













#### OPA348, OPA2348, OPA4348

SBOS213H-NOVEMBER 2001-REVISED NOVEMBER 2016

## OPAx348 1-MHz, 45-μA, CMOS, Rail-to-Rail Operational Amplifiers

#### 1 Features

- Low I<sub>O</sub>: 45 μA (Typical)
- Rail-To-Rail Input and Output
- Single Supply: 2.1 V to 5.5 V
- Input Bias Current: 0.5 pA
- Micro Size Packages:
  - 5-Pin SC70
  - 8-Pin SOT-23
  - 14-Pin TSSOP
- Excellent Bandwidth-to-Power Consumption Trade-off
- Number of Channels:

OPA348: 1OPA2348: 2OPA4348: 4

#### 2 Applications

- Portable Equipment
- Battery-Powered Equipment
- Smoke Alarms
- CO Detectors
- Medical Instrumentation

#### 3 Description

The OPAx348 series of amplifiers are single-supply, low-power, CMOS operational amplifiers. Featuring an extended bandwidth of 1 MHz, and a supply current of 45  $\mu$ A, the OPAx348 series is useful for low-power applications on single supplies of 2.1 V to 5.5 V.

A low supply current of  $45~\mu A$  and an input bias current of 0.5~pA, makes the OPAx348 series an optimal candidate for low-power applications such as smoke detectors and other high-impedance sensors.

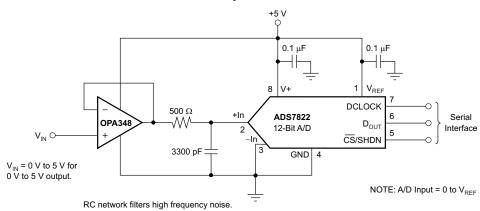
The OPA348 is available in the miniature 5-pin SC70 (SOT), 5-pin SOT-23 (SOT), and 8-pin SO (SOIC) packages. The OPA2348 is available in 8-pin SOT-23 (SOT) and 8-pin SO (SOIC) packages, and the OPA4348 is offered in space-saving 14-pin TSSOP and 14-pin SO (SOIC) packages. The extended temperature range of -40°C to +125°C over all supply voltages offers design flexibility.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)				
	SOIC (8)	4.90 mm × 3.91 mm				
OPA348	SOT-23 (5)	2.90 mm × 1.60 mm				
	SC70 (5)	2.00 mm × 1.25 mm				
	SOIC (8)	4.90 mm × 3.91 mm				
OPA2348	SOT-23 (8)	2.90 mm × 1.63 mm				
	VSSOP (8)	3.00 mm × 3.00 mm				
OPA4348	SOIC (14)	8.65 mm × 3.91 mm				
UPA4346	TSSOP (14)	5.00 mm × 4.40 mm				

(1) For all available packages, see the orderable addendum at the end of the data sheet.

#### **ADC Input Driver**



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#### 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

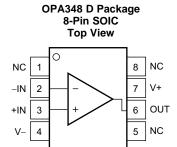
CI	nanges from Revision G (March 2013) to Revision H	Page
•	Added ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section	
•	Changed OPA348 DCK package designator from SOT to SC70 to match Package Option Addendum information	3
•	Deleted Lead temperature specification from Absolute Maximum Ratings table	<mark>6</mark>
•	Reformatted Thermal Information table note	<mark>7</mark>
<u>•</u>	Changed second and third paragraphs of <i>Driving A/D Converters</i> section to eliminate redundancy	17
CI	nanges from Revision F (October 2012) to Revision G	Page
<u>.</u>	Changed 2nd footnote for Absolute Maximum Ratings table	6
CI	nanges from Revision E (September 2012) to Revision F	Page
•	Deleted Packaging and Ordering information table data	1

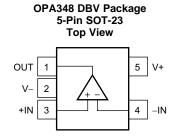
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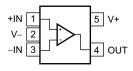


#### 5 Pin Configuration and Functions





OPA348 DCK Package 5-Pin SC70 (*Micro* size) Top View



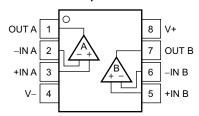
#### **Pin Functions: OPA348**

		PIN				
NAME	DBV (SOT-23)	DCK (SC70)	D (SOIC)	I/O DESCRIPTION		I/O
-IN	4	3	2	1	Negative (inverting) input	
+IN	3	1	3	- 1	Positive (noninverting) input	
NC	_	_	1, 5, 8	_	No internal connection (can be left floating)	
OUT	1	4	6	0	Output	
V-	2	2	4	_	Negative (lowest) power supply	
V+	5	5	7	_	Positive (highest) power supply	

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#### OPA2348 D, DCN, and DGK Packages 8-Pin SOIC, SOT, and VSSOP Top View

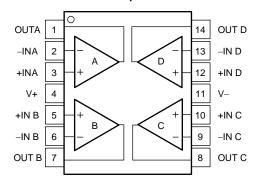


#### Pin Functions: OPA2348

PIN						
NAME	D (SOIC)	DCN (SOT-23)	DGK (VSSOP)	I/O	DESCRIPTION	
-IN A	2	2	2	I	Inverting input, channel A	
–IN B	6	6	6	ı	Inverting input, channel B	
+IN A	3	3	3	I	Noninverting input, channel A	
+IN B	5	5	5	I	Noninverting input, channel B	
OUT A	1	1	1	0	Output, channel A	
OUT B	7	7	7	0	Output, channel B	
V-	4	4	4	_	Negative (lowest) power supply	
V+	8	8	8	_	Positive (highest) power supply	



#### OPA4348 D and PW Packages 14-Pin SOIC and TSSOP Top View



#### Pin Functions: OPA4348

	PIN		-	In anothers. Of A4940	
NAME	D (SOIC)	PW (TSSOP)	I/O	DESCRIPTION	
-IN A	2	2	I	Inverting input, channel A	
–IN B	6	6	I	Inverting input, channel B	
-IN C	9	9	I	Inverting input, channel C	
–IN D	13	13	I	Inverting input, channel D	
+IN A	3	3	I	Noninverting input, channel A	
+IN B	5	5	I	Noninverting input, channel B	
+IN C	10	10	I	Noninverting input, channel C	
+IN D	12	12	I	Noninverting input, channel D	
OUT A	1	1	0	Output, channel A	
OUT B	7	7	0	Output, channel B	
OUT C	8	8	0	Output, channel C	
OUT D	14	14	0	Output, channel D	
V-	11	11	_	Negative (lowest) power supply	
V+	4	4	_	Positive (highest) power supply	

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#### 6 Specifications

#### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
Voltage	Supply voltage, $V_S = (V+) - (V-)$		7.5	V
Voltage	Signal input terminals, voltage <sup>(2)</sup>	(V-) - 0.5	(V+) + 0.5	V
Current	Signal input terminals, current <sup>(2)</sup>		10	mA
Current	Output short-circuit <sup>(3)</sup>	Conti	nuous	
	Junction, T <sub>J</sub>		150	
Temperature	Operating, T <sub>A</sub>	-65	150	°C
	Storage, T <sub>stg</sub>	-65	150	

<sup>(1)</sup> Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only. Functional operation of the device at these conditions, or beyond the specified operating conditions, is not implied.

#### 6.2 ESD Ratings

			VALUE	UNIT
V	Floatroatatio diacharga	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	\/
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±500	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

#### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
Supply voltage	2.1	5.5	V
Specified temperature	-40	125	°C

<sup>(2)</sup> Input terminals are not diode-clamped to the power-supply rails. Input signals that can swing more than 0.5 V beyond the supply rails must be current-limited to 10 mA or less.

<sup>(3)</sup> Short-circuit to ground, one amplifier per package.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



#### 6.4 Thermal Information: OPA348

			OPA348			
	THERMAL METRIC <sup>(1)</sup>	DBV (SOT-23)	DCK (SC70)	D (SOIC)	UNIT	
		5 PINS	5 PINS	8 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance	229	267	142	°C/W	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	99	81	90	°C/W	
$R_{\theta JB}$	Junction-to-board thermal resistance	55	55	83	°C/W	
ΨЈТ	Junction-to-top characterization parameter	7.7	1.2	40	°C/W	
ΨЈВ	Junction-to-board characterization parameter	54	54	82	°C/W	
R <sub>θ</sub> JC(bot)	Junction-to-case (bottom) thermal resistance	_	_	_	°C/W	

<sup>(1)</sup> For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics (SPRA953).

#### 6.5 Thermal Information: OPA2348

			OPA2348			
	THERMAL METRIC <sup>(1)</sup>	D (SOIC)	DGK (VSSOP)	DCN (SOT-23)	UNIT	
		8 PINS	8 PINS	8 PINS		
$R_{\theta JA}$	Junction-to-ambient thermal resistance	134	191	147	°C/W	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	90	83	115	°C/W	
$R_{\theta JB}$	Junction-to-board thermal resistance	79	112	32	°C/W	
ΨЈТ	Junction-to-top characterization parameter	30	18	38	°C/W	
ΨЈВ	Junction-to-board characterization parameter	78	110	33	°C/W	
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	_	°C/W	

<sup>(1)</sup> For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics (SPRA953).

#### 6.6 Thermal Information: OPA4348

		OP	OPA4348			
	THERMAL METRIC <sup>(1)</sup>	D (SOIC)	PW (TSSOP)	UNIT		
		14 PINS	14 PINS			
$R_{\theta JA}$	Junction-to-ambient thermal resistance	78	121	°C/W		
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	35	49	°C/W		
$R_{\theta JB}$	Junction-to-board thermal resistance	33	63	°C/W		
ΨЈТ	Junction-to-top characterization parameter	7	5.9	°C/W		
ΨЈВ	Junction-to-board characterization parameter	33	62	°C/W		
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	°C/W		

<sup>(1)</sup> For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics (SPRA953).

#### 6.7 Electrical Characteristics

at  $V_S = 2.5$  V to 5.5 V,  $T_A = 25$ °C,  $R_L = 100$  k $\Omega$  connected to  $V_S / 2$ , and  $V_{OUT} = V_S / 2$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT			
OFFSET '	VOLTAGE								
		$V_S = 5 \text{ V}, V_{CM} = (V-) + 0.8 \text{ V}$		1	5				
V <sub>OS</sub>	'OS Input offset voltage	$V_S = 5 \text{ V}, V_{CM} = (V-) + 0.8 \text{ V},$ at $T_A = -40^{\circ}\text{C}$ to 125°C			6	mV			
dV <sub>OS</sub> /dT	Input offset voltage drift	At $T_A = -40$ °C to 125°C		4		μV/°C			
		$V_S = 2.5 \text{ V to } 5.5 \text{ V}, V_{CM} < (V+) - 1.7 \text{ V}$		60	175				
PSRR	Input offset voltage versus power supply	At $T_A = -40^{\circ}\text{C}$ to 125°C, $V_S = 2.5 \text{ V}$ to 5.5 V, $V_{CM} < (V+) - 1.7 \text{ V}$			300	μV/V			
	Channel concretion	At dc		0.2		μV/V			
Channel separation		At f = 1 kHz		134		dB			
INPUT VO	DLTAGE RANGE								
V <sub>CM</sub>	Common-mode voltage range		(V-) - 0.2		(V+) + 0.2	V			

Product Folder Links: OPA348 OPA2348 OPA4348



#### **Electrical Characteristics (continued)**

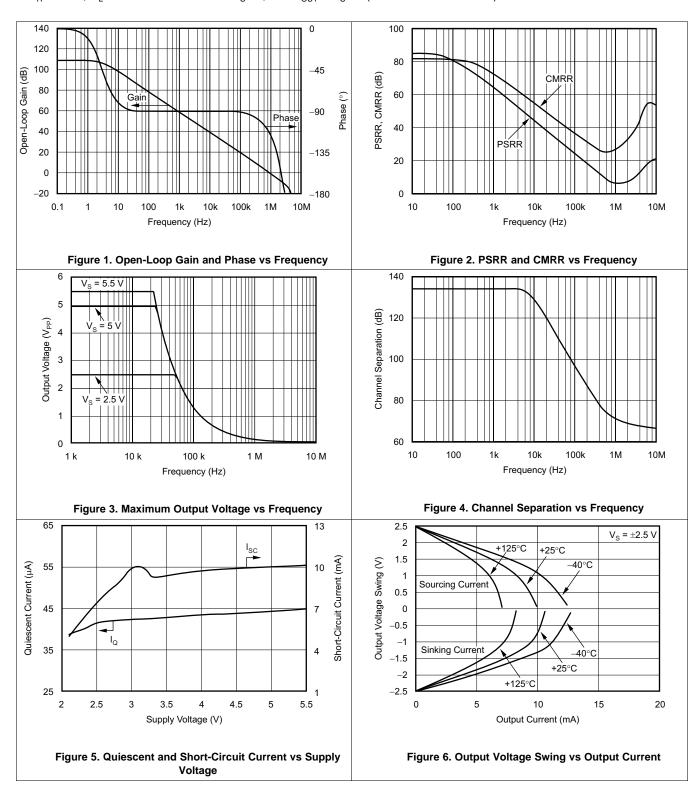
 $\underline{at~V_S} = 2.5~V~to~5.5~V,~T_A = 25^{\circ}C,~R_L = 100~k\Omega~connected~to~V_S~/~2,~and~V_{OUT} = V_S~/~2~(unless~otherwise~noted)$ 

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		(V-) - 0.2 V < V <sub>CM</sub> < (V+) - 1.7 V	70	82		
OMPR		$(V-) < V_{CM} < (V+) - 1.7 V$ , at $T_A = -40$ °C to 125°C	66			15
CMRR	Common-mode rejection ratio	$V_S = 5.5 \text{ V}, (V-) - 0.2 \text{ V} < V_{CM} < (V+) + 0.2 \text{ V}$	60	71		dB
		$V_S = 5.5 \text{ V}, (V-) < V_{CM} < (V+),$ at $T_A = -40^{\circ}\text{C}$ to 125°C	56			
INPUT B	IAS CURRENT					
$I_{B}$	Input bias current			±0.5	±10	pA
Ios	Input offset current			±0.5	±10	pA
INPUT IN	MPEDANCE					
	Differential			10 <sup>13</sup>    3		$\Omega \parallel pF$
	Common-mode			10 <sup>13</sup>    6		$\Omega \parallel pF$
NOISE						
	Input voltage noise	V <sub>CM</sub> < (V+) - 1.7 V, f = 0.1 Hz to 10 Hz		10		$\mu V_{PP}$
e <sub>n</sub>	Input voltage noise density	V <sub>CM</sub> < (V+) - 1.7 V, f = 1 kHz		35		nV/ <del>Hz</del>
i <sub>n</sub>	Input current noise density	V <sub>CM</sub> < (V+) - 1.7 V, f = 1 kHz		4		fA/Hz
OPEN-LO	OOP GAIN					
		$V_S = 5 \text{ V}, R_L = 100 \text{ k}\Omega, \\ 0.025 \text{ V} < V_O < 4.975 \text{ V}$	94	108		
A <sub>OL</sub>	Open-loop voltage gain	$\begin{array}{c} V_S=5~V,~R_L=100~k\Omega,\\ 0.025~V< V_O<4.975~V,\\ \text{at}~T_A=-40^{\circ}\text{C}~\text{to}~125^{\circ}\text{C} \end{array}$ en-loop voltage gain			dB	
OL.		$V_S = 5 \text{ V}, R_L = 5 \text{ k}\Omega,$ 0.125 V < $V_O$ < 4.875 V	90	98		
		$V_S = 5 \text{ V}, R_L = 5 \text{ k}\Omega, 0.125 \text{ V} < V_O < 4.875 \text{ V},$ at $T_A = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$	88			
OUTPUT	•				·	
		$R_L = 100 \text{ k}\Omega, A_{OL} > 94 \text{ dB}$		18	25	
	W. II	$R_L$ = 100 kΩ, $A_{OL}$ > 90 dB, at $T_A$ = -40°C to 125°C				\/
	Voltage output swing from rail	$R_L = 5 \text{ k}\Omega, A_{OL} > 90 \text{ dB}$		100	125	mV
		$R_L = 5 \text{ k}\Omega$ , $A_{OL} > 88 \text{ dB}$ , at $T_A = -40^{\circ}\text{C}$ to 125°C			125	
I <sub>SC</sub>	Short-circuit current			±10		mA
C <sub>LOAD</sub>	Capacitive load drive		See Typic	al Characteristics		
FREQUE	NCY RESPONSE					
GBP	Gain-bandwidth product	C <sub>L</sub> = 100 pF		1		MHz
SR	Slew rate	C <sub>L</sub> = 100 pF, G = +1		0.5		V/µs
	Settling time, 0.1%	C <sub>L</sub> = 100 pF, V <sub>S</sub> = 5.5 V, 2-V Step, G = +1		5		
t <sub>S</sub>	Settling time, 0.01%	C <sub>L</sub> = 100 pF, V <sub>S</sub> = 5.5 V, 2-V Step, G = +1		7		μs
	Overload recovery time	C <sub>L</sub> = 100 pF, V <sub>IN</sub> × Gain > V <sub>S</sub>		1.6		μs
THD+N	Total harmonic distortion + noise	$C_L = 100 \text{ pF}, V_S = 5.5 \text{ V}, V_O = 3 \text{ V}_{PP},$ G = +1, f = 1  kHz		0.0023%		
POWER	SUPPLY					
Vs	Specified voltage		2.5		5.5	V
	Operating voltage			2.1	5.5	V
	Ovies and suggest (	I <sub>O</sub> = 0 mA		45	65	4
IQ	Quiescent current (per amplifier)	$I_{O} = 0$ mA, at $T_{A} = -40$ °C to 125°C			75	μΑ



#### 6.8 Typical Characteristics

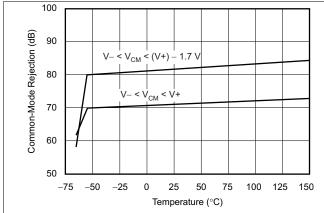
at  $T_A$  = 25°C,  $R_L$  = 100 k $\Omega$  connected to  $V_S$  / 2, and  $V_{OUT}$  =  $V_S$  / 2 (unless otherwise noted)



### TEXAS INSTRUMENTS

#### **Typical Characteristics (continued)**

at  $T_A$  = 25°C,  $R_L$  = 100 k $\Omega$  connected to  $V_S$  / 2, and  $V_{OUT}$  =  $V_S$  / 2 (unless otherwise noted)



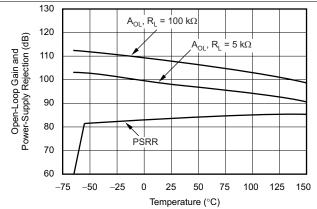
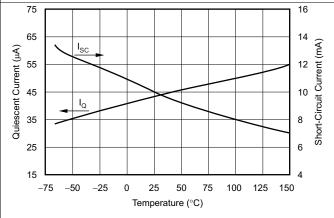


Figure 7. Common-Mode Rejection vs Temperature

Figure 8. Open-Loop Gain and PSRR vs Temperature



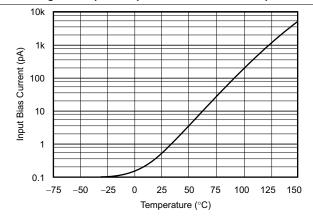
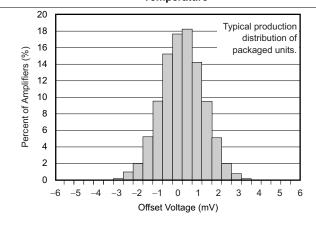


Figure 9. Quiescent and Short-Circuit Current vs
Temperature

Figure 10. Input Bias (I<sub>B</sub>) Current vs Temperature



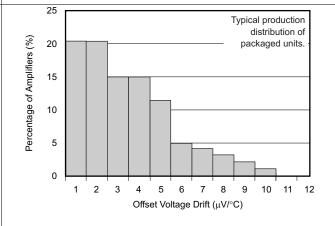


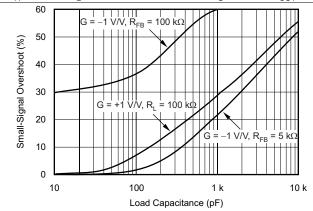
Figure 11. Offset Voltage Production Distribution

Figure 12. Offset Voltage Drift Magnitude Production Distribution



#### **Typical Characteristics (continued)**

at  $T_A$  = 25°C,  $R_L$  = 100 k $\Omega$  connected to  $V_S$  / 2, and  $V_{OUT}$  =  $V_S$  / 2 (unless otherwise noted)



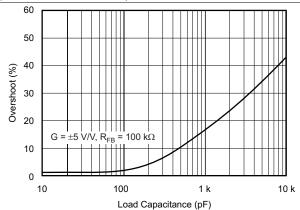
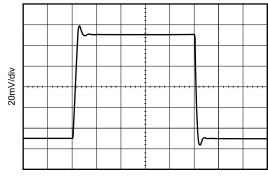
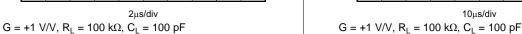
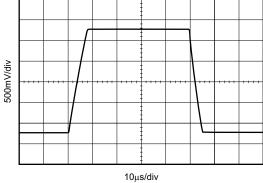


Figure 13. Small-Signal Overshoot vs Load Capacitance

Figure 14. Percent Overshoot vs Load Capacitance

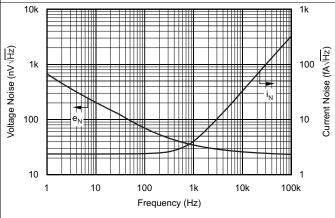






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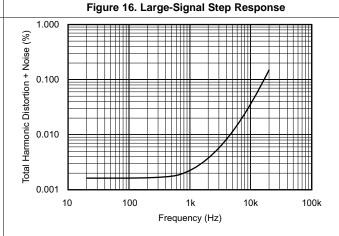


Figure 18. Total Harmonic Distortion + Noise vs Frequency

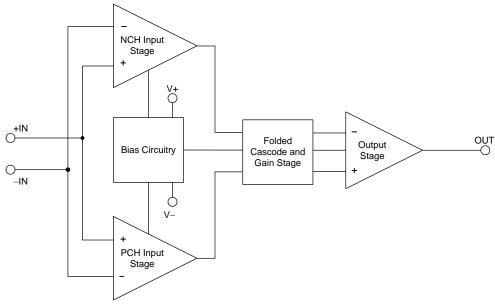


#### 7 Detailed Description

#### 7.1 Overview

The OPAx348 series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The OPAx348 series features wide bandwidth with rail-to-rail input and output for increased dynamic range.

#### 7.2 Functional Block Diagram



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#### 7.3 Feature Description

#### 7.3.1 Operating Voltage

The OPAx348 series op amps are fully specified and tested from 2.5 V to 5.5 V. However, supply voltage may range from 2.1 V to 5.5 V. Parameters are tested over the specified supply range which is an unique feature of the OPAx348 series. All temperature specifications apply from –40°C to +125°C. Most behavior remains virtually unchanged throughout the full operating voltage range. Parameters that vary significantly with operating voltages or temperature are shown in the *Typical Characteristics* section.



#### **Feature Description (continued)**

#### 7.3.2 Common-Mode Voltage Range

The input common-mode voltage range of the OPA348 series extends 200 mV beyond the supply rails. This extended range is achieved with a complementary input stage which is a N-channel input differential pair in parallel with a P-channel differential pair. The N-channel pair is active for input voltages close to the positive rail, typically (V+) - 1.2 V to 300 mV above the positive supply, while the P-channel pair is on for inputs from 300 mV below the negative supply to approximately (V+) - 1.4 V. There is a small transition region, typically (V+) - 1.4 V to (V+) - 1.2 V, in which both pairs are on. This 200-mV transition region, shown in Figure 19, can vary  $\pm 300 \text{ mV}$  with process variation. Thus, the transition region (both stages on) ranges from (V+) - 1.7 V to (V+) - 1.5 V on the low end, up to (V+) - 1.1 V to (V+) - 0.9 V on the high end. Within the 200-mV transition region PSRR, CMRR, offset voltage, offset drift, and THD may be degraded compared to operation outside this region.

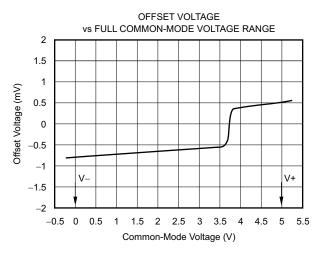


Figure 19. Behavior of Typical Transition Region at Room Temperature

#### 7.3.3 Rail-To-Rail Input

The input common-mode range extends from (V-) - 0.2 V to (V+) + 0.2 V. For normal operation, inputs must be limited to this range. The absolute maximum input voltage is 500 mV beyond the supplies. Inputs greater than the input common-mode range but less than the maximum input voltage, while not valid, do not cause any damage to the op amp. Unlike some other op amps, if input current is limited the inputs may go beyond the power supplies without phase inversion; see Figure 20.

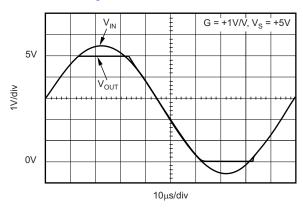
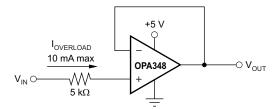


Figure 20. OPA348: No Phase Inversion with Inputs Greater Than The Power-Supply Voltage



#### **Feature Description (continued)**

Normally, input currents are 0.5 pA. However, large inputs (greater than 500 mV beyond the supply rails) can cause excessive current to flow in or out of the input pins. Therefore, as well as keeping the input voltage below the maximum rating, it is also important to limit the input current to less than 10 mA. This limiting is easily accomplished with an input voltage resistor, as shown in Figure 21.



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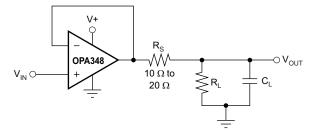
Figure 21. Input Current Protection for Voltages Exceeding the Supply Voltage

#### 7.3.4 Rail-To-Rail Output

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. This output stage is capable of driving 5-k $\Omega$  loads connected to any potential between V+ and ground. For light resistive loads (> 100 k $\Omega$ ), the output voltage can typically swing to within 18 mV from supply rail. With moderately resistive loads (10 k $\Omega$  to 50 k $\Omega$ ), the output voltage can typically swing to within 100 mV of the supply rails while maintaining high open-loop gain (see Figure 6).

#### 7.3.5 Capacitive Load and Stability

The OPA348 in a unity-gain configuration can directly drive up to 250 pF pure capacitive load. Increasing the gain enhances the ability of the amplifier to drive greater capacitive loads (see Figure 13). In unity-gain configurations, capacitive load drive can be improved by inserting a small (10- $\Omega$  to 20- $\Omega$ ) resistor, R<sub>S</sub>, in series with the output, as shown in Figure 22. This small resistor significantly reduces ringing while maintaining dc performance for purely capacitive loads. However, if there is a resistive load in parallel with the capacitive load, a voltage divider is created, introducing a direct current (dc) error at the output and slightly reducing the output swing. The error introduced is proportional to the ratio R<sub>S</sub> / R<sub>L</sub>, and is generally negligible.



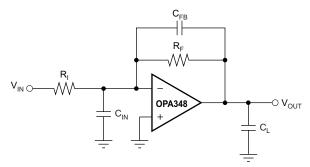
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Figure 22. Series Resistor in Unity-Gain Buffer Configuration Improves Capacitive Load Drive



#### **Feature Description (continued)**

In unity-gain inverter configuration, phase margin can be reduced by the reaction between the capacitance at the op amp input, and the gain setting resistors, thus degrading capacitive load drive. Best performance is achieved by using small valued resistors. For example, when driving a 500-pF load, reducing the resistor values from 100 k $\Omega$  to 5 k $\Omega$  decreases overshoot from 55% to 13% (see Figure 13). However, when large valued resistors cannot be avoided, a small (4-pF to 6-pF) capacitor,  $C_{FB}$ , can be inserted in the feedback, as shown in Figure 23. This configuration significantly reduces overshoot by compensating the effect of capacitance,  $C_{IN}$ , which includes the amplifier input capacitance and printed circuit board (PCB) parasitic capacitance.



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Figure 23. Improving Capacitive Load Drive

#### 7.4 Device Functional Modes

The OPAx348 has a single functional mode and is operational when the power-supply voltage is greater than 2.1 V (±1.05 V). The maximum power supply voltage for the OPAx348 is 5.5 V (±2.75 V).

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#### 8 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

#### 8.1 Application Information

The OPA348 amplifier is a single-supply, CMOS op amp with 1-MHz unity-gain bandwidth and supply current of only 45  $\mu$ A. Its performance is optimized for a lower power (2.1 V to 5.5 V), single-supply application, with its input common-mode voltage linear range extending 200 mV beyond the rails and the output voltage swing within 25 mV of either rail.

The OPA348 series features wide bandwidth and unity-gain stability with rail-to-rail input and output for increased dynamic range. Figure 24 shows the input and output waveforms for the OPA348 in unity-gain configuration. Operation is from a single 5-V supply with a  $100-k\Omega$  load connected to  $V_S$  / 2. The input is a  $5-V_{PP}$  sinusoid. Output voltage is approximately  $4.98~V_{PP}$ .

Power-supply pins must be bypassed with 0.01-µF ceramic capacitors.

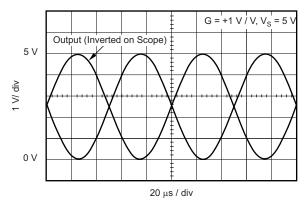


Figure 24. OPA348 Features Rail-to-Rail Input and Output

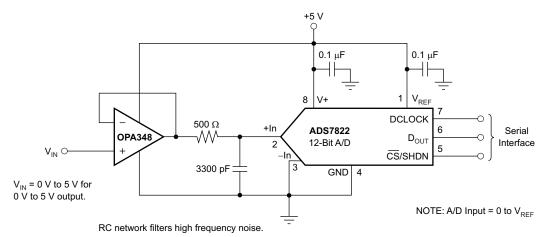


#### **Application Information (continued)**

#### 8.1.1 Driving A/D Converters

The OPA348 series op amps are optimized for driving medium-speed sampling analog-to-digital converters (ADCs). The OPA348 op amps buffer the ADC input capacitance and resulting charge injection while providing signal gain.

Figure 25 shows the OPA348 in a basic noninverting configuration driving the ADS7822. The ADS7822 is a 12-bit, *micro*POWER sampling converter in the MSOP-8 package. When used with the low-power, miniature packages of the OPA348, the combination is ideal for space-limited, low-power applications. In this configuration, an RC network at the ADC input can be used to provide both an anti-aliasing filter and charge injection current.



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Figure 25. OPA348 in Noninverting Configuration Driving ADS7822

Figure 26 illustrates the OPA2348 driving an ADS7822 in a speech-bandpass filtered data acquisition system. This small, low-cost solution provides the necessary amplification and signal conditioning to interface directly with an electret microphone. This circuit operates with  $V_S = 2.7 \text{ V}$  to 5 V with less than 250- $\mu$ A typical quiescent current.

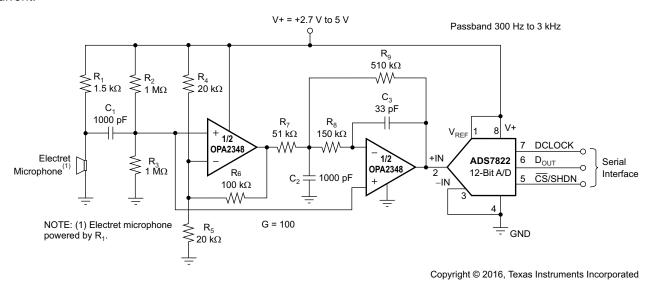
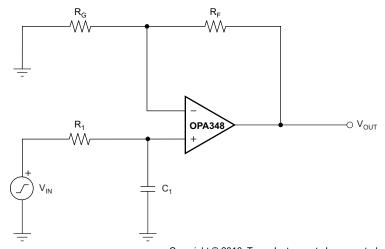


Figure 26. OPA2348 as a Speech-Bandpass Filtered Data Acquisition System



#### 8.2 Typical Application

Figure 27 shows the OPA348 in a typical noninverting application with input signal bandwidth limited by the input low-pass filter.



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Figure 27. Single-Pole, Low-Pass Filter

Equation 1 and Equation 2 show the relationships for the low-pass cutoff frequency and the low frequency gain and the passive elements surrounding the amplifier.

$$f_{-3 dB} = \frac{1}{2\pi R_1 C_1} \tag{1}$$

$$\frac{V_{OUT}}{V_{IN}} = \left(1 + \frac{R_F}{R_G}\right) \left(\frac{1}{1 + sR_1C_1}\right) \tag{2}$$

#### 8.2.1 Design Requirements

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier, as shown in Figure 27. If a steeper attenuation level is required, a two-pole or higher-order filter may be used.

#### 8.2.2 Detailed Design Procedure

The design goals for this circuit include these parameters:

- A noninverting gain of 10 V/V (20 dB)
- Design a single-pole response circuit with -3-dB roll-off at 15.9 kHz and 159 Hz
- Modify the design to increase attenuation level to -40-dB/decade (Sallen-Key Filter)

Use these design values:

- $C_1 = 0 \text{ nF}, 10 \text{ nF}, 1 \mu\text{F}$
- $R_1 = 1 k\Omega$
- $R_G = 10 \text{ k}\Omega$
- R<sub>F</sub> = 90 kΩ



#### Typical Application (continued)

Figure 28 shows how the output voltage of OPA348 changes over frequency depending on the value of C<sub>1</sub> with a constant  $R_1$  of 1 k $\Omega$ . Without any filtering of the input signal ( $C_1 = 0$ ), the -3-dB effective bandwidth is a function of the OPA348 unity-gain bandwidth and closed-loop gain,  $f_{(-3dB)} = UGBW/A_{CL}$ , where  $A_{CL}$  is closed-loop gain and UGBW denotes unity-gain bandwidth. Thus, for a closed-loop gain = 10, f<sub>(-3dB)</sub> = 1 MHz/10 =100 kHz; refer

To further limit the output bandwidth, an appropriate choice of  $C_1$  must be made: for  $C_1 = 10$  nF,  $f_C = \frac{1}{2\pi \times R_1 C_1} = \frac{1}{2\pi \times 1^3 \times 1^{-8}} = 15.9 \text{ kHz}.$ 

To further limit the bandwidth, a larger C₁ must be used:

choosing 
$$C_1 = 1 \ \mu F$$
,  $f_C = \frac{1}{2\pi \times R_1 C_1} = \frac{1}{2\pi \times 1^3 \times 1^{-6}} = 159 \ Hz$ . See Figure 28.

#### 8.2.3 Application Curve

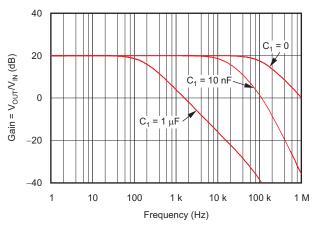


Figure 28. OPA348 Single-Pole AC Gain vs Frequency Response

If even more attenuation is required, a multiple pole filter is required. The Sallen-Key filter may be used for this task, as shown in Figure 29. For best results, the amplifier must have effective bandwidth that is at least 10 times higher than the filter cutoff frequency. Failure to follow this guideline results in a phase shift of the amplifier, which in turn leads to lower precision of the filter bandwidth. Additionally, in order to minimize the loading effect between multiple RC pairs on overall the filter cutoff frequency, choose R = 10  $\times$  R<sub>1</sub> and C<sub>2</sub> = C1/10; see Figure 29.

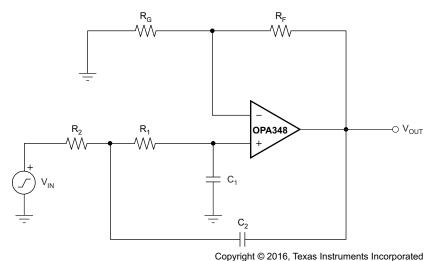


Figure 29. Two-Pole, Low-Pass Sallen-Key Filter



#### **Typical Application (continued)**

Equation 3, Equation 4, and Equation 5 show the relationships for low-pass cutoff frequency, filter transfer function, and low frequency gain, and the surrounding passive elements.

$$f_{\rm C} = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}}$$
 (3)

$$\frac{V_{OUT(s)}}{V_{IN(s)}} = \frac{G(2\pi f_c)^2}{s^2 + 2\zeta(2\pi f_c)s + (2\pi f_c)^2}$$
(4)

$$G = \frac{R_G + R_F}{R_G} \tag{5}$$

Use these design values:

- C<sub>1</sub> = 10 nF and C<sub>2</sub> = 1 nF
- $R_1 = 1 \text{ k}\Omega \text{ and } R_2 = 10 \text{ k}\Omega$
- $R_G = 10 \text{ k}\Omega$
- $R_F = 90 \text{ k}\Omega$

Figure 30 shows the Sallen-Key filter second-order response for different RC values: for R and C values above, 1 1

$$f_C = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}} = \frac{1}{2\pi\sqrt{1^3 \times 1^{-8} \times 1^4 \times 1^{-9}}} = 15.9 \text{ kHz.}$$

To further limit the bandwidth, a larger RC value must be used: increasing C values 100 times, such as  $C_1 = 1$   $\mu F$  and  $C_2 = 0.1$   $\mu F$ , with unchanged resistors, results in the second-order roll-off at  $f_C = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}} = \frac{1}{2\pi\sqrt{1^3\times 1^{-6}\times 1^4\times 1^{-7}}} = 159$  Hz. Refer to Figure 30.

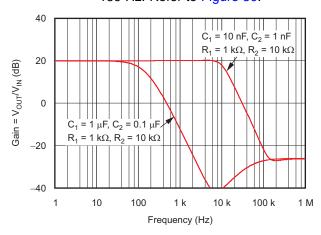


Figure 30. OPA348 Two-Pole, Low-Pass Sallen-Key AC Gain vs Frequency Response



#### 9 Power Supply Recommendations

The OPAx348 is specified for operation from 2.1 V to 5.5 V ( $\pm 1.05$  V to  $\pm 2.75$  V); many specifications apply from  $-40^{\circ}$ C to  $+125^{\circ}$ C. Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in the *Typical Characteristics*.

#### **CAUTION**

Supply voltages larger than 5.5 V can permanently damage the device (see the *Absolute Maximum Ratings* table).

Place 0.1-µF bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high impedance power supplies. For more detailed information on bypass capacitor placement, see *Layout*.

#### 10 Layout

#### 10.1 Layout Guidelines

For best operational performance of the device, use good PCB layout practices, including:

- Connect low equivalent series resistance (ESR), 0.1-µF ceramic bypass capacitors between each supply pin
  and ground, placed as close to the device as possible. A single bypass capacitor from V+ to ground is
  applicable for single-supply applications.
  - Noise can propagate into analog circuitry through the power pins of the circuit as a whole and op amp itself. Bypass capacitors are used to reduce the coupled noise by providing low-impedance power sources local to the analog circuitry.
- Make sure to physically separate digital and analog grounds paying attention to the flow of the ground current. Separate grounding for analog and digital portions of circuitry is one of the simplest and mosteffective methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes. A ground plane helps distribute heat and reduces EMI noise pickup.
- In order to reduce parasitic coupling, run the input traces as far away from the supply or output traces as
  possible. If these traces cannot be kept separate, crossing the sensitive trace perpendicular is much better as
  opposed to in parallel with the noisy trace.
- Place the external components as close to the device as possible. As shown in Figure 31, keeping RF and RG close to the inverting input minimizes parasitic capacitance.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.
- Clean the PCB following board assembly for best performance.
- Any precision integrated circuit may experience performance shifts due to moisture ingress into the plastic package. After any aqueous PCB cleaning process, bake the PCB assembly to remove moisture introduced into the device packaging during the cleaning process. A low-temperature, post-cleaning bake at 85°C for 30 minutes is sufficient for most circumstances.



#### 10.2 Layout Example

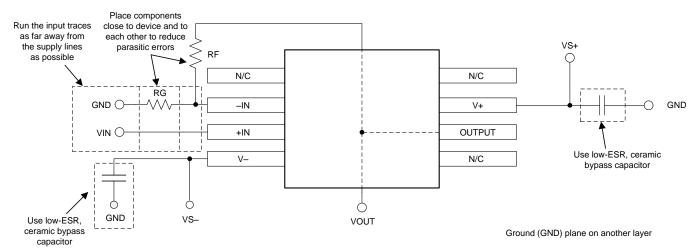


Figure 31. Operational Amplifier Board Layout for Noninverting Configuration



#### 11 Device and Documentation Support

#### 11.1 Device Support

#### 11.1.1 Development Support

#### 11.1.1.1 TINA-TI™ (Free Software Download)

TINA<sup>TM</sup> is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI<sup>TM</sup> is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional dc, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a free download from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic guick-start tool.

#### NOTE

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the TINA-TI folder.

#### 11.1.1.2 DIP Adapter EVM

The DIP Adapter EVM tool provides an easy, low-cost way to prototype small surface mount devices. The evaluation tool these TI packages: D or U (SOIC-8), PW (TSSOP-8), DGK (MSOP-8), DBV (SOT23-6, SOT23-5 and SOT23-3), DCK (SC70-6 and SC70-5), and DRL (SOT563-6). The DIP Adapter EVM may also be used with terminal strips or may be wired directly to existing circuits.

#### 11.1.1.3 Universal Op Amp EVM

The Universal Op Amp EVM is a series of general-purpose, blank circuit boards that simplify prototyping circuits for a variety of device package types. The evaluation module board design allows many different circuits to be constructed easily and quickly. Five models are offered, with each model intended for a specific package type. PDIP, SOIC, MSOP, TSSOP and SOT-23 packages are all supported.

#### NOTE

These boards are unpopulated, so users must provide their own devices. TI recommends requesting several op amp device samples when ordering the Universal Op Amp EVM.

#### 11.1.1.4 TI Precision Designs

TI Precision Designs are analog solutions created by TI's precision analog applications experts and offer the theory of operation, component selection, simulation, complete PCB schematic and layout, bill of materials, and measured performance of many useful circuits. TI Precision Designs are available online at <a href="http://www.ti.com/ww/en/analog/precision-designs/">http://www.ti.com/ww/en/analog/precision-designs/</a>.

#### 11.1.1.5 WEBENCH® Filter Designer

WEBENCH® Filter Designer is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

Available as a web-based tool from the WEBENCH® Design Center, WEBENCH® Filter Designer allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

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#### 11.2 Documentation Support

#### 11.2.1 Related Documentation

The following documents are relevant to using the OPAx348, and recommended for reference. All are available for download at www.ti.com unless otherwise noted.

- Hardware Pace Using Slope Detection (SLAU511).
- Mobile Phone Bank Card Reader Application Report (TIDU399).
- TPS61040 Inverter Design (SLVA008).
- Op Amp Performance Analysis (SBOA054).
- Single-Supply Operation of Operational Amplifiers (SBOA059).
- Tuning in Amplifiers (SBOA067).

#### 11.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 11.4 Related Links

Table 1 lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 1. Related Links

PARTS	PARTS PRODUCT FOLDER		TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
OPA348	Click here	Click here	Click here	Click here	Click here
OPA2348	Click here	Click here	Click here	Click here	Click here
OPA4348	Click here	Click here	Click here	Click here	Click here

#### 11.5 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on Alert me to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 11.6 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community T's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support TI's Design Support Quickly find helpful E2E forums along with design support tools and contact information for technical support.

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#### 11.8 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### 11.9 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

#### 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Product Folder Links: OPA348 OPA2348 OPA4348

www.ti.com

17-Jun-2025

#### **PACKAGING INFORMATION**

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
OPA2348AID	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA2348AID.B	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA2348AIDCNR	Active	Production	SOT-23 (DCN)   8	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDCNR.B	Active	Production	SOT-23 (DCN)   8	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDCNRG4	Active	Production	SOT-23 (DCN)   8	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDCNT	Active	Production	SOT-23 (DCN)   8	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDCNT.B	Active	Production	SOT-23 (DCN)   8	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDCNTG4	Active	Production	SOT-23 (DCN)   8	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	B48
OPA2348AIDG4	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA2348AIDGK	Active	Production	VSSOP (DGK)   8	80   TUBE	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OUTQ
OPA2348AIDGK.B	Active	Production	VSSOP (DGK)   8	80   TUBE	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OUTQ
OPA2348AIDGKR	Active	Production	VSSOP (DGK)   8	2500   LARGE T&R	Yes	NIPDAU   SN   NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OUTQ
OPA2348AIDGKR.B	Active	Production	VSSOP (DGK)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OUTQ
OPA2348AIDR	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA2348AIDR.B	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA2348AIDRG4	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 2348A
OPA348AID	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 348A
OPA348AID.B	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 348A
OPA348AIDBVR	Active	Production	SOT-23 (DBV)   5	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48
OPA348AIDBVR.B	Active	Production	SOT-23 (DBV)   5	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48
OPA348AIDBVRG4	Active	Production	SOT-23 (DBV)   5	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48
OPA348AIDBVT	Active	Production	SOT-23 (DBV)   5	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48





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Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
OPA348AIDBVT.B	Active	Production	SOT-23 (DBV)   5	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48
OPA348AIDBVTG4	Active	Production	SOT-23 (DBV)   5	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	A48
OPA348AIDCKR	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDCKR.B	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDCKRG4	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDCKT	Active	Production	SC70 (DCK)   5	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDCKT.B	Active	Production	SC70 (DCK)   5	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDCKTG4	Active	Production	SC70 (DCK)   5	250   SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	S48
OPA348AIDR	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 348A
OPA348AIDR.B	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 348A
OPA4348AID	Active	Production	SOIC (D)   14	50   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AID.B	Active	Production	SOIC (D)   14	50   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AIDG4	Active	Production	SOIC (D)   14	50   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AIDG4.B	Active	Production	SOIC (D)   14	50   TUBE	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AIDR	Active	Production	SOIC (D)   14	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AIDR.B	Active	Production	SOIC (D)   14	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA4348A
OPA4348AIPWR	Active	Production	TSSOP (PW)   14	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A
OPA4348AIPWR.B	Active	Production	TSSOP (PW)   14	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A
OPA4348AIPWRG4	Active	Production	TSSOP (PW)   14	2500   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A
OPA4348AIPWT	Active	Production	TSSOP (PW)   14	250   SMALL T&R	Yes	NIPDAU   NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A
OPA4348AIPWT.B	Active	Production	TSSOP (PW)   14	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A
OPA4348AIPWTG4	Active	Production	TSSOP (PW)   14	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPA 4348A

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

#### PACKAGE OPTION ADDENDUM

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- (2) Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.
- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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#### OTHER QUALIFIED VERSIONS OF OPA2348, OPA348, OPA4348:

Automotive: OPA2348-Q1, OPA348-Q1, OPA4348-Q1

NOTE: Qualified Version Definitions:

Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects



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#### TAPE AND REEL INFORMATION



# TAPE DIMENSIONS KO P1 BO BO Cavity AO

A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2348AIDCNR	SOT-23	DCN	8	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2348AIDCNT	SOT-23	DCN	8	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2348AIDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2348AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA348AIDBVR	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA348AIDBVT	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA348AIDBVT	SOT-23	DBV	5	250	178.0	9.0	3.23	3.17	1.37	4.0	8.0	Q3
OPA348AIDCKR	SC70	DCK	5	3000	179.0	8.4	2.2	2.5	1.2	4.0	8.0	Q3
OPA348AIDCKT	SC70	DCK	5	250	179.0	8.4	2.2	2.5	1.2	4.0	8.0	Q3
OPA348AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4348AIDR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
OPA4348AIPWR	TSSOP	PW	14	2500	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
OPA4348AIPWT	TSSOP	PW	14	250	180.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1



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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
	<b>-</b>				<b>-</b> ` '	. ,	
OPA2348AIDCNR	SOT-23	DCN	8	3000	200.0	183.0	25.0
OPA2348AIDCNT	SOT-23	DCN	8	250	200.0	183.0	25.0
OPA2348AIDGKR	VSSOP	DGK	8	2500	353.0	353.0	32.0
OPA2348AIDR	SOIC	D	8	2500	353.0	353.0	32.0
OPA348AIDBVR	SOT-23	DBV	5	3000	180.0	180.0	18.0
OPA348AIDBVT	SOT-23	DBV	5	250	195.0	200.0	45.0
OPA348AIDBVT	SOT-23	DBV	5	250	445.0	220.0	345.0
OPA348AIDCKR	SC70	DCK	5	3000	200.0	183.0	25.0
OPA348AIDCKT	SC70	DCK	5	250	200.0	183.0	25.0
OPA348AIDR	SOIC	D	8	2500	353.0	353.0	32.0
OPA4348AIDR	SOIC	D	14	2500	353.0	353.0	32.0
OPA4348AIPWR	TSSOP	PW	14	2500	353.0	353.0	32.0
OPA4348AIPWT	TSSOP	PW	14	250	213.0	191.0	35.0

#### **PACKAGE MATERIALS INFORMATION**

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#### **TUBE**



\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
OPA2348AID	D	SOIC	8	75	506.6	8	3940	4.32
OPA2348AID.B	D	SOIC	8	75	506.6	8	3940	4.32
OPA2348AIDG4	D	SOIC	8	75	506.6	8	3940	4.32
OPA2348AIDGK	DGK	VSSOP	8	80	330	6.55	500	2.88
OPA2348AIDGK.B	DGK	VSSOP	8	80	330	6.55	500	2.88
OPA348AID	D	SOIC	8	75	506.6	8	3940	4.32
OPA348AID.B	D	SOIC	8	75	506.6	8	3940	4.32
OPA4348AID	D	SOIC	14	50	506.6	8	3940	4.32
OPA4348AID.B	D	SOIC	14	50	506.6	8	3940	4.32
OPA4348AIDG4	D	SOIC	14	50	506.6	8	3940	4.32
OPA4348AIDG4.B	D	SOIC	14	50	506.6	8	3940	4.32



SMALL OUTLINE TRANSISTOR



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
  3. Reference JEDEC MO-203.

- 4. Support pin may differ or may not be present.5. Lead width does not comply with JEDEC.
- 6. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25mm per side



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

7. Publication IPC-7351 may have alternate designs.8. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 9. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 10. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE INTEGRATED CIRCUIT



#### NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.







- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE TRANSISTOR



- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
  3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.







PowerPAD is a trademark of Texas Instruments.

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.





- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 9. Size of metal pad may vary due to creepage requirement.





- 11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 12. Board assembly site may have different recommendations for stencil design.



DCN (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE (DIE DOWN)



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Package outline exclusive of metal burr & dambar protrusion/intrusion.
- D. Package outline inclusive of solder plating.
- E. A visual index feature must be located within the Pin 1 index area.
- F. Falls within JEDEC MO-178 Variation BA.
- G. Body dimensions do not include flash or protrusion. Mold flash and protrusion shall not exceed 0.25 per side.



DCN (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE (DIE DOWN)



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.







- 1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm, per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.43 mm, per side.
- 5. Reference JEDEC registration MS-012, variation AB.





NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.





- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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