			5V		3.7	5.1	
t _{PZL}	— OE	Υ					ns
t _{PZH}			5V		4.2	5.8	



24244555	FROM	то	.,	-40°(-40°C to 125°C		
PARAMETER	(INPUT)	(OUTPUT)	V _{cc}	MIN	TYP	MAX	UNIT
t _r		Υ	5V		1.4	2.2	
L _f		ĭ	5V		1.4	2	ns
t _{sk(o)}		Υ	5V		0.1	0.3	ns
t _{PLH}		V	6V		3.2	4.3	
t _{PHL}	A	Υ	6V		3.3	4.3	ns
t _{PLZ}	0.5	.,	6V		2.1	2.6	
t _{PHZ}	OE	Υ	6V		3.3	4.3	ns
t _{PZL}		.,	6V		3.1	4.3	
t _{PZH}	OE OE	Υ	6V		3.5	4.8	ns
t _r			6V		1.1	1.7	
t _f		Υ	6V		1.1	1.6	ns
$t_{sk(o)}$		Υ	6V		0.1	0.3	ns
C _L = 50 pF		1	I	I			
t _{PLH}			1.5V		15.6	24.1	
t _{PHL}	⊢ A	Y	1.5V		14.9	22.9	ns
t _{PLZ}			1.5V		13.3	17.7	
t _{PHZ}	OE	Y	1.5V		23.1	29.7	ns
t _{PZL}			1.5V		15.1	23.7	
t _{PZH}	OE	Υ	1.5V		17.3	26.5	ns
t _r			1.5V		7.5	12	
t _f	_	Y	1.5V		7.2	11.9	ns
		Υ	1.5V		0.5	1.1	ns
t _{sk(o)}		1	1.8V		11.4	17.1	115
t _{PLH}	— A	Y	1.8V		11.1	16.6	ns
t _{PHL}			1.8V		10.1	13.1	
t _{PLZ}	OE	Y	1.8V		17.8	22.4	ns
t _{PHZ}							
t _{PZL}	— OE	Υ	1.8V		11	16.8	ns
t _{PZH}			1.8V		12.6	18.8	
t _r	_	Y	1.8V		5.7	8.9	ns
t _f			1.8V		5.5	8.6	
t _{sk(o)}		Υ	1.8V		0.3	0.7	ns
t _{PLH}	A	Υ	2.5V		7.5	11	ns
t _{PHL}			2.5V		7.2	10.4	
t _{PLZ}	OE	Y	2.5V		5	6.8	ns
t _{PHZ}	-		2.5V		8.5	11.2	
t _{PZL}	OE	Υ	2.5V		7.6	11.1	ns
t _{PZH}			2.5V		8.4	12.2	110
t _r		Υ	2.5V		4.3	6.8	ns
t _f			2.5V		3.8	5.8	113
t _{sk(o)}		Υ	2.5V		0.2	0.4	ns
t _{PLH}	^	Υ	3.3V		6.3	9	
t _{PHL}	A	ľ	3.3V		6.1	8.7	ns
t _{PLZ}	05	V	3.3V		4	5.4	
t _{PHZ}	OE	Y	3.3V		7	9	ns



DADAMETER	FROM	то	W	-40°	C to 125°C		UNIT
PARAMETER	(INPUT)	(OUTPUT)	V _{cc}	MIN	TYP	MAX	
PZL	OE	Υ	3.3V		6.4	9.2	ne
t _{PZH}		1	3.3V		7	10	ns
t _r		Υ	3.3V		3.6	5.7	ne
t _f		1	3.3V		3.2	4.9	ns
t _{sk(o)}		Υ	3.3V		0.2	0.3	ns
t _{PLH}		Υ	5V		4.7	6.5	20
t _{PHL}	A	Ť	5V		4.7	6.2	ns
t _{PLZ}	OE	Υ	5V		3.2	3.6	
t _{PHZ}	OE .	Ť	5V		5.6	6.1	ns
t _{PZL}	05	V	5V		5.7	6.5	
t _{PZH}	OE	Υ	5V		6.4	7.3	ns
t _r			5V		2.7	4.1	
t _f		Y	5V		2.5	3.6	ns
t _{sk(o)}		Υ	5V		0.1	0.3	ns
t _{PLH}			6V		4	5.5	
t _{PHL}	A	Υ	6V		4	5.2	ns
t _{PLZ}		.,	6V		2.4	3	
t _{PHZ}	OE	Υ	6V		3.8	4.8	ns
t _{PZL}			6V		4.1	5.5	
t _{PZH}	OE	Υ	6V		4.5	6.2	ns
t _r			6V		2.3	3.5	
t _f		Y	6V		2.2	3	ns
$t_{sk(o)}$		Υ	6V		0.1	0.3	ns
C _L = 150 pF							
t _{PLH}			1.5V		20.8	32	
t _{PHL}	— A	Y	1.5V		19.4	30.1	ns
t _{PLZ}			1.5V		25	30	
t _{PHZ}	OE	Y	1.5V		45.7	52.6	ns
t _{PZL}			1.5V		20.5	32.8	
t _{PZH}	OE OE	Y	1.5V		22.6	34.4	ns
t _r			1.5V		18	29.6	
t _f		Y	1.5V		15	25.8	ns
t _{sk(o)}		Υ	1.5V		0.5	1.1	ns
t _{PLH}			1.8V		15.1	22.8	
t _{PHL}	— A	Y	1.8V		14.5	21.8	ns
t _{PLZ}			1.8V		19.7	22.8	
t _{PHZ}	OE	Y	1.8V		35.9	40.8	ns
t _{PZL}			1.8V		15.1	23.1	
t _{PZH}	OE	Y	1.8V		16.6	24.6	ns
t _r			1.8V		13.3	21.7	
t _f	\dashv	Υ	1.8V		11.1	18.5	ns
		Υ	1.8V		0.3	0.6	ns
t _{sk(o)}			2.5V		10.2	15	110
t _{PLH}	<u> </u> А	Υ	2.5V		9.5	13.7	ns
t _{PHL}			2.0 V		9.0	13.7	



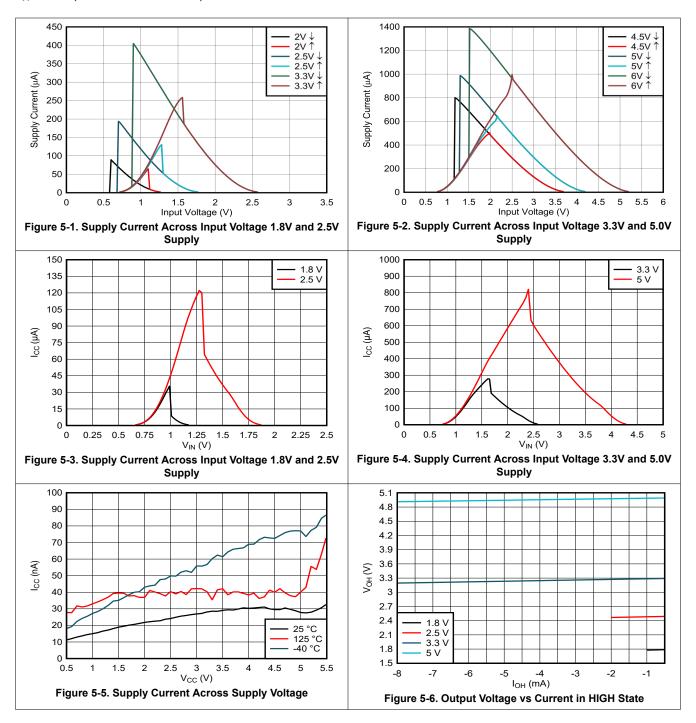
	FROM	то		-40°	-40°C to 125°C		
PARAMETER	(INPUT)	(OUTPUT)	V _{CC}	MIN	TYP	MAX	UNIT
t _{PLZ}	05	V	2.5V		8.3	10.1	
t _{PHZ}	OE	Y	2.5V		14.8	17.6	ns
t _{PZL}	OE	Υ	2.5V		10.4	15.1	
t _{PZH}	OE .	ĭ	2.5V		11.4	16.2	ns
t _r		Υ	2.5V		9.9	16.6	
t _f		ĭ	2.5V		7.6	12	ns
t _{sk(o)}		Y	2.5V		0.2	0.4	ns
t _{PLH}	A	Υ	3.3V		8.7	12.5	ns
t _{PHL}			3.3V		8.2	11.5	115
t _{PLZ}	- OE	Υ	3.3V		7	8.3	ns
t _{PHZ}			3.3V		12.1	14.2	115
t _{PZL}	OE	Υ	3.3V		8.9	12.6	ns
t _{PZH}			3.3V		9.7	13.6	115
t _r		Υ	3.3V		8.4	13.7	20
t _f		Ţ	3.3V		6.5	9.9	ns
t _{sk(o)}		Υ	3.3V		0.2	0.3	ns
t _{PLH}	A	Υ	5V		6.6	9.2	ns
t _{PHL}			5V		6.4	8.4	115
t _{PLZ}	OE	Υ	5V		4.7	5.4	ns
t _{PHZ}		1	5V		7.8	9	115
t _{PZL}	OE	Υ	5V		6.8	9.2	ns
t _{PZH}			5V		7.5	10.2	113
t_r		Υ	5V		6.2	9.8	ns
t_{f}			5V		5	7.2	113
$t_{sk(o)}$		Υ	5V		0.1	0.4	ns
t _{PLH}	A	Υ	6V		5.7	7.9	ns
t _{PHL}		'	6V		5.4	7.1	113
t_{PLZ}	OE	Υ	6V		3.8	4.4	ns
t _{PHZ}			6V		5.7	6.8	113
t _{PZL}	- OE	Υ	6V		5.8	7.7	ns
t _{PZH}			6V		6.5	8.7	113
t _r		Υ	6V		5.1	7.9	ns
t _f			6V		4.4	6.1	113
t _{sk(o)}		Υ	6V		0.1	0.3	ns
C _{PD} (1)	Any		5V		17		pF

⁽¹⁾ Power dissipation capacitance is calculated using the method described in CMOS Power Consumption and Cpd Calculation.



5.7 Typical Characteristics

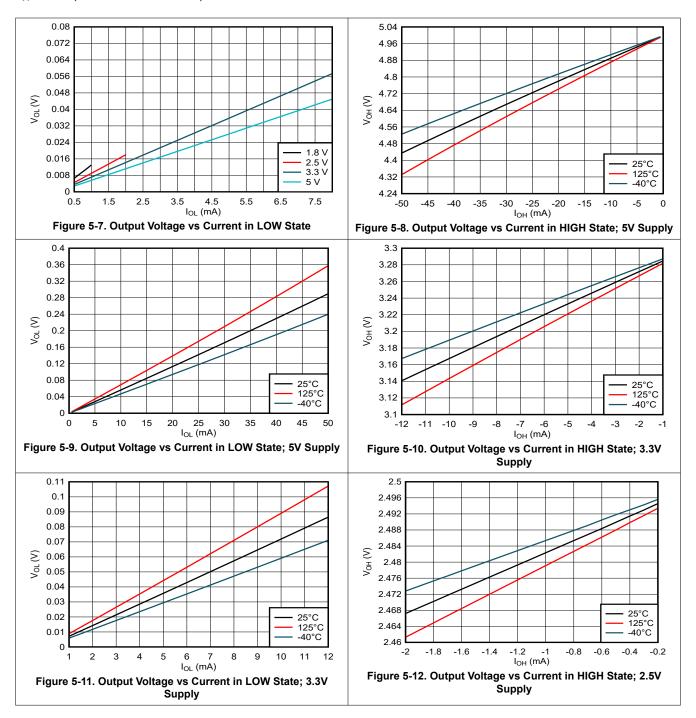
T_A = 25°C (unless otherwise noted)





5.7 Typical Characteristics (continued)

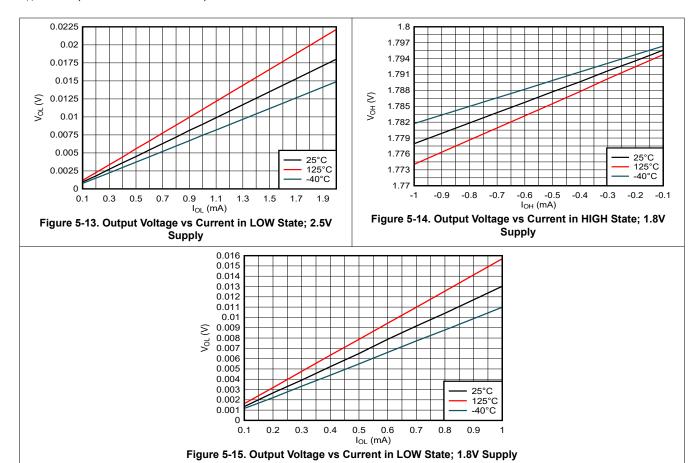
T_A = 25°C (unless otherwise noted)





5.7 Typical Characteristics (continued)

T_A = 25°C (unless otherwise noted)



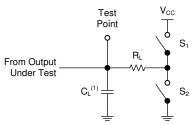


6 Parameter Measurement Information

Phase relationships between waveforms were chosen arbitrarily for the examples listed in the following table. All input pulses are supplied by generators having the following characteristics: PRR \leq 1MHz, $Z_O = 50\Omega$, $t_t < 2.5$ ns.

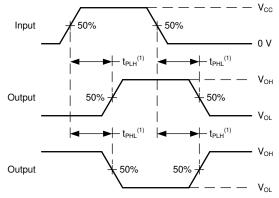
The outputs are measured individually with one input transition per measurement.

TEST	S1	S2	R _L	C _L	ΔV	V _{cc}
t _{PLH} , t _{PHL}	OPEN	OPEN	_	50pF	_	ALL
t _{PLZ} , t _{PZL}	CLOSED	OPEN	500Ω	50pF	0.15V	≤ 2.5V
t _{PHZ} , t _{PZH}	OPEN	CLOSED	500Ω	50pF	0.15V	≤ 2.5V
t _{PLZ} , t _{PZL}	CLOSED	OPEN	500Ω	50pF	0.3V	> 2.5V
t _{PHZ} , t _{PZH}	OPEN	CLOSED	500Ω	50pF	0.3V	> 2.5V



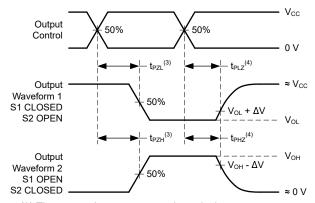
(1) C₁ includes probe and test-fixture capacitance.

Figure 6-1. Load Circuit for 3-State Outputs



(1) The greater between t_{PLH} and t_{PHL} is the same as t_{pd} .

Figure 6-2. Voltage Waveforms Propagation Delays



- (3) The greater between $t_{\mbox{\scriptsize PZL}}$ and $t_{\mbox{\scriptsize PZH}}$ is the same as $t_{\mbox{\scriptsize en}}$.
- (4) The greater between t_{PLZ} and t_{PHZ} is the same as t_{dis}

Input $\frac{10\%}{10\%} - \frac{10\%}{t_r^{(1)}} = \frac{10\%}{t_r^{(1)}} = 0 \text{ V}$ Output $\frac{10\%}{10\%} - \frac{10\%}{t_r^{(1)}} = 0 \text{ V}$ Ottput $\frac{10\%}{10\%} - \frac{10\%}{t_r^{(1)}} = 0 \text{ V}$

Figure 6-4. Voltage Waveforms, Input and Output Transition Times

Figure 6-3. Voltage Waveforms Propagation Delays

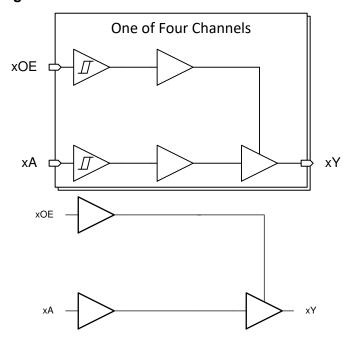


7 Detailed Description

7.1 Overview

This device contains four independent buffers with 3-state outputs and Schmitt-trigger inputs. Each gate performs the Boolean function Y = A in positive logic.

7.2 Functional Block Diagram



One of four channels



7.3 Feature Description

7.3.1 Balanced CMOS 3-State Outputs

This device includes balanced CMOS 3-state outputs. Driving high, driving low, and high impedance are the three states that these outputs can be in. The term *balanced* indicates that the device can sink and source similar currents. The drive capability of this device may create fast edges into light loads, so routing and load conditions should be considered to prevent ringing. Additionally, the outputs of this device can drive larger currents than the device can sustain without being damaged. It is important for the output power of the device to be limited to avoid damage due to overcurrent. The electrical and thermal limits defined in the *Absolute Maximum Ratings* must be followed at all times.

When placed into the high-impedance state, the output will neither source nor sink current, with the exception of minor leakage current as defined in the *Electrical Characteristics* table. In the high-impedance state, the output voltage is not controlled by the device and is dependent on external factors. If no other drivers are connected to the node, then this is known as a floating node and the voltage is unknown. A pull-up or pull-down resistor can be connected to the output to provide a known voltage at the output while it is in the high-impedance state. The value of the resistor will depend on multiple factors, including parasitic capacitance and power consumption limitations. Typically, a $10k\Omega$ resistor can be used to meet these requirements.

Unused 3-state CMOS outputs should be left disconnected.

7.3.2 CMOS Schmitt-Trigger Inputs

This device includes inputs with the Schmitt-trigger architecture. These inputs are high impedance and are typically modeled as a resistor in parallel with the input capacitance given in the *Electrical Characteristics* table from the input to ground. The worst case resistance is calculated with the maximum input voltage, given in the *Absolute Maximum Ratings* table, and the maximum input leakage current, given in the *Electrical Characteristics* table, using Ohm's law ($R = V \div I$).

The Schmitt-trigger input architecture provides hysteresis as defined by ΔV_T in the *Electrical Characteristics* table, which makes this device extremely tolerant to slow or noisy inputs. While the inputs can be driven much slower than standard CMOS inputs, it is still recommended to properly terminate unused inputs. Driving the inputs with slow transitioning signals will increase dynamic current consumption of the device. For additional information regarding Schmitt-trigger inputs, please see *Understanding Schmitt Triggers*.

7.3.3 Standard CMOS Inputs

This device includes standard CMOS inputs. Standard CMOS inputs are high impedance and are typically modeled as a resistor in parallel with the input capacitance given in the *Electrical Characteristics*. The worst case resistance is calculated with the maximum input voltage, given in the *Absolute Maximum Ratings*, and the maximum input leakage current, given in the *Electrical Characteristics*, using Ohm's law $(R = V \div I)$.

Standard CMOS inputs require that input signals transition between valid logic states quickly, as defined by the input transition time or rate in the *Recommended Operating Conditions* table. Failing to meet this specification will result in excessive power consumption and could cause oscillations. More details can be found in *Implications of Slow or Floating CMOS Inputs*.

Do not leave standard CMOS inputs floating at any time during operation. Unused inputs must be terminated at V_{CC} or GND. If a system will not be actively driving an input at all times, then a pull-up or pull-down resistor can be added to provide a valid input voltage during these times. The resistor value will depend on multiple factors; a $10k\Omega$ resistor, however, is recommended and will typically meet all requirements.

7.3.4 Clamp Diode Structure

As shown in Figure 7-1, the inputs and outputs to this device have both positive and negative clamping diodes.

Product Folder Links: SN74AC126



CAUTION

Voltages beyond the values specified in the *Absolute Maximum Ratings* table can cause damage to the device. The input and output voltage ratings may be exceeded if the input and output clamp-current ratings are observed.

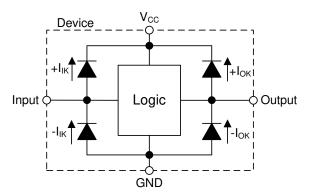


Figure 7-1. Electrical Placement of Clamping Diodes for Each Input and Output



7.4 Device Functional Modes

Table 7-1. Function Table

INP	UTS	OUTPUT				
OE	Α	Y				
L	Х	Z				
Н	L	L				
Н	Н	Н				



8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

The SN74AC126 can be used to drive signals over relatively long traces or transmission lines. A series damping resistor placed in series with the transmitter's output can be used to reduce ringing caused by impedance mismatches between the driver, transmission line, and receiver. The figure in the *Application Curve* section shows the received signal with three separate resistor values. Just a small amount of resistance can make a significant impact on signal integrity in this type of application.

8.2 Typical Application

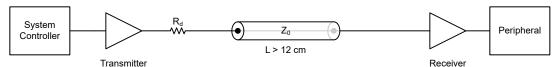


Figure 8-1. Typical Application Block Diagram



8.2.1 Design Requirements

8.2.1.1 Power Considerations

Ensure the desired supply voltage is within the range specified in the *Electrical Characteristics*. The supply voltage sets the device's electrical characteristics of the device as described in the *Electrical Characteristics* section.

The positive voltage supply must be capable of sourcing current equal to the total current to be sourced by all outputs of the SN74AC126 plus the maximum static supply current, I_{CC} , listed in the *Electrical Characteristics*, and any transient current required for switching. The logic device can only source as much current that is provided by the positive supply source. Ensure the maximum total current through V_{CC} listed in the *Absolute Maximum Ratings* is not exceeded.

The ground must be capable of sinking current equal to the total current to be sunk by all outputs of the SN74AC126 plus the maximum supply current, I_{CC}, listed in the *Electrical Characteristics*, and any transient current required for switching. The logic device can only sink as much current that can be sunk into its ground connection. Ensure the maximum total current through GND listed in the *Absolute Maximum Ratings* is not exceeded.

The SN74AC126 can drive a load with a total capacitance less than or equal to 50pF while still meeting all of the data sheet specifications. Larger capacitive loads can be applied; however, it is not recommended to exceed 50pF.

The SN74AC126 can drive a load with total resistance described by $R_L \ge V_O / I_O$, with the output voltage and current defined in the *Electrical Characteristics* table with V_{OH} and V_{OL} . When outputting in the HIGH state, the output voltage in the equation is defined as the difference between the measured output voltage and the supply voltage at the V_{CC} pin.

Total power consumption can be calculated using the information provided in the CMOS Power Consumption and Cpd Calculation application note.

Thermal increase can be calculated using the information provided in the *Thermal Characteristics of Standard Linear and Logic (SLL) Packages and Devices* application note.

CAUTION

The maximum junction temperature, $T_{J(max)}$ listed in the *Absolute Maximum Ratings*, is an additional limitation to prevent damage to the device. Do not violate any values listed in the *Absolute Maximum Ratings*. These limits are provided to prevent damage to the device.

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8.2.1.2 Input Considerations

Input signals must cross $V_{IL(max)}$ $V_{t-(min)}$ to be considered a logic LOW, and $V_{IH(min)}$ $V_{t+(max)}$ to be considered a logic HIGH. Do not exceed the maximum input voltage range found in the *Absolute Maximum Ratings*.

Unused inputs must be terminated to either V_{CC} or ground. The unused inputs can be directly terminated if the input is completely unused, or they can be connected with a pull-up or pull-down resistor if the input will be used sometimes, but not always. A pull-up resistor is used for a default state of HIGH, and a pull-down resistor is used for a default state of LOW. The drive current of the controller, leakage current into the SN74AC126 (as specified in the *Electrical Characteristics*), and the desired input transition rate limits the resistor size. A $10k\Omega$ resistor value is often used due to these factors.

The SN74AC126 has CMOS inputs and thus requires fast input transitions to operate correctly, as defined in the *Electrical Characteristics* table. Slow input transitions can cause oscillations, additional power consumption, and reduction in device reliability.

The SN74AC126 has no input signal transition rate requirements because it has Schmitt-Trigger inputs.

Another benefit to having Schmitt-Trigger inputs is the ability to reject noise. Noise with a large enough amplitude can still cause issues. To know how much noise is too much, please refer to the $\Delta V_{T(min)}$ in the *Electrical Characteristics*. This hysteresis value will provide the peak-to-peak limit.

Unlike what happens with standard CMOS inputs, Schmitt-Trigger inputs can be held at any valid value without causing huge increases in power consumption. The typical additional current caused by holding an input at a value other than V_{CC} or ground is plotted in the *Typical Characteristics*.

Refer to the *Feature Description* section for additional information regarding the inputs for this device.

8.2.1.3 Output Considerations

The positive supply voltage is used to produce the output HIGH voltage. Drawing current from the output will decrease the output voltage as specified by the V_{OH} specification in the *Electrical Characteristics*. The ground voltage is used to produce the output LOW voltage. Sinking current into the output will increase the output voltage as specified by the V_{OL} specification in the *Electrical Characteristics*.

Push-pull outputs that could be in opposite states, even for a very short time period, should never be connected directly together. This can cause excessive current and damage to the device.

Two channels within the same device with the same input signals can be connected in parallel for additional output drive strength.

Unused outputs can be left floating. Do not connect outputs directly to V_{CC} or ground.

Refer to the *Feature Description* section for additional information regarding the outputs for this device.

8.2.2 Detailed Design Procedure

- Add a decoupling capacitor from V_{CC} to GND. The capacitor needs to be placed physically close to the device and electrically close to both the V_{CC} and GND pins. An example layout is shown in the *Layout* section.
- 2. Verify that the capacitive load at the output is ≤ 50pF. This is not a hard limit; by design, however, it will optimize performance. This can be accomplished by providing short, appropriately sized traces from the SN74AC126 to one or more of the receiving devices.
- 3. Verify that the resistive load at the output is larger than $(V_{CC} / I_{O(max)})\Omega$. Doing this prevents the maximum output current from the *Absolute Maximum Ratings* from being violated. Most CMOS inputs have a resistive load measured in M Ω ; much larger than the minimum calculated previously.
- 4. Thermal issues are rarely a concern for logic gates; the power consumption and thermal increase, however, can be calculated using the steps provided in the CMOS Power Consumption and Cpd Calculation application note.

8.2.3 Application Curve

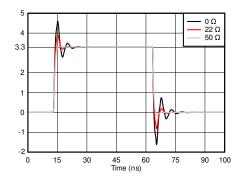


Figure 8-2. Simulated Signal Integrity at the Receiver with Different Damping Resistor (R_d) Values

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8.3 Power Supply Recommendations

The power supply can be any voltage between the minimum and maximum supply voltage rating located in the *Recommended Operating Conditions*. Each V_{CC} terminal should have a good bypass capacitor to prevent power disturbance.

A $0.1\mu F$ capacitor is recommended for this device. It is acceptable to parallel multiple bypass capacitors to reject different frequencies of noise. The $0.1\mu F$ and $1\mu F$ capacitors are commonly used in parallel. The bypass capacitor should be installed as close to the power terminal as possible for best results.

8.4 Layout

8.4.1 Layout Guidelines

- · Bypass capacitor placement
 - Place near the positive supply terminal of the device
 - Provide an electrically short ground return path
 - Use wide traces to minimize impedance
 - Keep the device, capacitors, and traces on the same side of the board whenever possible
- Signal trace geometry
 - 8mil to 12mil trace width
 - Lengths less than 12cm to minimize transmission line effects
 - Avoid 90° corners for signal traces
 - Use an unbroken ground plane below signal traces
 - Flood fill areas around signal traces with ground
 - Parallel traces must be separated by at least 3x dielectric thickness
 - For traces longer than 12cm
 - Use impedance controlled traces
 - Source-terminate using a series damping resistor near the output
 - Avoid branches; buffer each signal that must branch separately

8.4.2 Layout Example

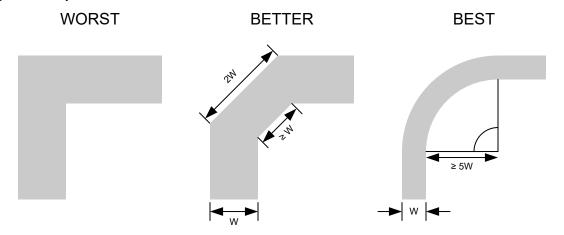


Figure 8-3. Example Trace Corners for Improved Signal Integrity

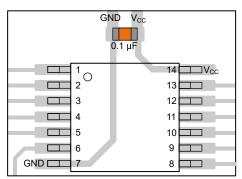


Figure 8-4. Example Bypass Capacitor Placement for TSSOP and Similar Packages

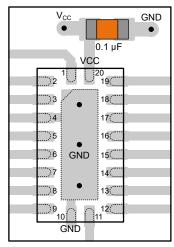


Figure 8-5. Example Bypass Capacitor Placement for WQFN and Similar Packages

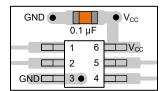


Figure 8-6. Example Bypass Capacitor Placement for SOT, SC70 and Similar Packages

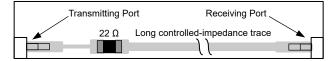


Figure 8-7. Example Damping Resistor Placement for Improved Signal Integrity

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9 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

9.1 Documentation Support

9.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, CMOS Power Consumption and Cpd Calculation application note
- · Texas Instruments, Designing With Logic application note
- Texas Instruments, *Thermal Characteristics of Standard Linear and Logic (SLL) Packages and Devices* application note

9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

9.3 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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9.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

9.6 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
August 2025	*	Initial Release



11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
SN74AC126BQAR	Active	Production	WQFN (BQA) 14	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-	AC126
SN74AC126PWR	Active	Production	TSSOP (PW) 14	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-	AC126

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF SN74AC126:

Automotive : SN74AC126-Q1

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No. RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



PACKAGE OPTION ADDENDUM

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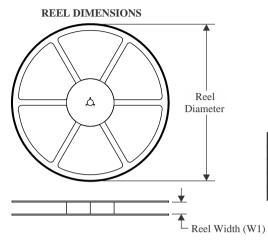
NOTE: Qualified Version Definition	วทร
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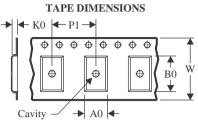
• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

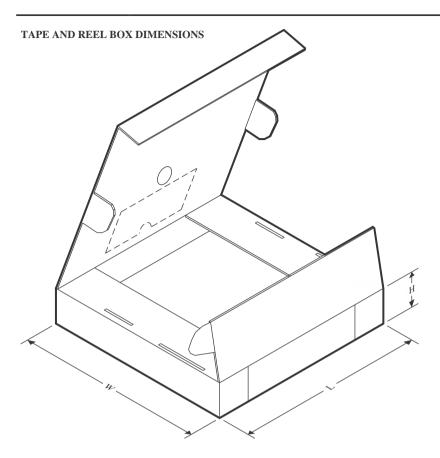
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN74AC126BQAR	WQFN	BQA	14	3000	180.0	12.4	2.8	3.3	1.1	4.0	12.0	Q1
SN74AC126PWR	TSSOP	PW	14	3000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

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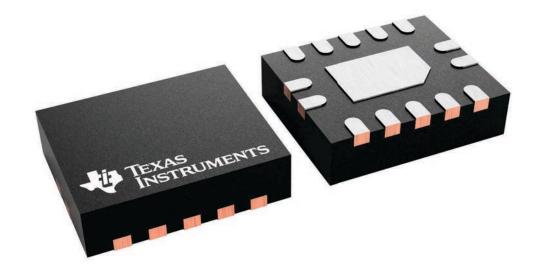
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SN74AC126BQAR	WQFN	BQA	14	3000	210.0	185.0	35.0
SN74AC126PWR	TSSOP	PW	14	3000	353.0	353.0	32.0

2.5 x 3, 0.5 mm pitch

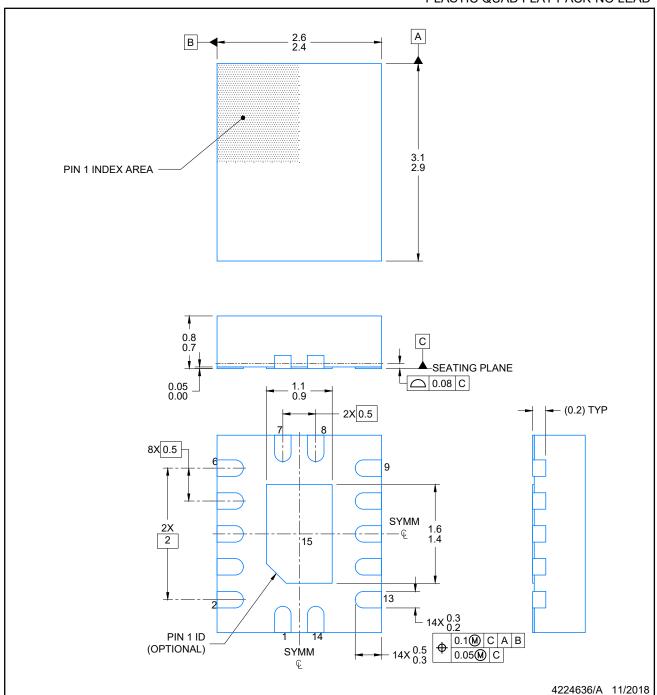
PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



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PLASTIC QUAD FLAT PACK-NO LEAD

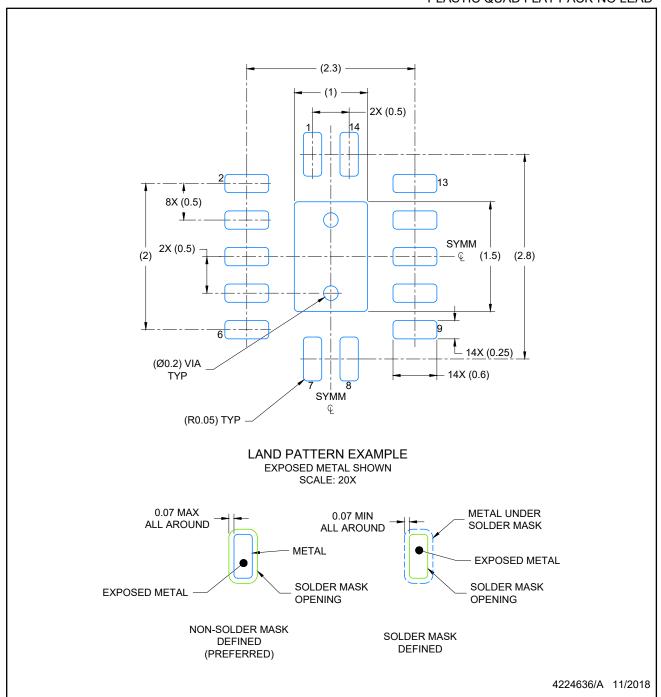


NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



PLASTIC QUAD FLAT PACK-NO LEAD

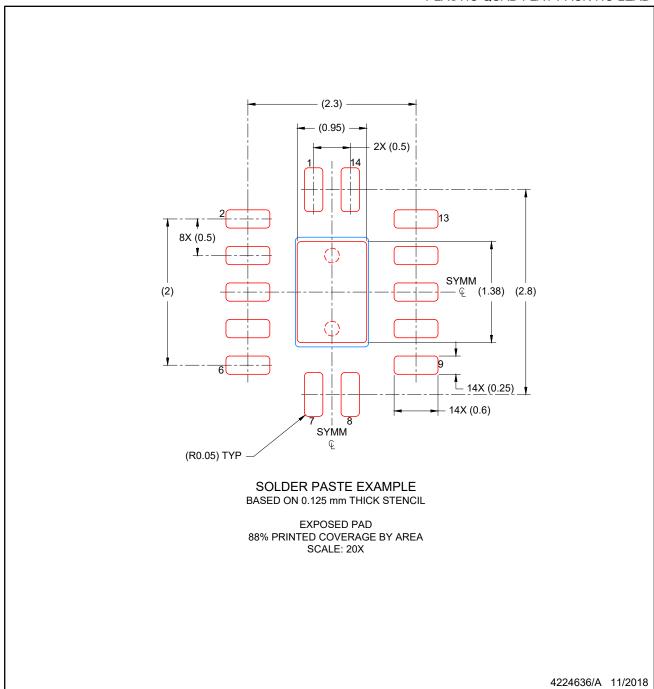


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLAT PACK-NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.





SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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