



bq40z50-R1

SLUSCB3-JULY 2015

bq40z50-R1 1-Series, 2-Series, 3-Series, and 4-Series Li-Ion Battery Pack Manager

Technical

Documents

Sample &

Buy

1 Features

- Fully Integrated 1-Series, 2-Series, 3-Series, and 4-Series Li-Ion or Li-Polymer Cell Battery Pack Manager and Protection
- Next-Generation Patented Impedance Track[™] Technology Accurately Measures Available Charge in Li-Ion and Li-Polymer Batteries
- High Side N-CH Protection FET Drive
- Integrated Cell Balancing While Charging or At Rest
- Full Array of Programmable Protection Features
 - Voltage
 - Current
 - Temperature
 - **Charge Timeout** _
 - CHG/DSG FETs
 - AFE
- Sophisticated Charge Algorithms
 - JEITA
 - Enhanced Charging
 - Adaptive Charging
 - Cell Balancing
- Supports TURBO BOOST Mode
- Supports Battery Trip Point (BTP)
- Diagnostic Lifetime Data Monitor and Black Box Recorder
- LED Display
- Supports Two-Wire SMBus v1.1 Interface
- SHA-1 Authentication
- Compact Package: 32-Lead QFN (RSM)

Applications 2

- Notebook/Netbook PCs
- Medical and Test Equipment
- Portable Instrumentation

3 Description

Tools &

Software

The bq40z50-R1 device, incorporating patented Impedance Track[™] technology, is a fully integrated, single-chip, pack-based solution that provides a rich array of features for gas gauging, protection, and authentication for 1-series, 2-series, 3-series, and 4series cell Li-lon and Li-Polymer battery packs.

Support &

Community

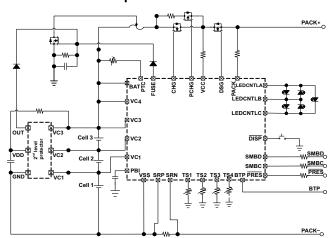
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Usina its integrated high-performance analog peripherals, the bq40z50-R1 device measures and maintains an accurate record of available capacity, voltage, current, temperature, and other critical parameters in Li-Ion or Li-Polymer batteries, and reports this information to the system host controller over an SMBus v1.1 compatible interface.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
bq40z50-R1	VQFN (32)	4.00 mm × 4.00 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.



An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.

Simplified Schematic





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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

DATE	REVISION	NOTES
July 2015	*	Initial Release



5 Description (continued)

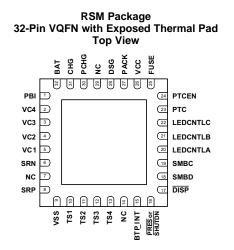
The bq40z50-R1 device supports TURBO BOOST mode by providing the available max power and max current to the host system. The device also supports Battery Trip Point to send a BTP interrupt signal to the host system at the pre-set state of charge thresholds.

The bq40z50-R1 provides software-based 1st- and 2nd-level safety protection against overvoltage, undervoltage, overcurrent, short-circuit current, overload, and overtemperature conditions, as well as other pack- and cell-related faults.

SHA-1 authentication, with secure memory for authentication keys, enables identification of genuine battery packs.

The compact 32-lead QFN package minimizes solution cost and size for smart batteries while providing maximum functionality and safety for battery gauging applications.

6 Pin Configuration and Functions



Pin Functions

PIN NAME	NUMBER	TYPE ⁽¹⁾	DESCRIPTION	
PBI	1	Р	Power supply backup input pin	
VC4	2	IA	Sense voltage input pin for most positive cell, and balance current input for most positive cell	
VC3	3	IA	Sense voltage input pin for second most positive cell, balance current input for second most positive cell, and return balance current for most positive cell	
VC2	4	IA	Sense voltage input pin for third most positive cell, balance current input for third most positive cell, and return balance current for second most positive cell	
VC1	5	IA	Sense voltage input pin for least positive cell, balance current input for least positive cell, and return balance current for third most positive cell	
SRN	6	Ι	Analog input pin connected to the internal coulomb counter peripheral for integrating a small voltage between SRP and SRN where SRP is the top of the sense resistor.	
NC	7	—	Not internally connected. Connect to VSS.	
SRP	8	Ι	Analog input pin connected to the internal coulomb counter peripheral for integrating a small voltage between SRP and SRN where SRP is the top of the sense resistor.	
VSS	9	Р	Device ground	
TS1	10	IA	Temperature sensor 1 thermistor input pin	
TS2	11	IA	Temperature sensor 2 thermistor input pin	
TS3	12	IA	Temperature sensor 3 thermistor input pin	
TS4	13	IA	Temperature sensor 4 thermistor input pin	
NC	14	_	Not internally connected	

(1) P = Power Connection, O = Digital Output, AI = Analog Input, I = Digital Input, I/OD = Digital Input/Output

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Pin Functions (continued)

PIN NAME	NUMBER	TYPE ⁽¹⁾	DESCRIPTION
BTP_INT	15	0	Battery Trip Point (BTP) interrupt output
PRES or SHUTDN	16	I	Host system present input for removable battery pack or emergency system shutdown input for embedded pack
DISP	17	—	Display control for LEDs
SMBD	18	I/OD	SMBus data pin
SMBC	19	I/OD	SMBus clock pin
LEDCNTLA	20	_	LED display segment that drives the external LEDs depending on the firmware configuration
LEDCNTLB	21	_	LED display segment that drives the external LEDs depending on the firmware configuration
LEDCNTLC	22	_	LED display segment that drives the external LEDs depending on the firmware configuration
PTC	23	IA	Safety PTC thermistor input pin. To disable, connect both PTC and PTCEN to VSS.
PTCEN	24	IA	Safety PTC thermistor enable input pin. Connect to BAT. To disable, connect both PTC and PTCEN to VSS.
FUSE	25	0	Fuse drive output pin
VCC	26	Р	Secondary power supply input
PACK	27	IA	Pack sense input pin
DSG	28	0	NMOS Discharge FET drive output pin
NC	29	_	Not internally connected
PCHG	30	0	PMOS Precharge FET drive output pin
CHG	31	0	NMOS Charge FET drive output pin
BAT	32	Р	Primary power supply input pin



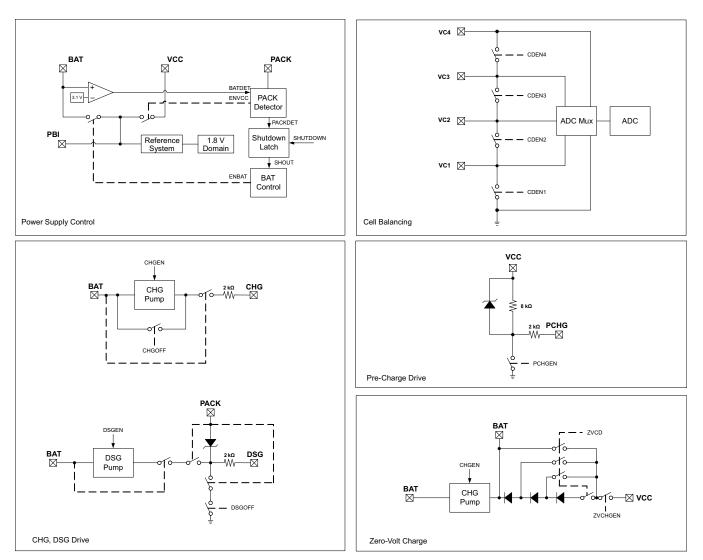


Figure 1. Pin Equivalent Diagram 1

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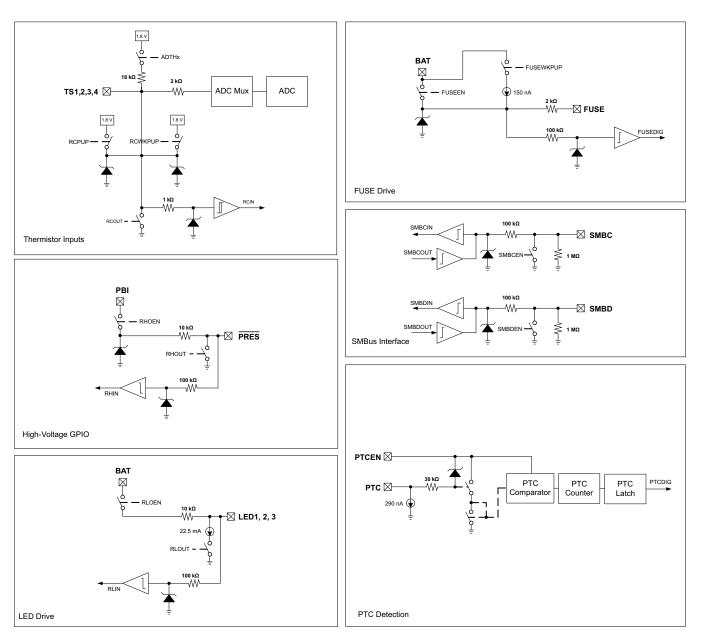


Figure 2. Pin Equivalent Diagram 2



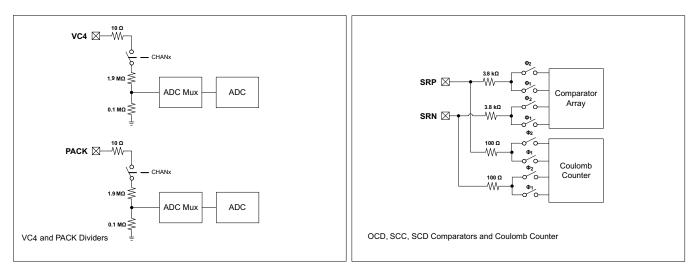


Figure 3. Pin Equivalent Diagram 3

7 Specifications

7.1 Absolute Maximum Ratings

Over-operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Supply voltage range, V_{CC}	BAT, VCC, PBI	-0.3	30	V
	PACK, SMBC, SMBD, PRES or SHUTDN, BTP_INT, DISP	-0.3	30	V
Input voltage range,	TS1, TS2, TS3, TS4	-0.3	V _{REG} + 0.3	V
	PTC, PTCEN, LEDCNTLA, LEDCNTLB, LEDCNTLC	-0.3	V _{BAT} + 0.3	V
	SRP, SRN	-0.3	0.3	V
	VC4	VC3 – 0.3	VC3 + 8.5, or VSS + 30	V
V _{IN}	VC3	VC2 – 0.3	30 30 V _{REG} + 0.3 V _{BAT} + 0.3 0.3 VC3 + 8.5, or	V
	VC2	VC1 – 0.3		V
	VC1	VSS – 0.3		$\begin{array}{c cccc} 0 & V \\ \hline 0 & V \\ + 0.3 & V \\ + 0.3 & V \\ \hline 3 & V \\ \hline 3 & V \\ \hline 8.5, or \\ + 30 \\ \hline 0 \\ $
Output voltage range,	CHG, DSG	-0.3	32	
Vo	PCHG, FUSE	$\begin{array}{c c c c c c c c c c c c c c c c c c c $	V	
Maximum VSS current,	I _{SS}		50	mA
Storage temperature, T	STG	-65	150	°C
Lead temperature (sold	ering, 10 s), T _{SOLDER}		300	°C

(1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

			VALUE	UNIT
	Flectrostatio	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 $^{\left(2\right) }$	±500	V

JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
 JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V _{CC}	Supply voltage	BAT, VCC, PBI	2.2		26	V
V _{SHUTDOWN-}	Shutdown voltage	V _{PACK} < V _{SHUTDOWN-}	1.8	2.0	2.2	V
V _{SHUTDOWN+}	Start-up voltage	V _{PACK} > V _{SHUTDOWN-} + V _{HYS}	2.05	2.25	2.45	V
V _{HYS}	Shutdown voltage hysteresis	V _{SHUTDOWN+} – V _{SHUTDOWN-}		250		mV
		PACK, SMBC, SMBD, PRES, BTP_IN, DISP			26	
		TS1, TS2, TS3, TS4			V_{REG}	
	Input voltage range	PTC, PTCEN, LEDCNTLA, LEDCNTLB, LEDCNTLC			V_{BAT}	
N		SRP, SRN	-0.2		0.2	V
V _{IN}		VC4	V _{VC3}		V _{VC3} + 5	v
		VC3	V _{VC2}		V _{VC2} + 5	
		VC2	V _{VC1}		V _{VC1} + 5	
		VC1	V _{VSS}		V _{VSS} + 5	
Vo	Output voltage range	CHG, DSG, PCHG, FUSE			26	V
C _{PBI}	External PBI capacitor		2.2			μF
T _{OPR}	Operating temperature		-40		85	°C

7.4 Thermal Information

	THERMAL METRIC ⁽¹⁾	RSM (QFN)	
		32 PINS	UNIT
R _{θJA, High K}	Junction-to-ambient thermal resistance	47.4	
R _{0JC(top)}	Junction-to-case(top) thermal resistance	40.3	
R _{θJB}	Junction-to-board thermal resistance	14.7	00AM
Ψ _{JT}	Junction-to-top characterization parameter	0.8	°C/W
Ψ _{ЈВ}	Junction-to-board characterization parameter	14.4	
R _{0JC(bottom)}	Junction-to-case(bottom) thermal resistance	3.8	

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

7.5 Electrical Characteristics: Supply Current

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 20 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
INORMAL	NORMAL mode	CHG on. DSG on, no Flash write		336		μA
		CHG off, DSG on, no SBS communication		75		۵
ISLEEP	SLEEP mode	CHG off, DSG off, no SBS communication		52		μA
I _{SHUTDOWN}	SHUTDOWN mode			1.6		μA

7.6 Electrical Characteristics: Power Supply Control

PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
V _{SWITCHOVER-} BAT to V _{CC} switchover voltage	V _{BAT} < V _{SWITCHOVER-}	1.95	2.1	2.2	V



Electrical Characteristics: Power Supply Control (continued)

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{SWITCHOVER+}	V _{CC} to BAT switchover voltage	$V_{BAT} > V_{SWITCHOVER-} + V_{HYS}$	2.9	3.1	3.25	V
V _{HYS}	Switchover voltage hysteresis	V _{SWITCHOVER+} – V _{SWITCHOVER-}		1000		mV
		BAT pin, BAT = 0 V, VCC = 25 V, PACK = 25 V			1	
luvo	Input Leakage	PACK pin, BAT = 25 V, VCC = 0 V, PACK = 0 V			1	μA
ILKG	current	BAT and PACK terminals, BAT = 0 V, VCC = 0 V, PACK = 0 V, PBI = 25 V			1	μ
R _{PD}	Internal pulldown resistance	PACK	30	40	50	kΩ

7.7 Electrical Characteristics: AFE Power-On Reset

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
V _{REGIT-}	Negative-going voltage input	V _{REG}	1.51	1.55	1.59	V
V _{HYS}	Power-on reset hysteresis	V _{REGIT+} – V _{REGIT-}	70	100	130	mV
t _{RST}	Power-on reset time		200	300	400	μs

7.8 Electrical Characteristics: AFE Watchdog Reset and Wake Timer

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		t _{WDT} = 500	372	500	628	ms
t _{WDT}	AFE watchdog	t _{WDT} = 1000	744	1000	1256	
	timeout	t _{WDT} = 2000	1488	2000	2512	
		t _{WDT} = 4000	2976	4000	5024	
		$t_{WAKE} = 250$	186	250	314	
+	AFE wake timer	$t_{WAKE} = 500$	372	500	00 628	
t _{WAKE}	AFE Wake unlei	t _{WAKE} = 1000	744	1000	1256	ms
		t _{WAKE} = 512	1488	2000	2512	
t _{FETOFF}	FET off delay after reset	$t_{FETOFF} = 512$	409	512	614	ms

7.9 Electrical Characteristics: Current Wake Comparator

PA	RAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
N	Wake voltage threshold	$V_{WAKE} = \pm 0.625 \text{ mV}$	±0.3	±0.625	±0.9	
		$V_{WAKE} = \pm 1.25 \text{ mV}$	±0.6	±1.25	±1.8	mV
V _{WAKE}		$V_{WAKE} = \pm 2.5 \text{ mV}$	±1.2	±2.5	±3.6	
		$V_{WAKE} = \pm 5 \text{ mV}$	±2.4	±5.0	±7.2	
V _{WAKE(DRIFT)}	Temperature drift of V _{WAKE} accuracy			0.5%		°C

EXAS

Electrical Characteristics: Current Wake Comparator (continued)

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

Р	ARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{WAKE}	Time from application of current to wake interrupt				700	μs
t _{WAKE(SU)}	Wake comparator startup time			500	1000	μs

7.10 Electrical Characteristics: VC1, VC2, VC3, VC4, BAT, PACK

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
		VC1-VSS, VC2-VC1, VC3-VC2, VC4-VC3	0.1980	0.2000	0.2020	
к	Scaling factor	BAT-VSS, PACK-VSS	0.049	0.050	0.051	—
		V _{REF2}	0.490	0.500	0.510	
V	Input voltago rango	VC1-VSS, VC2-VC1, VC3-VC2, VC4-VC3	-0.2		5	5 20 V
V _{IN}	Input voltage range	BAT-VSS, PACK-VSS	-0.2		20	
I _{LKG}	Input leakage current	VC1, VC2, VC3, VC4, cell balancing off, cell detach detection off, ADC multiplexer off			1	μA
R _{CB}	Internal cell balance resistance	$R_{\text{DS}(\text{ON})}$ for internal FET switch at 2 V < V_{\text{DS}} < 4 V			200	Ω
I _{CD}	Internal cell detach check current	VCx > VSS + 0.8 V	30	50	70	μA

7.11 Electrical Characteristics: SMBD, SMBC

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
VIH	Input voltage high	SMBC, SMBD, $V_{REG} = 1.8 V$	1.3			V
VIL	Input voltage low	SMBC, SMBD, $V_{REG} = 1.8 V$			0.8	V
V _{OL}	Output low voltage	SMBC, SMBD, V_{REG} = 1.8 V, I_{OL} = 1.5 mA			0.4	V
C _{IN}	Input capacitance			5		pF
I _{LKG}	Input leakage current				1	μA
R _{PD}	Pulldown resistance		0.7	1.0	1.3	MΩ

7.12 Electrical Characteristics: PRES, BTP_INT, DISP

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input		1.3			V
VIL	Low-level input				0.55	V
V		$V_{BAT} > 5.5 \text{ V}, I_{OH} = -0 \ \mu\text{A}$	3.5			V
V _{OH}	Output voltage high	$V_{BAT} > 5.5 \text{ V}, I_{OH} = -10 \ \mu\text{A}$	1.8			v
V _{OL}	Output voltage low	I _{OL} = 1.5 mA			0.4	V
C _{IN}	Input capacitance			5		pF
I _{LKG}	Input leakage current				1	μA
R _O	Output reverse resistance	Between \overline{PRES} or BTP_INT or \overline{DISP} and PBI	8			kΩ



7.13 Electrical Characteristics: LEDCNTLA, LEDCNTLB, LEDCNTLC

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IH}	High-level input		1.45			V
V _{IL}	Low-level input				0.55	V
V _{OH}	Output voltage high	V _{BAT} > 3.0 V, I _{OH} = -22.5 mA	V _{BAT} – 1.6			V
V _{OL}	Output voltage low	I _{OL} = 1.5 mA			0.4	V
I _{SC}	High level output current protection		-30	-45	-6 0	mA
I _{OL}	Low level output current	V _{BAT} > 3.0 V, V _{OH} = 0.4 V	15.75	22.5	29.25	mA
ILEDCNTLX	Current matching between LEDCNTLx	$V_{BAT} = V_{LEDCNTLx} + 2.5 V$		±1%		
CIN	Input capacitance			20		pF
I _{LKG}	Input leakage current				1	μA
f _{LEDCNTLx}	Frequency of LED pattern			124		Hz

7.14 Electrical Characteristics: Coulomb Counter

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage range		-0.1		0.1	V
Full scale range		-V _{REF1} /10		V _{REF1} /10	V
Integral nonlinearity ⁽¹⁾	16-bit, best fit over input voltage range		±5.2	±22.3	LSB
Offset error	16-bit, Post-calibration		±5	±10	μV
Offset error drift	15-bit + sign, Post-calibration		0.2	0.3	μV/°C
Gain error	15-bit + sign, over input voltage range		±0.2%	±0.8%	FSR
Gain error drift	15-bit + sign, over input voltage range			150	PPM/°C
Effective input resistance		2.5			MΩ

(1) 1 LSB = $V_{REF1}/(10 \times 2^N) = 1.215/(10 \times 2^{15}) = 3.71 \ \mu V$

7.15 Electrical Characteristics: CC Digital Filter

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	ΤΥΡ ΜΑΧ	UNIT
Conversion time	Single conversion		250	ms
Effective resolution	Single conversion	15		Bits

7.16 Electrical Characteristics: ADC

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
Input voltage range	Internal reference (V _{REF1})	-0.2	1	V
	External reference (V _{REG})	-0.2	0.8 x V _{REG}	v
Full scale range	$V_{FS} = V_{REF1}$ or V_{REG}	-V _{FS}	V _{FS}	V
late and a sull a sull (1)	16-bit, best fit, –0.1 V to 0.8 x V _{REF1}		±6.6	LSB
Integral nonlinearity ⁽¹⁾	16-bit, best fit, -0.2 V to -0.1 V		±13.1	LOD

(1) 1 LSB = $V_{REF1}/(2^N)$ = 1.225/(2¹⁵) = 37.4 µV (when t_{CONV} = 31.25 ms)



Electrical Characteristics: ADC (continued)

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT			
Offset error ⁽²⁾	16-bit, Post-calibration, $V_{FS} = V_{REF1}$		±67	±157	μV			
Offset error drift	16-bit, Post-calibration, $V_{FS} = V_{REF1}$		0.6	3	µV/°C			
Gain error	16-bit, -0.1 V to 0.8 x V _{FS}		±0.2%	±0.8%	FSR			
Gain error drift	16-bit, -0.1 V to 0.8 x V _{FS}			150	PPM/°C			
Effective input resistance		8			MΩ			

(2) For VC1–VSS, VC2–VC1, VC3–VC2, VC4–VC3, VC4–VSS, PACK–VSS, and V_{REF1}/2, the offset error is multiplied by (1/ADC multiplexer scaling factor (K)).

7.17 Electrical Characteristics: ADC Digital Filter

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Single conversion		31.25		
Conversion time	Single conversion		15.63		ms
	Single conversion		7.81		1115
	Single conversion		1.95		
Resolution	No missing codes	16			Bits
	With sign, t _{CONV} = 31.25 ms	14	15		
Effective resolution	With sign, t _{CONV} = 15.63 ms	13	14		Bits
Enective resolution	With sign, t _{CONV} = 7.81 ms	11	12		DIIS
	With sign, $t_{CONV} = 1.95$ ms	9	10		

7.18 Electrical Characteristics: CHG, DSG FET Drive

PA	RAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Output voltage	Ratio_{DSG} = (V_{DSG} - V_{BAT})/V_{BAT}, 2.2 V < V_{BAT} < 4.92 V, 10 M\Omega between PACK and DSG	2.133	2.333	2.433	
	ratio	Ratio _{CHG} = (V _{CHG} – V _{BAT})/V _{BAT} , 2.2 V < V _{BAT} < 4.92 V, 10 MΩ between BAT and CHG	2.133	2.333	2.433	—
V	Output voltage,	$V_{DSG(ON)}$ = V_{DSG} – $V_{BAT},$ V_{BAT} ≥ 4.92 V, 10 M Ω between PACK and DSG, V_{BAT} = 18 V	10.5	11.5	12	V
V _(FETON)	CHG and DSG on	$V_{CHG(ON)}$ = V_{CHG} – $V_{BAT},$ V_{BAT} ≥ 4.92 V, 10 M Ω between BAT and CHG, V_{BAT} = 18 V	10.5	11.5	12	v
V _(FETOFF)	Output voltage, CHG and DSG off	$V_{DSG(OFF)}$ = V_{DSG} – $V_{PACK},$ 10 M Ω between PACK and DSG	-0.4		0.4	0.4 0.4
(,		$V_{CHG(OFF)}$ = V_{CHG} – $V_{BAT},$ 10 $M\Omega$ between BAT and CHG	-0.4		0.4	
		V_{DSG} from 0% to 35% $V_{DSG(ON)(TYP)}$, $V_{BAT} ≥ 2.2$ V, $C_L = 4.7$ nF between DSG and PACK, 5.1 kΩ between DSG and C_L , 10 MΩ between PACK and DSG		200	500	
t _R	Rise time	V_{CHG} from 0% to 35% $V_{CHG(ON)(TYP)}$, $V_{BAT} ≥ 2.2$ V, $C_L = 4.7$ nF between CHG and BAT, 5.1 kΩ between CHG and C_L , 10 MΩ between BAT and CHG		200	500	μs
		V_{DSG} from $V_{DSG(ON)(TYP)}$ to 1 V, $V_{BAT} \ge 2.2$ V, $C_L = 4.7$ nF between DSG and PACK, 5.1 k Ω between DSG and C_L , 10 M Ω between PACK and DSG		40	300	
t _F	Fall time	V_{CHG} from $V_{CHG(ON)(TYP)}$ to 1 V, $V_{BAT} ≥ 2.2$ V, $C_L = 4.7$ nF between CHG and BAT, 5.1 kΩ between CHG and C_L , 10 MΩ between BAT and CHG		40	200	μs



7.19 Electrical Characteristics: PCHG FET Drive

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _(FETON)	Output voltage, PCHG on	$V_{PCHG(ON)}$ = VV_{CC} – $V_{PCHG},$ 10 $M\Omega$ between V_{CC} and PCHG	6	7	8	V
V _(FETOFF)	Output voltage, PCHG off	$V_{PCHG(OFF)}$ = VV_{CC} – $V_{PCHG},$ 10 $M\Omega$ between V_{CC} and PCHG	-0.4		0.4	V
t _R	Rise time	V_{PCHG} from 10% to 90% $V_{PCHG(ON)(TYP)}$, $VV_{CC} \ge 8$ V, C_L = 4.7 nF between PCHG and V_{CC} , 5.1 k Ω between PCHG and C_L , 10 M Ω between V_{CC} and CHG		40	200	μs
t _F	Fall time	V_{PCHG} from 90% to 10% $V_{PCHG(ON)(TYP)}$, $V_{CC} \ge 8$ V, C_L = 4.7 nF between PCHG and V_{CC} , 5.1 k Ω between PCHG and C_L , 10 M Ω between V_{CC} and CHG		40	200	μs

7.20 Electrical Characteristics: FUSE Drive

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PA	RAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
M	Output voltage	$V_{BAT} \ge 8 \text{ V}, \text{ C}_{L} = 1 \text{ nF}, \text{ I}_{AFEFUSE} = 0 \ \mu\text{A}$	6	7	8.65	N/
V _{OH}	high	V_{BAT} < 8 V, C_L = 1 nF, $I_{AFEFUSE}$ = 0 μ A	V _{BAT} – 0.1		V_{BAT}	V
V _{IH}	High-level input		1.5	2.0	2.5	V
I _{AFEFUSE(PU)}	Internal pullup current	V _{BAT} ≥ 8 V, V _{AFEFUSE} = VSS		150	330	nA
R _{AFEFUSE}	Output impedance		2	2.6	3.2	kΩ
C _{IN}	Input capacitance			5		pF
t _{DELAY}	Fuse trip detection delay		128		256	μs
t _{RISE}	Fuse output rise time	$V_{BAT} \ge 8 \text{ V}, \text{ C}_{L} = 1 \text{ nF}, \text{ V}_{OH} = 0 \text{ V to 5 V}$		5	20	μs

7.21 Electrical Characteristics: Internal Temperature Sensor

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Internal	Internal	V _{TEMPP}	-1.9	-2.0	-2.1	
V _{TEN}	MP sensor voltage drift	V _{TEMPP} - V _{TEMPN} , assured by design	0.177	0.178	0.179	mV/°C

7.22 Electrical Characteristics: TS1, TS2, TS3, TS4

PAR	AMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Input voltage	TS1, TS2, TS3, TS4, $V_{BIAS} = V_{REF1}$	-0.2		0.8 x V_{REF1}	V
V _{IN}	range	TS1, TS2, TS3, TS4, $V_{BIAS} = V_{REG}$	-0.2		$0.8 ext{ x V}_{\text{REG}}$	V
R _{NTC(PU)}	Internal pullup resistance	TS1, TS2, TS3, TS4	14.4	18	21.6	kΩ
R _{NTC(DRIFT)}	Resistance drift over temperature	TS1, TS2, TS3, TS4	-360	-280	-200	PPM/°C



7.23 Electrical Characteristics: PTC, PTCEN

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

PA	RAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
R _{PTC(TRIP)}	PTC trip resistance		1.2	2.5	3.95	MΩ
V _{PTC(TRIP)}	PTC trip voltage	$V_{PTC(TRIP)} = V_{PTCEN} - V_{PTC}$	200	500	890	mV
I _{PTC}	Internal PTC current bias	$T_A = -40^{\circ}C$ to $110^{\circ}C$	200	290	350	nA
t _{PTC(DELAY)}	PTC delay time	$T_A = -40^{\circ}C$ to 110°C	40	80	145	ms

7.24 Electrical Characteristics: Internal 1.8-V LDO

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

P	ARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{REG}	Regulator voltage		1.6	1.8	2.0	V
$\Delta V_{O(TEMP)}$	Regulator output over temperature	$\Delta V_{REG} / \Delta T_A$, I _{REG} = 10 mA		±0.25%		
$\Delta V_{O(LINE)}$	Line regulation	$\Delta V_{REG} / \Delta V_{BAT}$, V_{BAT} = 10 mA	-0 .6%		0.5%	
$\Delta V_{O(LOAD)}$	Load regulation	$\Delta V_{REG}/\Delta I_{REG}$, I_{REG} = 0 mA to 10 mA	-1.5%		1.5%	
I _{REG}	Regulator output current limit	$V_{REG} = 0.9 \text{ x } V_{REG(NOM)}, V_{IN} > 2.2 \text{ V}$	20			mA
I _{SC}	Regulator short- circuit current limit	$V_{REG} = 0 \times V_{REG(NOM)}$	25	40	55	mA
PSRR _{REG}	Power supply rejection ratio	$\Delta V_{BAT} / \Delta V_{REG}$, I _{REG} = 10 mA ,V _{IN} > 2.5 V, f = 10 Hz		40		dB
V _{SLEW}	Slew rate enhancement voltage threshold	V _{REG}	1.58	1.65		V

7.25 Electrical Characteristics: High-Frequency Oscillator

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
f _{HFO}	Operating frequency			16.78		MHz
f _{HFO(ERR)} Frequency error	$T_A = -20^{\circ}C$ to 70°C, includes frequency drift	-2.5%	±0.25%	2.5%		
	Frequency error	$T_A = -40^{\circ}C$ to 85°C, includes frequency drift	-3.5%	±0.25%	3.5%	
	Start-up time	$T_A = -20^{\circ}C$ to 85°C, oscillator frequency within +/-3% of nominal			4	ms
111 0(00)	·	oscillator frequency within +/-3% of nominal			100	μs

7.26 Electrical Characteristics: Low-Frequency Oscillator

Р	ARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
f _{LFO}	Operating frequency			262.144		kHz
		$T_A = -20^{\circ}C$ to 70°C, includes frequency drift	-1.5%	±0.25%	1.5%	
ILFO(ERR)	Frequency error	$T_A = -40^{\circ}C$ to 85°C, includes frequency drift	-2.5	±0.25	2.5	
f _{LFO(FAIL)}	Failure detection frequency		30	80	100	kHz



7.27 Electrical Characteristics: Voltage Reference 1

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

P	ARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{REF1}	Internal reference voltage	$T_A = 25^{\circ}C$, after trim	1.21	1.215	1.22	V
M	Internal reference ^{T)} voltage drift	$T_A = 0^{\circ}C$ to 60°C, after trim		±50		
V _{REF1} (DRIFT)		$T_A = -40^{\circ}C$ to 85°C, after trim		±80		PPM/°C

7.28 Electrical Characteristics: Voltage Reference 2

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

P	PARAMETER	TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
V _{REF2}	Internal reference voltage	$T_A = 25^{\circ}C$, after trim	1.22	1.225	1.23	V
V	Internal reference	$T_A = 0^{\circ}C$ to 60°C, after trim		±50		PPM/°C
VREF2(DRIFT)	voltage drift	$T_A = -40^{\circ}C$ to 85°C, after trim		±80		PPIM/ C

7.29 Electrical Characteristics: Instruction Flash

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

P	ARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Data retention		10			Years
	Flash programming write cycles		1000			Cycles
t _{PROGWORD}	Word programming time	$T_A = -40^{\circ}C$ to $85^{\circ}C$			40	μs
t _{MASSERASE}	Mass-erase time	$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$			40	ms
t _{PAGEERASE}	Page-erase time	$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$			40	ms
I _{FLASHREAD}	Flash-read current	$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$			2	mA
I _{FLASHWRITE}	Flash-write current	$T_A = -40^{\circ}C$ to $85^{\circ}C$			5	mA
I _{FLASHERASE}	Flash-erase current	$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$			15	mA

7.30 Electrical Characteristics: Data Flash

F	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Data retention		10			Years
	Flash programming write cycles		20000			Cycles
t _{PROGWORD}	Word programming time	$T_A = -40^{\circ}C$ to $85^{\circ}C$			40	μs
t _{MASSERASE}	Mass-erase time	$T_A = -40^{\circ}C$ to $85^{\circ}C$			40	ms
t _{PAGEERASE}	Page-erase time	$T_A = -40^{\circ}C$ to $85^{\circ}C$			40	ms
I _{FLASHREAD}	Flash-read current	$T_A = -40^{\circ}C$ to $85^{\circ}C$			1	mA
I _{FLASHWRITE}	Flash-write current	$T_A = -40^{\circ}C$ to $85^{\circ}C$			5	mA
I _{FLASHERASE}	Flash-erase current	$T_A = -40^{\circ}C$ to $85^{\circ}C$			15	mA

TEXAS INSTRUMENTS

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7.31 Electrical Characteristics: OCD, SCC, SCD1, SCD2 Current Protection Thresholds

Typical values stated where $T_A = 25^{\circ}C$ and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}C$ to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
N/	OCD detection	$V_{OCD} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1	-16.6		-100	m)/
V _{OCD} threshold voltage range	$V_{OCD} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0	-8.3		-50	mV	
A\/	OCD detection	$V_{OCD} = V_{SRP} - V_{SRN,}$ AFE PROTECTION CONTROL[RSNS] = 1		-5.56		mV
ΔV _{OCD} threshold voltage program step	$V_{OCD} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0		-2.78		mv	
V	SCC detection	$V_{SCC} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1	44.4		200	m)/
V _{SCC}	threshold voltage range	$V_{SCC} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0	22.2		100	mV
A) (SCC detection	$V_{SCC} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1		22.2		m)/
ΔV _{SCC}	threshold voltage program step	$V_{SCC} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0		11.1		mV
N/	SCD1 detection	$V_{SCD1} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1	-44.4		-200	
V _{SCD1}	threshold voltage range	$V_{SCD1} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0	-22.2		-100	mV
A) (SCD1 detection	$V_{SCD1} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1		-22.2		
∆V _{SCD1}	threshold voltage program step	$V_{SCD1} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0		-11.1		mV
	SCD2 detection	$V_{SCD2} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1	-44.4		-200	
V _{SCD2}	threshold voltage range	$V_{SCD2} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0	-22.2		-100	mV
SCD2 detection		$V_{SCD2} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 1		-22.2		m)/
ΔV _{SCD2}	threshold voltage program step	$V_{SCD2} = V_{SRP} - V_{SRN}$, AFE PROTECTION CONTROL[RSNS] = 0		-11.1		mV
V _{OFFSET}	OCD, SCC, and SCDx offset error	Post-trim	-2.5		2.5	mV
	OCD, SCC, and SCDx	No trim	-10%		10%	
V _{SCALE}	scale error	Post-trim	-5%		5%	_

7.32 Timing Requirements: OCD, SCC, SCD1, SCD2 Current Protection Timing

			MIN	NOM	MAX	UNIT
t _{OCD}	OCD detection delay time		1		31	ms
Δt _{OCD}	OCD detection delay time program step			2		ms
t _{SCC}	SCC detection delay time		0		915	μs
Δt _{SCC}	SCC detection delay time program step			61		μs
+	SCD1 detection	AFE PROTECTION CONTROL[SCDDx2] = 0	0		915	110
t _{SCD1}	delay time	AFE PROTECTION CONTROL[SCDDx2] = 1	0		1850	μs



Timing Requirements: OCD, SCC, SCD1, SCD2 Current Protection Timing (continued)

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

			MIN	NOM	MAX	UNIT	
	SCD1 detection	AFE PROTECTION CONTROL[SCDDx2] = 0		61			
∆t _{SCD1}	delay time program step	AFE PROTECTION CONTROL[SCDDx2] = 1		121		μs	
	SCD2 detection	AFE PROTECTION CONTROL[SCDDx2] = 0	0 458				
t _{SCD2}	delay time	AFE PROTECTION CONTROL[SCDDx2] = 1	0		915	915 µs	
	SCD2 detection	AFE PROTECTION CONTROL[SCDDx2] = 0		30.5			
∆t _{SCD2}	delay time program step	AFE PROTECTION CONTROL[SCDDx2] = 1		61		μs	
t _{DETECT}	Current fault detect time	V_{SRP} – V_{SRN} = V_{T} – 3 mV for OCD, SCD1, and SC2, V_{SRP} – V_{SRN} = V_{T} + 3 mV for SCC			160	μs	
t _{ACC}	Current fault delay time accuracy	Max delay setting	-10%		10%		

7.33 Timing Requirements: SMBus

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

			MIN	NOM	MAX	UNIT
f _{SMB}	SMBus operating frequency	SLAVE mode, SMBC 50% duty cycle	10		100	kHz
f _{MAS}	SMBus master clock frequency	MASTER mode, no clock low slave extend		51.2		kHz
t _{BUF}	Bus free time between start and stop		4.7			μs
t _{HD(START)}	Hold time after (repeated) start		4.0			μs
t _{SU(START)}	Repeated start setup time		4.7			μs
t _{SU(STOP)}	Stop setup time		4.0			μs
t _{HD(DATA)}	Data hold time		300			ns
t _{SU(DATA)}	Data setup time		250			ns
t _{TIMEOUT}	Error signal detect time		25		35	ms
t _{LOW}	Clock low period		4.7			μs
t _{HIGH}	Clock high period		4.0		50	μs
t _R	Clock rise time	10% to 90%			1000	ns
t _F	Clock fall time	90% to 10%			300	ns
t _{LOW(SEXT)}	Cumulative clock low slave extend time				25	ms
t _{LOW(MEXT)}	Cumulative clock low master extend time				10	ms

7.34 Timing Requirements: SMBus XL

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

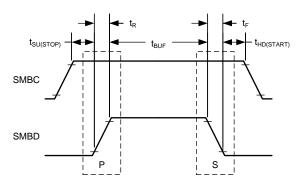
			MIN	NOM MAX	UNIT
f _{SMBXL}	SMBus XL operating frequency	SLAVE mode	40	400	kHz
t _{BUF}	Bus free time between start and stop		4.7		μs
t _{HD(START)}	Hold time after (repeated) start		4.0		μs

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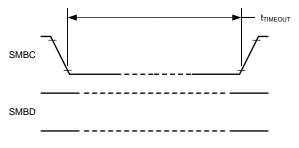
Timing Requirements: SMBus XL (continued)

Typical values stated where $T_A = 25^{\circ}$ C and VCC = 14.4 V, Min/Max values stated where $T_A = -40^{\circ}$ C to 85°C and VCC = 2.2 V to 26 V (unless otherwise noted)

		MIN	NOM	MAX	UNIT
t _{SU(START)}	Repeated start setup time	4.7			μs
t _{SU(STOP)}	Stop setup time	4.0			μs
t _{TIMEOUT}	Error signal detect time	5		20	ms
t _{LOW}	Clock low period			20	μs
t _{HIGH}	Clock high period			20	μs



Start and Stop Condition

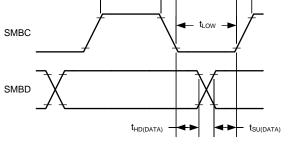


Timeout Condition

Figure 4. SMBus Timing Diagram



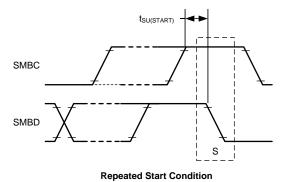




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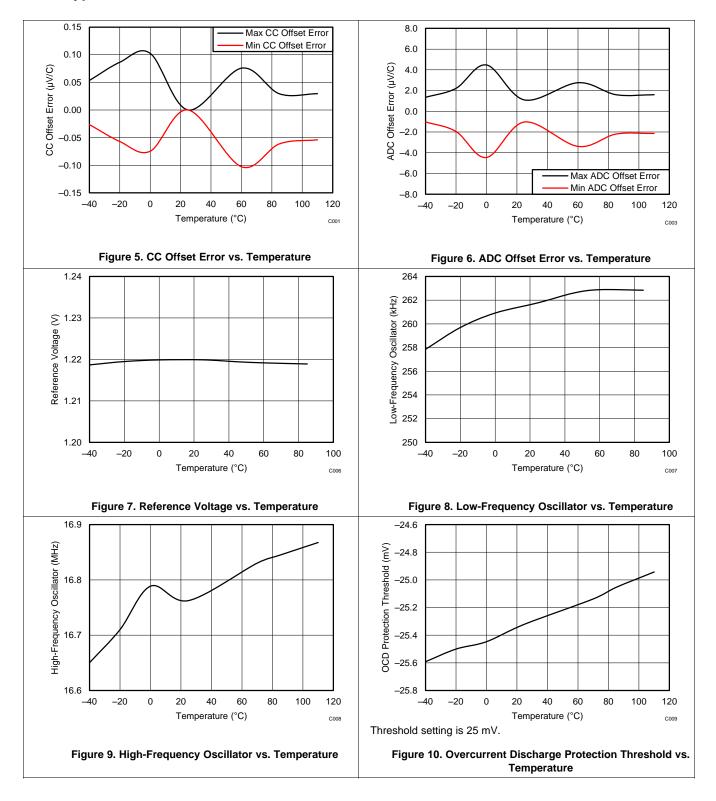
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Wait and Hold Condition



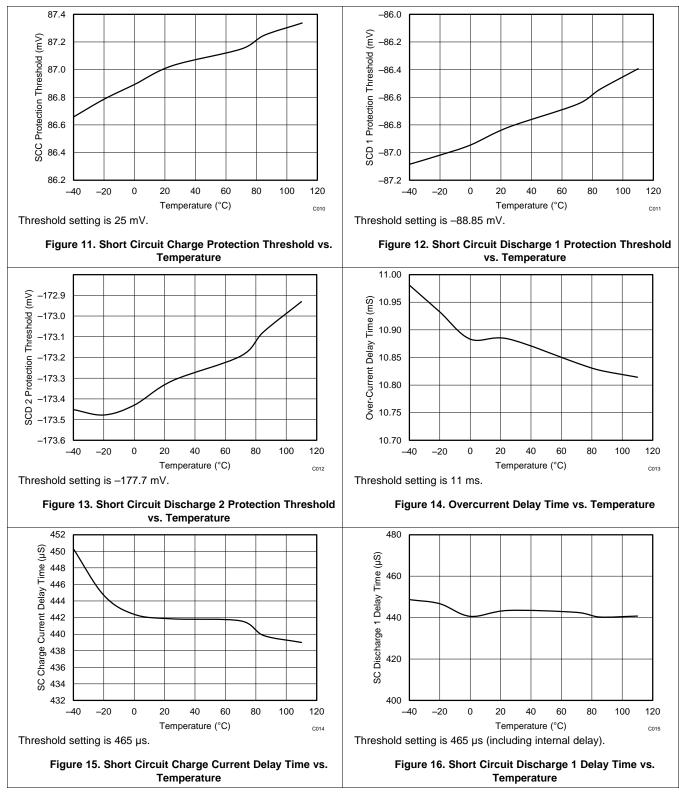


7.35 Typical Characteristics



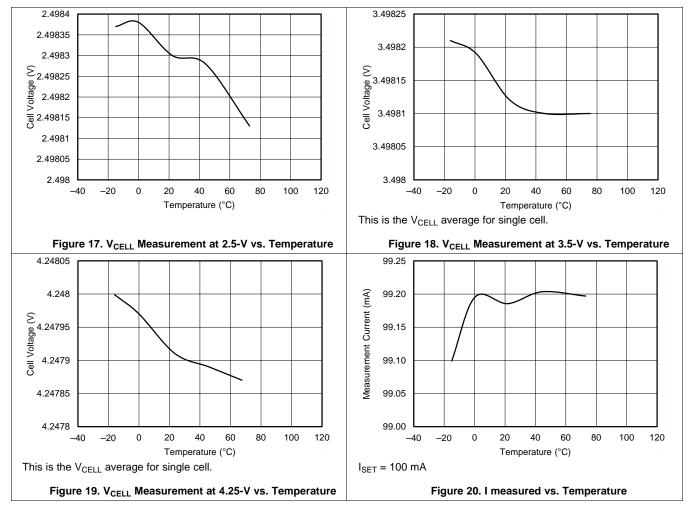


Typical Characteristics (continued)





Typical Characteristics (continued)



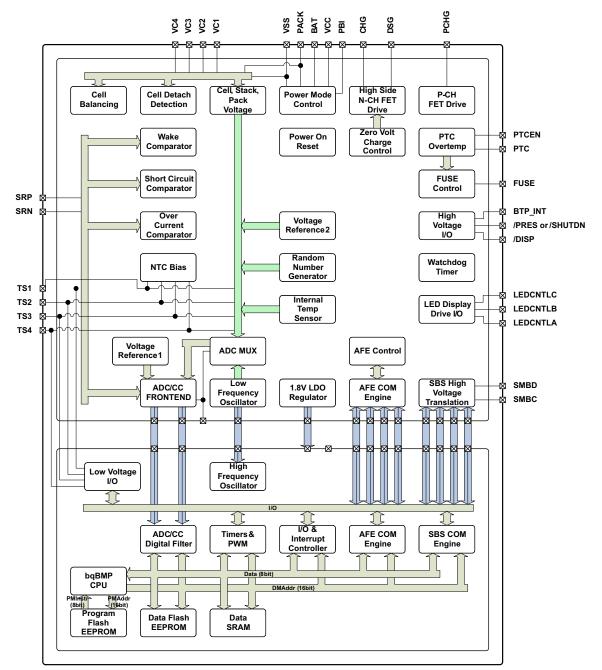


8 Detailed Description

8.1 Overview

The bq40z50-R1 device, incorporating patented Impedance Track[™] technology, provides cell balancing while charging or at rest. This fully integrated, single-chip, pack-based solution provides a rich array of features for gas gauging, protection, and authentication for 1-series, 2-series, 3-series, and 4-series cell Li-Ion and Li-Polymer battery packs, including a diagnostic lifetime data monitor and black box recorder.

8.2 Functional Block Diagram





8.3 Feature Description

8.3.1 Primary (1st Level) Safety Features

The bq40z50-R1 supports a wide range of battery and system protection features that can easily be configured. See the *bq40z50-R1 Technical Reference Manual* (SLUUBC1) for detailed descriptions of each protection function.

The primary safety features include:

- Cell Overvoltage Protection
- Cell Undervoltage Protection
- Cell Undervoltage Protection Compensated
- Overcurrent in Charge Protection
- Overcurrent in Discharge Protection
- Overload in Discharge Protection
- Short Circuit in Charge Protection
- Short Circuit in Discharge Protection
- Overtemperature in Charge Protection
- Overtemperature in Discharge Protection
- Undertemperature in Charge Protection
- Undertemperature in Discharge Protection
- Overtemperature FET protection
- Precharge Timeout Protection
- Host Watchdog Timeout Protection
- Fast Charge Timeout Protection
- Overcharge Protection
- Overcharging Voltage Protection
- Overcharging Current Protection
- Over Precharge Current Protection

8.3.2 Secondary (2nd Level) Safety Features

The secondary safety features of the bq40z50-R1 can be used to indicate more serious faults via the FUSE pin. This pin can be used to blow an in-line fuse to permanently disable the battery pack from charging or discharging. See the *bq40z50-R1 Technical Reference Manual* (SLUUBC1) for detailed descriptions of each protection function.

The secondary safety features provide protection against:

- Safety Overvoltage Permanent Failure
- Safety Undervoltage Permanent Failure
- Safety Overtemperature Permanent Failure
- Safety FET Overtemperature Permanent Failure
- Qmax Imbalance Permanent Failure
- Impedance Imbalance Permanent Failure
- Capacity Degradation Permanent Failure
- Cell Balancing Permanent Failure
- Fuse Failure Permanent Failure
- PTC Permanent Failure
- Voltage Imbalance at Rest Permanent Failure
- Voltage Imbalance Active Permanent Failure
- Charge FET Permanent Failure
- Discharge FET Permanent Failure
- AFE Register Permanent Failure
- AFE Communication Permanent Failure



Feature Description (continued)

- Second Level Protector Permanent Failure
- Instruction Flash Checksum Permanent Failure
- Open Cell Connection Permanent Failure
- Data Flash Permanent Failure
- Open Thermistor Permanent Failure

8.3.3 Charge Control Features

The bq40z50-R1 charge control features include:

- Supports JEITA temperature ranges. Reports charging voltage and charging current according to the active temperature range
- Handles more complex charging profiles. Allows for splitting the standard temperature range into two subranges and allows for varying the charging current according to the cell voltage
- Reports the appropriate charging current needed for constant current charging and the appropriate charging voltage needed for constant voltage charging to a smart charger using SMBus broadcasts
- Reduces the charge difference of the battery cells in fully charged state of the battery pack gradually using a
 voltage-based cell balancing algorithm during charging. A voltage threshold can be set up for cell balancing to
 be active. This prevents fully charged cells from overcharging and causing excessive degradation and also
 increases the usable pack energy by preventing premature charge termination.
- Supports pre-charging/zero-volt charging
- Supports charge inhibit and charge suspend if battery pack temperature is out of temperature range
- Reports charging fault and also indicates charge status via charge and discharge alarms

8.3.4 Gas Gauging

The bq40z50-R1 uses the Impedance Track algorithm to measure and calculate the available capacity in battery cells. The bq40z50-R1 accumulates a measure of charge and discharge currents and compensates the charge current measurement for the temperature and state-of-charge of the battery. The bq40z50-R1 estimates self-discharge of the battery and also adjusts the self-discharge estimation based on temperature. The device also has TURBO BOOST mode support, which enables the bq40z50-R1 to provide the necessary data for the MCU to determine what level of peak power consumption can be applied without causing a system reset or transient battery voltage level spike to trigger termination flags. See the *bq40z50-R1 Technical Reference Manual* (SLUUBC1) for further details.

8.3.5 Configuration

8.3.5.1 Oscillator Function

The bq40z50-R1 fully integrates the system oscillators and does not require any external components to support this feature.

8.3.5.2 System Present Operation

The bq40z50-R1 checks the PRES pin periodically (1 s). If PRES input is pulled to ground by the external system, the bq40z50-R1 detects this as system present.

8.3.5.3 Emergency Shutdown

For battery maintenance, the emergency shutdown feature enables a push button action connecting the SHUTDN pin to shutdown an embedded battery pack system before removing the battery. A high-to-low transition of the SHUTDN pin signals the bq40z50-R1 to turn off both CHG and DSG FETs, disconnecting the power from the system to safely remove the battery pack. The CHG and DSG FETs can be turned on again by another high-to-low transition detected by the SHUTDN pin or when a data flash configurable timeout is reached.

8.3.5.4 1-Series, 2-Series, 3-Series, or 4-Series Cell Configuration

In a 1-series cell configuration, VC4 is shorted to VC, VC2 and VC1. In a 2-series cell configuration, VC4 is shorted to VC3 and VC2. In a 3-series cell configuration, VC4 is shorted to VC3.



Feature Description (continued)

8.3.5.5 Cell Balancing

The device supports cell balancing by bypassing the current of each cell during charging or at rest. If the device's internal bypass is used, up to 10 mA can be bypassed and multiple cells can be bypassed at the same time. Higher cell balance current can be achieved by using an external cell balancing circuit. In external cell balancing mode, only one cell at a time can be balanced.

The cell balancing algorithm determines the amount of charge needed to be bypassed to balance the capacity of all cells.

8.3.6 Battery Parameter Measurements

8.3.6.1 Charge and Discharge Counting

The bq40z50-R1 uses an integrating delta-sigma analog-to-digital converter (ADC) for current measurement, and a second delta-sigma ADC for individual cell and battery voltage and temperature measurement.

The integrating delta-sigma ADC measures the charge/discharge flow of the battery by measuring the voltage drop across a small-value sense resistor between the SRP and SRN terminals. The integrating ADC measures bipolar signals from -0.1 V to 0.1 V. The bq40z50-R1 detects charge activity when $V_{SR} = V_{(SRP)} - V_{(SRN)}$ is positive, and discharge activity when $V_{SR} = V_{(SRP)} - V_{(SRN)}$ is negative. The bq40z50-R1 continuously integrates the signal over time, using an internal counter. The fundamental rate of the counter is 0.26 nVh.

8.3.7 Battery Trip Point (BTP)

Required for WIN8 OS, the battery trip point (BTP) feature indicates when the RSOC of a battery pack has depleted to a certain value set in a DF register. This feature allows a host to program two capacity-based thresholds that govern the triggering of a BTP interrupt on the BTP_INT pin and the setting or clearing of the *OperationStatus[BTP_INT]* on the basis of *RemainingCapacity()*.

An internal weak pull-up is applied when the BTP feature is active. Depending on the system design, an external pull-up may be required to put on the BTP_INT pin. See *Electrical Characteristics: PRES*, *BTP_INT*, *DISP* for details.

8.3.8 Lifetime Data Logging Features

The bq40z50-R1 offers lifetime data logging for several critical battery parameters. The following parameters are updated every 10 hours if a difference is detected between values in RAM and data flash:

- Maximum and Minimum Cell Voltages
- Maximum Delta Cell Voltage
- Maximum Charge Current
- Maximum Discharge Current
- Maximum Average Discharge Current
- Maximum Average Discharge Power
- Maximum and Minimum Cell Temperature
- Maximum Delta Cell Temperature
- Maximum and Minimum Internal Sensor Temperature
- Maximum FET Temperature
- Number of Safety Events Occurrences and the Last Cycle of the Occurrence
- Number of Valid Charge Termination and the Last Cycle of the Valid Charge Termination
- Number of Qmax and Ra Updates and the Last Cycle of the Qmax and Ra Updates
- Number of Shutdown Events
- Cell Balancing Time for Each Cell
 - (This data is updated every 2 hours if a difference is detected.)
- Total FW Runtime and Time Spent in Each Temperature Range (This data is updated every 2 hours if a difference is detected.)



Feature Description (continued)

8.3.9 Authentication

The bq40z50-R1 supports authentication by the host using SHA-1.

8.3.10 LED Display

The bq40z50-R1 can drive a 3-, 4-, or 5- segment LED display for remaining capacity indication and/or a permanent fail (PF) error code indication.

8.3.11 Voltage

The bq40z50-R1 updates the individual series cell voltages at 0.25-second intervals. The internal ADC of the bq40z50-R1 measures the voltage, and scales and calibrates it appropriately. This data is also used to calculate the impedance of the cell for the Impedance Track gas gauging.

8.3.12 Current

The bq40z50-R1 uses the SRP and SRN inputs to measure and calculate the battery charge and discharge current using a 1-m Ω to 3-m Ω typ. sense resistor.

8.3.13 Temperature

The bq40z50-R1 has an internal temperature sensor and inputs for four external temperature sensors. All five temperature sensor options can be individually enabled and configured for cell or FET temperature usage. Two configurable thermistor models are provided to allow the monitoring of cell temperature in addition to FET temperature, which use a different thermistor profile.

8.3.14 Communications

The bq40z50-R1 uses SMBus v1.1 with MASTER mode and packet error checking (PEC) options per the SBS specification.

8.3.14.1 SMBus On and Off State

The bq40z50-R1 detects an SMBus off state when SMBC and SMBD are low for two or more seconds. Clearing this state requires that either SMBC or SMBD transition high. The communication bus will resume activity within 1 ms.

8.3.14.2 SBS Commands

See the *bq40z50-R1 Technical Reference Manual* (SLUUBC1) for further details.

8.4 Device Functional Modes

The bq40z50-R1 supports three power modes to reduce power consumption:

- In NORMAL mode, the bq40z50-R1 performs measurements, calculations, protection decisions, and data updates in 250-ms intervals. Between these intervals, the bq40z50-R1 is in a reduced power stage.
- In SLEEP mode, the bq40z50-R1 performs measurements, calculations, protection decisions, and data updates in adjustable time intervals. Between these intervals, the bq40z50-R1 is in a reduced power stage. The bq40z50-R1 has a wake function that enables exit from SLEEP mode when current flow or failure is detected.
- In SHUTDOWN mode, the bq40z50-R1 is completely disabled.



9 Applications and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The bq40z50-R1 is a gas gauge with primary protection support, and that can be used with a 1-series to 4-series Li-Ion/Li Polymer battery pack. To implement and design a comprehensive set of parameters for a specific battery pack, users need the Battery Management Studio (bqSTUDIO) graphical user-interface tool installed on a PC during development. The firmware installed on the bqSTUDIO tool has default values for this product, which are summarized in the *bq40z50-R1 Technical Reference Manual* (SLUUBC1). Using the bqSTUDIO tool, these default values can be changed to cater to specific application requirements during development once the system parameters, such as fault trigger thresholds for protection, enable/disable of certain features for operation, configuration of cells, chemistry that best matches the cell used, and more are known. This data is referred to as the "golden image."



9.2 Typical Applications

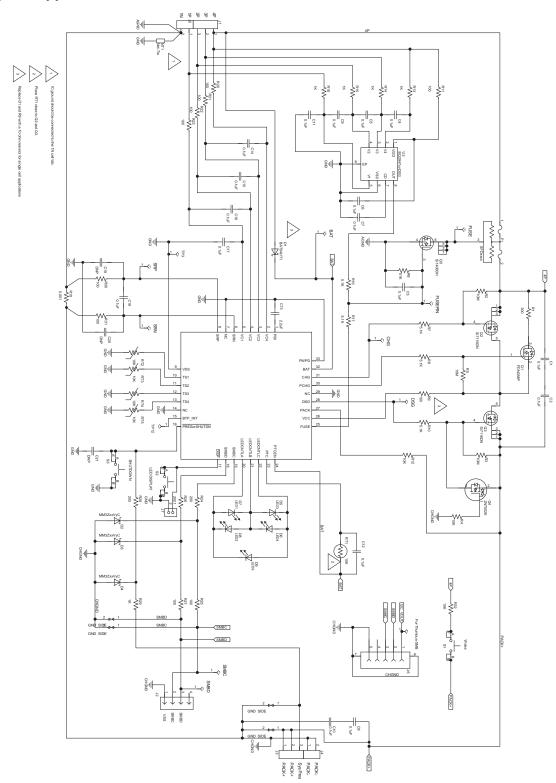


Figure 21. Application Schematic



Typical Applications (continued)

9.2.1 Design Requirements

Table 1 shows the default settings for the main parameters. Use the bqSTUDIO tool to update the settings to meet the specific application or battery pack configuration requirements.

The device should be calibrated before any gauging test. Follow the bqSTUDIO **Calibration** page to calibrate the device, and use the bqSTUDIO **Chemistry** page to update the match chemistry profile to the device.

DESIGN PARAMETER	EXAMPLE
Cell Configuration	3s1p (3-series with 1 Parallel) ⁽¹⁾
Design Capacity	4400 mAh
Device Chemistry	1210 (LiCoO2/graphitized carbon)
Cell Overvoltage at Standard Temperature	4300 mV
Cell Undervoltage	2500 mV
Shutdown Voltage	2300 mV
Overcurrent in CHARGE Mode	6000 mA
Overcurrent in DISCHARGE Mode	–6000 mA
Short Circuit in CHARGE Mode	0.1 V/Rsense across SRP, SRN
Short Circuit in DISCHARGE Mode	0.1 V/Rsense across SRP, SRN
Safety Overvoltage	4500 mV
Cell Balancing	Disabled
Internal and External Temperature Sensor	External Temperature Sensor is used.
Undertemperature Charging	0°C
Undertemperature Discharging	0°C
BROADCAST Mode	Disabled
Battery Trip Point (BTP) with active high interrupt	Disabled

Table 1. Design Parameters

(1) When using the device the first time, if the a 1-s or 2-s battery pack is used, then a charger or power supply should be connected to the PACK+ terminal to prevent device shutdown. Then update the cell configuration (see the bq40z50-R1 Technical Reference Manual [SLUUBC1] for details) before removing the charger connection.

9.2.2 Detailed Design Procedure

9.2.2.1 High-Current Path

The high-current path begins at the PACK+ terminal of the battery pack. As charge current travels through the pack, it finds its way through protection FETs, a chemical fuse, the lithium-ion cells and cell connections, and the sense resistor, and then returns to the PACK– terminal (see Figure 22). In addition, some components are placed across the PACK+ and PACK– terminals to reduce effects from electrostatic discharge.

9.2.2.1.1 Protection FETs

Select the N-channel charge and discharge FETs for a given application. Most portable battery applications are a good match for the CSD17308Q3. The TI CSD17308Q3 is a 47A, 30-V device with Rds(on) of 8.2 m Ω when the gate drive voltage is 8 V.

If a precharge FET is used, R1 is calculated to limit the precharge current to the desired rate. Be sure to account for the power dissipation of the series resistor. The precharge current is limited to $(V_{CHARGER} - V_{BAT})/R1$ and maximum power dissipation is $(V_{charger} - V_{bat})^2/R1$.

The gates of all protection FETs are pulled to the source with a high-value resistor between the gate and source to ensure they are turned off if the gate drive is open.

Capacitors C1 and C2 help protect the FETs during an ESD event. Using two devices ensures normal operation if one becomes shorted. To have good ESD protection, the copper trace inductance of the capacitor leads must be designed to be as short and wide as possible. Ensure that the voltage rating of both C1 and C2 are adequate to hold off the applied voltage if one of the capacitors becomes shorted.

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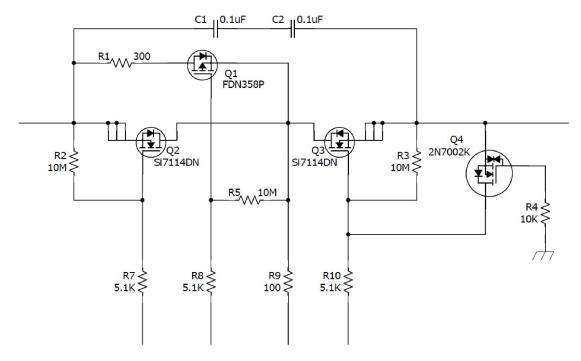


Figure 22. bq40z50-R1 Protection FETs

9.2.2.1.2 Chemical Fuse

The chemical fuse (Dexerials, Uchihashi, and so forth) is ignited under command from either the bq294700 secondary voltage protection IC or from the FUSE pin of the gas gauge. Either of these events applies a positive voltage to the gate of Q5, shown in Figure 23, which then sinks current from the third terminal of the fuse, causing it to ignite and open permanently.

It is important to carefully review the fuse specifications and match the required ignition current to that available from the N-channel FET. Ensure that the proper voltage, current, and Rds(on) ratings are used for this device. The fuse control circuit is discussed in detail in *FUSE Circuitry*.



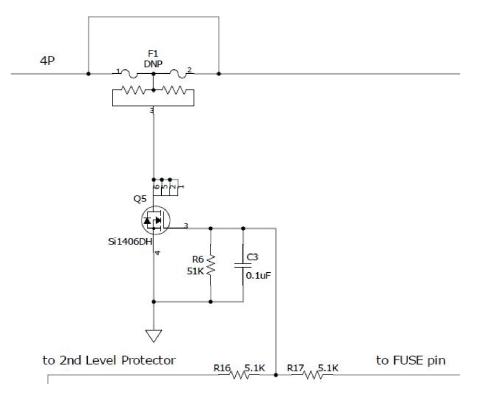


Figure 23. FUSE Circuit

9.2.2.1.3 Lithium-Ion Cell Connections

The important part to remember about the cell connections is that high current flows through the top and bottom connections; therefore, the voltage sense leads at these points must be made with a Kelvin connection to avoid any errors due to a drop in the high-current copper trace. The location marked 4P in Figure 24 indicates the Kelvin connection of the most positive battery node. The connection marked 1N is equally important. The VC5 pin (a ground reference for cell voltage measurement), which is in the older generation devices, is not in the bq40z50-R1 device. Therefore, the single-point connection at 1N to the low-current ground is needed to avoid an undesired voltage drop through long traces while the gas gauge is measuring the bottom cell voltage.

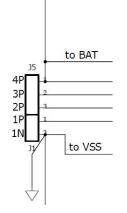


Figure 24. Lithium-Ion Cell Connections



9.2.2.1.4 Sense Resistor

As with the cell connections, the quality of the Kelvin connections at the sense resistor is critical. The sense resistor must have a temperature coefficient no greater than 50 ppm in order to minimize current measurement drift with temperature. Choose the value of the sense resistor to correspond to the available overcurrent and short-circuit ranges of the bq40z50-R1. Select the smallest value possible to minimize the negative voltage generated on the bq40z50-R1 V_{SS} node(s) during a short circuit. This pin has an absolute minimum of –0.3 V. Parallel resistors can be used as long as good Kelvin sensing is ensured. The device is designed to support a 1-m Ω to 3-m Ω sense resistor.

The ground scheme of bq40z50-R1 is different from the older generation devices. In previous devices, the device ground (or low current ground) is connected to the SRN side of the Rsense resistor pad. The bq40z50-R1, however, connects the low-current ground on the SRP side of the Rsense resistor pad, close to the battery 1N terminal (see *Lithium-Ion Cell Connections*). This is because the bq40z50-R1 has one less VC pin (a ground reference pin VC5) compared to the previous devices. The pin was removed and was internally combined to SRP.

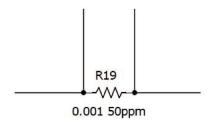


Figure 25. Sense Resistor

9.2.2.1.5 ESD Mitigation

A pair of series 0.1-µF ceramic capacitors is placed across the PACK+ and PACK– terminals to help in the mitigation of external electrostatic discharges. The two devices in series ensure continued operation of the pack if one of the capacitors becomes shorted.

Optionally, a tranzorb such as the SMBJ2A can be placed across the terminals to further improve ESD immunity.

9.2.2.2 Gas Gauge Circuit

The Gas Gauge Circuit includes the bq40z50-R1 and its peripheral components. These components are divided into the following groups: Differential Low-Pass Filter, PBI, System Present, SMBus Communication, FUSE circuit, and LED.

9.2.2.2.1 Coulomb-Counting Interface

The bq40z50-R1 uses an integrating delta-sigma ADC for current measurements. Add a $100-\Omega$ resistor from the sense resistor to the SRP and SRN inputs of the device. Place a $0.1-\mu$ F (C18) filter capacitor across the SRP and SRN inputs. Optional $0.1-\mu$ F filter capacitors (C19 and C20) can be added for additional noise filtering, if required for your circuit.



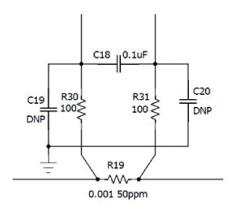


Figure 26. Differential Filter

9.2.2.2.2 Power Supply Decoupling and PBI

The bq40z50-R1 has an internal LDO that is internally compensated and does not require an external decoupling capacitor.

The PBI pin is used as a power supply backup input pin providing power during brief transient power outages. A standard 2.2-µF ceramic capacitor is connected from the PBI pin to ground as shown in Figure 27.

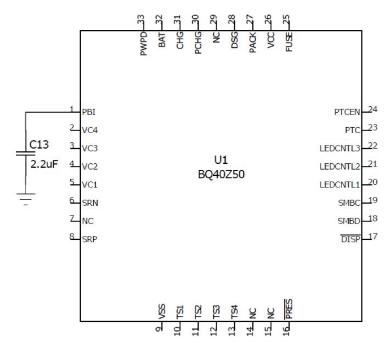


Figure 27. Power Supply Decoupling

9.2.2.2.3 System Present

The System Present signal is used to inform the gas gauge whether the pack is installed into or removed from the system. In the host system, this pin is grounded. The PRES pin of the bq40z50-R1 is occasionally sampled to test for system present. To save power, an internal pullup is provided by the gas gauge during a brief 4- μ s sampling pulse once per second. A resistor can be used to pull the signal low and the resistance must be 20 k Ω or lower to insure that the test pulse is lower than the VIL limit. The pull-up current source is typically 10 μ A to 20 μ A.

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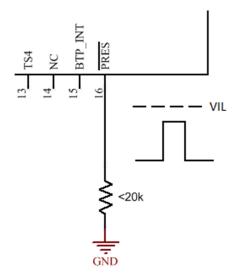
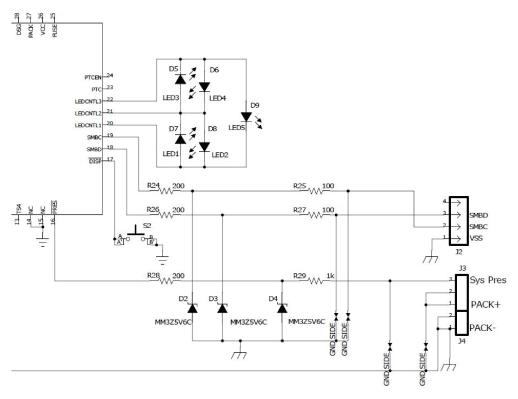


Figure 28. System Present Pull-Down Resistor

Because the System Present signal is part of the pack connector interface to the outside world, it must be protected from external electrostatic discharge events. An integrated ESD protection on the PRES device pin reduces the external protection requirement to just R29 for an 8-kV ESD contact rating. However, if it is possible that the System Present signal may short to PACK+, then R28 and D4 must be included for high-voltage protection.





9.2.2.2.4 SMBus Communication

The SMBus clock and data pins have integrated high-voltage ESD protection circuits, however, adding a Zener diode (D2 and D3) and series resistor (R24 and R26) provides more robust ESD performance.



The SMBus clock and data lines have internal pulldown. When the gas gauge senses that both lines are low (such as during removal of the pack), the device performs auto-offset calibration and then goes into SLEEP mode to conserve power.

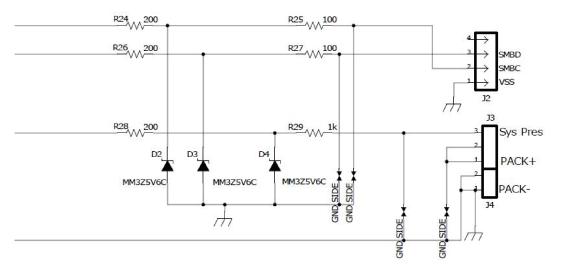


Figure 30. ESD Protection for SMB Communication

9.2.2.2.5 FUSE Circuitry

The FUSE pin of the bq40z50-R1 is designed to ignite the chemical fuse if one of the various safety criteria is violated. The FUSE pin also monitors the state of the secondary-voltage protection IC. Q5 ignites the chemical fuse when its gate is high. The 7-V output of the bq294700 is divided by R16 and R6, which provides adequate gate drive for Q5 while guarding against excessive back current into the bq294700 if the FUSE signal is high.

Using C3 is generally a good practice, especially for RFI immunity. C3 may be removed, if desired, because the chemical fuse is a comparatively slow device and is not affected by any sub-microsecond glitches that come from the FUSE output during the cell connection process.

TEXAS INSTRUMENTS

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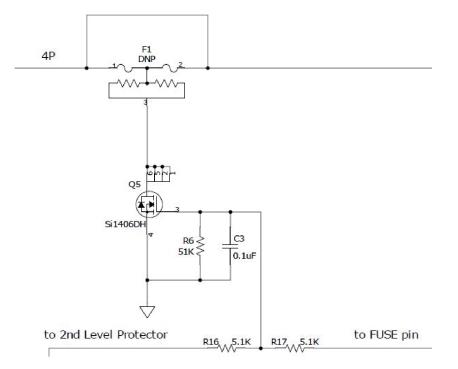


Figure 31. FUSE Circuit

When the bq40z50-R1 is commanded to ignite the chemical fuse, the FUSE pin activates to give a typical 8-V output. The new design makes it possible to use a higher Vgs FET for Q5. This improves the robustness of the system, as well as widens the choices for Q5.

9.2.2.3 Secondary-Current Protection

The bq40z50-R1 provides secondary overcurrent and short-circuit protection, cell balancing, cell voltage multiplexing, and voltage translation. The following discussion examines Cell and Battery Inputs, Pack and FET Control, Temperature Output, and Cell Balancing.

9.2.2.3.1 Cell and Battery Inputs

Each cell input is conditioned with a simple RC filter, which provides ESD protection during cell connect and acts to filter unwanted voltage transients. The resistor value allows some trade-off for cell balancing versus safety protection.

The integrated cell balancing FETs allow the AFE to bypass cell current around a given cell or numerous cells, effectively balancing the entire battery stack. External series resistors placed between the cell connections and the VCx I/O pins set the balancing current magnitude. The internal FETs provide a 200- Ω resistance (2 V < VDS < 4 V). Series input resistors between 100 Ω and 1 k Ω are recommended for effective cell balancing.

The BAT input uses a diode (D1) to isolate and decouple it from the cells in the event of a transient dip in voltage caused by a short-circuit event.

Also, as described in *High-Current Path*, the top and bottom nodes of the cells must be sensed at the battery connections with a Kelvin connection to prevent voltage sensing errors caused by a drop in the high-current PCB copper.



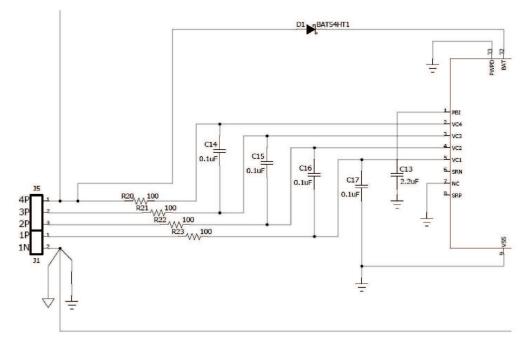


Figure 32. Cell and BAT Inputs

9.2.2.3.2 External Cell Balancing

Internal cell balancing can only support up to 10 mA. External cell balancing provide as another option for faster cell balancing. For details, refer to the application note, *Fast Cell Balancing Using External MOSFET* (SLUA420).

9.2.2.3.3 PACK and FET Control

The PACK and V_{CC} inputs provide power to the bq40z50-R1 from the charger. The PACK input also provides a method to measure and detect the presence of a charger. The PACK input uses a 100- Ω resistor; whereas, the V_{CC} input uses a diode to guard against input transients and prevents mis-operation of the date driver during short-circuit events.

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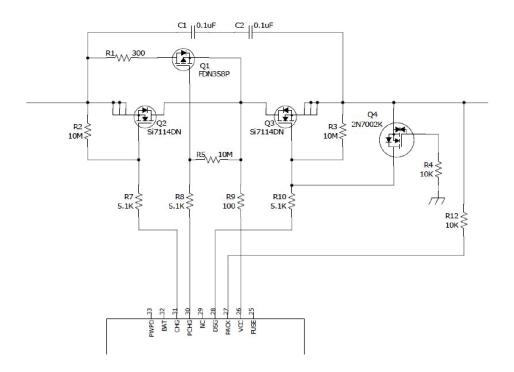


Figure 33. bq40z50-R1 PACK and FET Control

The N-channel charge and discharge FETs are controlled with $5.1-k\Omega$ series gate resistors, which provide a switching time constant of a few microseconds. The $10-M\Omega$ resistors ensure that the FETs are off in the event of an open connection to the FET drivers. Q4 is provided to protect the discharge FET (Q3) in the event of a reverse-connected charger. Without Q4, Q3 can be driven into its linear region and suffer severe damage if the PACK+ input becomes slightly negative.

Q4 turns on in that case to protect Q3 by shorting its gate to source. To use the simple ground gate circuit, the FET must have a low gate turn-on threshold. If it is desired to use a more standard device, such as the 2N7002 as the reference schematic, the gate should be biased up to 3.3 V with a high-value resistor. The bq40z50-R1 device has the capability to provide a current-limited charging path typically used for low battery voltage or low temperature charging. The bq40z50-R1 device uses an external P-channel, pre-charge FET controlled by PCHG.

9.2.2.3.4 Temperature Output

For the bq40z50-R1 device, TS1, TS2, TS3, and TS4 provide thermistor drive-under program control. Each pin can be enabled with an integrated 18-k Ω (typical) linearization pullup resistor to support the use of a 10-k Ω at 25°C (103) NTC external thermistor such as a Mitsubishi BN35-3H103. The reference design includes four 10-k Ω thermistors: RT1, RT2, RT3, and RT4. The bq40z50-R1 device supports up to four external thermistors. Connect unused thermistor pins to V_{SS}.



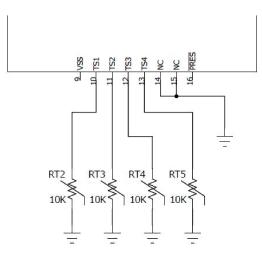
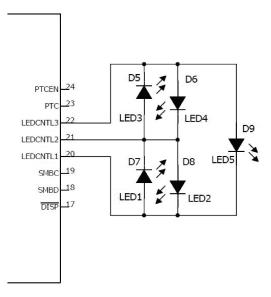
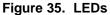


Figure 34. Thermistor Drive

9.2.2.3.5 LEDs

Three LED control outputs provide constant current sinks for the driving external LEDs. These outputs are configured to provide voltage and control for up to 5 LEDs. No external bias voltage is required. Unused LEDCNTL pins can remain open or they can be connected to V_{SS} . The DISP pin should be connected to V_{SS} , if the LED feature is not used.





9.2.2.3.6 Safety PTC Thermistor

The bq40z50-R1 device provides support for a safety PTC thermistor. The PTC thermistor is connected between the PTC pin and V_{SS} . It can be placed close to the CHG/DSG FETs to monitor the temperature. The PTC pin outputs a very small current, typical ~370 nA, and the PTC fault will be triggered at ~0.7 V typical. A PTC fault is one of the permanent failure modes. It can only be cleared by a POR.

To disable this feature, connect a 10-k Ω resistor between PTC and V_{SS}.



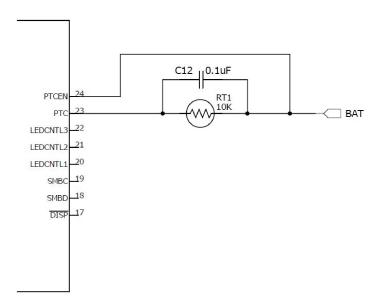
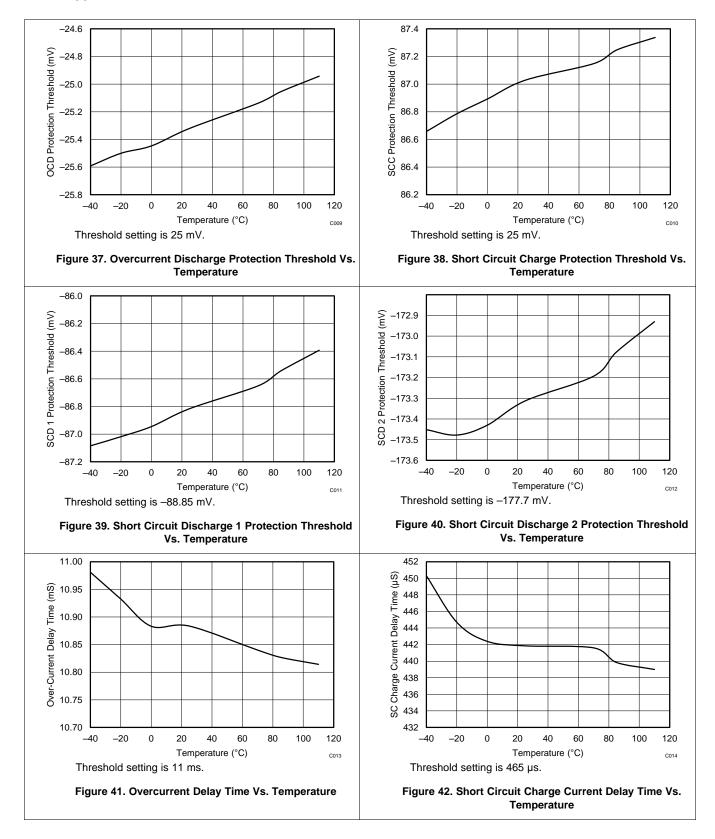


Figure 36. PTC Thermistor



9.2.3 Application Curves





10 Power Supply Recommendations

The device manages its supply voltage dynamically according to the operation conditions. Normally, the BAT input is the primary power source to the device. The BAT pin should be connected to the positive termination of the battery stack. The input voltage for the BAT pin ranges from 2.2 V to 26 V.

The VCC pin is the secondary power input, which activates when the BAT voltage falls below minimum Vcc. This allows the device to source power from a charger (if present) connected to the PACK pin. The VCC pin should be connected to the common drain of the CHG and DSG FETs. The charger input should be connected to the PACK pin.



11 Layout

11.1 Layout Guidelines

A battery fuel gauge circuit board is a challenging environment due to the fundamental incompatibility of highcurrent traces and ultra-low current semiconductor devices. The best way to protect against unwanted trace-totrace coupling is with a component placement, such as that shown in Figure 43, where the high-current section is on the opposite side of the board from the electronic devices. Clearly this is not possible in many situations due to mechanical constraints. Still, every attempt should be made to route high-current traces away from signal traces, which enter the bq40z50-R1 directly. IC references and registers can be disturbed and in rare cases damaged due to magnetic and capacitive coupling from the high-current path. Note that during surge current and ESD events, the high-current traces appear inductive and can couple unwanted noise into sensitive nodes of the gas gauge electronics, as illustrated in Figure 44.

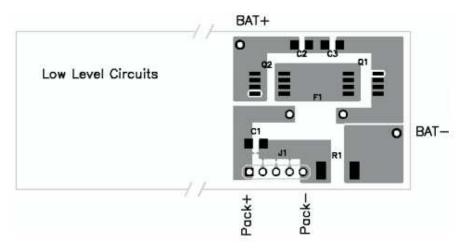


Figure 43. Separating High- and Low-Current Sections Provides an Advantage in Noise Immunity

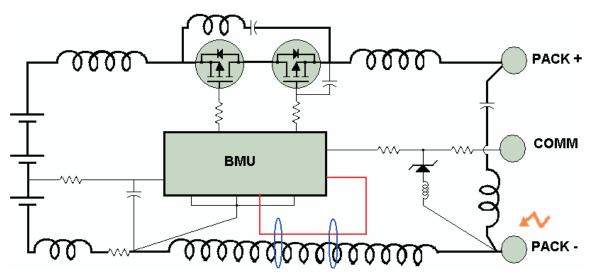


Figure 44. Avoid Close Spacing Between High-Current and Low-Level Signal Lines

Kelvin voltage sensing is extremely important in order to accurately measure current and top and bottom cell voltages. Place all filter components as close as possible to the device. Route the traces from the sense resistor in parallel to the filter circuit. Adding a ground plane around the filter network can add additional noise immunity. Figure 45 and Figure 46 demonstrates correct kelvin current sensing.

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Layout Guidelines (continued)

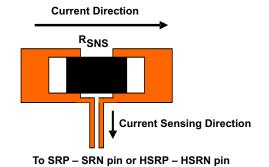


Figure 45. Sensing Resistor PCB Layout

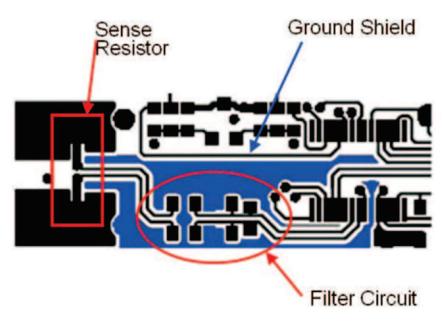
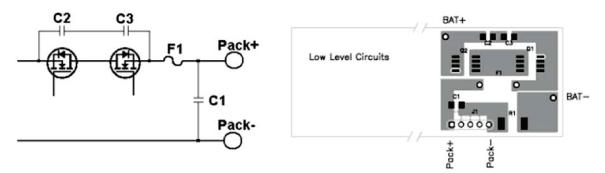
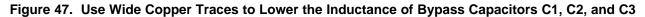


Figure 46. Sense Resistor, Ground Shield, and Filter Circuit Layout

11.1.1 Protector FET Bypass and Pack Terminal Bypass Capacitors

The general principle is to use wide copper traces to lower the inductance of the bypass capacitor circuit. In Figure 47, an example layout demonstrates this technique.







Layout Guidelines (continued)

11.1.2 ESD Spark Gap

Protect SMBus Clock, Data, and other communication lines from ESD with a spark gap at the connector. The pattern in Figure 48 recommended, with 0.2-mm spacing between the points.

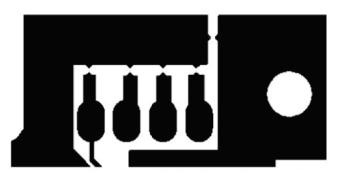


Figure 48. Recommended Spark-Gap Pattern Helps Protect Communication Lines from ESD

11.2 Layout Example

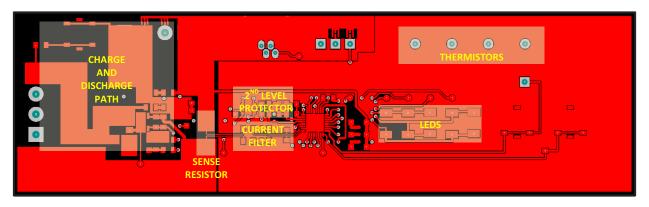
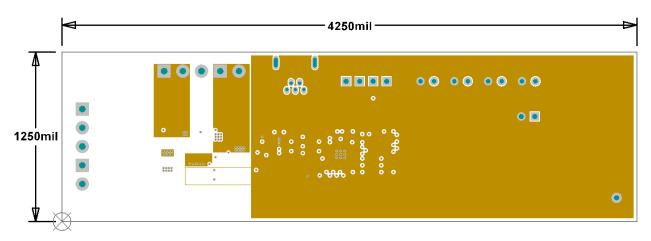


Figure 49. Top Layer





TEXAS INSTRUMENTS

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Layout Example (continued)

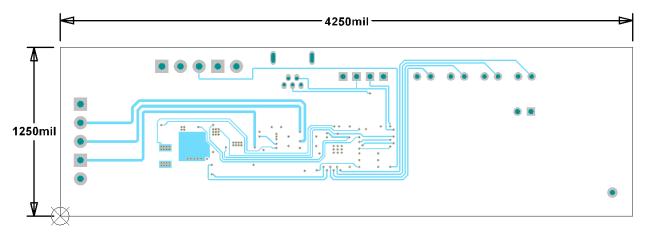


Figure 51. Internal Layer 2

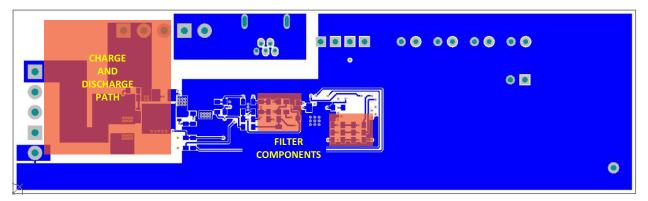


Figure 52. Bottom Layer



12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For related documentation, see the bq40z50-R1 Technical Reference Manual (SLUUBC1).

12.2 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E[™] Online Community *TI's Engineer-to-Engineer (E2E) Community.* Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support TI's Design Support Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.3 Trademarks

Impedance Track, E2E are trademarks of Texas Instruments. All other trademarks are the property of their respective owners.

12.4 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

E SI

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.5 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
BQ40Z50RSMR-R1	Active	Production	VQFN (RSM) 32	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMR-R1.A	Active	Production	VQFN (RSM) 32	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMR-R1.B	Active	Production	VQFN (RSM) 32	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMR-R1G4.A	Active	Production	VQFN (RSM) 32	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMR-R1G4.B	Active	Production	VQFN (RSM) 32	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMT-R1	Active	Production	VQFN (RSM) 32	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMT-R1.A	Active	Production	VQFN (RSM) 32	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50
BQ40Z50RSMT-R1.B	Active	Production	VQFN (RSM) 32	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ40Z50

⁽¹⁾ Status: For more details on status, see our product life cycle.

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

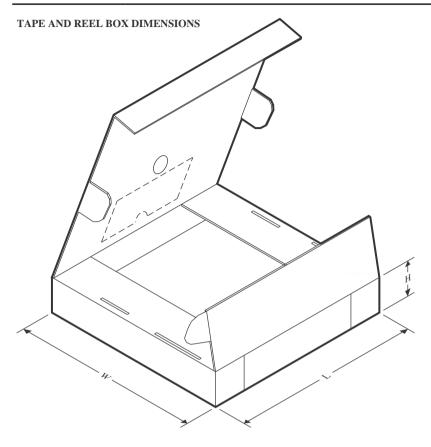


*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ40Z50RSMR-R1	VQFN	RSM	32	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
BQ40Z50RSMT-R1	VQFN	RSM	32	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2



PACKAGE MATERIALS INFORMATION

25-Sep-2024



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ40Z50RSMR-R1	VQFN	RSM	32	3000	367.0	367.0	35.0
BQ40Z50RSMT-R1	VQFN	RSM	32	250	210.0	185.0	35.0

RSM 32

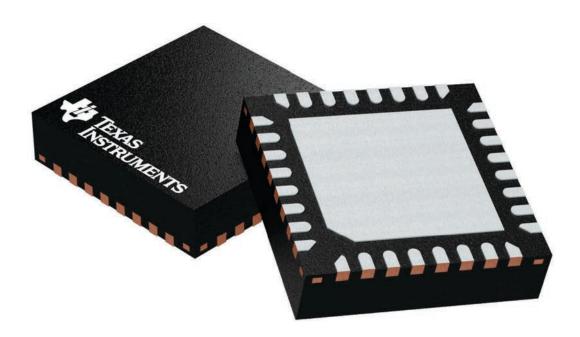
4 x 4, 0.4 mm pitch

GENERIC PACKAGE VIEW

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





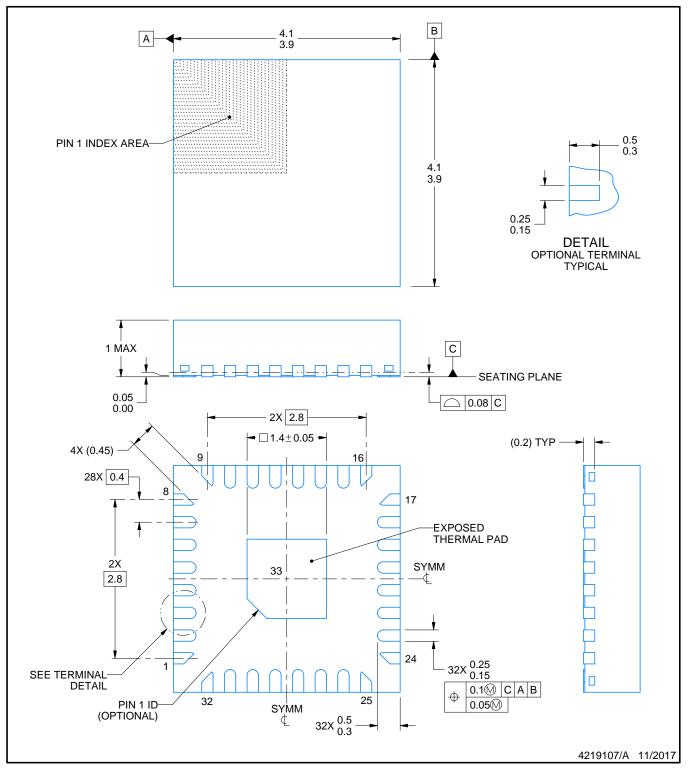
RSM0032A



PACKAGE OUTLINE

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

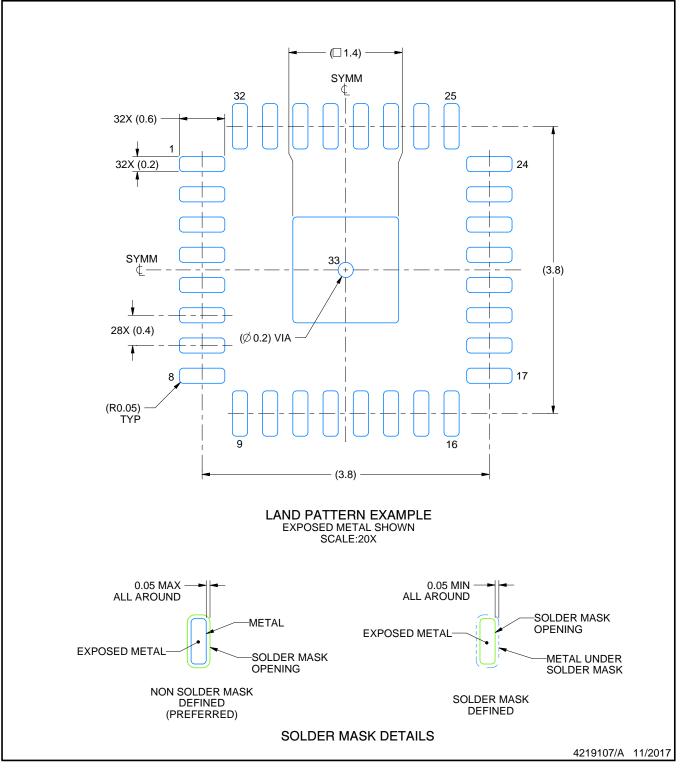


RSM0032A

EXAMPLE BOARD LAYOUT

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

 This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).

 Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

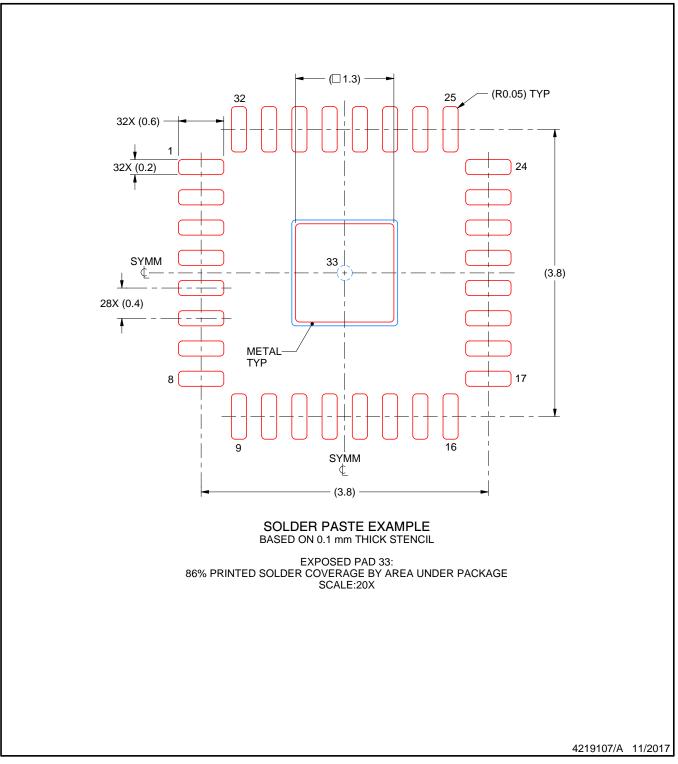


RSM0032A

EXAMPLE STENCIL DESIGN

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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