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# MIXED SIGNAL MICROCONTROLLER

# FEATURES

- Low Supply Voltage Range 1.8 V to 3.6 V
- Ultra-Low Power Consumption
  - Active Mode: 220 µA at 1 MHz, 2.2 V
  - Standby Mode: 0.5 μA
  - Off Mode (RAM Retention): 0.1  $\mu$ A
- Five Power-Saving Modes
- Ultra-Fast Wake-Up From Standby Mode in Less Than 1  $\mu s$
- 16-Bit RISC Architecture, 62.5-ns Instruction Cycle Time
- Basic Clock Module Configurations:
  - Internal Frequencies up to 16 MHz With Four Calibrated Frequencies to ±1%
  - Internal Very Low-Power Low-Frequency Oscillator
  - 32-kHz Crystal (1)
  - External Digital Clock Source
- 16-Bit Timer\_A With Two Capture/Compare Registers
- 16-Bit Sigma-Delta A/D Converter With Differential PGA Inputs and Internal Reference <sup>(2)</sup>
- Universal Serial Interface (USI) Supporting SPI and I2C
- (1) Crystal oscillator cannot be operated beyond 105°C.
- (2) ADC performance characterized up to 105°C only.

- Brownout Detector
- Serial Onboard Programming, No External Programming Voltage Needed, Programmable Code Protection by Security Fuse
- On-Chip Emulation Logic With Spy-Bi-Wire Interface
- 2KB + 256B Flash Memory; 128B RAM
- Available in a 16-Pin QFN Package
- For Complete Module Descriptions, See the MSP430x2xx Family User's Guide (SLAU144)

### SUPPORTS DEFENSE, AEROSPACE, AND MEDICAL APPLICATIONS

- Controlled Baseline
- One Assembly/Test Site
- One Fabrication Site
- Available in Extended (-40°C/125°C) Temperature Range<sup>(3)</sup>
- Extended Product Life Cycle
- Extended Product-Change Notification
- Product Traceability
- (3) Custom temperature ranges available

# DESCRIPTION

The Texas Instruments MSP430 family of ultra-low-power microcontrollers consist of several devices featuring different sets of peripherals targeted for various applications. The architecture, combined with five low-power modes is optimized to achieve extended battery life in portable measurement applications. The device features a powerful 16-bit RISC CPU, 16-bit registers, and constant generators that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows wake-up from low-power modes to active mode in less than 1 µs.

The MSP430F2013 is an ultra-low-power mixed signal microcontroller with a built-in 16-bit timer and ten I/O pins. In addition, the MSP430F2013 has a built-in communication capability using synchronous protocols (SPI or I2C) and a 16-bit sigma-delta A/D converter.

Typical applications include sensor systems that capture analog signals, convert them to digital values, and then process the data for display or for transmission to a host system. Stand alone RF sensor front end is another area of application.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

# MSP430F2013-EP



#### SLAS774A - JULY 2011-REVISED OCTOBER 2011

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### Table 1. ORDERING INFORMATION<sup>(1)</sup>

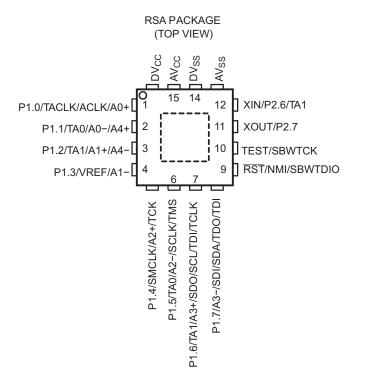
T <sub>A</sub>	PACKAGE <sup>(2)</sup>	ORDERABLE PART NUMBER	VID NUMBER
-40°C to 125°C	QFN (RSA)	MSP430F2013QRSATEP	V62/11613-01XE

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI Web site at www.ti.com.

(2) Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

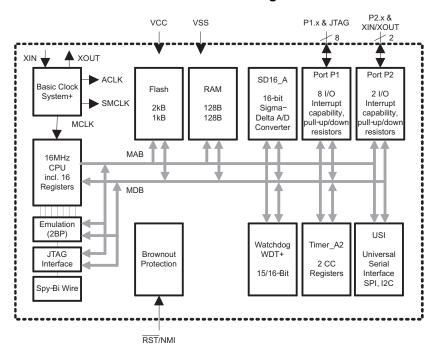
### **Device Pinout**

See port schematics section for detailed I/O information.





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#### **Functional Block Diagram**



Table 2. Terminal Functio	
	ne

TERMINAL			DECODIDION				
NAME	NO.	I/O	DESCRIPTION				
P1.0/TACLK/ACLK/A0+	1	I/O	General-purpose digital I/O pin Timer_A, clock signal TACLK input ACLK signal output SD16_A positive analog input A0				
P1.1/TA0/A0-/A4+	2	I/O	General-purpose digital I/O pin Timer_A, capture: CCI0A input, compare: Out0 output SD16_A negative analog input A0 SD16_A positive analog input A4				
P1.2/TA1/A1+/A4-	3	I/O	General-purpose digital I/O pin Timer_A, capture: CCI1A input, compare: Out1 output SD16_A positive analog input A1 SD16_A negative analog input A4				
P1.3/VREF/A1-	4	I/O	General-purpose digital I/O pin Input for an external reference voltage/internal reference voltage output (can be used a mid-voltage) SD16_A negative analog input A1				
P1.4/SMCLK/A2+/TCK	5	I/O	General-purpose digital I/O pin SMCLK signal output SD16_A positive analog input A2 JTAG test clock, input terminal for device programming and test				
P1.5/TA0/A2-/SCLK/TMS	6	I/O	General-purpose digital I/O pin Timer_A, compare: Out0 output SD16_A negative analog input A2 USI: external clock input in SPI or I2C mode; clock output in SPI mode JTAG test mode select, input terminal for device programming and test				
P1.6/TA1/A3+/SDO/SCL/ TDI/TCLK	7	I/O	General-purpose digital I/O pin Timer_A, capture: CCI1B input, compare: Out1 output SD16_A positive analog input A3 USI: Data output in SPI mode; I2C clock in I2C mode JTAG test data input or test clock input during programming and test				
P1.7/A3-/SDI/SDA/ TDO/TDI <sup>(1)</sup>	8	I/O	General-purpose digital I/O pin SD16_A negative analog input A3 USI: Data input in SPI mode; I2C data in I2C mode JTAG test data output terminal or test data input during programming and test				
XIN/P2.6/TA1	12	I/O	Input terminal of crystal oscillator General-purpose digital I/O pin Timer_A, compare: Out1 output				
XOUT/P2.7	11	I/O	Output terminal of crystal oscillator General-purpose digital I/O pin <sup>(2)</sup>				
RST/NMI/SBWTDIO	9	I	Reset or nonmaskable interrupt input Spy-Bi-Wire test data input/output during programming and test				
TEST/SBWTCK	10	I	Selects test mode for JTAG pins on Port 1. The device protection fuse is connected to TEST. Spy-Bi-Wire test clock input during programming and test				
DV <sub>CC</sub>	16		Digital supply voltage				
AV <sub>CC</sub>	15		Analog supply voltage				
DV <sub>SS</sub>	14		Digital ground reference				
AV <sub>SS</sub>	13		Analog ground reference				
QFN Pad	Pad	NA	QFN package pad. Connection to VSS is recommended.				

 TDO or TDI is selected via JTAG instruction.
 If XOUT/P2.7 is used as an input, excess current flows until P2SEL.7 is cleared. This is due to the oscillator output driver connection to this pad after reset.



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## SHORT-FORM DESCRIPTION

### CPU

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The MSP430 CPU has a 16-bit RISC architecture that is highly transparent to the application. All operations, other than program-flow instructions, are performed as register operations in conjunction with seven addressing modes for source operand and four addressing modes for destination operand.

The CPU is integrated with 16 registers that provide reduced instruction execution time. The register-to-register operation execution time is one cycle of the CPU clock.

Four of the registers, R0 to R3, are dedicated as program counter, stack pointer, status register, and constant generator respectively. The remaining registers are general-purpose registers.

Peripherals are connected to the CPU using data, address, and control buses, and can be handled with all instructions.

#### **Instruction Set**

The instruction set consists of 51 instructions with three formats and seven address modes. Each instruction can operate on word and byte data. Table 3 shows examples of the three types of instruction formats; Table 4 shows the address modes.

Program Counter	PC/R0
Stack Pointer	SP/R1
Status Register	SR/CG1/R2
Constant Generator	CG2/R3
General-Purpose Register	R4
General-Purpose Register	R5
General-Purpose Register	R6
General-Purpose Register	R7
General-Purpose Register	R8
General-Purpose Register	R9
General-Purpose Register	R10
General-Purpose Register	R11
General-Purpose Register	R12
General-Purpose Register	R13
General-Purpose Register	R14
General-Purpose Register	R15

#### Table 3. Instruction Word Formats

INSTRUCTION FORMAT	EXAMPLE	OPERATION
Dual operands, source-destination	ADD R4,R5	R4 + R5> R5
Single operands, destination only	CALL R8	PC>(TOS), R8> PC
Relative jump, un/conditional	JNE	Jump-on-equal bit = 0

#### Table 4. Address Mode Descriptions

ADDRESS MODE	S <sup>(1)</sup>	D <sup>(1)</sup>	SYNTAX	EXAMPLE	OPERATION
Register	$\checkmark$	1	MOV Rs,Rd	MOV R10,R11	R10> R11
Indexed	$\checkmark$	1	MOV X(Rn),Y(Rm)	MOV 2(R5),6(R6)	M(2+R5)> M(6+R6)
Symbolic (PC relative)	$\checkmark$	1	MOV EDE, TONI		M(EDE)> M(TONI)
Absolute	~	~	MOV &MEM,&TCDAT		M(MEM)> M(TCDAT)
Indirect	~		MOV @Rn,Y(Rm)	MOV @R10,Tab(R6)	M(R10)> M(Tab+R6)
Indirect autoincrement	utoincrement  V MOV @Rn+,Rm		MOV @R10+,R11	M(R10)> R11 R10 + 2> R10	
Immediate	$\checkmark$		MOV #X,TONI	MOV #45,TONI	#45> M(TONI)

(1) S = source, D = destination



SLAS774A - JULY 2011 - REVISED OCTOBER 2011

### **Operating Modes**

The MSP430 has one active mode and five software-selectable low-power modes of operation. An interrupt event can wake the device from any of the five low-power modes, service the request, and restore back to the low-power mode on return from the interrupt program.

The following six operating modes can be configured by software:

- Active mode (AM)
- All clocks are active
- Low-power mode 0 (LPM0)
  - CPU is disabled
  - ACLK and SMCLK remain active
  - MCLK is disabled
- Low-power mode 1 (LPM1)
  - CPU is disabled
  - ACLK and SMCLK remain active. MCLK is disabled
  - DCO's dc-generator is disabled if DCO not used in active mode
- Low-power mode 2 (LPM2)
  - CPU is disabled
  - MCLK and SMCLK are disabled
  - DCO's dc-generator remains enabled
  - ACLK remains active
- Low-power mode 3 (LPM3)
  - CPU is disabled
  - MCLK and SMCLK are disabled
  - DCO's dc-generator is disabled
  - ACLK remains active
- Low-power mode 4 (LPM4)
  - CPU is disabled
  - ACLK is disabled
  - MCLK and SMCLK are disabled
  - DCO's dc-generator is disabled
  - Crystal oscillator is stopped



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#### **Interrupt Vector Addresses**

The interrupt vectors and the power-up starting address are located in the address range of 0FFFFh to 0FFC0h. The vector contains the 16-bit address of the appropriate interrupt handler instruction sequence.

If the reset vector (located at address 0FFFEh) contains 0FFFFh (for example, flash is not programmed) the CPU goes into LPM4 immediately after power-up.

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY	
Power-up External reset Watchdog Timer+ Flash key violation PC out-of-range <sup>(1)</sup>	PORIFG RSTIFG WDTIFG KEYV See <sup>(2)</sup>	Reset	0FFFEh	31, highest	
NMI Oscillator fault Flash memory access violation	NMIIFG OFIFG ACCVIFG <sup>(2)(3)</sup>	(non)-maskable, (non)-maskable, (non)-maskable	0FFFCh	30	
			0FFFAh	29	
			0FFF8h	28	
Watchdog Timer+	WDTIFG	maskable	0FFF4h	26	
Timer_A2	TACCR0 CCIFG <sup>(4)</sup>	maskable	0FFF2h	25	
Timer_A2	TACCR1 CCIFG.TAIFG <sup>(2)(4)</sup>	maskable	0FFF0h	24	
			0FFEEh	23	
			0FFECh	22	
SD16_A	SD16CCTL0 SD16OVIFG, SD16CCTL0 SD16IFG <sup>(2)(4)</sup>	maskable			
USI	USIIFG, USISTTIFG <sup>(2)(4)</sup>	maskable	0FFE8h	20	
I/O Port P2 (two flags)	P2IFG.6 to P2IFG.7 <sup>(2)(4)</sup>	maskable	0FFE6h	19	
I/O Port P1 (eight flags)	P1IFG.0 to P1IFG.7 <sup>(2)(4)</sup>	maskable	0FFE4h	18	
			0FFE2h	17	
			0FFE0h	16	
See <sup>(5)</sup>			0FFDEh to 0FFC0h	15 to 0, lowest	

#### **Table 5. Interrupt Sources**

(1) A reset is generated if the CPU tries to fetch instructions from within the module register memory address range (0h to 01FFh) or from within unused address ranges.

(2) Multiple source flags

(3) (non)-maskable: the individual interrupt-enable bit can disable an interrupt event, but the general interrupt enable cannot.

(4) Interrupt flags are located in the module.

(5) The interrupt vectors at addresses 0FFDEh to 0FFC0h are not used in this device and can be used for regular program code if necessary.

#### **Special Function Registers**

Most interrupt and module enable bits are collected into the lowest address space. Special function register bits not allocated to a functional purpose are not physically present in the device. Simple software access is provided with this arrangement.

Legend	rw:	Bit can be read and written.
	rw-0,1:	Bit can be read and written. It is reset or set by PUC.
	rw-(0,1):	Bit can be read and written. It is reset or set by POR.
		SFR bit is not present in device.

#### Table 6. Interrupt Enable Register 1 and 2

Address	7	6	5	4	3	2	1	0		
00h			ACCVIE	NMIIE			OFIE	WDTIE		
			rw-0	rw-0			rw-0	rw-0		
WDTIE		Watchdog Timer interrupt enable. Inactive if watchdog mode is selected. Active if Watchdog Timer is configured in interval timer mode.								
OFIE	Oscillator	fault interrupt e	enable							
NMIIE	(Non)mas	(Non)maskable interrupt enable								
ACCVIE	Flash acc	ess violation in	terrupt enable							

Address	7	6	5	4	3	2	1	0
01h								

#### Table 7. Interrupt Flag Register 1 and 2

Address	7	6	5	4	3	2	1	0		
02h				NMIIFG	RSTIFG	PORIFG	OFIFG	WDTIFG		
				rw-0	rw-(0)	rw-(1)	rw-1	rw-(0)		
WDTIFG			verflow (in watch or a reset conditi							
OFIFG	Flag set or	Flag set on oscillator fault.								
PORIFG	Power-On	Reset interrup	t flag. Set on V <sub>C</sub>	<sub>C</sub> power-up.						
RSTIFG	External re	eset interrupt fla	ag. Set on a res	et condition at F	ST/NMI pin in r	eset mode. Res	et on V <sub>CC</sub> powe	er-up.		
NMIIFG	Set via RS	T/NMI pin	-		·					
	7	C	F	4	2	2	4	0		

Address	7	6	5	4	3	2	1	0
03h								



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### Memory Organization

· · · · · · · · · · · · · · · · · · ·							
		MSP430F200x	MSP430F201x				
Memory	Size	1KB Flash	2KB Flash				
Main: interrupt vector	Flash	0FFFFh-0FFC0h	0FFFFh-0FFC0h				
Main: code memory	Flash	0FFFFh-0FC00h	0FFFFh-0F800h				
Information memory	Size	256 Byte	256 Byte				
	Flash	010FFh - 01000h	010FFh - 01000h				
RAM	Size	128 Byte 027Fh - 0200h	128 Byte 027Fh - 0200h				
Peripherals	16-bit	01FFh - 0100h	01FFh - 0100h				
	8-bit	0FFh - 010h	0FFh - 010h				
	8-bit SFR	0Fh - 00h	0Fh - 00h				

### Table 8. Memory Organization

## **Flash Memory**

The flash memory can be programmed via the Spy-Bi-Wire/JTAG port, or in-system by the CPU. The CPU can perform single-byte and single-word writes to the flash memory. Features of the flash memory include:

- Flash memory has n segments of main memory and four segments of information memory (A to D) of 64 bytes each. Each segment in main memory is 512 bytes in size.
- Segments 0 to n may be erased in one step, or each segment may be individually erased.
- Segments A to D can be erased individually, or as a group with segments 0 to n. Segments A to D are also called *information memory*.
- Segment A contains calibration data. After reset segment A is protected against programming and erasing. It
  can be unlocked but care should be taken not to erase this segment if the device-specific calibration data is
  required.

#### SLAS774A – JULY 2011 – REVISED OCTOBER 2011

#### Peripherals

Peripherals are connected to the CPU through data, address, and control busses and can be handled using all instructions. For complete module descriptions, refer to the *MSP430F2xx Family User's Guide*.

#### **Oscillator and System Clock**

The clock system is supported by the basic clock module that includes support for a 32768-Hz watch crystal oscillator, an internal very-low-power low-frequency oscillator and an internal digitally-controlled oscillator (DCO). The basic clock module is designed to meet the requirements of both low system cost and low power consumption. The internal DCO provides a fast turn-on clock source and stabilizes in less than 1  $\mu$ s. The basic clock module provides the following clock signals:

- Auxiliary clock (ACLK), sourced either from a 32768-Hz watch crystal or the internal LF oscillator.
- Main clock (MCLK), the system clock used by the CPU.
- Sub-Main clock (SMCLK), the sub-system clock used by the peripheral modules.

Memory Segment A)								
DCO FREQUENCY	CALIBRATION REGISTER	SIZE	ADDRESS					
1 MHz	CALBC1_1MHZ	byte	010FFh					
	CALDCO_1MHZ	byte	010FEh					
8 MHz	CALBC1_8MHZ	byte	010FDh					
	CALDCO_8MHZ	byte	010FCh					
12 MHz	CALBC1_12MHZ	byte	010FBh					
	CALDCO_12MHZ	byte	010FAh					
16 MHz	CALBC1_16MHZ	byte	010F9h					
	CALDCO_16MHZ	byte	010F8h					

#### Table 9. DCO Calibration Data (Provided From Factory in Flash Information Memory Segment A)

#### Brownout

The brownout circuit is implemented to provide the proper internal reset signal to the device during power on and power off.

### **Digital I/O**

There is one 8-bit I/O port implemented—port P1—and two bits of I/O port P2:

- All individual I/O bits are independently programmable.
- Any combination of input, output, and interrupt condition is possible.
- Edge-selectable interrupt input capability for all the eight bits of port P1 and the two bits of port P2.
- Read/write access to port-control registers is supported by all instructions.
- Each I/O has an individually programmable pullup/pulldown resistor.

#### Watchdog Timer (WDT+)

The primary function of the watchdog timer (WDT+) module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be disabled or configured as an interval timer and can generate interrupts at selected time intervals.



### Timer\_A2

Timer\_A2 is a 16-bit timer/counter with two capture/compare registers. Timer\_A2 can support multiple capture/compares, PWM outputs, and interval timing. Timer\_A2 also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

INPUT PIN	NUMBER	DEVICE INPUT	MODULE	MODULE	MODULE	OUTPUT P	N NUMBER
PW, N	RSA	SIGNAL	INPUT NAME	BLOCK	OUTPUT SIGNAL	PW, N	RSA
2 - P1.0	1 - P1.0	TACLK	TACLK	Timer	NA		
		ACLK	ACLK				
		SMCLK	SMCLK				
2 - P1.0	1 - P1.0	TACLK	INCLK				
3 - P1.1	2 - P1.1	TA0	CCI0A	CCR0	TA0	3 - P1.1	2 - P1.1
7 - P1.5	6 - P1.5	ACLK (internal)	CCI0B			7 - P1.5	6 - P1.5
		V <sub>SS</sub>	GND				
		V <sub>CC</sub>	V <sub>CC</sub>				
4 - P1.2	3 - P1.2	TA1	CCI1A	CCR1	TA1	4 - P1.2	3 - P1.2
8 - P1.6	7 - P1.6	TA1	CCI1B			8 - P1.6	7 - P1.6
		V <sub>SS</sub>	GND			13 - P2.6	12 - P2.6
		V <sub>CC</sub>	V <sub>CC</sub>				

Table 10. Timer\_A2 Signal Connections



### USI

The universal serial interface (USI) module is used for serial data communication and provides the basic hardware for synchronous communication protocols like SPI and I2C.

### SD16\_A

The SD16\_A module supports 16-bit analog-to-digital conversions. The module implements a 16-bit sigma-delta core and reference generator. In addition to external analog inputs, internal  $V_{CC}$  sense and temperature sensors are also available.

#### **Peripheral File Map**

#### **Table 11. Peripherals With Word Access**

SD16_A	General Control	SD16CTL	0100h
	Channel 0 Control	SD16CCTL0	0102h
	Interrupt vector word register	SD16IV	0110h
	Channel 0 conversion memory	SD16MEM0	0112h
Timer_A	Capture/compare register	TACCR1	0174h
	Capture/compare register	TACCR0	0172h
	Timer_A register	TAR	0170h
	Capture/compare control	TACCTL1	0164h
	Capture/compare control	TACCTL0	0162h
	Timer_A control	TACTL	0160h
	Timer_A interrupt vector	TAIV	012Eh
Flash Memory	Flash control 3	FCTL3	012Ch
	Flash control 2	FCTL2	012Ah
	Flash control 1	FCTL1	0128h
Watchdog Timer+	Watchdog/timer control	WDTCTL	0120h

#### **Table 12. Peripherals With Byte Access**

SD16_A	Channel 0 Input Control	SD16INCTL0	0B0h
	Analog Enable	SD16AE	0B7h
USI	USI control 0	USICTL0	078h
	USI control 1	USICTL1	079h
	USI clock control	USICKCTL	07Ah
	USI bit counter	USICNT	07Bh
	USI shift register	USISR	07Ch
Basic Clock System+	Basic clock system control 3	BCSCTL3	053h
	Basic clock system control 2	BCSCTL2	058h
	Basic clock system control 1	BCSCTL1	057h
	DCO clock frequency control	DCOCTL	056h
Port P2	Port P2 resistor enable Port P2 selection Port P2 interrupt enable Port P2 interrupt edge select Port P2 interrupt flag Port P2 direction Port P2 output Port P2 input	P2REN P2SEL P2IE P2IES P2IFG P2DIR P2OUT P2IN	02Fh 02Eh 02Dh 02Ch 02Bh 02Ah 02Ah 029h 028h
Port P1	Port P1 resistor enable	P1REN	027h
	Port P1 selection	P1SEL	026h
	Port P1 interrupt enable	P1IE	025h
	Port P1 interrupt edge select	P1IES	024h
	Port P1 interrupt flag	P1IFG	023h
	Port P1 direction	P1DIR	022h
	Port P1 output	P1OUT	021h
	Port P1 input	P1IN	020h
Special Function	SFR interrupt flag 2	IFG2	003h
	SFR interrupt flag 1	IFG1	002h
	SFR interrupt enable 2	IE2	001h
	SFR interrupt enable 1	IE1	000h



#### SLAS774A-JULY 2011-REVISED OCTOBER 2011

#### Absolute Maximum Ratings<sup>(1)</sup>

	Voltage applied at $V_{CC}$ to $V_{SS}$	-0.3 V to 4.1 V	
	Voltage applied to any pin <sup>(2)</sup>	-0.3 V to V <sub>CC</sub> + 0.3 V	
	Diode current at any device terminal	±2 mA	
-	Otomo no tomo noturo (3)	Unprogrammed device	-55°C to 150°C
I stg	Storage temperature <sup>(3)</sup>	Programmed device	-40°C to 150°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages referenced to V<sub>SS</sub>. The JTAG fuse-blow voltage, V<sub>FB</sub>, is allowed to exceed the absolute maximum rating. The voltage is applied to the TEST pin when blowing the JTAG fuse.

(3) Higher temperature may be applied during board soldering according to the current JEDEC J-STD-020 specification with peak reflow temperatures not higher than classified on the device label on the shipping boxes or reels.

### THERMAL INFORMATION

		MSP430F2013-EP	
	THERMAL METRIC <sup>(1)</sup>	RSA	UNITS
		16 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance <sup>(2)</sup>	38.1	
θ <sub>JCtop</sub>	Junction-to-case (top) thermal resistance <sup>(3)</sup>	26	
$\theta_{JB}$	Junction-to-board thermal resistance <sup>(4)</sup>	7.5	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter <sup>(5)</sup>	0.3	C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter <sup>(6)</sup>	5.7	
θ <sub>JCbot</sub>	Junction-to-case (bottom) thermal resistance <sup>(7)</sup>	1.9	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, SPRA953.

(2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as specified in JESD51-7, in an environment described in JESD51-2a.

(3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter,  $\psi_{JT}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter,  $\psi_{JB}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).

(7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

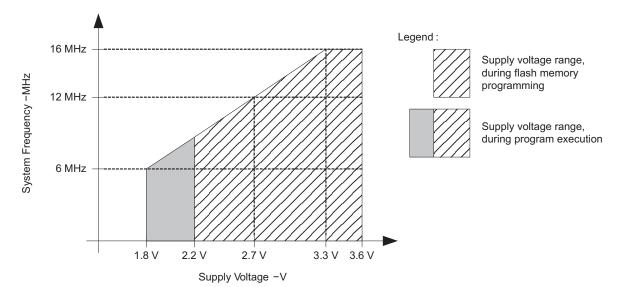
### **Recommended Operating Conditions**

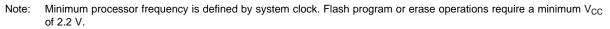
			MIN	NOM	MAX	UNIT
V	Supply voltage	During program execution	1.8		3.6	V
V <sub>CC</sub>		During flash program/erase	2.2		3.6	v
V <sub>SS</sub>	S Supply voltage			0		V
T <sub>A</sub>	Operating free-air temperature		-40		125	°C
		$V_{CC} = 1.8 V,$ Duty cycle = 50% ± 10%	dc		6	
<b>f</b> SYSTEM	Processor frequency (maximum MCLK frequency) <sup>(1)(2)</sup>	$V_{CC} = 2.7 V,$ Duty cycle = 50% ± 10%	dc		12	MHz
		$V_{CC} \ge 3.3 \text{ V},$ Duty cycle = 50% ± 10%	dc		16	

(1) The MSP430 CPU is clocked directly with MCLK. Both the high and low phase of MCLK must not exceed the pulse width of the specified maximum frequency.

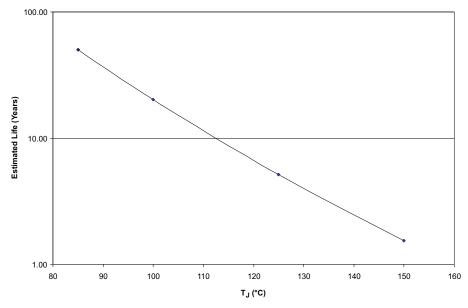
(2) Modules might have a different maximum input clock specification. See the specification of the respective module in this data sheet.











- (1) See data sheet for absolute maximum and minimum recommended operating conditions.
- (2) Silicon operating life design goal is 10 years at 110°C junction temperture (does not include package interconnect life).
- (3) The predicted operating lifetime vs. junction temperature is based on reliability modeling using electromigration as the dominant failure mechanism affecting device wearout for the specific device process and design characteristics.

Figure 2. Operating Life Derating Chart

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### **Electrical Characteristics**

### Active Mode Supply Current Into V<sub>cc</sub> Excluding External Current

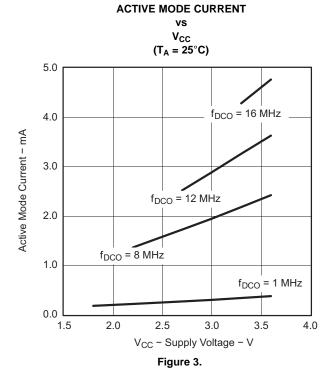
over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)<sup>(1)(2)</sup>

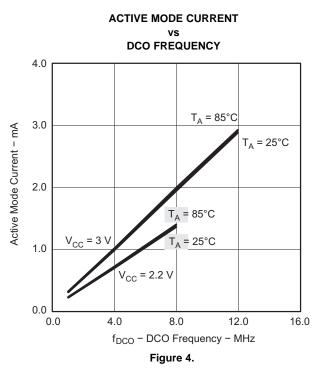
	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	Vcc	MIN	TYP	MAX	UNIT
I <sub>AM,1MHz</sub>	Active mode (AM) current (1 MHz)	$\label{eq:f_DCO} \begin{split} &f_{DCO} = f_{MCLK} = f_{SMCLK} = 1 \text{ MHz}, \\ &f_{ACLK} = 32768 \text{ Hz}, \\ &\text{Program executes in flash}, \\ &\text{BCSCTL1} = CALBC1_1MHZ, \\ &\text{DCOCTL} = CALDCO_1MHZ, \\ &\text{CPUOFF} = 0, \text{ SCG0} = 0, \\ &\text{SCG1} = 0, \text{ OSCOFF} = 0 \end{split}$		2.2 V 3 V		220 310	280 380	μA
		$f_{DCO} = f_{MCLK} = f_{SMCLK} = 1 \text{ MHz},$		2.2 V		190		
I <sub>AM,1MHz</sub>	Active mode (AM) current (1 MHz)	$\label{eq:faclk} \begin{array}{l} f_{ACLK} = 32768 \mbox{ Hz}, \\ \mbox{Program executes in RAM}, \\ \mbox{BCSCTL1} = CALBC1_1MHZ, \\ \mbox{DCOCTL} = CALDCO_1MHZ, \\ \mbox{CPUOFF} = 0, \mbox{SCG0} = 0, \\ \mbox{SCG1} = 0, \mbox{OSCOFF} = 0 \end{array}$		3 V		265		μA
		$f_{MCLK} = f_{SMCLK} = f_{ACLK} = 32768 \text{ Hz/8}$	-40°C to 85°C	2.2 V		1.2	3	
		= 4096 Hz, f <sub>DCO</sub> = 0 Hz,	125°C	2.2 V			6	
1	Active mode (AM)	Program executes in flash,	-40°C to 85°C	3 V		1.6	4	μA
IAM,4kHz	current (4 kHz)	SELMx = 11, SELS = 1, DIVMx = DIVSx = DIVAx = 11, CPUOFF = 0, SCG0 = 1, SCG1 = 0, OSCOFF = 0	125°C	3 V			7	μ
		$f_{MCLK} = f_{SMCLK} = f_{DCO(0, 0)} \approx 100 \text{ kHz},$	-40°C to 85°C	2.2 V		37	50	
	Active mode (AM)	f <sub>ACLK</sub> = 0 Hz, Program executes in flash,	125°C	2.2 V			65	
I <sub>AM,100kHz</sub>	current (100 kHz)	RSELx = 0, DCOx = 0,	-40°C to 85°C	3 V		40	55	μA
	·····(·····//	CPUOFF = 0, SCG0 = 0, SCG1 = 0, OSCOFF = 1	125°C	3 V			70	·

All inputs are tied to 0 V or to V<sub>CC</sub>. Outputs do not source or sink any current. (1) (2)

External crystal not used. The currents are characterized with a clock derived from alternate external clock source.

Typical Characteristics - Active Mode Supply Current (Into V<sub>cc</sub>)





# Low-Power Mode Supply Currents (Into V<sub>cc</sub>) Excluding External Current

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)<sup>(1) (2)</sup>

P/	ARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN TYP	MAX	UNIT
		f <sub>MCLK</sub> = 0 MHz,		2.2 V	65	86	
I <sub>LPM0,1MHz</sub>	Low-power mode 0 (LPM0) current <sup>(3)</sup>	$ \begin{array}{l} f_{SMCLK} = f_{DCO} = 1 \mbox{ MHz}, \\ f_{ACLK} = 32,768 \mbox{ Hz}, \\ BCSCTL1 = CALBC1_1MHZ, \\ DCOCTL = CALDCO_1MHZ, \\ CPUOFF = 1, \mbox{ SCG0} = 0, \\ \mbox{ SCG1} = 0, \mbox{ OSCOFF} = 0 \end{array} $		3 V	85	108	μA
		f <sub>MCLK</sub> = 0 MHz,		2.2 V	37	52	
I <sub>LPM0,100kHz</sub>	Low-power mode 0 (LPM0) current <sup>(3)</sup>			3 V	41	56	μA
		$f_{MCLK} = f_{SMCLK} = 0 MHz,$	-40°C to 85°C	2.2 V	22	29	
		f <sub>DCO</sub> = 1 MHz, f <sub>ACLK</sub> = 32,768 Hz,	125°C	2.2 V		34	
I <sub>LPM2</sub> Low-power mode (LPM2) current <sup>(4</sup>	Low-power mode 2 (LPM2) current <sup>(4)</sup>	$BCSCTL1 = CALBC1_1MHZ$ ,	-40°C to 85°C		25	32	μA
	(LI MZ) current	$DCOCTL = CALDCO_1MHZ,$ CPUOFF = 1, SCG0 = 0, SCG1 = 1, OSCOFF = 0	125°C	3 V		37	
			-40°C	2.2 V	0.7	1.2	μΑ
	Low-power mode 3 (LPM3) current <sup>(3)</sup>		25°C		0.7	1	
			85°C		1.4	2.3	
			125°C		3	6.5	
LPM3,LFXT1			-40°C	3 V	0.9	1.2	
			25°C		0.9	1.2	
LPM3,LFXT1			85°C		1.6	2.8	
			125°C		3	7.6	
			-40°C		0.4	0.7	
			25°C	2.2 V	0.5	0.7	
		$f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 \text{ MHz},$	85°C	2.2 V	1	1.6	
ILPM3.VLO	Low-power mode 3	f <sub>ACLK</sub> from internal LF oscillator (VLO),	125°C		2	5.5	μA
IPM3,VLO	(LPM3) current <sup>(4)</sup>	CPUOFF = 1, SCG0 = 1,	-40°C		0.5	0.9	μд
		SCG1 = 1, OSCOFF = 0	25°C	3 V	0.6	0.9	
			85°C		1.3	1.8	-
			125°C		2.5	6.5	
		$f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 MHz,$	-40°C		0.1	0.5	
I <sub>LPM4</sub>	Low-power mode 4	$f_{ACLK} = 0 Hz,$	25°C	2.2 V/3 V	0.1	0.5	- 11A
'LPM4	(LPM4) current <sup>(5)</sup>	CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 1	85°C		0.8	1.5	
			125°C		2	4.4	

All inputs are tied to 0 V or to V<sub>CC</sub>. Outputs do not source or sink any current. External crystal not used. The currents are characterized with a clock derived from alternate external clock source. Current for brownout and WDT clocked by SMCLK included. Current for brownout and WDT clocked by ACLK included.

(1)
(2)
(3)
(4)
(5)

Current for brownout included.





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### Schmitt-Trigger Inputs (Ports P1 and P2)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
				$0.45 V_{CC}$		0.75 V <sub>CC</sub>	
V <sub>IT+</sub>	Positive-going input threshold voltage		2.2 V	1.00		1.65	V
VIT+Positive-going input threshold voltage $0.45 V_{CC}$ VIT+Positive-going input threshold voltage $2.2 V$ $1.00$ $3 V$ $1.35$ $0.25 V_{CC}$ VIT-Negative-going input threshold voltage $2.2 V$ $0.25 V_{CC}$ $V_{IT-}$ Negative-going input threshold voltage $2.2 V$ $0.25 V_{CC}$ $V_{IT-}$ $2.2 V$ $0.25 V_{CC}$ $3 V$ $0.75$ $V_{hys}$ Input voltage hysteresis ( $V_{IT+} - V_{IT-}$ ) $2.2 V$ $0.2$ $V_{hys}$ Nullup/oulldown residentFor pullup: $V_{IN} = V_{SS}$ $20$		2.25					
V <sub>IT-</sub>	Negative-going input threshold voltage			0.25 V <sub>CC</sub>		0.55 V <sub>CC</sub>	
			2.2 V	0.55		1.20	V
			$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$				
			2.2 V	0.2		1.0	V
V <sub>hys</sub>	input voltage hysteresis (v <sub>IT+</sub> - v <sub>IT-</sub> )		3 V	0.3		1.0	V
R <sub>Pull</sub>	Pullup/pulldown resistor	For pullup: $V_{IN} = V_{SS}$ , For pulldown: $V_{IN} = V_{CC}$		20	35	50	kΩ
CI	Input capacitance	$V_{IN} = V_{SS}$ or $V_{CC}$			5		pF

### Inputs (Ports P1 and P2)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	Т	EST CONDITIONS	V <sub>cc</sub>	MIN	TYP I	МАХ	UNIT
t <sub>(int)</sub> External interrupt	timing Port P1, P2: P1.x width to set intern	to P2.x, External trigger pulse 2 upt flag <sup>(1)</sup>	2.2 V/3 V	25			ns

An external signal sets the interrupt flag every time the minimum interrupt pulse width t<sub>(int)</sub> is met. It may be set even with trigger signals shorter than t<sub>(int)</sub>.

### Leakage Current (Ports P1 and P2)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN MAX	UNIT
I <sub>lkg(Px.y)</sub>	High-impedance leakage current	See $^{(1)}$ and $^{(2)}$	2.2 V/3 V	±50	nA

(1) The leakage current is measured with V<sub>SS</sub> or V<sub>CC</sub> applied to the corresponding pin(s), unless otherwise noted.

(2) The leakage of the digital port pins is measured individually. The port pin is selected for input and the pullup/pulldown resistor is disabled.

#### **Outputs (Ports P1 and P2)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		$I_{(OHmax)} = -1.5 \text{ mA}^{(1)}$	2.2 V	V <sub>CC</sub> - 0.25		V <sub>CC</sub>	
V	Lich lovel output voltoge	$I_{(OHmax)} = -6 \text{ mA}^{(2)}$	2.2 V	V <sub>CC</sub> - 0.6		V <sub>CC</sub>	V
V <sub>OH</sub>	High-level output voltage	$I_{(OHmax)} = -1.5 \text{ mA}^{(1)}$	3 V	V <sub>CC</sub> - 0.25		V <sub>CC</sub>	v
		$I_{(OHmax)} = -6 \text{ mA}^{(2)}$	3 V	V <sub>CC</sub> - 0.6		V <sub>CC</sub>	
		$I_{(OLmax)} = 1.5 \text{ mA}^{(1)}$	2.2 V	V <sub>SS</sub>	V	<sub>SS</sub> + 0.25	
v		$I_{(OLmax)} = 6 \text{ mA}^{(2)}$	2.2 V	V <sub>SS</sub>	١	/ <sub>SS</sub> + 0.6	V
V <sub>OL</sub>	Low-level output voltage	$I_{(OLmax)} = 1.5 \text{ mA}^{(1)}$	3 V	V <sub>SS</sub>	Vs	<sub>SS</sub> + 0.25	v
		$I_{(OLmax)} = 6 \text{ mA}^{(2)}$	3 V	V <sub>SS</sub>	١	/ <sub>SS</sub> + 0.6	

The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined should not exceed ±12 mA to hold the maximum voltage drop (1) specified.

The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined should not exceed ±48 mA to hold the maximum voltage drop (2)specified.

### **Output Frequency (Ports P1 and P2)**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

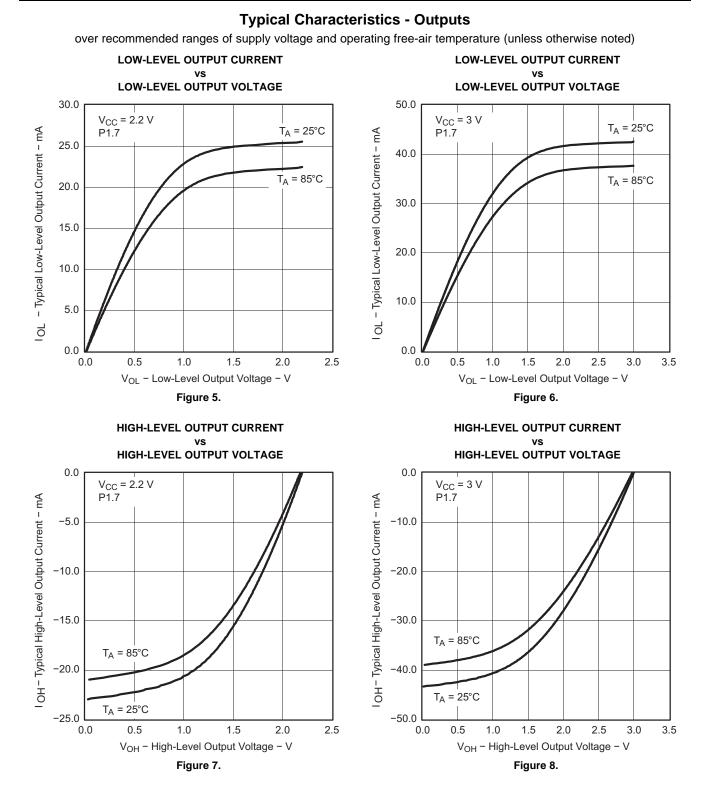
	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
4	Port output frequency	P1.4/SMCLK, $C_L = 20 \text{ pF}$ , $R_L = 1 \text{ k}\Omega^{(1)}$ (2)	2.2 V			10	MHz
T <sub>Px.y</sub>	(with load)	P1.4/SMCLK, $C_L = 20 \text{ pr}$ , $R_L = 1 \text{ k}\Omega^{-1}$	3 V			12	IVIEZ
1	Clock output froquency	P2.0/ACLK, P1.4/SMCLK, C <sub>1</sub> = 20 pF <sup>(2)</sup>	2.2 V			12	
<sup>T</sup> Port°CLK	Clock output frequency	$P2.0/ACLR, P1.4/SWICLR, C_L = 20 \text{ pr}^{-7}$	3 V			16	MHz

(1) A resistive divider with  $2 \times 0.5 \text{ k}\Omega$  between V<sub>CC</sub> and V<sub>SS</sub> is used as load. The output is connected to the center tap of the divider. (2) The output voltage reaches at least 10% and 90% V<sub>CC</sub> at the specified toggle frequency.

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### POR/Brownout Reset (BOR)<sup>(1)</sup>

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>CC(start)</sub>	See Figure 9	dV <sub>CC</sub> /dt ≤ 3 V/s			0.7 × V <sub>(B_IT-)</sub>		V
V <sub>(B_IT-)</sub>	See Figure 9 through Figure 11	$dV_{CC}/dt \le 3 V/s$				1.71	V
V <sub>hys(B_IT-)</sub>	See Figure 9	dV <sub>CC</sub> /dt ≤ 3 V/s		70	155	210	mV
t <sub>d(BOR)</sub>	See Figure 9 <sup>(2)</sup>					2000	μs
t <sub>(reset)</sub>	Pulse length needed at $\overline{\text{RST}}$ /NMI pin to accepted reset internally <sup>(2)</sup>		2.2 V/3 V	2			μs

The current consumption of the brownout module is already included in the I<sub>CC</sub> current consumption data. The voltage level V<sub>(B\_IT-)</sub> + (1)  $V_{hys(B_{IT})}$ is  $\leq 1.8$  V. Minimum and maximum parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

(2)

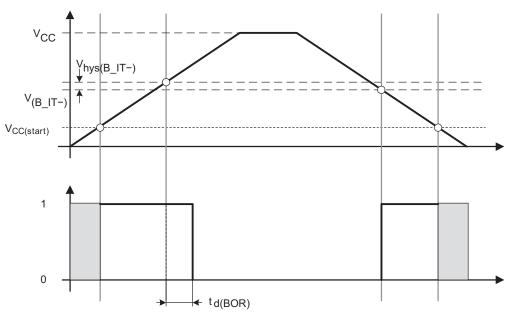


Figure 9. POR/Brownout Reset (BOR) vs Supply Voltage

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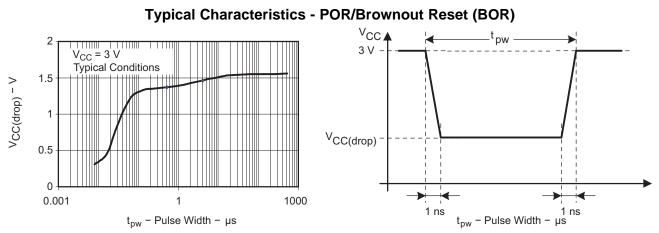


Figure 10. V<sub>CC(drop)</sub> Level With a Square Voltage Drop to Generate a POR/Brownout Signal

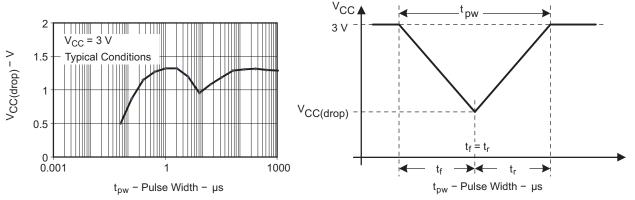


Figure 11. V<sub>CC(drop)</sub> Level With a Triangle Voltage Drop to Generate a POR/Brownout Signal



# **Main DCO Characteristics**

- All ranges selected by RSELx overlap with RSELx + 1: RSELx = 0 overlaps RSELx = 1, ... RSELx = 14 overlaps RSELx = 15.
- DCO control bits DCOx have a step size as defined by parameter  $S_{\text{DCO}}$ .
- Modulation control bits MODx select how often  $f_{DCO(RSEL,DCO+1)}$  is used within the period of 32 DCOCLK cycles. The frequency  $f_{DCO(RSEL,DCO)}$  is used for the remaining cycles. The frequency is an average equal to:  $f_{DCO(RSEL,DCO)} = \frac{32 \times f_{DCO(RSEL,DCO+1)}}{5}$

 $f_{average} = \frac{DOO(ROEL, DOO)}{MOD \times f_{DCO(RSEL, DCO)} + (32 - MOD) \times f_{DCO(RSEL, DCO+1)}}$ 

### **DCO Frequency**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		RSELx < 14		1.8		3.6	
V <sub>CC</sub>	Supply voltage	RSELx = 14		2.2		3.6	V
		RSELx = 15		3.0		3.6	
f <sub>DCO(0,0)</sub>	DCO frequency (0, 0)	RSELx = 0, $DCOx = 0$ , $MODx = 0$	2.2 V/3 V	0.06		0.14	MHz
f <sub>DCO(0,3)</sub>	DCO frequency (0, 3)	RSELx = 0, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.07		0.17	MHz
f <sub>DCO(1,3)</sub>	DCO frequency (1, 3)	RSELx = 1, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.10		0.20	MHz
f <sub>DCO(2,3)</sub>	DCO frequency (2, 3)	RSELx = 2, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.14		0.28	MHz
f <sub>DCO(3,3)</sub>	DCO frequency (3, 3)	RSELx = 3, DCOx = 3, MODx = 0	2.2 V/3 V	0.20		0.40	MHz
f <sub>DCO(4,3)</sub>	DCO frequency (4, 3)	RSELx = 4, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.28		0.54	MHz
f <sub>DCO(5,3)</sub>	DCO frequency (5, 3)	RSELx = 5, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.39		0.77	MHz
f <sub>DCO(6,3)</sub>	DCO frequency (6, 3)	RSELx = 6, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.54		1.06	MHz
f <sub>DCO(7,3)</sub>	DCO frequency (7, 3)	RSELx = 7, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	0.80		1.50	MHz
f <sub>DCO(8,3)</sub>	DCO frequency (8, 3)	RSELx = 8, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	1.10		2.10	MHz
f <sub>DCO(9,3)</sub>	DCO frequency (9, 3)	RSELx = 9, $DCOx = 3$ , $MODx = 0$	2.2 V/3 V	1.60		3.00	MHz
f <sub>DCO(10,3)</sub>	DCO frequency (10, 3)	RSELx = 10, DCOx = 3, MODx = 0	2.2 V/3 V	2.50		4.30	MHz
f <sub>DCO(11,3)</sub>	DCO frequency (11, 3)	RSELx = 11, DCOx = 3, MODx = 0	2.2 V/3 V	3.00		5.50	MHz
f <sub>DCO(12,3)</sub>	DCO frequency (12, 3)	RSELx = 12, DCOx = 3, MODx = 0	2.2 V/3 V	4.30		7.30	MHz
f <sub>DCO(13,3)</sub>	DCO frequency (13, 3)	RSELx = 13, DCOx = 3, MODx = 0	2.2 V/3 V	6.00		9.60	MHz
f <sub>DCO(14,3)</sub>	DCO frequency (14, 3)	RSELx = 14, DCOx = 3, MODx = 0	2.2 V/3 V	8.60		13.9	MHz
f <sub>DCO(15,3)</sub>	DCO frequency (15, 3)	RSELx = 15, DCOx = 3, MODx = 0	3 V	12.0		18.5	MHz
f <sub>DCO(15,7)</sub>	DCO frequency (15, 7)	RSELx = 15, DCOx = 7, MODx = 0	3 V	16.0		26.0	MHz
S <sub>RSEL</sub>	Frequency step between range RSEL and RSEL+1	$S_{RSEL} = f_{DCO(RSEL+1,DCO)}/f_{DCO(RSEL,DCO)}$	2.2 V/3 V			1.55	ratio
S <sub>DCO</sub>	Frequency step between tap DCO and DCO+1	$S_{DCO} = f_{DCO(RSEL, DCO+1)}/f_{DCO(RSEL, DCO)}$	2.2 V/3 V	1.03	1.08	1.14	
	Duty cycle	Measured at P1.4/SMCLK	2.2 V/3 V	40	50	60	%

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#### **Calibrated DCO Frequencies - Tolerance at Calibration**

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	Vcc	MIN	TYP	MAX	UNIT
	Frequency tolerance at calibration		25°C	3 V	-1	±0.2	+1	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	25°C	3 V	0.990	1	1.010	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	25°C	3 V	7.920	8	8.080	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	25°C	3 V	11.88	12	12.12	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	25°C	3 V	15.84	16	16.16	MHz

# Calibrated DCO Frequencies - Tolerance Over Temperature -40°C to 125°C

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	Vcc	MIN	TYP	MAX	UNIT
	1-MHz tolerance over temperature		-40°C to 125°C	3 V	-2.5	±1.25	+2.5	%
	8-MHz tolerance over temperature		-40°C to 125°C	3 V	-5	±1.25	+5	%
	12-MHz tolerance over temperature		-40°C to 125°C	3 V	-5	±1.25	+2.5	%
	16-MHz tolerance over temperature		-40°C to 125°C	3 V	-6.25	±2.0	+3	%
		BCSCTL1 = CALBC1_1MHZ,		2.2 V	0.97	1	1.03	
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	-40°C to 125°C	3 V	0.975	1	1.025	MHz
				3.6 V	0.97	1	1.03	ĺ
		BCSCTL1 = CALBC1 8MHZ,		2.2 V	7.6	8	8.4	
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	$DCOCTL = CALDCO_8MHZ,$	-40°C to 125°C	3 V	7.6	8	8.4	MHz
		Gating time: 5 ms		3.6 V	7.6	8	8.4	ĺ
		BCSCTL1 = CALBC1_12MHZ,		2.2 V	11.6	12	12.3	
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	DCOCTL = CALDCO_12MHZ,	-40°C to 125°C	3 V	11.6	12	12.3	MHz
		Gating time: 5 ms		3.6 V	11.6	12	12.3	ĺ
		BCSCTL1 = CALBC1_16MHZ,		3 V	15	16	16.48	
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	-40°C to 125°C	3.6 V	15	16	16.48	MHz

# Calibrated DCO Frequencies - Tolerance Over Supply Voltage $V_{cc}$

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	1-MHz tolerance over $V_{CC}$		25°C	1.8 V to 3.6 V	-3	±2	+3	%
	8-MHz tolerance over $V_{CC}$		25°C	1.8 V to 3.6 V	-3	±2	+3	%
	12-MHz tolerance over $V_{CC}$		25°C	2.2 V to 3.6 V	-4	±2	+3	%
	16-MHz tolerance over $V_{CC}$		25°C	3 V to 3.6 V	-6.25	±2	+3	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	25°C	1.8 V to 3.6 V	0.97	1	1.03	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	25°C	1.8 V to 3.6 V	7.76	8	8.24	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	25°C	2.2 V to 3.6 V	11.64	12	12.36	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	25°C	3 V to 3.6 V	15	16	16.48	MHz

# **Calibrated DCO Frequencies - Overall Tolerance**

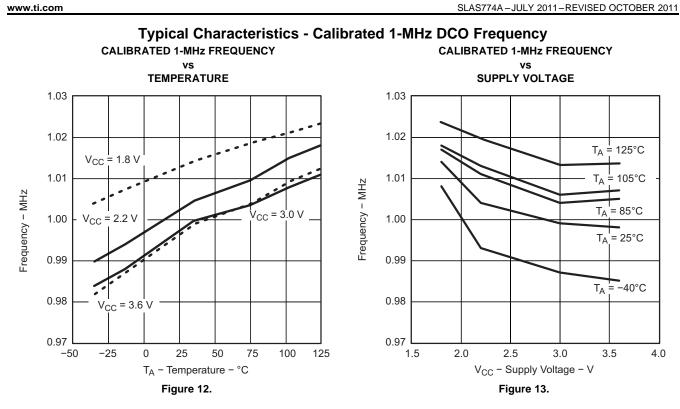
over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

P/	ARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	1-MHz tolerance overall		1.8 V to 3.6 V	-5	±2.5	+5	%
	8-MHz tolerance overall		1.8 V to 3.6 V	-5	±2.5	+5	%
	12-MHz tolerance overall		2.2 V to 3.6 V	-5	±2.5	+5	%
	16-MHz tolerance overall		3 V to 3.6 V	-6.25	±3	+6.25	%
f <sub>CAL(1MHz)</sub>	1-MHz calibration value	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, Gating time: 5 ms	1.8 V to 3.6 V	0.95	1	1.05	MHz
f <sub>CAL(8MHz)</sub>	8-MHz calibration value	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, Gating time: 5 ms	1.8 V to 3.6 V	7.6	8	8.4	MHz
f <sub>CAL(12MHz)</sub>	12-MHz calibration value	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, Gating time: 5 ms	2.2 V to 3.6 V	11.4	12	12.6	MHz
f <sub>CAL(16MHz)</sub>	16-MHz calibration value	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, Gating time: 2 ms	3 V to 3.6 V	15	16	17	MHz

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#### ÈXAS NSTRUMENTS

#### SLAS774A-JULY 2011-REVISED OCTOBER 2011



## Wake-Up From Lower-Power Modes (LPM3/4)<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

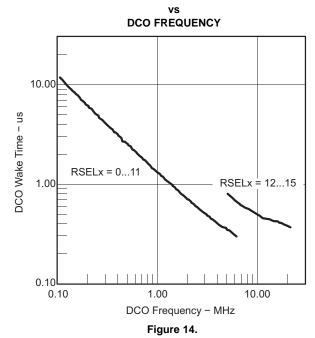
	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN TYP MA	X UNIT
		BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ			2
	DCO clock wake-up time	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ	2.2 V/3 V	1	.5
<sup>t</sup> DCO,LPM3/4	from LPM3/4 <sup>(2)</sup>	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ			μs 1
		BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ	3 V		1
t <sub>CPU,LPM3/4</sub>	CPU wake-up time from LPM3/4 <sup>(3)</sup>			1 / f <sub>MCLK</sub> + <sup>t</sup> Clock,LPM3/4	

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

(2) The DCO clock wake-up time is measured from the edge of an external wake-up signal (for example, port interrupt) to the first clock edge observable externally on a clock pin (MCLK or SMCLK).

(3) Parameter applicable only if DCOCLK is used for MCLK.

#### Typical Characteristics - DCO Clock Wake-Up Time From LPM3/4 DCO WAKE-UP TIME FROM LPM3





#### SLAS774A-JULY 2011-REVISED OCTOBER 2011

# Crystal Oscillator, XT1, Low-Frequency Mode<sup>(1)(2)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
LFXT1 oscillator crystal frequency, LF mode 0, 1	XTS = 0, LFXT1Sx = 0 or 1	1.8 V to 3.6 V		32768		Hz
LFXT1 oscillator logic level square wave input frequency, LF mode	XTS = 0, LFXT1Sx = 3	1.8 V to 3.6 V	10000	32768	50000	Hz
Oscillation allowance for	$\begin{split} XTS &= 0, \ LFXT1Sx = 0, \\ f_{LFXT1,LF} &= 32768 \ Hz, \ C_{L,eff} = 6 \ pF \end{split}$			500		kΩ
LF crystals	XTS = 0, LFXT1Sx = 0, f <sub>LFXT1,LF</sub> = 32768 Hz, C <sub>L,eff</sub> = 12 pF			200		K12
	XTS = 0, XCAPx = 0			1		
Integrated effective load	XTS = 0, XCAPx = 1			5.5		~ <b>Г</b>
capacitance, LF mode <sup>(3)</sup>	XTS = 0, XCAPx = 2			8.5		pF
	XTS = 0, XCAPx = 3			11		
Duty cycle, LF mode	$XTS = 0$ , Measured at P1.0/ACLK, $f_{LFXT1,LF} = 32768 \text{ Hz}$	2.2 V/3 V	30	50	70	%
Oscillator fault frequency, LF mode <sup>(4)</sup>	XTS = 0, LFXT1Sx = 3 <sup>(5)</sup>	2.2 V/3 V	10		10000	Hz
	LFXT1 oscillator crystal frequency, LF mode 0, 1 LFXT1 oscillator logic level square wave input frequency, LF mode Oscillation allowance for LF crystals Integrated effective load capacitance, LF mode <sup>(3)</sup> Duty cycle, LF mode Oscillator fault frequency,	LFXT1 oscillator crystal frequency, LF mode 0, 1XTS = 0, LFXT1Sx = 0 or 1LFXT1 oscillator logic level square wave input frequency, LF modeXTS = 0, LFXT1Sx = 3Oscillation allowance for LF crystalsXTS = 0, LFXT1Sx = 0, f_LFXT1,LF = 32768 Hz, CL,eff = 6 pFXTS = 0, LFXT1Sx = 0, f_LFXT1,LF = 32768 Hz, CL,eff = 12 pFIntegrated effective load capacitance, LF modeXTS = 0, XCAPx = 0Integrated effective load capacitance, LF modeXTS = 0, XCAPx = 1Duty cycle, LF modeXTS = 0, Measured at P1.0/ACLK, f_LFXT1,LF = 32768 HzOscillator fault frequency, VTS = 0, LFXT42x = 2(5)	LFXT1 oscillator crystal frequency, LF mode 0, 1XTS = 0, LFXT1Sx = 0 or 11.8 V to 3.6 VLFXT1 oscillator logic level square wave input frequency, LF modeXTS = 0, LFXT1Sx = 31.8 V to 3.6 VOscillation allowance for LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 6 pF1.8 V to 3.6 VMathematical distribution allowance for LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 12 pF1.8 V to 3.6 VIntegrated effective load capacitance, LF modeXTS = 0, XCAPx = 0XTS = 0, XCAPx = 1Mathematical distributionXTS = 0, XCAPx = 3XTS = 0, XCAPx = 3Duty cycle, LF modeXTS = 0, Measured at P1.0/ACLK, fLFXT1,LF = 32768 Hz2.2 V/3 VOscillator fault frequency, 	LFXT1 oscillator crystal frequency, LF mode 0, 1XTS = 0, LFXT1Sx = 0 or 11.8 V to 3.6 VLFXT1 oscillator logic level square wave input frequency, LF modeXTS = 0, LFXT1Sx = 31.8 V to 3.6 VOscillation allowance for LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 6 pF10000NTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 12 pFXTS = 0, XCAPx = 010000Integrated effective load capacitance, LF modeXTS = 0, XCAPx = 010000Ntrs = 0, XCAPx = 2 XTS = 0, XCAPx = 3XTS = 0, XCAPx = 310000Duty cycle, LF modeXTS = 0, Measured at P1.0/ACLK, fLFXT1,LF = 32768 Hz2.2 V/3 V30Oscillator fault frequency, VTS = 0, LFXT4Sx = 0, fLFXT1,LF = 32768 Hz2.2 V/3 V40	LFXT1 oscillator crystal frequency, LF mode 0, 1XTS = 0, LFXT1Sx = 0 or 11.8 V to 3.6 V32768LFXT1 oscillator logic level square wave input frequency, LF modeXTS = 0, LFXT1Sx = 31.8 V to 3.6 V1000032768Oscillation allowance for LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 6 pF1.8 V to 3.6 V1000032768Integrated effective load capacitance, LF modeXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 12 pF200NTS = 0, XCAPx = 0XTS = 0, XCAPx = 01Integrated effective load capacitance, LF modeXTS = 0, XCAPx = 15.5XTS = 0, XCAPx = 3XTS = 0, XCAPx = 311Duty cycle, LF modeXTS = 0, Measured at P1.0/ACLK, fLFXT1,LF = 32768 Hz2.2 V/3 V3050Oscillator fault frequency, VTS = 0, LFXT4Sw = 0/(5)XTS = 0, XCAPx = 0/(5)0.0 V/0 V/040	LFXT1 oscillator crystal frequency, LF mode 0, 1XTS = 0, LFXT1Sx = 0 or 11.8 V to 3.6 V32768LFXT1 oscillator logic level square wave input frequency, LF modeXTS = 0, LFXT1Sx = 31.8 V to 3.6 V100003276850000Oscillation allowance for LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 6 pF1.8 V to 3.6 V10000327685000Mathematical difference LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 6 pF500500Name Mathematical difference LF crystalsXTS = 0, LFXT1Sx = 0, fLFXT1,LF = 32768 Hz, CL,eff = 12 pF2001Integrated effective load capacitance, LF modeXTS = 0, XCAPx = 011XTS = 0, XCAPx = 1 XTS = 0, XCAPx = 3XTS = 0, XCAPx = 311Duty cycle, LF modeXTS = 0, Measured at P1.0/ACLK, fLFXT1,LF = 32768 Hz2.2 V/3 V305070Oscillator fault frequency, VTS = 0, LFXT4SE = 0(5)XTS = 0, V/3 V4040000

(1) To improve EMI on the XT1 oscillator, the following guidelines should be observed.

(a) Keep the trace between the device and the crystal as short as possible.

- (b) Design a good ground plane around the oscillator pins.
- (c) Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
- (d) Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
- (e) Use assembly materials and praxis to avoid any parasitic load on the oscillator XIN and XOUT pins.
- (f) If conformal coating is used, ensure that it does not induce capacitive/resistive leakage between the oscillator pins.
   (g) Do not route the XOUT line to the JTAG header to support the serial programming adapter as shown in other documentation. This signal is no longer required for the serial programming adapter.
- (2) Crystal oscillator cannot be operated beyond 105°C. Parameters are characterized up to T<sub>A</sub> = 105°C unless otherwise noted.
   (3) Includes parasitic bond and package capacitance (approximately 2 pF per pin).

Since the PCB adds additional capacitance, it is recommended to verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.

- (4) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- (5) Measured with logic-level input frequency but also applies to operation with crystals.

# Internal Very-Low-Power Low-Frequency Oscillator (VLO)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
4	V/I O frequency	-40°C to 85°C	2.2 V/3 V	4	12	20	kHz
t <sub>VLO</sub>	VLO frequency	125°C	2.2 V/3 V			22	KI IZ
df <sub>VLO</sub> /dT	VLO frequency temperature drift <sup>(1)</sup>	-40°C to 125°C	2.2 V/3 V		0.68		%/°C
$df_{VLO}/dV_{CC}$	VLO frequency supply voltage drift <sup>(2)</sup>	25°C	1.8 V to 3.6 V		4		%/V

(1) Calculated using the box method:

(MAX(-40 to 125°C) - MIN(-40 to 125°C)) / MIN(-40 to 125°C) / (125°C - (-40°C))

(2) Calculated using the box method: (MAX(1.8 to 3.6 V) - MIN(1.8 to 3.6 V)) / MIN(1.8 to 3.6 V) / (3.6 V - 1.8 V)

### Timer\_A

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		Internal: SMCLK, ACLK	2.2 V			10	
f <sub>TA</sub>	Timer_A clock frequency	External: TACLK, INCLK Duty cycle = $50\% \pm 10\%$	3 V			16	MHz
t <sub>TA,cap</sub>	Timer_A capture timing <sup>(1)</sup>	TA0, TA1	2.2 V/3 V	20			ns

(1) Parameter characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

XAS STRUMENTS

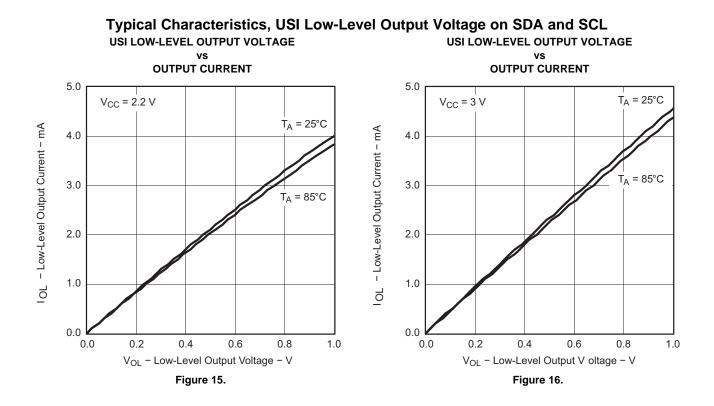
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# USI, Universal Serial Interface<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP MAX	UNIT
		External: SCLK,	2.2 V		10	
f <sub>USI</sub>	USI clock frequency	Duty cycle = 50% ±10%, SPI slave mode	3 V		16	MHz
V <sub>OL,I2C</sub>	Low-level output voltage on SDA and SCL	USI module in I2C mode, $I_{(OLmax)} = 1.5 \text{ mA}$	2.2 V/3 V	$V_{SS}$	V <sub>SS</sub> + 0.4	V

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.







SLAS774A-JULY 2011-REVISED OCTOBER 2011

# SD16\_A, Power Supply and Recommended Operating Conditions<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CO	NDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT		
AV <sub>CC</sub>	Analog supply voltage range					2.5		3.6	V		
			GAIN: 1,2	-40°C to 85°C				730	1050		
			GAIN. 1,2	105°C				1170			
		SD16LP = 0,		-40°C to 85°C			810	1150			
		f <sub>SD16</sub> = 1 MHz, SD16OSR = 256	GAIN: 4,8,16	105°C				1300			
	Analog supply current including internal		GAIN: 32	-40°C to 85°C	2.1/		1160	1700			
	reference		GAIN: 32	105°C	3 V			1850	μA		
				-40°C to 85°C			720	1030			
				SD16LP = 1,	GAIN: 1	105°C				1160	
		f <sub>SD16</sub> = 0.5 MHz, SD16OSR = 256	0.4101 00	-40°C to 85°C			810	1150			
			GAIN: 32	105°C				1300			
1	SD16 input clock	SD16LP = 0 (Low power mode c	lisabled)		3 V	0.03	1	1.1	MHz		
Ionia	frequency	SD16LP = 1 (Low power mode e	enabled)		3 V	0.03	0.5		IVIHZ		

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

# SD16\_A, Input Range<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CO	NDITIONS	Vcc	MIN	TYP	MAX	UNIT			
	Differential full scale input voltage	Bipolar mode, SD	16UNI = 0		-(V <sub>REF</sub> /2)/ GAIN		+(V <sub>REF</sub> /2)/ GAIN				
V <sub>ID,FSR</sub>	range <sup>(2)</sup>	Unipolar mode, SI	Unipolar mode, SD16UNI = 1		0		+(V <sub>REF</sub> /2)/ GAIN	mV			
			SD16GAINx = 1			±500					
			SD16GAINx = 2			±250					
V	Differential input voltage range for	SD16REFON = 1	SD16GAINx = 4			±125		mV			
	specified performance <sup>(2)</sup>	SDIGREFON = 1	SD16GAINx = 8			±62		mv			
						SD16GAINx = 16			±31		
			SD16GAINx = 32			±15					
7	Input impedance		SD16GAINx = 1	3 V		200		kΩ			
ZI	(one input pin to AV <sub>SS</sub> )	$f_{SD16} = 1 \text{ MHz}$	SD16GAINx = 32	3 V		75		K12			
7	Differential input impedance	£ 1 MIL-	SD16GAINx = 1	3 V	300	400		kΩ			
Z <sub>ID</sub>	(IN+ to IN-)	$f_{SD16} = 1 MHz$	SD16GAINx = 32	3 V	100	150		K12			
VI	Absolute input voltage range				AV <sub>SS</sub> - 0.1		$AV_{CC}$	V			
V <sub>IC</sub>	Common-mode input voltage range				AV <sub>SS</sub> - 0.1		$AV_{CC}$	V			

(1)

Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. The analog input range depends on the reference voltage applied to  $V_{REF}$ . If  $V_{REF}$  is sourced externally, the full-scale range is defined by  $V_{FSR+} = +(V_{REF}/2)/GAIN$  and  $V_{FSR-} = -(V_{REF}/2)/GAIN$ . The analog input range should not exceed 80% of  $V_{FSR+}$  or  $V_{FSR-}$ . (2)

SLAS774A – JULY 2011 – REVISED OCTOBER 2011

# SD16\_A, SINAD Performance ( $f_{SD16} = 1 \text{ MHz}$ , SD16OSRx = 1024, SD16REFON = 1)<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_{A} = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	UNIT
		SD16GAINx = 1, Signal amplitude: $V_{IN}$ = 500 mV, Signal frequency: $f_{IN}$ = 100 Hz		86	87	
		SD16GAINx = 2, Signal amplitude: $V_{IN}$ = 250 mV, Signal frequency: $f_{IN}$ = 100 Hz		82	83	
SINAD <sub>1024</sub> Signal-to-noise + distortion ratio (OSR = 1024)	SD16GAINx = 4, Signal amplitude: $V_{IN}$ = 125 mV, Signal frequency: $f_{IN}$ = 100 Hz		78	79		
		SD16GAINx = 8, Signal amplitude: $V_{IN}$ = 62 mV, Signal frequency: $f_{IN}$ = 100 Hz	3 V	73	74	dB
		SD16GAINx = 16, Signal amplitude: $V_{IN}$ = 31 mV, Signal frequency: $f_{IN}$ = 100 Hz		68	69	
		SD16GAINx = 32, Signal amplitude: $V_{IN}$ = 15 mV, Signal frequency: $f_{IN}$ = 100 Hz		62	63	

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

# SD16\_A, SINAD Performance ( $f_{SD16} = 1 \text{ MHz}$ , SD16OSRx = 256, SD16REFON = 1)<sup>(1)</sup>

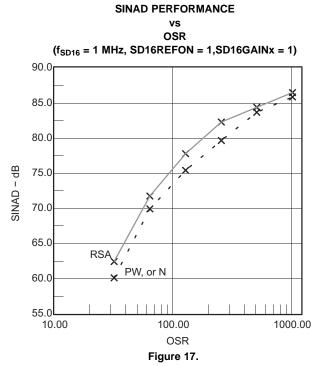
over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	UNIT
		$\begin{array}{l} SD16GAINx = 1,\\ Signal \ amplitude: \ V_{IN} = 500 \ mV,\\ Signal \ frequency: \ f_{IN} = 100 \ Hz \end{array}$		82	83	
		$ \begin{array}{l} \text{SD16GAINx} = 2,\\ \text{Signal amplitude: } V_{\text{IN}} = 250 \text{ mV},\\ \text{Signal frequency: } f_{\text{IN}} = 100 \text{ Hz} \end{array} $		76	77	
SINAD	Signal-to-noise + distortion ratio	$ \begin{array}{l} SD16GAINx=4,\\ Signal \ amplitude: \ V_{IN}=125 \ mV,\\ Signal \ frequency: \ f_{IN}=100 \ Hz \end{array} $	3∨	71	72	٩D
	(OSR = 256)	$ \begin{array}{l} SD16GAINx = 8,\\ Signal \ amplitude: \ V_{IN} = 62 \ mV,\\ Signal \ frequency: \ f_{IN} = 100 \ Hz \end{array} $	3 V	67	68	dB
		$\begin{array}{l} SD16GAINx=16,\\ Signal \ amplitude: \ V_{IN}=31 \ mV,\\ Signal \ frequency: \ f_{IN}=100 \ Hz \end{array}$		63	64	
		SD16GAINx = 32, Signal amplitude: $V_{IN}$ = 15 mV, Signal frequency: $f_{IN}$ = 100 Hz		57	58	

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.







# Typical Characteristics, SD16\_A SINAD Performance Over OSR

# SD16\_A, Performance ( $f_{SD16} = 1 \text{ MHz}$ , SD16OSRx = 256, SD16REFON = 1)<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		SD16GAINx = 1		0.97	1.00	1.02	
		SD16GAINx = 2		1.90	1.96	2.02	
<u> </u>	Nominal agin	SD16GAINx = 4	3 V	3.76	3.86	3.96	
G	Nominal gain	SD16GAINx = 8	3V	7.36	7.62	7.84	
		SD16GAINx = 16		14.56	15.04	15.52	
		SD16GAINx = 32		27.20	28.35	29.76	
ΔG/ΔT	Gain temperature drift	SD16GAINx = 1 <sup>(2)</sup>	3 V		15		ppm/°C
F	Offeet error	SD16GAINx = 1	3 V			±0.2	%FSR
E <sub>OS</sub>	Offset error	SD16GAINx = 32	3 V			±1.5	%F3R
	Offset error temperature	SD16GAINx = 1	3 V		±4	±20	
$\Delta E_{OS} / \Delta T$	coefficient	SD16GAINx = 32	3 V		±20	±100	ppm FSR/°C
01100	Common-mode rejection	SD16GAINx = 1, Common-mode input signal: V <sub>ID</sub> = 500 mV, f <sub>IN</sub> = 50 Hz, 100 Hz	0.14		>90		10
CMRR	ratio	$\begin{array}{l} \text{SD16GAINx} = 32,\\ \text{Common-mode input signal:}\\ \text{V}_{\text{ID}} = 16 \text{ mV},  f_{\text{IN}} = 50 \text{ Hz}, 100 \text{ Hz} \end{array}$	3 V		>75		dB
DC PSR	DC power supply rejection	SD16GAINx = 1, $V_{IN}$ = 500 mV, $V_{CC}$ = 2.5 V to 3.6 V <sup>(3)</sup>	2.5 V to 3.6 V		0.35		%/V
AC PSRR	AC power supply rejection ratio	SD16GAINx = 1, V <sub>CC</sub> = 3 V ± 100 mV, f <sub>IN</sub> = 50 Hz	3 V		>80		dB

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

(2) Calculated using the box method: (MAX(-40°C to 85°C) - MIN(-40°C to 85°C)) / MIN(-40°C to 85°C) / (85°C - (-40°C))

(3) Calculated using the ADC output code and the box method:

(MAX-code(2.5 V to 3.6 V) - MIN-code(2.5 V to 3.6 V)) / MIN-code(2.5 V to 3.6 V) / (3.6 V - 2.5 V)

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# SD16\_A, Built-In Voltage Reference<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_{A} = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
$V_{REF}$	Internal reference voltage	SD16REFON = 1, SD16VMIDON = 0		3 V	1.14	1.20	1.26	V
	Deference cumply current	SD16REFON = 1,	-40°C to 85°C	3 V		190	280	
I <sub>REF</sub>	Reference supply current	SD16VMIDON = 0	105°C	3 V			295	μA
тс	Temperature coefficient	SD16REFON = 1, SD16VMIDON = 0		3 V		18	50	ppm/°C
$C_{REF}$	V <sub>REF</sub> load capacitance	SD16REFON = 1, SD16VMIDON = $0^{(2)}$				100		nF
I <sub>LOAD</sub>	V <sub>REF(I)</sub> maximum load current	SD16REFON = 1, SD16VMIDON = 0		3 V			±200	nA
t <sub>ON</sub>	Turn-on time	$\begin{array}{l} \text{SD16REFON} = 0 \rightarrow 1, \\ \text{SD16VMIDON} = 0, \\ \text{C}_{\text{REF}} = 100 \text{ nF} \end{array}$		3 V		5		ms
DC PSR	DC power supply rejection $\Delta V_{REF} / \Delta V_{CC}$	$\begin{array}{l} \text{SD16REFON} = 1,\\ \text{SD16VMIDON} = 0,\\ \text{V}_{\text{CC}} = 2.5 \text{ V to } 3.6 \text{ V} \end{array}$		2.5 V to 3.6 V		100		μV/V

(1)

Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. There is no capacitance required on  $V_{REF}$ . However, a capacitance of at least 100 nF is recommended to reduce any reference voltage (2)noise.

# SD16\_A, Reference Output Buffer<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	Vcc	MIN	TYP	MAX	UNIT
V <sub>REF,BUF</sub>	Reference buffer output voltage	SD16REFON = 1, SD16VMIDON = 1		3 V		1.2		V
	Reference supply + reference	SD16REFON = 1,	-40°C to 85°C	2.1/		385	600	
I <sub>REF,BUF</sub>	output buffer quiescent current	SD16VMIDON = 1	105°C	3 V			660	μA
C <sub>REF(O)</sub>	Required load capacitance on V <sub>REF</sub>	SD16REFON = 1, SD16VMIDON = 1			470			nF
I <sub>LOAD,Max</sub>	Maximum load current on $V_{REF}$	SD16REFON = 1, SD16VMIDON = 1		3 V			±1	mA
	Maximum voltage variation vs load current	$ I_{LOAD}  = 0$ to 1 mA		3 V	-15		+15	mV
t <sub>ON</sub>	Turn on time	$\begin{array}{l} \text{SD16REFON} = 0 \rightarrow 1, \\ \text{SD16VMIDON} = 1, \\ \text{C}_{\text{REF}} = 470 \text{ nF} \end{array}$		3 V		100		μs

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.

# SD16\_A, External Reference Input<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

PARAMETER	TEST CONDITIONS	Vcc	MIN	TYP	MAX	UNIT
V <sub>REF(I)</sub> Input voltage range	SD16REFON = 0	3 V	1	1.25	1.5	V
I <sub>REF(I)</sub> Input current	SD16REFON = 0	3 V			50	nA

(1) Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted.



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### SD16\_A, Temperature Sensor<sup>(1)(2)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_{A} = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
TC <sub>Sensor</sub>	Sensor temperature coefficient			1.18	1.32	1.46	mV/°C
V <sub>Offset,Sensor</sub>	Sensor offset voltage			-100		100	mV
		Temperature sensor voltage at $T_A = 85^{\circ}C$		435	475	515	
V <sub>Sensor</sub>	Sensor output voltage <sup>(3)</sup>	Temperature sensor voltage at $T_A = 25^{\circ}C$	3 V	355	395	435	mV
		Temperature sensor voltage at $T_A = 0^{\circ}C$		320	360	400	

Values are not based on calculations using  $TC_{Sensor}$  or  $V_{Offset,sensor}$  but on measurements. Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. (1)

(2)

(3) The following formula can be used to calculate the temperature sensor output voltage:

V<sub>Sensor,typ</sub> = TC<sub>Sensor</sub> ( 273 + T [°C] ) + V<sub>Offset,sensor</sub> [mV] or  $V_{\text{Sensor,typ}} = TC_{\text{Sensor}} T [^{\circ}C] + V_{\text{Sensor}}(T_A = 0^{\circ}C) [mV]$ 

# Flash Memory<sup>(1)(2)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>CC(PGM/ERASE)</sub>	Program and erase supply voltage			2.2		3.6	V
f <sub>FTG</sub>	Flash timing generator frequency			257		476	kHz
I <sub>PGM</sub>	Supply current from V <sub>CC</sub> during program		2.2 V/3.6 V		1	5	mA
I <sub>ERASE</sub>	Supply current from $V_{CC}$ during erase		2.2 V/3.6 V		1	7	mA
t <sub>CPT</sub>	Cumulative program time <sup>(3)</sup>		2.2 V/3.6 V			10	ms
t <sub>CMErase</sub>	Cumulative mass erase time		2.2 V/3.6 V	20			ms
	Program/erase endurance	$-40^{\circ}C \le T_{J} \le 105^{\circ}C$		10 <sup>4</sup>	10 <sup>5</sup>		cycles
t <sub>Retention</sub>	Data retention duration	$T_J = 25^{\circ}C$		100			years
t <sub>Word</sub>	Word or byte program time	See (4)			30		t <sub>FTG</sub>
t <sub>Block, 0</sub>	Block program time for first byte or word	See (4)			25		t <sub>FTG</sub>
t <sub>Block, 1-63</sub>	Block program time for each additional byte or word	See <sup>(4)</sup>			18		t <sub>FTG</sub>
t <sub>Block, End</sub>	Block program end-sequence wait time	See (4)			6		t <sub>FTG</sub>
t <sub>Mass Erase</sub>	Mass erase time	See (4)			10593		t <sub>FTG</sub>
t <sub>Seg Erase</sub>	Segment erase time	See (4)			4819		t <sub>FTG</sub>

Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. (1)

Additional flash retention documentation located in application report (SLAA392). (2)

(3)The cumulative program time must not be exceeded when writing to a 64-byte flash block. This parameter applies to all programming methods: individual word/byte write and block write modes.

(4)These values are hardwired into the Flash Controller's state machine ( $t_{FTG} = 1/f_{FTG}$ ).

### RAM<sup>(1)</sup>

over recommended ranges of supply voltage and up to operating free-air temperature  $T_A = 105^{\circ}C$ 

PARAMETER		TEST CONDITIONS	MIN MAX	UNIT	
V <sub>(RAMh)</sub>	RAM retention supply voltage <sup>(2)</sup>		CPU halted	1.6	V

Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. (1)

This parameter defines the minimum supply voltage V<sub>CC</sub> when the data in RAM remains unchanged. No program execution should (2)happen during this supply voltage condition.

### JTAG and Spy-Bi-Wire Interface

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>SBW</sub>	Spy-Bi-Wire input frequency	2.2 V/3 V	0		20	MHz
t <sub>SBW,Low</sub>	Spy-Bi-Wire low clock pulse length <sup>(1)</sup>	2.2 V/3 V	0.025		15	μs
t <sub>SBW,En</sub>	Spy-Bi-Wire enable time (TEST high to acceptance of first clock edge <sup>(2)</sup> )	2.2 V/3 V			1	μs
t <sub>SBW,Ret</sub>	Spy-Bi-Wire return to normal operation time	2.2 V/3 V	15		100	μs
f <sub>TCK</sub>	TCK input frequency <sup>(3)</sup>	2.2 V	0		5	MHz
		3 V	0		10	MHz
R <sub>Internal</sub>	Internal pulldown resistance on TEST	2.2 V/3 V	25	60	90	kΩ

(1)

Parameters are characterized up to  $T_A = 105^{\circ}C$  unless otherwise noted. Tools accessing the Spy-Bi-Wire interface need to wait for the maximum  $t_{SBW,En}$  time after pulling the TEST/SBWCLK pin high before (2)applying the first SBWCLK clock edge.

(3) f<sub>TCK</sub> may be restricted to meet the timing requirements of the module selected.

### JTAG Fuse<sup>(1)</sup>

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
V <sub>CC(FB)</sub>	Supply voltage during fuse-blow condition	$T_A = 25^{\circ}C$	2.5		V
V <sub>FB</sub>	Voltage level on TEST for fuse blow		6	7	V
I <sub>FB</sub>	Supply current into TEST during fuse blow			100	mA
t <sub>FB</sub>	Time to blow fuse			1	ms

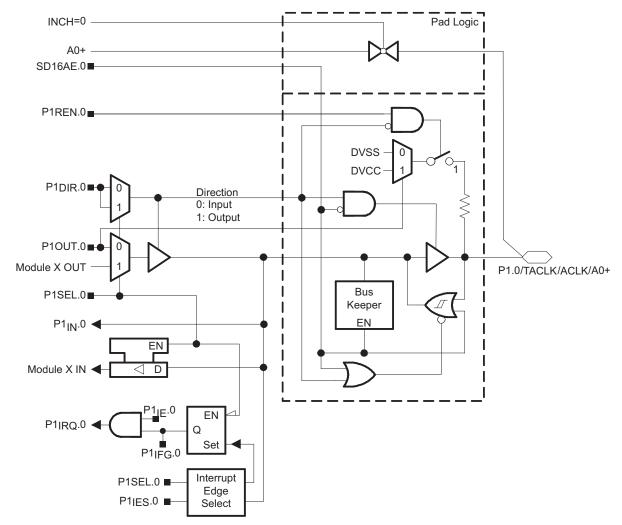
Once the fuse is blown, no further access to the JTAG/Test, Spy-Bi-Wire, and emulation feature is possible, and JTAG is switched to (1) bypass mode.

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#### **APPLICATION INFORMATION**

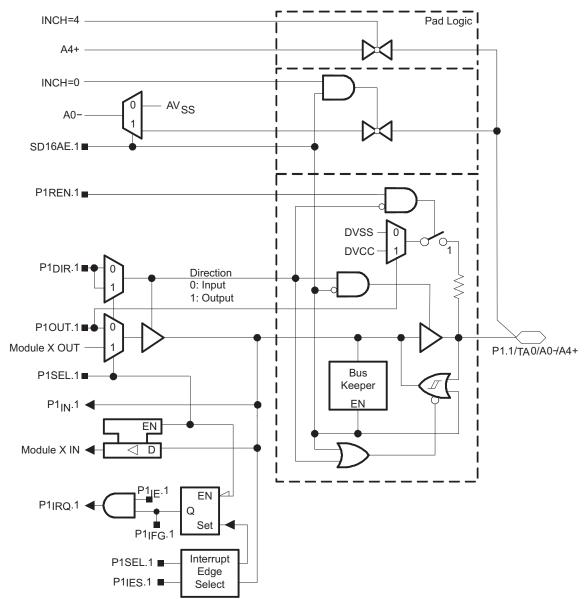


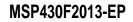
# Port P1 (P1.0) Pin Schematics



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TEXAS INSTRUMENTS

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#### Pad Logic INCH=1 -A1+ -INCH=4 AVSS 0 A4-SD16AE.2 P1REN.2 DVSS 0 DVCC P1DIR.2 Direction I 0: Input 1 1: Output P10UT.2 Module X OUT P1.2/TA1/A1+/A4-Bus P1SEL.2 Keeper L P1<sub>IN</sub>.2 ◀ ΕN L L ΕN I Module X IN ┥ D 1 1 1<sub>IE</sub>.2 ΕN P1IRQ.2 Q Set Ċ. P1IFG.2 Interrupt P1SEL.2 Edge P1IES.2 Select

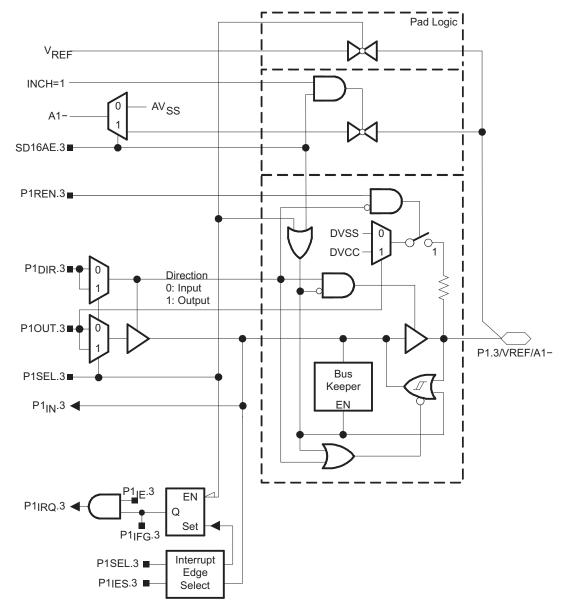
## Port P1 (P1.2) Pin Schematics



#### SLAS774A-JULY 2011-REVISED OCTOBER 2011

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SLAS774A-JULY 2011-REVISED OCTOBER 2011

		FUNCTION	CONTROL BITS / SIGNALS <sup>(1)(2)</sup>						
PIN NAME (P1.x)	x	FUNCTION	P1DIR.x         P1SEL.x         SD16/           0/1         0         0           JCLK         0         1         0           JCLK         0         1         0           X         X         1         0           X         X         1         0           0/1         0         0         0           X         X         1         0           0         1         0         0           X         X         1         0           X         X         1         0           X         X         1         0           0/1         0         0         0           0/1         0         0         0           0/1         0         0         0	SD16AE.x	INCHx				
		P1.0 <sup>(3)</sup> input/output	0/1	0	0	N/A			
	0	Timer_A2.TACLK/INCLK	0	1	0	N/A			
P1.0/TACLK/ACLK/A0+	0	ACLK	1	1	0	N/A			
		A0+ <sup>(4)</sup>	Х	Х	SD16AE.x           0         0           1         0           1         0           X         1           0         0           1         0           X         1           0         0           1         0           1         0           1         0           X         1           X         1           X         1           0         0           1         0           1         0           1         0	0			
P1.1/TA0/A0-/A4+		P1.1 <sup>(3)</sup> input/output	0/1	0	0	N/A			
	1	Timer_A2.CCI0A	0	1	0	N/A			
		Timer_A2.TA0	1	1	0	N/A			
		A0- <sup>(4)(5)</sup>	Х	Х	1	0			
		A4+ <sup>(4)</sup>	Х	Х	1	4			
		P1.2 <sup>(3)</sup> input/output	0/1	0	0	N/A			
		Timer_A2.CCI1A	0	1	0	N/A			
P1.2/TA1/A1+/A4-	2	Timer_A2.TA1	1	1	0	N/A			
		A1+ <sup>(4)</sup>	Х	Х	1	1			
		A4- <sup>(4)(5)</sup>	Х	Х	1	4			
		P1.3 <sup>(3)</sup> input/output	0/1	0	0	N/A			
P1.3/VREF/A1-	3	VREF	Х	1	0	N/A			
		A1- <sup>(4)(5)</sup>	Х	Х	1	1			

### Table 13. Port P1 (P1.0 to P1.3) Pin Functions

X = Don't care (1)

(2) (3) N/A = Not available or not applicable

Default after reset (PUC/POR)

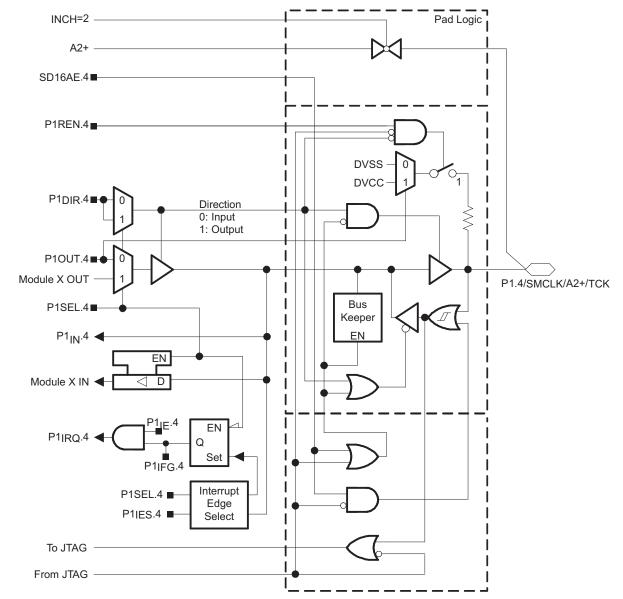
Setting the SD16AE.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying (4) analog signals.

(5) With SD16AE.x = 0 the negative inputs are connected to VSS if the corresponding input is selected.



SLAS774A - JULY 2011 - REVISED OCTOBER 2011

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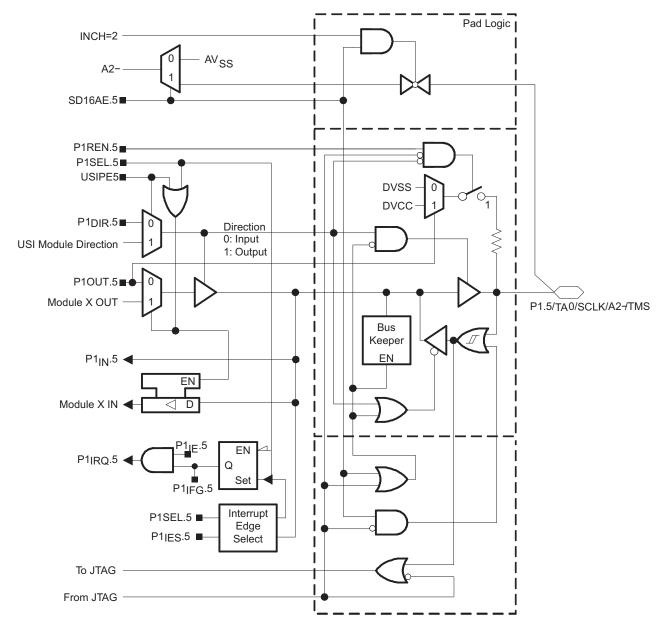


## Port P1 (P1.4) Pin Schematics



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### Port P1 (P1.5) Pin Schematics

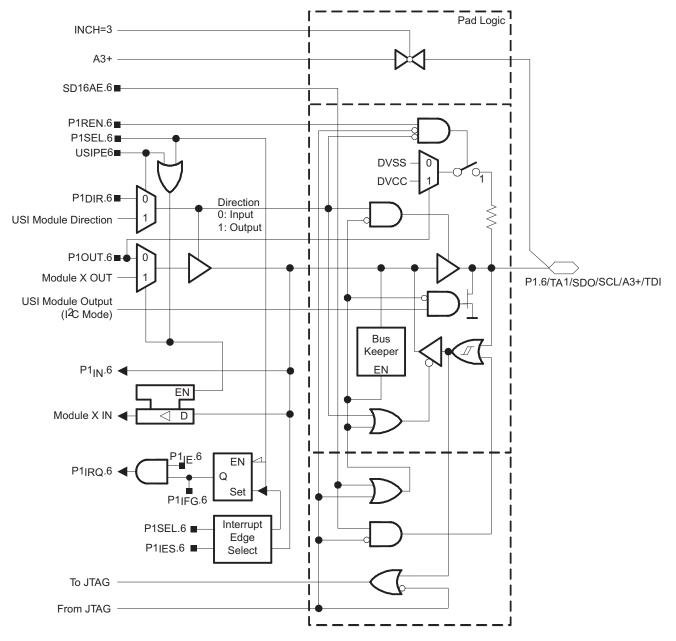




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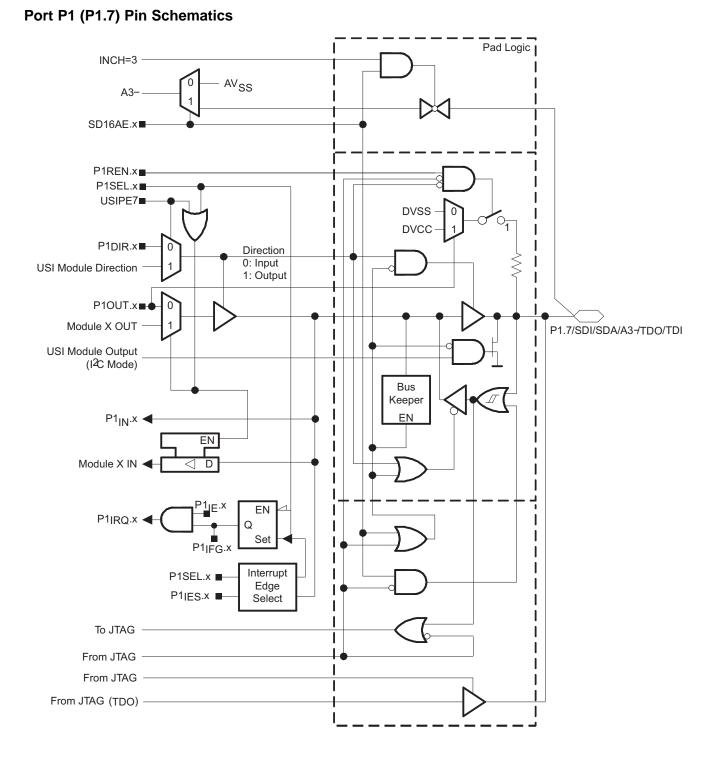
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## Port P1 (P1.6) Pin Schematics





SLAS774A-JULY 2011-REVISED OCTOBER 2011



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PIN NAME (P1.x)		FUNCTION	CONTROL BITS / SIGNALS <sup>(1)(2)</sup>								
PIN NAME (P1.X)	x	FUNCTION	P1DIR.x	P1SEL.x	USIP.x	SD16AE.x	INCHx	JTAG Mode			
		P1.4 <sup>(3)</sup> input/output	0/1	0	N/A	0	N/A	0			
P1.4/SMCLK/A2+/TCK P1.5/TA0/SCLK/A2-/TMS P1.6/TA1/SDO/SCL/ A3+/TDI		N/A	0	1	N/A	0	N/A	0			
	4	SMCLK	1	1	N/A	0	N/A	0			
		A2+ <sup>(4)</sup>	Х	Х	N/A	1	2	0			
		TCK <sup>(5)</sup>	х	х	N/A	Х	Х	1			
		P1.5 <sup>(3)</sup> input/output	0/1	0	0	0	N/A	0			
		N/A	0	1	0	0	N/A	0			
	-	Timer_A2.TA0	1	1	0	0	N/A	0			
P1.5/TAU/SCLK/A2-/TMS	5	SCLK	Х	Х	1	0	N/A	0			
		A2- <sup>(4)(6)</sup>	х	х	Х	1	2	0			
		TMS <sup>(5)</sup>	Х	Х	Х	Х	Х	1			
		P1.6 <sup>(3)</sup> input/output	0/1	0	0	0	N/A	0			
		Timer_A2.CCI1B	0	1	0	0	N/A	0			
P1.6/TA1/SDO/SCL/	6	Timer_A2.TA1	1	1	0	0	N/A	0			
	6	SDO (SPI) / SCL (I2C)	Х	Х	1	0	N/A	0			
		A3+ <sup>(4)</sup>	Х	х	Х	1	3	0			
		TDI <sup>(5)</sup>	Х	Х	Х	Х	Х	1			
		P1.7 <sup>(3)</sup> input/output	0/1	0	0	0	N/A	0			
		N/A	0	1	0	0	N/A	0			
P1.7/SDI/SDA/A3-/	7	DVSS	1	1	0	0	N/A	0			
TDO/TDI		SDI (SPI) / SDA (I2C)	Х	Х	1	0	N/A	0			
		A3- <sup>(4)(6)</sup>	Х	Х	Х	1	3	0			
		TDO/TDI <sup>(5)(7)</sup>	Х	Х	Х	Х	Х	1			

#### Table 14. Port P1 (P1.4 to P1.7) Pin Functions

(1) X = Don't care

(1) X = Don't care
(2) N/A = Not available or not applicable
(3) Default after reset (PUC/POR)

(4) Setting the SD16AE.x bit disables the output driver as well as the input Schmitt trigger to prevent parasitic cross currents when applying analog signals.

(5)

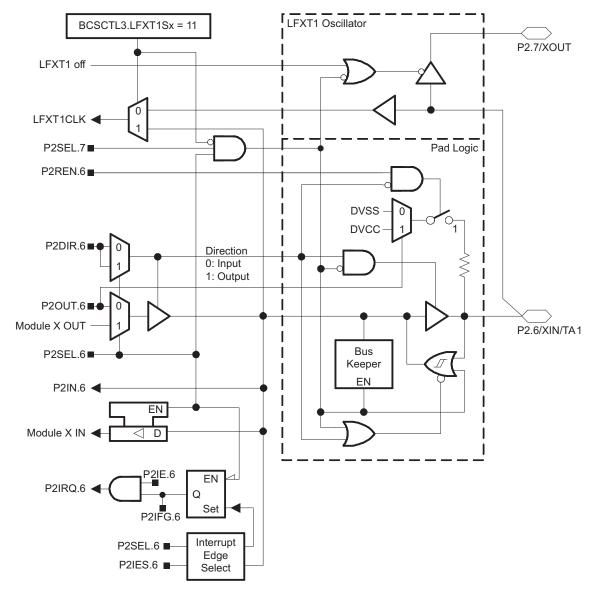
In JTAG mode the internal pullup/down resistors are disabled. With SD16AE.x = 0 the negative inputs are connected to VSS if the corresponding input is selected. (6)

(7) Function controlled by JTAG



SLAS774A-JULY 2011-REVISED OCTOBER 2011

### Port P2 (P2.6) Pin Schematics



### Table 15. Port P2 (P2.6) Pin Functions

		FUNCTION	<b>CONTROL BITS / SIGNAL</b>			
PIN NAME (P2.x)	x	FONCTION	P2DIR.x P2SEL.x			
		P2.6 input/output	0/1	0		
P2.6/XIN/TA1		XIN <sup>(1)(2)</sup>	0	1		
		Timer_A2.TA1	1	1		

(1) Default after reset (PUC/POR)

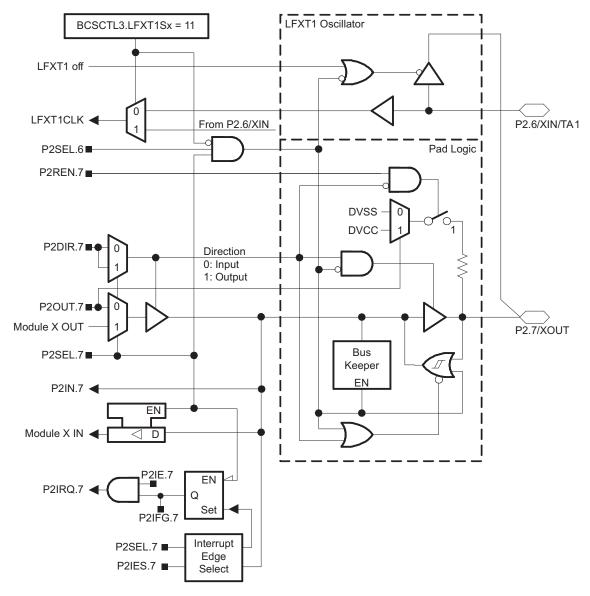
(2) XIN is used as digital clock input if the bits LFXT1Sx in register BCSCTL3 are set to 11.



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#### Port P2 (P2.7) Pin Schematics



#### Table 16. Port P2 (P2.7) Pin Functions

		FUNCTION	<b>CONTROL BITS / SIGNALS</b>		
FIN NAME (F2.X)	PIN NAME (P2.x) x	FONCTION	P2DIR.x	P2SEL.x	
	7	P2.7 input/output	0/1	0	
P2.7/XOUT		DVSS	0	1	
		XOUT <sup>(1)(2)</sup>	1	1	

(1) Default after reset (PUC/POR)

(2) If the pin P2.7/XOUT is used as an input a current can flow until P2SEL.7 is cleared due to the oscillator output driver connection to this pin after reset.



### **PACKAGING INFORMATION**

Orderable part number	Status (1)	Material type	Package   Pins	Package qty   Carrier	<b>RoHS</b> (3)	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
MSP430F2013QRSATEP	Active	Production	QFN (RSA)   16	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	M430F 2013Q
V62/11613-01XE	Active	Production	QFN (RSA)   16	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	M430F 2013Q

<sup>(1)</sup> **Status:** For more details on status, see our product life cycle.

<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF MSP430F2013-EP :



20-May-2025

• Catalog : MSP430F2013

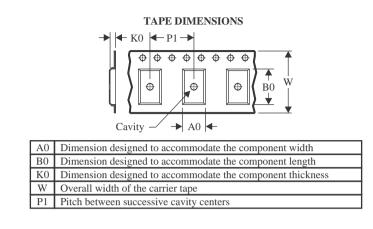
NOTE: Qualified Version Definitions:

• Catalog - TI's standard catalog product



## TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

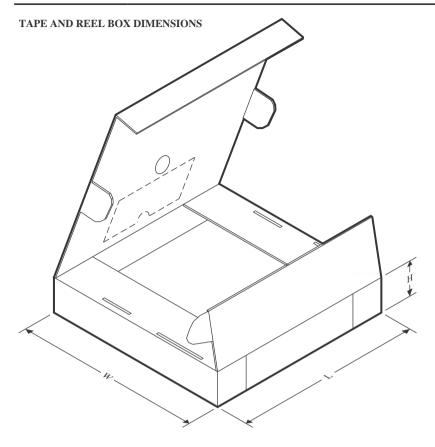


*All dimensions are nomina	l											
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430F2013QRSATE	P QFN	RSA	16	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2



# PACKAGE MATERIALS INFORMATION

15-Jul-2025



*All	dimensions	are	nominal
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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430F2013QRSATEP	QFN	RSA	16	250	213.0	191.0	35.0

# **GENERIC PACKAGE VIEW**

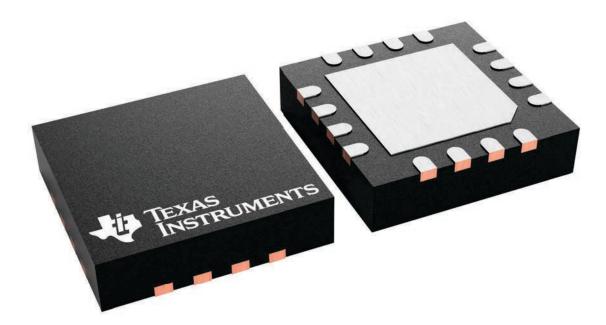
## VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD

4 x 4, 0.65 mm pitch

**RSA 16** 

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





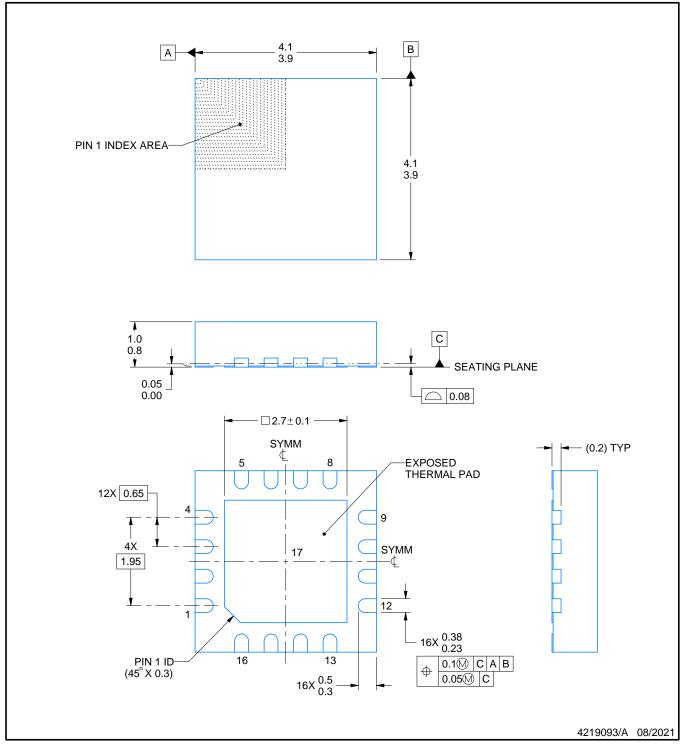
# **RSA0016B**



# **PACKAGE OUTLINE**

## VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice.
- The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.
   Reference JEDEC registration MO-220.

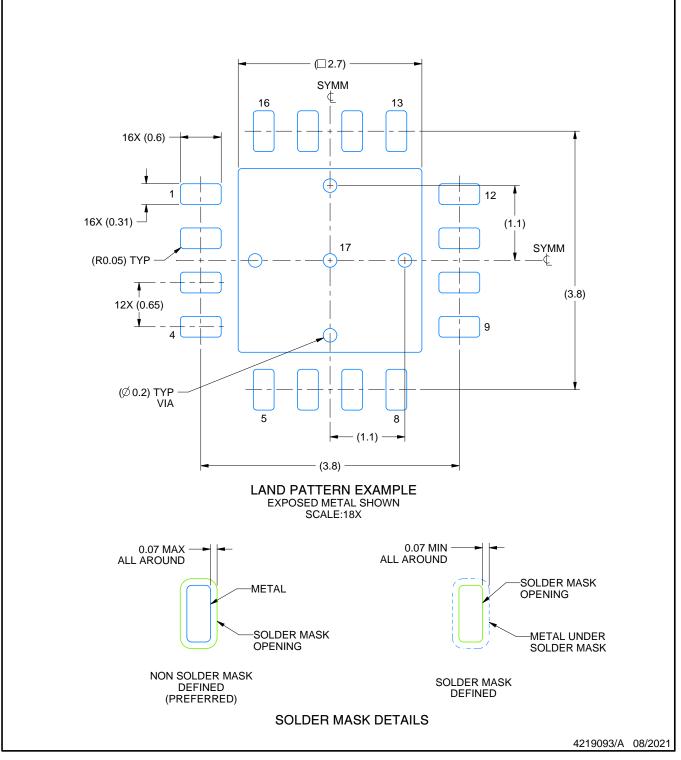


# **RSA0016B**

# **EXAMPLE BOARD LAYOUT**

## VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

5. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).

6. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

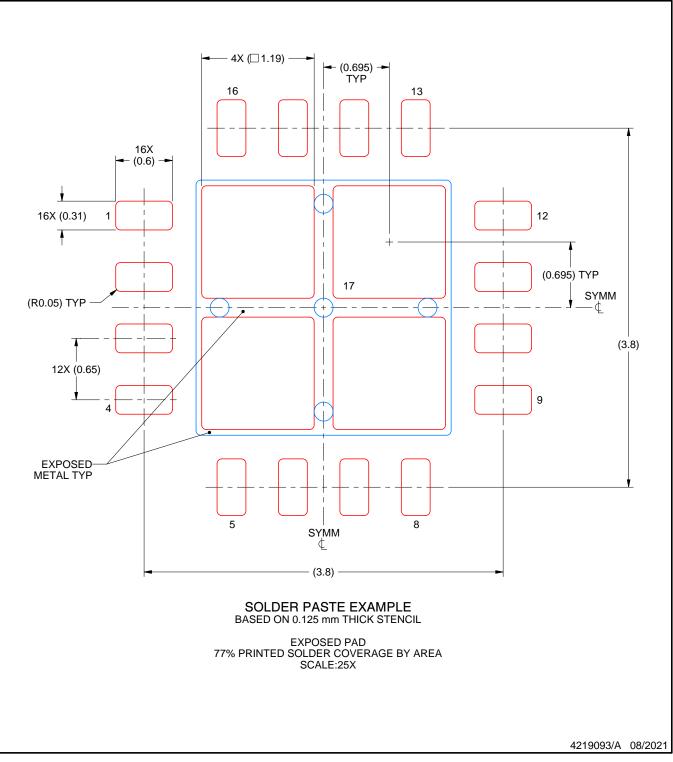


# **RSA0016B**

# **EXAMPLE STENCIL DESIGN**

## VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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