

**LMT86-Q1** 

SNIS201A - OCTOBER 2017 - REVISED JUNE 2022

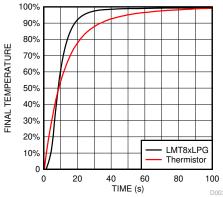
# LMT86-Q1 2.2-V, SC70, **Analog Temperature Sensor**

#### 1 Features

- LMT86-Q1-Q1 is AEC-Q100 Qualified for Automotive Applications:
  - Device Temperature Grade 0: –40°C to +150°C
  - Device HBM ESD Classification Level 2
  - Device CDM ESD Classification Level C6
- Functional Safety-Capable
  - Documentation available to aid functional safety system design
- Very Accurate: ±0.4°C Typical
- Low 2.2-V Operation
- Average Sensor Gain of -10.9 mV/°C
- Low 5.4-µA Quiescent Current
- Wide Temperature Range: -50°C to 150°C
- Output is Short-Circuit Protected
- Push-Pull Output With ±50-µA Drive Capability
- Footprint Compatible With the Industry-Standard LM20/19 and LM35 Temperature Sensors
- Cost-Effective Alternative to Thermistors

## 2 Applications

- Automotive
- Infotainment and Cluster
- Powertrain Systems
- Smoke and Heat Detectors
- **Drones**
- **Appliances**



<sup>\*</sup> Fast thermal response NTC

## **Thermal Time Constant**

## 3 Description

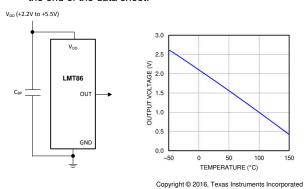
The LMT86-Q1 are precision CMOS temperature sensors with ±0.4°C typical accuracy (±2.7°C maximum) and a linear analog output voltage that is inversely proportional to temperature. The 2.2-V supply voltage operation, 5.4-µA quiescent current, and 0.7-ms power-on time enable effective powercycling architectures to minimize power consumption for battery-powered applications such as drones and sensor nodes. The LMT86-Q1-Q1 device is AEC-Q100 Grade 0 qualified and maintains ±2.7°C maximum accuracy over the full operating temperature range without calibration; this makes the LMT86-Q1-Q1 suitable for automotive applications such as infotainment, cluster, and powertrain systems. The accuracy over the wide operating range and other features make the LMT86-Q1 an excellent alternative to thermistors.

For devices with different average sensor gains and comparable accuracy, refer to Comparable Alternative Devices for alternative devices in the LMT8x family.

## Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LMT86-Q1	SOT (5)	2.00 mm × 1.25 mm

For all available packages, see the orderable addendum at the end of the data sheet.



**Output Voltage vs Temperature** 



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# **4 Revision History**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

С	hanges from Revision * (October 2017) to Revision A (June 2022)	Page
•	Updated the numbering format for tables, figures, and cross-references throughout the document	1
•	Added Functional Safety bullets to the Features section	1



## **5 Device Comparison**

## Table 5-1. Available Device Packages

ORDER NUMBER <sup>(1)</sup>	PACKAGE	PIN	BODY SIZE (NOM)	MOUNTING TYPE	
LMT86DCK	SOT (AKA <sup>(2)</sup> : SC70, DCK)	5	2.00 mm × 1.25 mm	Surface Mount	
LMT86LP	TO-92 (AKA <sup>(2)</sup> : LP)	3	4.30 mm × 3.50 mm	Through-hole; straight leads	
LMT86LPG	TO-92S (AKA <sup>(2)</sup> : LPG)	3	4.00 mm × 3.15 mm	Through-hole; straight leads	
LMT86LPM	TO-92 (AKA <sup>(2)</sup> : LPM)	3	4.30 mm × 3.50 mm	Through-hole; formed leads	
LMT86DCK-Q1	SOT (AKA <sup>(2)</sup> : SC70, DCK)	5	2.00 mm × 1.25 mm	Surface Mount	

- (1) For all available packages and complete order numbers, see the Package Option addendum at the end of the data sheet.
- (2) AKA = Also Known As

**Table 5-2. Comparable Alternative Devices** 

DEVICE NAME	AVERAGE OUTPUT SENSOR GAIN	POWER SUPPLY RANGE
LMT84-Q1	−5.5 mV/°C	1.5 V to 5.5 V
LMT85-Q1	−8.2 mV/°C	1.8 V to 5.5 V
LMT86-Q1	−10.9 mV/°C	2.2 V to 5.5 V
LMT87-Q1	−13.6 mV/°C	2.7 V to 5.5 V

## **6 Pin Configuration and Functions**

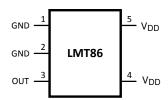


Figure 6-1. 5-Pin SOT (SC70) DCK Package (TOP VIEW)

Table 6-1. Pin Functions

PIN		TVDE	DESCRIPTION		
NAME	SOT (SC70)	TYPE	EQUIVALENT CIRCUIT	FUNCTION	
GND	1, 2 <sup>(1)</sup>	Ground	N/A	Power Supply Ground	
OUT	3	Analog Output	V <sub>DD</sub> GND	Outputs a voltage that is inversely proportional to temperature	
V <sub>DD</sub>	4, 5	Power	N/A	Positive Supply Voltage	

(1) Direct connection to the back side of the die



## 7 Specifications

## 7.1 Absolute Maximum Ratings

See (1) (2)

	MIN	MAX	UNIT
Supply voltage	-0.3	6	V
Voltage at output pin	-0.3	(V <sub>DD</sub> + 0.5)	V
Output current	<b>-7</b>	7	mA
Input current at any pin (3)	<b>-</b> 5	5	mA
Maximum junction temperature (T <sub>JMAX</sub> )		150	°C
Storage temperature, T <sub>stg</sub>	<b>–</b> 65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability
- (2) Soldering process must comply with TI's Reflow Temperature Profile specifications. Refer to www.ti.com/packaging. Reflow temperature profiles are different for lead-free and non-lead-free packages.
- (3) When the input voltage  $(V_1)$  at any pin exceeds power supplies  $(V_1 < GND \text{ or } V_1 > V)$ , the current at that pin should be limited to 5 mA.

## 7.2 ESD Ratings

			VALUE	UNIT
LMT86DCK-Q1 in SC70 package				
V	Floatrostatio discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2500	W
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per AEC Q100-011	±1000	V

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

#### 7.3 Recommended Operating Conditions

	MIN	MAX	UNIT
Specified temperature	$T_{MIN} \le T_A \le T_{MAX}$		°C
eu temperature	-50 ≤ T <sub>A</sub> ≤ 150		°C
Supply voltage (V <sub>DD</sub> )	2.2	5.5	V

### 7.4 Thermal Information

		LMT86-Q1	
	THERMAL METRIC <sup>(1)</sup> (2)	DCK (SOT/SC70)	UNIT
		5 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance (3) (4)	275	°C/W
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	84	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	56	°C/W
ΨЈТ	Junction-to-top characterization parameter	1.2	°C/W
ΨЈВ	Junction-to-board characterization parameter	55	°C/W

- (1) For information on self-heating and thermal response time, see section Mounting and Thermal Conductivity.
- (2) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report.
- (3) The junction to ambient thermal resistance (R<sub>θJA</sub>) under natural convection is obtained in a simulation on a JEDEC-standard, High-K board as specified in JESD51-7, in an environment described in JESD51-2. Exposed pad packages assume that thermal vias are included in the PCB, per JESD 51-5.
- (4) Changes in output due to self-heating can be computed by multiplying the internal dissipation by the thermal resistance.

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## 7.5 Accuracy Characteristics

These limits do not include DC load regulation. These stated accuracy limits are with reference to the values in Table 8-1.

PARAMETER	CONDITIONS	MIN <sup>(1)</sup>	TYP <sup>(2)</sup>	MAX <sup>(1)</sup>	UNIT
	40°C to 150°C; V <sub>DD</sub> = 2.2 V to 5.5 V	-2.7	±0.4	2.7	°C
	0°C to 40°C; V <sub>DD</sub> = 2.4 V to 5.5 V	-2.7	±0.7	2.7	°C
Temperature accuracy <sup>(3)</sup>	0°C to 70°C; V <sub>DD</sub> = 3.0 V to 5.5 V		±0.3		°C
	–50°C to 0°C; V <sub>DD</sub> = 3.0 V to 5.5 V	-2.7	±0.7	2.7	°C
	–50°C to 0°C; V <sub>DD</sub> = 3.6 V to 5.5 V		±0.25		°C

- (1) Limits are specified to TI's AOQL (Average Outgoing Quality Level).
- (2) Typicals are at  $T_J = T_A = 25^{\circ}$ C and represent most likely parametric norm.
- (3) Accuracy is defined as the error between the measured and reference output voltages, tabulated in the Transfer Table at the specified conditions of supply gain setting, voltage, and temperature (expressed in °C). Accuracy limits include line regulation within the specified conditions. Accuracy limits do not include load regulation; they assume no dc load.

### 7.6 Electrical Characteristics

Unless otherwise noted, these specifications apply for  $+V_{DD}$  = 2.2 V to 5.5 V. MIN and MAX limits apply for  $T_A$  =  $T_J$  =  $T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted; typical values apply for  $T_A$  =  $T_J$  = 25°C.

	PARAMETER	TEST CONDITIONS	MIN <sup>(1)</sup>	TYP <sup>(2)</sup>	MAX <sup>(1)</sup>	UNIT
	Average sensor gain (output transfer function slope)	-30°C and 90°C used to calculate average sensor gain		-10.9		mV/°C
	Load regulation <sup>(3)</sup>	Source $\leq$ 50 $\mu$ A, $(V_{DD} - V_{OUT}) \geq$ 200 mV	-1	-0.22		mV
	Load regulation (*)	Sink ≤ 50 μA, V <sub>OUT</sub> ≥ 200 mV		0.26	1	mV
	Line regulation <sup>(4)</sup>			200		μV/V
	Supply current	$T_A = 30^{\circ}C \text{ to } 150^{\circ}C, (V_{DD} - V_{OUT}) \ge 100 \text{ mV}$		5.4	8.1	μA
IS		$T_A = -50^{\circ}C \text{ to } 150^{\circ}C, (V_{DD} - V_{OUT}) \ge 100 \text{ mV}$		5.4	9	μA
CL	Output load capacitance			1100		pF
	Power-on time <sup>(5)</sup>	C <sub>L</sub> = 0 pF to 1100 pF		0.7	1.9	ms
	Output drive	$T_A = T_J = 25$ °C	-50		50	μA

- (1) Limits are specific to TI's AOQL (Average Outgoing Quality Level).
- (2) Typicals are at  $T_J = T_A = 25^{\circ}$ C and represent most likely parametric norm.
- (3) Source currents are flowing out of the LMT86-Q1. Sink currents are flowing into the LMT86-Q1.
- (4) Line regulation (DC) is calculated by subtracting the output voltage at the highest supply voltage from the output voltage at the lowest supply voltage. The typical DC line regulation specification does not include the output voltage shift discussed in Output Voltage Shift.
- (5) Specified by design and characterization.



## 7.7 Typical Characteristics

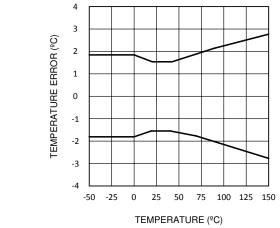


Figure 7-1. Temperature Error vs Temperature

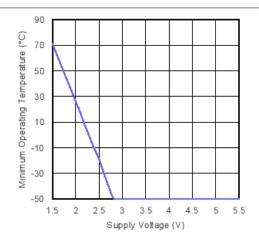


Figure 7-2. Minimum Operating Temperature vs Supply Voltage

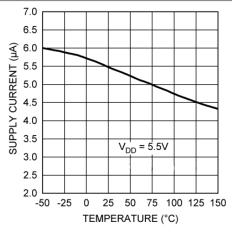


Figure 7-3. Supply Current vs Temperature

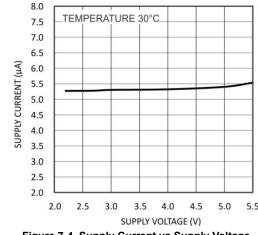
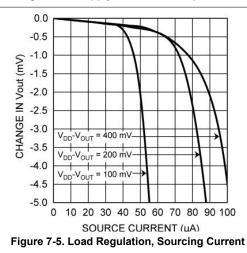


Figure 7-4. Supply Current vs Supply Voltage



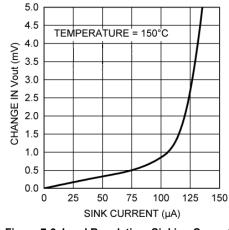
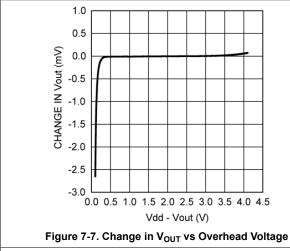


Figure 7-6. Load Regulation, Sinking Current



## 7.7 Typical Characteristics (continued)



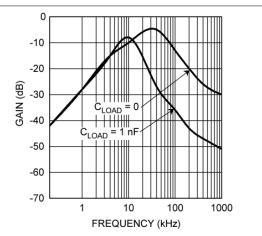


Figure 7-8. Supply-Noise Gain vs Frequency

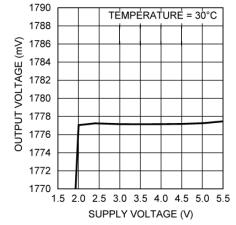


Figure 7-9. Output Voltage vs Supply Voltage



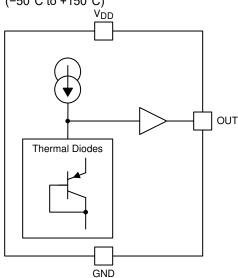
## 8 Detailed Description

#### 8.1 Overview

The LMT86-Q1 is an analog output temperature sensor. The temperature-sensing element is comprised of a simple base emitter junction that is forward biased by a current source. The temperature-sensing element is then buffered by an amplifier and provided to the OUT pin. The amplifier has a simple push-pull output stage thus providing a low impedance output source.

## 8.2 Functional Block Diagram

Full-Range Celsius Temperature Sensor (-50°C to +150°C)



## 8.3 Feature Description

#### 8.3.1 LMT86-Q1 Transfer Function

**TEMP** 

Table 8-1 shows the output voltage of the LMT86-Q1 across the complete operating temperature range. This table is the reference from which the LMT86-Q1 accuracy specifications (listed in the Accuracy Characteristics table) are determined. This table can be used, for example, in a host processor look-up table. A file containing this data is available for download at LMT86-Q1 product folder under Tools and Software Models.

V<sub>OUT</sub> (mV) V<sub>OUT</sub> (mV) V<sub>OUT</sub> (mV) V<sub>OUT</sub> (mV) V<sub>OUT</sub> (mV) (°C) (°C) (°C) (°C) (°C) -50 -10 -49 -9 -48 -8 -47 -7 -46 -6 -5 -45 -44 -4 -43 -3 -42 -2 -41 -1 -40 

Table 8-1. LMT86-Q1 Transfer Table

TEMP

**TEMP** 

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TEMP

-39

-38

-37

-36

-35

TEMP

Table 8-1. LMT86-Q1 Transfer Table (continued)

Table 8-1. LMT86-Q1 Transfer Table (continued)											
TEMP (°C)	V <sub>OUT</sub> (mV)	TEMP (°C)	V <sub>OUT</sub> (mV)	TEMP (°C)	V <sub>OUT</sub> (mV)	TEMP (°C)	V <sub>OUT</sub> (mV)	TEMP (°C)	V <sub>OUT</sub> (mV)		
-34	2460	6	2036	46	1602	86	1155	126	699		
-33	2449	7	2025	47	1591	87	1144	127	688		
-32	2439	8	2014	48	1580	88	1133	128	676		
-31	2429	9	2004	49	1569	89	1122	129	665		
-30	2418	10	1993	50	1558	90	1110	130	653		
-29	2408	11	1982	51	1547	91	1099	131	642		
-28	2397	12	1971	52	1536	92	1088	132	630		
-27	2387	13	1961	53	1525	93	1076	133	618		
-26	2376	14	1950	54	1514	94	1065	134	607		
-25	2366	15	1939	55	1503	95	1054	135	595		
-24	2355	16	1928	56	1492	96	1042	136	584		
-23	2345	17	1918	57	1481	97	1031	137	572		
-22	2334	18	1907	58	1470	98	1020	138	560		
-21	2324	19	1896	59	1459	99	1008	139	549		
-20	2313	20	1885	60	1448	100	997	140	537		
-19	2302	21	1874	61	1436	101	986	141	525		
-18	2292	22	1864	62	1425	102	974	142	514		
-17	2281	23	1853	63	1414	103	963	143	502		
-16	2271	24	1842	64	1403	104	951	144	490		
-15	2260	25	1831	65	1391	105	940	145	479		
-14	2250	26	1820	66	1380	106	929	146	467		
-13	2239	27	1810	67	1369	107	917	147	455		
-12	2228	28	1799	68	1358	108	906	148	443		
-11	2218	29	1788	69	1346	109	895	149	432		

Although the LMT86-Q1 is very linear, its response does have a slight umbrella parabolic shape. Table 8-1 very accurately reflects this shape. The Transfer Table can be calculated by using the parabolic equation (Equation 1).

$$V_{TEMP}(mV) = 1777.3mV - \left[10.888 \frac{mV}{^{\circ}C} (T - 30^{\circ}C)\right] - \left[0.00347 \frac{mV}{^{\circ}C^{2}} (T - 30^{\circ}C)^{2}\right]$$
(1)

The parabolic equation is an approximation of the transfer table and the accuracy of the equation degrades slightly at the temperature range extremes. Equation 1 can be solved for T resulting in:

$$T = \frac{10.888 - \sqrt{(-10.888)^2 + 4 \times 0.00347 \times (1777.3 - V_{TEMP} (mV))}}{2 \times (-0.00347)} + 30$$
 (2)

For an even less accurate linear approximation, a line can easily be calculated over the desired temperature range from the table using the two-point equation (Equation 3):

$$V - V_1 = \left(\frac{V_2 - V_1}{T_2 - T_1}\right) \times (T - T_1)$$
(3)

## where

- V is in mV.
- T is in °C,
- T<sub>1</sub> and V<sub>1</sub> are the coordinates of the lowest temperature,

and T<sub>2</sub> and V<sub>2</sub> are the coordinates of the highest temperature.

For example, if the user wanted to resolve this equation, over a temperature range of 20°C to 50°C, they would proceed as follows:

$$V - 1885 \text{ mV} = \left(\frac{1558 \text{ mV} - 1885 \text{ mV}}{50^{\circ}\text{C} - 20^{\circ}\text{C}}\right) \times (\text{T} - 20^{\circ}\text{C})$$
(4)

V - 1885 mV = 
$$(-10.9 \text{ mV} / {}^{\circ}\text{C}) \times (\text{T} - 20 {}^{\circ}\text{C})$$
 (5)

$$V = (-10.9 \text{ mV} / {}^{\circ}\text{C}) \times \text{T} + 2103 \text{ mV}$$
 (6)

Using this method of linear approximation, the transfer function can be approximated for one or more temperature ranges of interest.

#### 8.4 Device Functional Modes

## 8.4.1 Mounting and Thermal Conductivity

The LMT86-Q1 can be applied easily in the same way as other integrated-circuit temperature sensors. It can be glued or cemented to a surface.

To ensure good thermal conductivity, the backside of the LMT86-Q1 die is directly attached to the GND pin. The temperatures of the lands and traces to the other leads of the LMT86-Q1 will also affect the temperature reading.

Alternatively, the LMT86-Q1 can be mounted inside a sealed-end metal tube, and can then be dipped into a bath or screwed into a threaded hole in a tank. As with any IC, the LMT86-Q1 and accompanying wiring and circuits must be kept insulated and dry, to avoid leakage and corrosion. This is especially true if the circuit may operate at cold temperatures where condensation can occur. If moisture creates a short circuit from the output to ground or V<sub>DD</sub>, the output from the LMT86-Q1 will not be correct. Printed-circuit coatings are often used to ensure that moisture cannot corrode the leads or circuit traces.

The thermal resistance junction to ambient ( $R_{\theta JA}$  or  $\theta_{JA}$ ) is the parameter used to calculate the rise of a device junction temperature due to its power dissipation. Use Equation 7 to calculate the rise in the LMT86-Q1 die temperature:

$$T_{J} = T_{A} + \theta_{JA} \left[ (V_{DD}I_{S}) + (V_{DD} - V_{O}) I_{L} \right]$$

$$(7)$$

where

- T<sub>A</sub> is the ambient temperature,
- I<sub>S</sub> is the supply current,
- I<sub>I</sub> is the load current on the output,
- and V<sub>O</sub> is the output voltage.

For example, in an application where  $T_A = 30^{\circ}\text{C}$ ,  $V_{DD} = 5 \text{ V}$ ,  $I_S = 5.4 \,\mu\text{A}$ ,  $V_O = 1777 \,\text{mV}$  junction temp 30.014°C self-heating error of 0.014°C. Because the junction temperature of the LMT86-Q1 is the actual temperature being measured, take care to minimize the load current that the LMT86-Q1 is required to drive. The *Thermal Information* table shows the thermal resistance of the LMT86-Q1.

#### 8.4.2 Output Noise Considerations

A push-pull output gives the LMT86-Q1 the ability to sink and source significant current. This is beneficial when, for example, driving dynamic loads like an input stage on an analog-to-digital converter (ADC). In these applications the source current is required to quickly charge the input capacitor of the ADC. The LMT86-Q1 is ideal for this and other applications which require strong source or sink current.

The LMT86-Q1 supply-noise gain (the ratio of the AC signal on  $V_{OUT}$  to the AC signal on  $V_{DD}$ ) was measured during bench tests. Figure 7-8 shows the typical attenuation found in the *Typical Characteristics* section. A load capacitor on the output can help to filter noise.

For operation in very noisy environments, some bypass capacitance should be present on the supply within approximately 5 centimeters of the LMT86-Q1.

#### 8.4.3 Capacitive Loads

The LMT86-Q1 handles capacitive loading well. In an extremely noisy environment, or when driving a switched sampling input on an ADC, it may be necessary to add some filtering to minimize noise coupling. Without any precautions, Figure 8-1 shows how the LMT86-Q1 can drive a capacitive load less than or equal to 1100 pF. For capacitive loads greater than 1100 pF, Figure 8-2 shows how a series resistor may be required on the output.

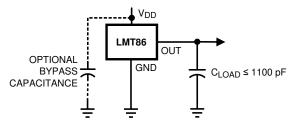


Figure 8-1. LMT86-Q1 No Decoupling Required for Capacitive Loads Less Than 1100 pF

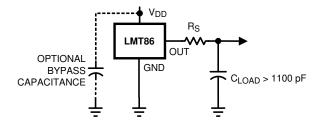


Figure 8-2. LMT86-Q1 With Series Resistor for Capacitive Loading Greater Than 1100 pF

C <sub>LOAD</sub>	MINIMUM R <sub>S</sub>
1.1 nF to 99 nF	3 kΩ
100 nF to 999 nF	1.5 kΩ
1 μF	800 Ω

**Table 8-2. Recommended Series Resistor Values** 

### 8.4.4 Output Voltage Shift

The LMT86-Q1 device is very linear over temperature and supply voltage range. Due to the intrinsic behavior of an NMOS/PMOS rail-to-rail buffer, a slight shift in the output can occur when the supply voltage is ramped over the operating range of the device. The location of the shift is determined by the relative levels of  $V_{DD}$  and  $V_{OUT}$ . The shift typically occurs when  $V_{DD} - V_{OUT} = 1 \text{ V}$ .

This slight shift (a few millivolts) takes place over a wide change (approximately 200 mV) in  $V_{DD}$  or  $V_{OUT}$ . Because the shift takes place over a wide temperature change of 5°C to 20°C,  $V_{OUT}$  is always monotonic. The accuracy specifications in the *Accuracy Characteristics* table already include this possible shift.

## 9 Application and Implementation

#### **Note**

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

## 9.1 Application Information

The LMT86-Q1 features make it suitable for many general temperature-sensing applications. It can operate down to 2.2-V supply with 5.4-µA power consumption, making it ideal for battery-powered devices.

## 9.2 Typical Applications

#### 9.2.1 Connection to an ADC

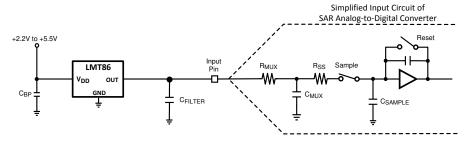


Figure 9-1. Suggested Connection to a Sampling Analog-to-Digital Converter Input Stage

#### 9.2.1.1 Design Requirements

Most CMOS ADCs found in microcontrollers and ASICs have a sampled data comparator input structure. When the ADC charges the sampling cap, it requires instantaneous charge from the output of the analog source such as the LMT86 temperature sensor and many op amps. This requirement is easily accommodated by the addition of a capacitor, C<sub>FILTER</sub>.

### 9.2.1.2 Detailed Design Procedure

The size of  $C_{\text{FILTER}}$  depends on the size of the sampling capacitor and the sampling frequency. Because not all ADCs have identical input stages, the charge requirements will vary. This general ADC application is shown as an example only.

## 9.2.1.3 Application Curve

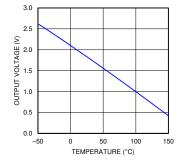


Figure 9-2. Analog Output Transfer Function

### 9.2.2 Conserving Power Dissipation With Shutdown

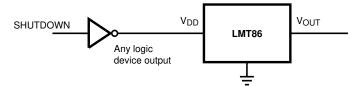


Figure 9-3. Conserving Power Dissipation With Shutdown

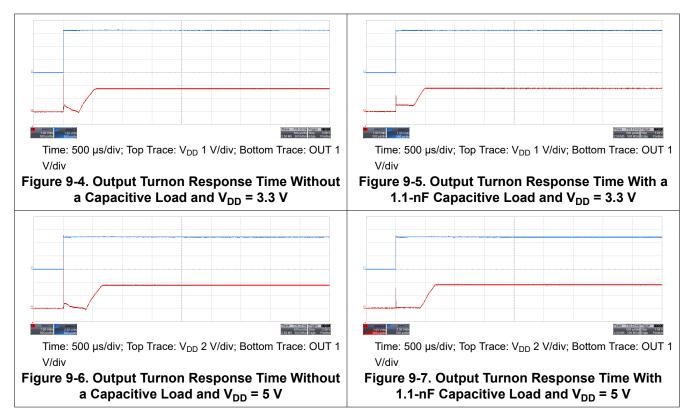
### 9.2.2.1 Design Requirements

Because the power consumption of the LMT86-Q1 is less than 9  $\mu$ A, it can simply be powered directly from any logic gate output and therefore not require a specific shutdown pin. The device can even be powered directly from a microcontroller GPIO. In this way, it can easily be turned off for cases such as battery-powered systems where power savings are critical.

#### 9.2.2.2 Detailed Design Procedure

Simply connect the V<sub>DD</sub> pin of the LMT86-Q1 directly to the logic shutdown signal from a microcontroller.

## 9.2.2.3 Application Curves



## 10 Power Supply Recommendations

The low supply current and supply range (2.2 V to 5.5 V) of the LMT86-Q1 allow the device to easily be powered from many sources. Power supply bypassing is optional and is mainly dependent on the noise on the power supply used. In noisy systems, it may be necessary to add bypass capacitors to lower the noise that is coupled to the output of the LMT86-Q1.

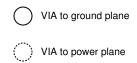


## 11 Layout

## 11.1 Layout Guidelines

The LMT86-Q1 is simple to layout. If a power-supply bypass capacitor is used, the *Layout Example* shows how to connect the capacitor to the device.

## 11.2 Layout Example



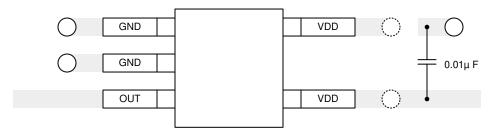


Figure 11-1. SC70 Package Recommended Layout



## 12 Device and Documentation Support

## 12.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

## 12.2 Support Resources

TI E2E<sup>™</sup> support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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#### 12.3 Trademarks

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## 12.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## 12.5 Glossary

TI Glossary

This glossary lists and explains terms, acronyms, and definitions.

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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24-Jul-2025

#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking
	(1)	(2)			(5)	(4)	(5)		(6)
LMT86QDCKRQ1	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-50 to 150	BTA
LMT86QDCKRQ1.A	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-50 to 150	BTA
LMT86QDCKRQ1.B	Active	Production	SC70 (DCK)   5	3000   LARGE T&R	-	SN	Level-1-260C-UNLIM	-50 to 150	BTA
LMT86QDCKTQ1	Obsolete	Production	SC70 (DCK)   5	-	-	Call TI	Call TI	-50 to 150	BTA

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### OTHER QUALIFIED VERSIONS OF LMT86-Q1:

<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

## **PACKAGE OPTION ADDENDUM**

www.ti.com 24-Jul-2025

NOTE: Qualified Version Definitions:

• Catalog - TI's standard catalog product

## **PACKAGE MATERIALS INFORMATION**

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## TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMT86QDCKRQ1	SC70	DCK	5	3000	178.0	8.4	2.25	2.45	1.2	4.0	8.0	Q3

# PACKAGE MATERIALS INFORMATION

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## \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMT86QDCKRQ1	SC70	DCK	5	3000	208.0	191.0	35.0



SMALL OUTLINE TRANSISTOR



#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
  3. Reference JEDEC MO-203.

- 4. Support pin may differ or may not be present.5. Lead width does not comply with JEDEC.
- 6. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25mm per side



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

7. Publication IPC-7351 may have alternate designs.8. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 9. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 10. Board assembly site may have different recommendations for stencil design.



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