

The Great Compromise: Voltage References



Jose Gonzalez Torres

Have you ever had to pick between two of your favorite desserts and thought to yourself, “Why can’t I have both?” Well, the same thing happens to engineers every day when designing with programmable voltage references (V_{REF}).

A very common goal for engineers is to come up with very-low-power designs that offer a set functionality: to sense temperature, power up a computer, or even serve you your preferred choice of sweets. But did you know that in order to get to that low-power operation, the engineer is also giving up other advantages? In order to achieve low power, engineers often have to design with V_{REF} s that offer very low current but suffer from a loss of accuracy across the operating temperature range. Is there a way for these engineers to have their cake and eat it too? I think you know the answer.

First, let’s take a look at what I mean by V_{REF} accuracy and the conditions that directly impact accuracy. For this example, I will use the commonly used [TL431](#) to drive my analysis. If you have a circuit similar to [Power-supply Current Limiter](#), you can set up R1 and R2 to get the desired V_{KA} output based on the V_{REF} . You can find more information about how to do so in this [application note](#).

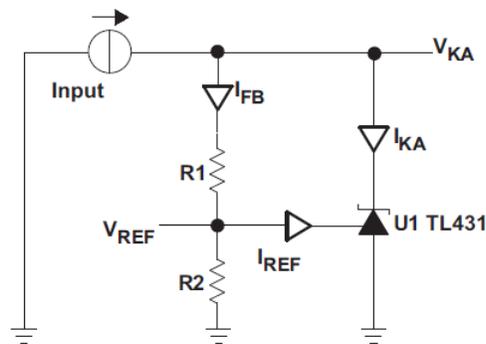


Figure 1. Power-supply Current Limiter

V_{REF} is not always at its nominal; in fact, it is guaranteed to have an offset based on the device’s operating conditions. [TL431 Electrical Specifications](#) shows the specifications table that directly affects V_{REF} .

PARAMETER	TEST CONDITIONS	TL431C, TL432C			UNIT
		MIN	TYP	MAX	
V_{ref} Reference voltage	$V_{KA} = V_{REF}, I_{KA} = 10 \text{ mA}$	2440	2495	2550	mV
$V_{I(dev)}$ Deviation of reference voltage over full temperature range	$V_{KA} = V_{REF}, I_{KA} = 10 \text{ mA}, T_A = 0^\circ\text{C to } 70^\circ\text{C}$	SOT23-3 and TL432 devices			mV
		All other devices			
$\frac{\Delta V_{ref}}{\Delta V_{KA}}$ Ratio of change in reference voltage to the change in cathode voltage	$I_{KA} = 10 \text{ mA}$	$\Delta V_{KA} = 10 \text{ V} - V_{REF}$			mV/V
		$\Delta V_{KA} = 36 \text{ V} - 10 \text{ V}$			
I_{ref} Reference input current	$I_{KA} = 10 \text{ mA}, R1 = 10 \text{ k}\Omega, R2 = \infty$				μA
$I_{I(dev)}$ Deviation of reference input current over full temperature range	$I_{KA} = 10 \text{ mA}, R1 = 10 \text{ k}\Omega, R2 = \infty, T_A = 0^\circ\text{C to } 70^\circ\text{C}$				μA
I_{min} Minimum cathode current for regulation	$V_{KA} = V_{REF}$				mA
I_{off} Off-state cathode current	$V_{KA} = 36 \text{ V}, V_{REF} = 0$				μA
$ Z_{KA} $ Dynamic impedance	$V_{KA} = V_{ref}, f \leq 1 \text{ kHz}, I_{KA} = 1 \text{ mA to } 100 \text{ mA}$				Ω

Figure 2. TL431 Electrical Specifications

You can use Equation [Figure 3](#) to calculate the effective V_{REF} by adding the collective effects of these parameters (typ), assuming operation of $V_{KA} = 5V$ and a cathode current of 2mA.

$$V_{REF} = V_{NOM} + (I_{KA} - I_{NOM}) \times Z_{KA} + (V_{KA} - V_{NOM}) \times 1.4mV/V$$

Figure 3. (1)

This tells you that for the TL431, the effective V_{REF} is now 2.4899V, or 0.2% accurate, which is not a significant difference in plain sight. But once you cross to the maximum value, which usually happens over high temperatures, you get an effective V_{REF} of 2.539V, or 1.78% accuracy.

How does this affect your system?

In an analog environment, where the overall voltage drift may be the necessary threshold to trigger an operational amplifier, a 44.5mV maximum/6mV minimum offset could mean the difference between regulation and standby, which could lead to system failures. But this becomes more of an issue when you take into consideration the use of the TL431 as a reference for an analog-to-digital converter (ADC). The least significant bit (LSB) voltage is determined based on the converter's accuracy in number of bits. Assuming the same conditions of 5V and an ADC of 8 bits, you get an LSB of 19.53mV, which should be fine during typical operation as you can see in equation [Figure 4](#). But across temperature the operation will change and the system might be reading wrong data or performing incorrectly.

$$LSB = \frac{5V}{2^8} = 19.53mV$$

Figure 3. (2)

So how can you resolve the accuracy issues and still maintain low-power operation? One solution is the [ATL431](#), which features lower power operation, but significantly improves accuracy. Using the ATL431 under the same conditions and design parameters as before, you'll get an effective V_{REF} of 2.499V (0.95mV), or 0.03% accuracy. This gives you a much larger margin of error when considering analog operation, but more so, you can now use a much higher resolution ADC (Equation [Figure 5](#)):

$$2^N = \frac{5V}{0.95mV} = 50263.16$$

$$N \sim 12 \text{ bits}$$

Figure 3. (3)

In the end, a small change in the right direction can yield results that are more of a compromise with our original design around TL431. The [ATL431](#) is one of those solutions that provide good-enough power savings to also have an improvement in accuracy, without having to sacrifice one for the other. In the end, even with compromise, it is possible to obtain the best of both worlds.

Additional Resources

- Learn more about the [ATL431](#).
- Learn how to [accurately set up your voltage reference](#).
- Read more in the "[Designing with the 'Advanced' TL431, ATL431](#)" application note.
- Start your design in [WEBENCH® Power Designer](#).

IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATA SHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#) or other applicable terms available either on [ti.com](https://www.ti.com) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

TI objects to and rejects any additional or different terms you may have proposed.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265
Copyright © 2023, Texas Instruments Incorporated