



3 Description





SN65HVD20, SN65HVD21, SN65HVD22, SN65HVD23, SN65HVD24 SLLS552G - DECEMBER 2002 - REVISED SEPTEMBER 2022

SN65HVD2x Extended Common-Mode RS-485 Transceivers

1 Features

- Common-mode voltage range (-20 V to 25 V) more than doubles TIA/EIA-485 requirement
- Receiver equalization extends cable length, signaling rate (SN65HVD2[3,4])
- Reduced unit-load for up to 256 nodes
- Bus I/O protection to over 16-kV HBM
- Failsafe receiver for open-circuit, short-circuit, and idle-bus conditions
- Low standby supply current 1 µA (maximum)
- More than 100 mV receiver hysteresis

2 Applications

- Long cable solutions
 - Factory automation
 - Security networks
 - Building HVAC
- Severe electrical environments
 - Electrical power inverters
 - Industrial drives
 - **Avionics**

protection, wide receiver hysteresis, and failsafe operation. This family of devices is designed for longcable networks, and other applications where the environment is too harsh for ordinary transceivers. These devices are designed for bidirectional data

The transceivers in the SN65HVD2x family offer

performance far exceeding typical RS-485 devices.

In addition to meeting all requirements of the TIA/EIA-485-A standard. The SN65HVD2x family

operates over an extended range of common-mode

voltages and has features such as high ESD

transmission on multipoint twisted-pair cables. Example applications are digital motor controllers, remote sensors and terminals, industrial process control, security stations, and environmental control systems.

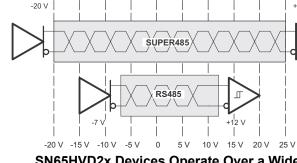
These devices combine a 3-state differential driver and a differential receiver that operate from a single 5-V power supply. The driver differential outputs and the receiver differential inputs are connected internally to form a differential bus port that offers minimum loading to the bus. This port features an extended common-mode voltage range, making the device suitable for multipoint applications over long cable runs.

Package Information

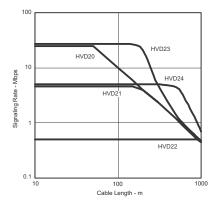
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PART NUMBER	PACKAGE ⁽¹⁾	BODY SIZE (NOM)		
SN65HVD2x	SOIC (8)	4.90 mm × 3.91 mm		
	PDIP (8)	9.81 mm × 6.35 mm		

For all available packages, see the orderable addendum at the end of the data sheet.

+25 V



SN65HVD2x Devices Operate Over a Wider Common-Mode Voltage Range



SN65HVD2x Application Space



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Changes from Revision A (March 2003) to Revision B (June 2003)	Page
Added V _{IK} Typical Value of 0.75 V (<i>Driver Electrical Characteristics</i> table)	8
 Deleted V_{IT(F+)} – VCM = –20 V to 25 V Minimum value (Receiver Electrical Characteristics table) 	9
Added the Receiver Equalization Characteristics table	
 Added Figure 8-6, Figure 8-7, and Figure 8-8 to the Typical Characteristics 	
Changed A Input circuit in the Equivalent Input and Output Schematic Diagrams	
Changed the Integrated Receiver Equalization Using the SN65HVD23 section	
Changes from Revision * (December 2002) to Revision A (March 2003)	Page
 Changed t_{PZH}, t_{PHZ}, t_{PZL}, and t_{PLZ} - From a maximum value of 120 to include typical and t	
 Changed t_{PZH}, t_{PHZ}, t_{PZL}, and t_{PLZ} - From a maximum value of 120 to include typical and typical a	alues for

5 Description (continued)

The SN65HVD20 device provides high signaling rate (up to 25 Mbps) for interconnecting networks of up to 64 nodes.

The SN65HVD21 device allows up to 256 connected nodes at moderate data rates (up to 5 Mbps). The driver output slew rate is controlled to provide reliable switching with shaped transitions which reduce high-frequency noise emissions.

The SN65HVD22 device has controlled driver output slew rate for low radiated noise in emission-sensitive applications and for improved signal quality with long stubs. Up to 256 SN65HVD22 nodes can be connected at signaling rates up to 500 kbps.

The SN65HVD23 device implements receiver equalization technology for improved jitter performance on differential bus applications with data rates up to 25 Mbps at cable lengths up to 160 meters.

The SN65HVD24 device implements receiver equalization technology for improved jitter performance on differential bus applications with data rates from 1 Mbps to 10 Mbps at cable lengths up to 1000 meters.

The receivers include a failsafe circuit that provides a high-level output within 250 microseconds after loss of the input signal. The most common causes of signal loss are disconnected cables, shorted lines, or the absence of any active transmitters on the bus. This feature prevents noise from being received as valid data under these fault conditions. This feature may be used for Wired-OR bus signaling.

The SN65HVD2x devices are characterized for operation temperatures from -40°C to 85°C.

6 Device Comparison

Table 6-1. Product Selection Guide

PART NUMBERS	CABLE LENGTH AND SIGNALING RATE ⁽¹⁾	NODES	MARKING			
SN65HVD20	Up to 50 m at 25 Mbps	Up to 64	D: VP20, P: 65HVD20			
SN65HVD21	Up to 150 m at 5 Mbps (with slew rate limit)	Up to 256	D: VP21, P: 65HVD21			
SN65HVD22	Up to1200 m at 500 kbps (with slew rate limit)	Up to 256	D: VP22, P: 65HVD22			
SN65HVD23	Up to 160 m at 25 Mbps (with receiver equalization)	Up to 64	D: VP23, P: 65HVD23			
SN65HVD24	Up to 500 m at 3 Mbps (with receiver equalization)	Up to 256	D: VP24, P: 65HVD24			

(1) Distance and signaling rate predictions based upon Belden 3105A cable and 15% eye pattern jitter.



7 Pin Configuration and Functions

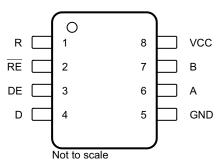


Figure 7-1. D or P Package, 8-Pin SOIC or PDIP (Top View)

Table 7-1. Pin Functions

PIN		- I/O	DESCRIPTION		
NAME	NO.	- 1/O	DESCRIPTION		
A	6	Bus input and output	Driver output or receiver input (complementary to B)		
В	7	Bus input and output	Driver output or receiver input (complementary to A)		
D	4	Digital input	Driver data input		
DE	3	Digital input	Driver enable, active high		
GND	5	Reference potential	Local device ground		
R	1	Digital output	Receive data output		
RE	2	Digital input	Receiver enable, active low		
VCC	8	Supply	4.5-V to 5.5-V supply		



8 Specifications

8.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)

			MIN	MAX	UNIT
Supply voltage ⁽²⁾			-0.5	7	V
Voltage at any bus I/O terminal			-27	27	V
Voltage input, transient pulse	A, B	(through 100 Ω, see Figure 9-16)	-60	60	V
Voltage input	D, DE, RE		-0.5	V _{CC} + 0.5	V
Receiver output current			-10	10	mA
Continuous total power dissipation				er Dissipation atings	
Junction temperature, T _J				150	°C
Storage temperature, T _{stg}				150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

8.2 ESD Ratings

	-			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/	All pins except 5, 6, and 7	±5000	
	Electrostatio discharge	JEDEC JS-001 ⁽¹⁾	Pins 5, 6, and 7	±16000	\/
V _(ESD)		Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾			V
		Machine Model (MM) (3)			

⁽¹⁾ Tested in accordance with JEDEC Standard 22, Test Method A114-A.

8.3 Recommended Operating Conditions

			MIN	NOM	MAX	UNIT
V _{CC}	Supply voltage		4.5	5	5.5	V
	Voltage at any bus I/O terminal	A, B	-20		25	V
V _{IH}	High-level input voltage	D, DE, RE	2		V _{CC}	V
V _{IL}	Low-level input voltage	TD, DE, RE	0		0.8	V
V _{ID}	Differential input voltage	A with respect to B	-25		25	V
	Output current	Driver	-110		110	A
		Receiver	-8		8	mA

⁽²⁾ Tested in accordance with JEDEC Standard 22, Test Method C101.

⁽³⁾ Tested in accordance with JEDEC Standard 22, Test Method A115-A



8.3 Recommended Operating Conditions (continued)

				MIN	NOM	MAX	UNIT
			SN65HVD20		6	9	
	Driver enabled (DE at V_{CC}),	SN65HVD21	,	8	12		
	Receiver enabled (RE at 0 V), No load,	SN65HVD22	,	6	9		
	$V_I = 0 \text{ V or } V_{CC}$	SN65HVD23		7	11		
			SN65HVD24		10	14	
			SN65HVD20		5	8	mA
		Driver enabled (DE at V_{CC}),	SN65HVD21	,	7	11	
		Receiver disabled (RE at V _{CC}), No load,	SN65HVD22	,	5	8	
cc	Supply current	$V_I = 0 \text{ V or } V_{CC}$	SN65HVD23		5	9	
			SN65HVD24		8	12	
			SN65HVD20		4	7	
			SN65HVD21	,	5	8	
		Driver disabled (DE at 0 V), Receiver enabled (RE at 0 V), No load	SN65HVD22	,	4	7	
		receiver chapies (NE at 0 V), No load	SN65HVD23		4.5	9	
		SN65HVD24		5.5	10		
		Driver disabled (DE at 0 V), Receiver disabled (RE at V _{CC}) D open	All SN65HVD2x			1	μΑ
A	Operating free-air tempera	perating free-air temperature ⁽¹⁾		-40		85	°C
J	Junction temperature			-40		130	°C

⁽¹⁾ Maximum free-air temperature operation is allowed as long as the device recommended junction temperature is not exceeded.



8.4 Thermal Information

		SN65HVD2x			
	THERMAL METRIC(1)	D (SOIC)	P (PDIP)	UNIT	
		8 PINS	8 PINS		
R _{0JA}	Junction-to-ambient thermal resistance	110.7	52.5	°C/W	
R _{0JC(top)}	Junction-to-case(top) thermal resistance	49.9	57.6	°C/W	
R _{0JB}	Junction-to-board thermal resistance	56.7	38.6	°C/W	
Ψлт	Junction-to-top characterization parameter	6.0	19.1	°C/W	
ΨЈВ	Junction-to-board characterization parameter	55.9	31.9	°C/W	

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

8.5 Driver Electrical Characteristics

over recommended operating conditions (unless otherwise noted).(1)

	PARAMETER	TEST	CONDITIONS	MIN	TYP	MAX	UNIT
V _{IK}	Input clamp voltage	I _I = -18 mA		-1.5	0.75		V
Vo	Open-circuit output voltage	A or B, No load		0		V _{CC}	V
		No load (open circuit)		3.3	4.2	V _{CC}	
$ V_{OD(SS)} $	(OD(SS)) Steady-state differential output voltage	R_L = 54 Ω , See Figure	9-1	1.8	2.5		V
		With common-mode lo	ading, See Figure 9-2	1.8			
$\Delta V_{OD(SS)} $	Change in steady-state differential output voltage between logic states	See Figure 9-1 and Fig	gure 9-3	-0.1		0.1	V
V _{OC(SS)}	Steady-state common-mode output voltage	See Figure 9-1		2.1	2.5	2.9	V
$\Delta V_{OC(SS)}$	Change in steady-state common-mode output voltage, $V_{\text{OC(H)}} - V_{\text{OC(L)}}$	See Figure 9-1 and Figure 9-4		-0.1		0.1	V
V _{OC(PP)}	Peak-to-peak common-mode output voltage, V _{OC(MAX)} – V _{OC(MIN)}	$R_L = 54 \Omega, C_L = 50 pF,$ 9-4	See Figure 9-1 and Figure	0.35			V
V _{OD(RING)}	Differential output voltage over and under shoot	$R_L = 54 \Omega, C_L = 50 pF$	See Figure 9-5			10%	
II	Input current	D, DE		-100		100	μA
		$V_{O} = -7 \text{ V to } 12 \text{ V},$	SN65HVD2[0,3]	-400		500	
	Output current with power off.	Other input = 0 V	SN65HVD2[1,2,4]	-100		125	
I _O	High impedance state output current.	$V_{O} = -20 \text{ V to } 25 \text{ V},$	SN65HVD2[0,3]	-800		1000	μA
		Other input = 0 V SN65HVD2[1,2,4]	-200		250		
Ios	Short-circuit output current	V _O = -20 V to 25 V, Se	ee Figure 9-9	-250		250	mA
C _{OD}	Differential output capacitance					20	pF

⁽¹⁾ All typical values are at $V_{CC} = 5 \text{ V}$ and 25°C .



8.6 Receiver Electrical Characteristics

over recommended operating conditions (unless otherwise noted)(1)

PARAMETER		TEST	TEST CONDITIONS		TYP	MAX	UNIT
V _{IT(+)}	Positive-going differential input voltage threshold	G Firms 0.40	V _O = 2.4 V, I _O = -8 mA		60	200	>/
V _{IT(-)}	Negative-going differential input voltage threshold	See Figure 9-10	V _O = 0.4 V, I _O = 8 mA	-200	-60		mV
V _{HYS}	Hysteresis voltage (V _{IT+} – V _{IT})			100	130		mV
	Positive-going differential input failsafe voltage	See Figure 0.45	VCM = -7 V to 12 V	40	120	200	mV
$V_{IT(F+)}$	threshold	See Figure 9-15	VCM = -20 V to 25 V		120	250	mv
	Negative-going differential input failsafe voltage		VCM = -7 V to 12 V	-200	-120	-40	>/
$V_{IT(F-)}$	threshold	See Figure 9-15	VCM = -20 V to 25 V	-250	-120		mV
V _{IK}	Input clamp voltage	I _I = -18 mA		-1.5			V
V _{OH}	High-level output voltage	V _{ID} = 200 mV, I _{OH} = -4	8 mA, See Figure 9-11	4			V
V _{OL}	Low-level output voltage	V _{ID} = -200 mV, I _{OL} = 8	3 mA, See Figure 9-11			0.4	V
		V _I = -7 to 12 V, Other input = 0 V	SN65HVD2[0,3]	-400		500	μA
	D		SN65HVD2[1,2,4]	-100		125	
I _{I(BUS)}	Bus input current (power on or power off)	$V_1 = -20 \text{ to } 25 \text{ V},$	SN65HVD2[0,3]	-800		1000	
		Other input = 0 V	SN65HVD2[1,2,4]	-200		250	
I _I	Input current	RE		-100		100	μA
Б	land a side a s	SN65HVD2[0,3]		24			1.0
R _I	Input resistance	SN65HVD2[1,2,4]		96			kΩ
C _{ID}	Differential input capacitance	V _{ID} = 0.5 + 0.4 sine (2	π × 1.5 × 10 ⁶ t)		20		pF

⁽¹⁾ All typical values are at $V_{CC} = 5 \text{ V}$ and 25°C .

8.7 Driver Switching Characteristics

over recommended operating conditions (unless otherwise noted)(1)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
			SN65HVD2[0,3]	6	10	20	
t _{PLH} , t _{PHL}	Differential output propagation delay, low-to-high and high-to-low	$R_L = 54 \Omega$, $C_L = 50 pF$, See Figure 9-3	SN65HVD2[1,4]	20	32	60	ns
	· · · · · · · · · · · · · · · · · · ·		SN65HVD22	160	280	500	
			SN65HVD2[0,3]	2	6	12	
t _r , t _f	Differential output rise time and fall time	$R_L = 54 \Omega$, $C_L = 50 pF$, See Figure 9-3	SN65HVD2[1,4]	20	40	60	ns
		gara a c	SN65HVD22	175	400	600	
	Propagation delay time,		SN65HVD2[0,3]			40	
t _{PZH} , t _{PHZ}	high-impedance-to-high-level output and	RE at 0 V, See Figure	SN65HVD2[1,4]			100	ns
	gh-level output-to-high-impedance		SN65HVD22			300	
	Propagation delay time,		SN65HVD2[0,3]			40	
t _{PZL} , t _{PLZ}	high-impedance-to-high-level output and	RE at 0 V, See Figure	SN65HVD2[1,4]		100		
	high-level output-to-high-impedance		SN65HVD22			300	
t _{d(standby)}	Time from an active differential output to standby					2	μs
t _{d(wake)}	Wake-up time from standby to an active differential output	RE at V _{CC} , See Figure 9	-8			8	μs
		SN65HVD2[0,3]			2		
t _{sk(p)}	Pulse skew t _{PLH} - t _{PHL}	SN65HVD2[1,4]			6	ns	
		SN65HVD22			50		

⁽¹⁾ All typical values are at V_{CC} = 5 V and 25°C



8.8 Receiver Switching Characteristics

over recommended operating conditions (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
t _{PLH}	Propagation delay time,	SN65HVD2[0,3]			16	35	ns
t _{PHL}	low-to-high level output and high-to low level output	See Figure 9-11	SN65HVD2[1,2,4]		25	50	
tr t _f	Receiver output rise time Receiver output fall time	See Figure 9-11			2	4	ns
t _{PZH}	Receiver output enable time to high level and	Soo Figure 0.12			90	120	20
t _{PHZ}	disable time from high level	See Figure 9-12	See Figure 9-12				ns
t _{PZL}	Receiver output enable time to low level and				90	120	
t _{PLZ}	disable time from low level	See Figure 9-13			16	35	ns
t _{r(standby)}	Time from an active receiver output to standby	See Figure 9-14, DE at 0	V			2	μs
t _{r(wake)}	Wake-up time from standby to an active receiver output	See Figure 9-14, DE at 0	V			8	μs
t _{sk(p)}	Pulse skew t _{PLH} - t _{PHL}					5	ns
t _{p(set)}	Delay time, bus fail to failsafe set	See Figure 9-15, pulse ra	ite = 1 kHz		250	350	μs
t _{p(reset)}	Delay time, bus recovery to failsafe reset	See Figure 9-15, pulse ra	ite = 1 kHz		50		ns

8.9 Receiver Equalization Characteristics

over recommended operating conditions (unless otherwise noted)⁽¹⁾ (2)

	PARAMETER	TEST	CONDITION	IS		TYP		
				0 m	SN65HVD23	2	ns	
				100 m	SN65HVD20	6		
				100 111	SN65HVD23	3	ns	
		2	25 Mbps 150	25 Mbps 150 m	150 m	SN65HVD20	15	ns
				130 111	SN65HVD23	4	115	
				200 m	SN65HVD20	27	ns	
				200 111	SN65HVD23	8	113	
				200 m	SN65HVD20	22	ns	
		Peudo-random NRZ code with a bit pattern length of $2^{16} - 1$, Beldon 3105A cable, See Figure 10-2		200 111	200 111	SN65HVD23	8	
			10 Mbps	250 m	SN65HVD20	34	ns	
$t_{j(pp)}$	Peak-to-peak eye-pattern jitter			200 111	SN65HVD23	15	110	
				300 m	SN65HVD20	49	ns	
				000 111	SN65HVD23	49 27	110	
			5 Mbps	500 m	SN65HVD21	128	ns	
			O WIDPO	000111	SN65HVD24	18	110	
					SN65HVD20	93		
			3 Mbps	500 m	SN65HVD21	103	ns	
			O WIDPS	000 111	SN65HVD23	90	113	
					SN65HVD24	16		
			1 Mbps	1000 m	SN65HVD21	216	ns	
			. Wibpo	1300 111	SN65HVD24	62		

⁽¹⁾ The SN65HVD20 and SN65HVD21 do not have receiver equalization, but are specified for comparison.

⁽²⁾ All typical values are at V_{CC} = 5 V, and temperature = 25°C.



8.10 Power Dissipation

	PARAMETERS		TEST CONDITIONS	VALUE	UNIT	
			V 5 V 7 0500	SN65HVD20: 25 Mbps	295	
P_D			$V_{CC} = 5 \text{ V}, T_J = 25^{\circ}\text{C},$ $R_1 = 54 \Omega, C_1 = 50 \text{ pF (driver)},$	SN65HVD21: 5 Mbps	260	
		Typical	C _L = 15 pF (receiver),	SN65HVD22: 500 kbps	233	mW
			50% Duty cycle square-wave signal, Driver and receiver enabled	SN65HVD23: 25 Mbps	302	
	Davisa nawar dissination		Briver and reserver enabled	SN65HVD24: 5 Mbps	267	
	Device power dissipation			SN65HVD20: 25 Mbps	408	
		$V_{CC} = 5.5 \text{ V}, T_J = 125^{\circ}\text{C},$ $R_I = 54 \Omega, C_I = 50 \text{ pF},$	SN65HVD21: 5 Mbps	342		
		Worst	C _L = 15 pF (receiver),	SN65HVD22: 500 kbps	300	mW
		5	50% Duty cycle square-wave signal, Driver and receiver enabled	SN65HVD23: 25 Mbps	417	
			Briver and reserver chabled	SN65HVD24: 5 Mbps	352	
T _{SD}	Thermal shut down junction temperature				170	°C



8.11 Typical Characteristics

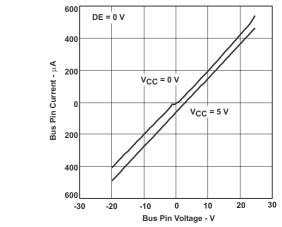


Figure 8-1. SN65HVD2[0,3] Bus Pin Current vs Bus Pin Voltage

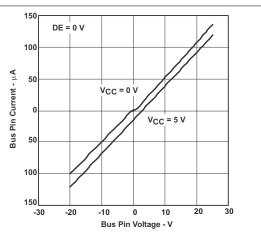


Figure 8-2. SN65HVD2[1,2,4] Bus Pin Current vs Bus Pin Voltage

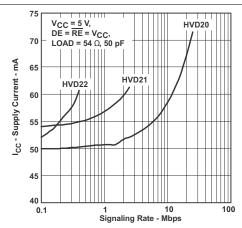


Figure 8-3. Supply Current vs Signaling Rate

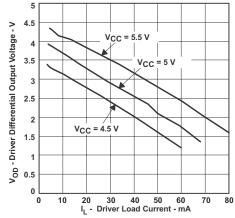


Figure 8-4. Driver Differential Output Voltage vs Driver Load Current

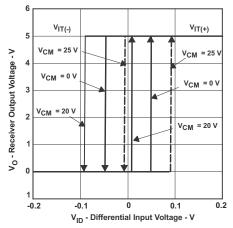
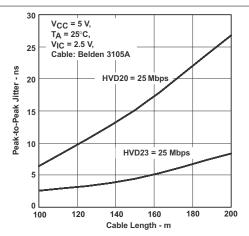


Figure 8-5. Receiver Output Voltage vs Differential Input Voltage Figure 8-6. SN65HVD2[0,3] Peak-to-Peak Jitter vs Cable Length





8.11 Typical Characteristics (continued)

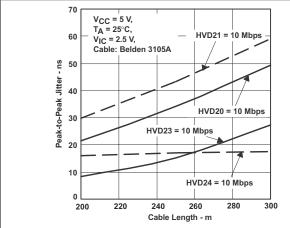


Figure 8-7. SN65HVD2[0,1,3,4] Peak-to-Peak Jitter vs Cable Length

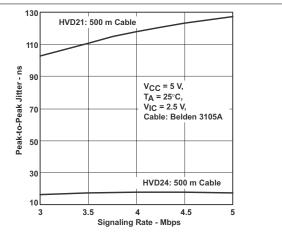


Figure 8-8. SN65HVD2[1,4] Peak-to-Peak Jitter vs Signaling Rate



9 Parameter Measurement Information

Note

Test load capacitance includes probe and jig capacitance (unless otherwise specified). Signal generator characteristics: rise and fall time < 6 ns, pulse rate 100 kHz, 50% duty cycle, $Z_O = 50$ (unless otherwise specified).

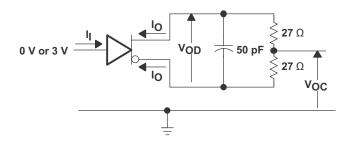


Figure 9-1. Driver Test Circuit, V_{OD} and V_{OC} Without Common-Mode Loading

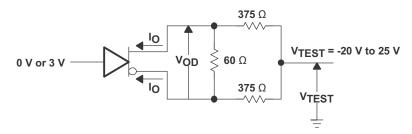


Figure 9-2. Driver Test Circuit, V_{OD} With Common-Mode Loading

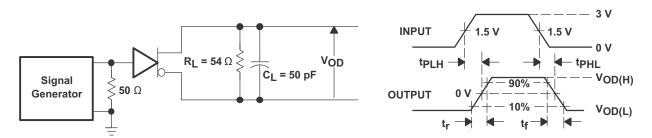


Figure 9-3. Driver Switching Test Circuit and Waveforms

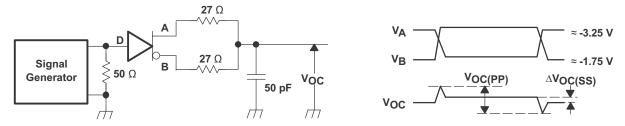
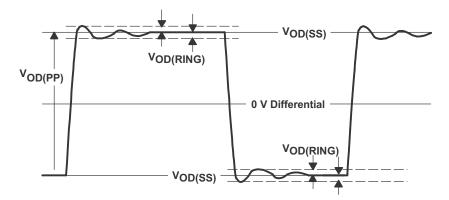


Figure 9-4. Driver V_{OC} Test Circuit and Waveforms



 $V_{OD(RING)}$ is measured at four points on the output waveform, corresponding to overshoot and undershoot from the $V_{OD(H)}$ and $V_{OD(L)}$ steady state values.

Figure 9-5. V_{OD(RING)} Waveform and Definitions

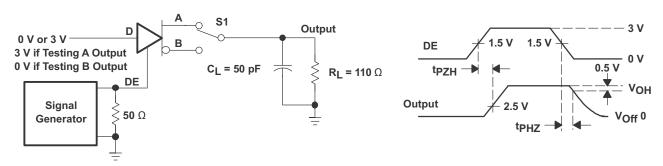


Figure 9-6. Driver Enable and Disable Test, High Output

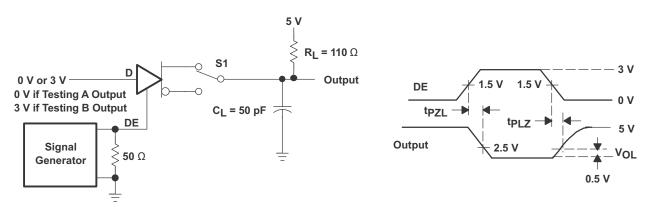


Figure 9-7. Driver Enable and Disable Test, Low Output

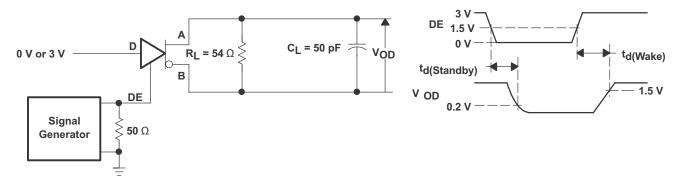


Figure 9-8. Driver Standby and Wake Test Circuit and Waveforms



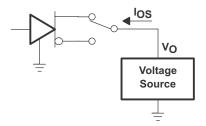


Figure 9-9. Driver Short-Circuit Test

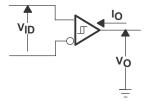


Figure 9-10. Receiver DC Parameter Definitions

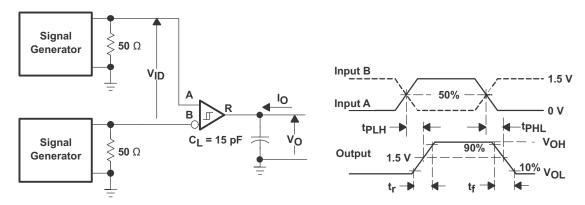


Figure 9-11. Receiver Switching Test Circuit and Waveforms

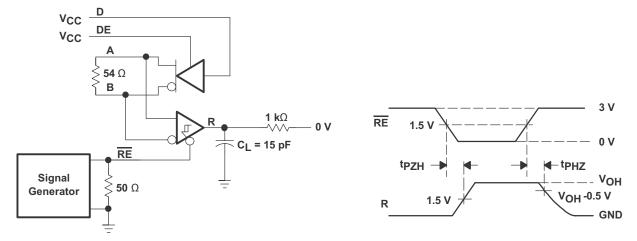


Figure 9-12. Receiver Enable Test Circuit and Waveforms, Data Output High

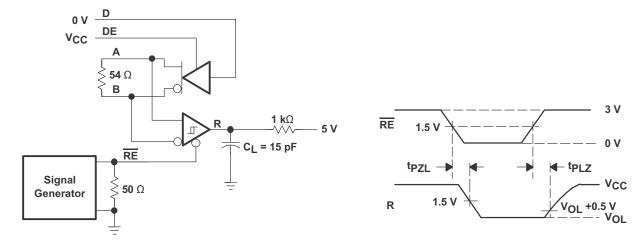


Figure 9-13. Receiver Enable Test Circuit and Waveforms, Data Output Low

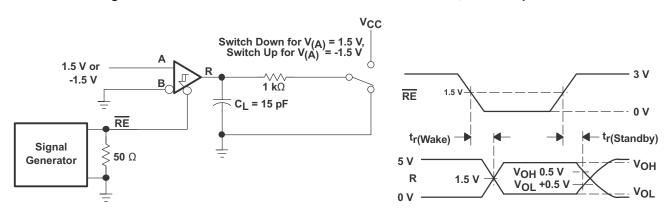


Figure 9-14. Receiver Standby and Wake Test Circuit and Waveforms

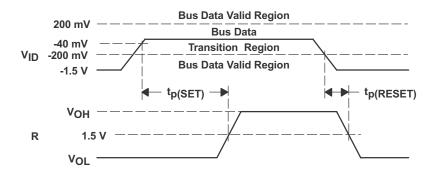


Figure 9-15. Receiver Active Failsafe Definitions and Waveforms

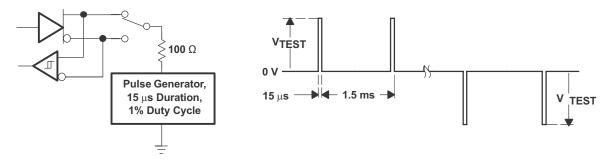


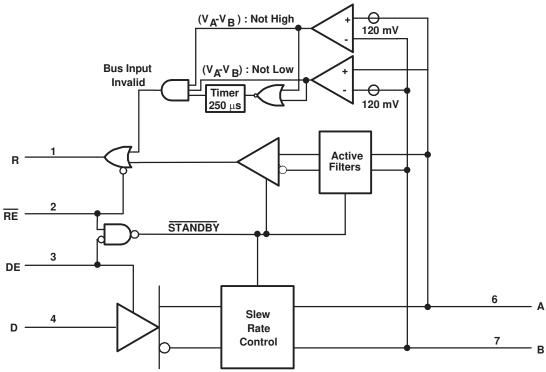
Figure 9-16. Test Circuit and Waveforms, Transient Overvoltage Test

10 Detailed Description

10.1 Overview

The SN65HVD2x family of devices are RS-485 compliant half-duplex transceivers designed for communication rates up to 500 kbps (SN65HVD22), 3 Mbps (SN65HVD24), 5 Mbps (SN65HVD21), or 25 Mpbs (SN65HVD20 and SN65HVD23). The devices feature extended common-mode range support, which provides immunity to larger ground potential differences that can occur between nodes that communicate over longer distances. The SN65HVD23 and the SN65HVD24 devices feature receiver equalization, which reduces the amount of data-dependent jitter that is introduced by the high-frequency losses associated with long cables.

10.2 Functional Block Diagram



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10.3 Feature Description

The SN65HVD2x family of devices integrates a differential receiver and differential driver with additional features for improved performance in electrically-noisy, long-cable, or other fault-intolerant applications.

The receiver hysteresis (typically 130 mV) is much larger than found in typical RS-485 transceivers. This helps reject spurious noise signals which would otherwise cause false changes in the receiver output state.

Slew rate limiting on the driver outputs (SN65HVD2[1,2,4]) reduces the high-frequency content of signal edges. This decreases reflections from bus discontinuities, and allows longer stub lengths between nodes and the main bus line. Designers must consider the maximum signaling rate and cable length required for a specific application, and choose the transceiver best matching those requirements.

When DE is low, the differential driver is disabled, and the A and B outputs are in high-impedance states. When DE is high, the differential driver is enabled, and drives the A and B outputs according to the state of the D inputs.

When \overline{RE} is high, the differential receiver output buffer is disabled, and the R output is in a high-impedance state. When \overline{RE} is low, the differential receiver is enabled, and the R output reflects the state of the differential bus inputs on the A and B pins.



If both the driver and receiver are disabled, (DE low and \overline{RE} high) then all nonessential circuitry, including auxiliary functions such as failsafe and receiver equalization is placed in a low-power standby state. This reduces power consumption to less than 5 μ W. When either enable input is asserted, the circuitry again becomes active.

In addition to the primary differential receiver, these devices incorporate a set of comparators and logic to implement an active receiver failsafe feature. These components determine whether the differential bus signal is valid. Whenever the differential signal is close to zero volts (neither high nor low), a timer initiates, If the differential input remains within the transition range for more than 250 μ s, the timer expires and set the receiver output to the high state. If a valid bus input (high or low) is received at any time, the receiver output reflects the valid bus state, and the timer is reset.

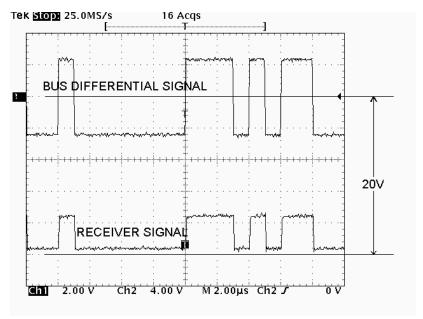


Figure 10-1. SN65HVD22 Receiver Operation With 20-V Offset on Input Signal

$H(s) = k_0 \left[(1-k_1) + \frac{k_1 p_1}{(s+p_1)} \right] \left[(1-k_2) + \frac{k_2 p_2}{(s+p_2)} \right] \left[(1-k_3) + \frac{k_3 p_3}{(s+p_3)} \right]$	k0 (DC loss)	p1 (MHz)	k1	p2 (MHz)	k2	p3 (MHz)	k3
Similar to 160m of Belden 3105A	0.95	0.25	0.3	3.5	0.5	15	1
Similar to 250m of Belden 3105A	0.9	0.25	0.4	3.5	0.7	12	1
Similar to 500m of Belden 3105A	0.8	0.25	0.6	2.2	1	8	1
Similar to 1000m of Belden 3105A	0.6	0.3	1	3	1	6	1



Figure 10-2. Cable Attenuation Model for Jitter Measurements

10.4 Device Functional Modes

The driver and receiver behavior for different input conditions are shown in Table 10-1 and Table 10-2, respectively.

Table 10-1	. Driver	Function	Table ^(*)	I)
-------------------	----------	-----------------	----------------------	----

INPUT ENABLE OUTPUTS											
DEVICE	INPUT	ENABLE	OUTI	PUTS							
DEVICE	D	DE	Α	В							
	Н	Н	Н	L							
SN65HVD2[0,1,2]	L	Н	L	Н							
	X	L	Z	Z							
	X	OPEN	Z	Z							
	OPEN	Н	Н	L							
	Н	Н	Н	L							
	L	Н	L	Н							
SN65HVD2[3,4]	Х	L	Z	Z							
	Х	OPEN	Z	Z							
	OPEN	Н	L	Н							

(1) Legend: H = high level, L = low level, X = don't care, Z = high impedance (off), ? = indeterminate

Table 10-2. Receiver Function Table⁽¹⁾

DIFFERENTIAL INPUT V _{ID} = (V _A - V _B)	ENABLE RE	OUTPUT R
0.2 V ≤ V _{ID}	L	Н
-0.2 V < V _{ID} < 0.2 V	L	H ⁽²⁾
V _{ID} ≤ -0.2 V	L	L
Х	Н	Z
Х	OPEN	Z
Open circuit	L	Н
Short Circuit	L	Н
Idle (terminated) bus	L	Н

- (1) H = high level, L = low level, Z = high impedance (off)
- (2) If the differential input V_{ID} remains within the transition range for more than 250 μs, the integrated failsafe circuitry detects a bus fault, and set the receiver output to a high state. See Figure 9-15.

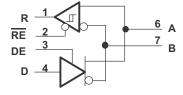


Figure 10-3. Logic Diagram



10.4.1 Test Mode Driver Disable

If the input signal to the D pin is such that:

- 1. the signal has signaling rate above 4 Mbps (for the SN65HVD21 and SN65HVD24),
- 2. the signal has signaling rate above 6 Mbps (for the SN65HVD20 and SN65HVD23),
- 3. the signal has average amplitude from 1.2 V to 1.6 V (1.4 V ± 200 mV), or
- 4. the average signal amplitude remains in this range for 100 µs or longer,

then the driver may activate a test-mode during which the driver outputs are temporarily disabled. This can cause loss of transmission of data during the period that the device is in the test-mode. The driver is re-enabled and resumes normal operation whenever the above conditions are not true. The device is not damaged by this test mode.

Although rare, there are combinations of specific voltage levels and input data patterns within the operating conditions of the SN65HVD2x family which may lead to a temporary state where the driver outputs are disabled for a period of time.

Observations:

- The conditions for inadvertently entering the test mode are dependent on the levels, duration, and duty cycle
 of the logic signal input to the D pin. Operating input levels are specified as greater than 2 V for a logic HIGH
 input, and less than 0.8 V for a logic LOW input. Therefore, a valid steady-state logic input does not cause
 the device to activate the test mode
- 2. Only input signals with frequency content above 2 MHz (4 Mbps) have a possibility of activating the test mode. Therefore, this issue should not affect the normal operation of the SN65HVD22 (500 kbps).
- 3. For operating signaling rates of 4 Mbps (or above), the conditions stated above must remain true over a period of: 4 Mbps $\times 100 \ \mu s = 400 \ bits$. Therefore, a normal short message does not inadvertently activate the test model.
- 4. One example of an input signal which may cause the test mode to activate is a clock signal with frequency 3 MHz and 50% duty cycle (symmetric HIGH and LOW half-cycles) with logic HIGH levels of 2.4 V and logic LOW levels of 0.4 V. This signal applied to the D pin as a driver input would meet the criteria listed above, and may cause the test-mode to activate, which would disable the driver. This example situation may occur if the clock signal is generated from a microcontroller or logic chip with a 2.7-V supply.

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10.4.2 Equivalent Input and Output Schematic Diagrams

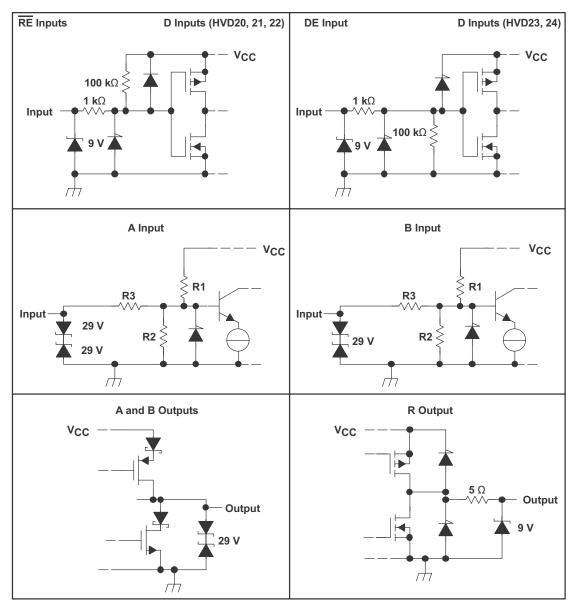


Figure 10-4. Equivalent Input and Output Schematic Diagrams

Table 10-3. Input and Output Resistor Values

DEVICE	R1, R2	R3
SN65HVD2[0,3]	9 kΩ	45 kΩ
SN65HVD2[1,2,4]	36 kΩ	180 kΩ



11 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

11.1 Application Information

The SN65HVD2x devices are half-duplex RS-485 transceivers that can be used for bidirectional, multipoint communication at various data rates over differential transmission lines. These devices support a wide common-mode range, allowing for robust communication even in the presence of voltage differences between the reference potentials of different nodes on a network.

11.2 Typical Application

Figure 11-1 shows a typical RS-485 application. Transceivers of different nodes are connected to one another over a shared bus. Twisted-pair cabling with a controlled differential impedance is used, and termination resistances are placed at the two ends of the cable to match the transmission line impedance and minimize signal reflections.

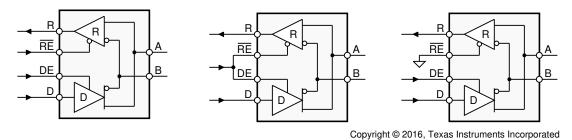


Figure 11-1. Half-Duplex Transceiver Configurations

11.2.1 Design Requirements

As the distances between nodes in an RS-485 network become greater and greater, it becomes more of a challenge to ensure reliable communication. The increased distance often means that the reference (ground) potentials has more of a difference between nodes. These ground potential differences give rise to differences in the common-mode voltages seen by the various transceivers on the bus. Standard RS-485 transceivers are typically specified to operate over a common-mode voltage from -7 V to 12 V, which may be insufficient for larger distances. The SN65HVD2x family of devices extends this range to -20 V to 25 V, allowing for greater communication distances between nodes.

Increased cable lengths can lead to increased jitter, especially in links operating at high data rates. This increased jitter is due to the attenuation of the cable, which tends to increase with frequency. Having unequal loss between higher and lower frequencies causes the RS-485 signal to distort, adding some timing deviation (jitter) to the edge crossings of the RS-485 data. If the jitter amplitude exceeds the jitter tolerance of the receiving MCU or UART, then bit errors are likely to result in the link. However, jitter can be reduced for a given link through the use of receiver equalization.

11.2.2 Detailed Design Procedure

11.2.2.1 Noise Considerations for Equalized Receivers

The simplest way of overcoming the effects of cable losses is to increase the sensitivity of the receiver. If the maximum attenuation of frequencies of interest is 20 dB, increasing the receiver gain by a factor of ten compensates for the cable. However, this means that signal and noise are amplified. Therefore, the receiver with

higher gain is more sensitive to noise and it is important to minimize differential noise coupling to the equalized receiver.

Differential noise is crated when conducted or radiated noise energy generates more voltage on one line of the differential pair than the other. For this to occur from conducted or electric far-field noise, the impedance to ground of the lines must differ.

For noise frequency out to 50 MHz, the input traces can be treated as a lumped capacitance if the receiver is approximately 10 inches or less from the connector. Therefore, matching impedance of the lines is accomplished by matching the lumped capacitance of each.

The primary factors that affect the capacitance of a trace are in length, thickness, width, dielectric material, distance from the signal return path, stray capacitance, and proximity to other conductors. It is difficult to match each of the variables for each line of the differential pair exactly, but a reasonable effort to do so keeps the lines balanced and less susceptible to differential noise coupling.

Another source of differential noise is from near-field coupling. In this situation, an assumption of equal noise-source impedance cannot be made as in the far-field. Familiarly known as crosstalk, more energy from a nearby signal is coupled to one line of the differential pair. Minimization of this differential noise is accomplished by keeping the signal pair close together and physical separation from high-voltage, high-current, or high-frequency signals.

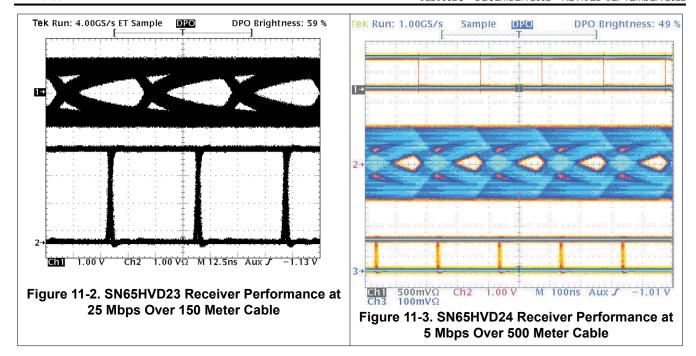
In summary, follow these guidelines in board layout for keeping differential noise to a minimum.

- · Keep the differential input traces short.
- Match the length, physical dimensions, and routing of each line of the pair.
- · Keep the lines close together.
- · Match components connected to each line.
- Separate the inputs from high-voltage, high-frequency, or high-current signals.

11.2.3 Application Curves

Figure 11-2 illustrates the benefits of integrated receiver equalization as implemented in the SN65HVD23 transceiver. In this test setup, a differential signal generator applied a signal voltage at one end of the cable, which was Belden 3105A twisted-pair shielded cable. The test signal was a pseudo-random bit stream (PRBS) of nonreturn-to-zero (NRZ) data. Channel 1 (top) shows the eye-pattern of the differential voltage at the receiver inputs (after the cable attenuation). Channel 2 (bottom) shows the output of the receiver.

Figure 11-3 illustrates the benefits of integrated receiver equalization as implemented in the SN65HVD24 transceiver. In this test setup, a differential signal generator applied a signal voltage at one end of the cable, which was Belden 3105A twisted-pair shielded cable. The test signal was a pseudo-random bit stream (PRBS) of nonreturn-to-zero (NRZ) data. Channel 1 (top) shows the eye-pattern of the bit stream. Channel 2 (middle) shows the eye-pattern of the differential voltage at the receiver inputs (after the cable attenuation). Channel 3 (bottom) shows the output of the receiver.



12 Power Supply Recommendations

To ensure reliable operation at all data rates and supply voltages, each supply must be decoupled with a 100-nF ceramic capacitor placed as close to the supply pins as possible. This helps to reduce supply voltage ripple present on the outputs of switched-mode power supplies and also helps to compensate for the resistance and inductance of the PCB power planes.



13 Layout

13.1 Layout Guidelines

In addition to the guidelines on differential trace matching given in Section 11.2.2, the layout guidelines below must be followed:

- Route power and ground nets as planes rather than traces, and keep their widths as large as possible to minimize resistance and inductance while maximizing parasitic capacitance.
- If external components (like transient voltage suppression diodes) are used for transient protection, place them close to the connector port and within the path of the signal lines. Make sure component capacitances are small enough not to impact the RS-485 signaling at the chosen data rate.
- Small-valued series pulse-proof resistances can be used to provide additional immunity to transients. This is needed to limit input currents if the clamping voltages of external transient protection devices exceed the absolute maximum ratings of the transceiver. These resistances must be less than 10 Ω so that the RS-485 signal is not overly attenuated.

13.2 Layout Example

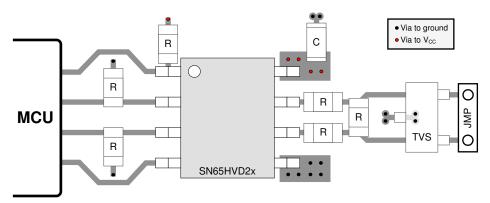


Figure 13-1. SN65HVD2x Layout Example



14 Device and Documentation Support

14.1 Receiving Notification of Documentation Updates

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

14.5 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

15 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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17-Jun-2025

PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
SN65HVD20DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP20
SN65HVD20DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP20
SN65HVD20DR1G4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP20
SN65HVD20DR1G4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP20
SN65HVD20P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD20
SN65HVD20P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD20
SN65HVD21D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP21
SN65HVD21DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP21
SN65HVD21DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP21
SN65HVD21DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP21
SN65HVD21DRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP21
SN65HVD21P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD21
SN65HVD21P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD21
SN65HVD22D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP22
SN65HVD22DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP22
SN65HVD22DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP22
SN65HVD22DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP22
SN65HVD22P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD22
SN65HVD22P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD22
SN65HVD23D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP23
SN65HVD23DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP23
SN65HVD23DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP23
SN65HVD23DRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP23
SN65HVD23DRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP23
SN65HVD23P	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD23
SN65HVD23P.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	65HVD23
SN65HVD24D	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	VP24
SN65HVD24DR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP24
SN65HVD24DR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	VP24

PACKAGE OPTION ADDENDUM

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- (1) Status: For more details on status, see our product life cycle.
- (2) Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.
- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
SN65HVD20DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD20DR1G4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD21DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD21DRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD22DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD23DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD23DRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
SN65HVD24DR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
SN65HVD20DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD20DR1G4	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD21DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD21DRG4	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD22DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD23DR	SOIC	D	8	2500	340.5	336.1	25.0
SN65HVD23DRG4	SOIC	D	8	2500	353.0	353.0	32.0
SN65HVD24DR	SOIC	D	8	2500	340.5	336.1	25.0

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
SN65HVD20P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD20P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD21P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD21P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD22P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD22P.A	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD23P	Р	PDIP	8	50	506	13.97	11230	4.32
SN65HVD23P.A	Р	PDIP	8	50	506	13.97	11230	4.32



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



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