

# Voltage and Current Sensing in HEV/EV Applications



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This post was co-authored by [Ryan Small](#).

The rapid increase of semiconductor content in [automotive](#) systems has prompted a need to manage key voltages and currents in each subsystem. Supervising supply voltages, load currents or other important system functions can help indicate fault conditions, prevent catastrophic failures and protect end users from potential harm.

Hybrid electric vehicles (HEVs) and electric vehicles (EVs) have some distinct challenges compared to conventional combustion-powered vehicles. HEV and EV vehicles feature numerous systems that require monitoring voltages and currents, including the on-board charger (OBC), battery-management system (BMS), DC/DC converters and inverters. In this post, we'll discuss the three fundamental circuits for monitoring current and voltage in HEV/EV systems: low-side current sensing with a difference amplifier (DA), in-line isolated current sensing and high-voltage sensing using an attenuating DA.

[Figure 1](#) is a simplified HEV/EV block diagram showing common systems that require voltage and current monitoring.

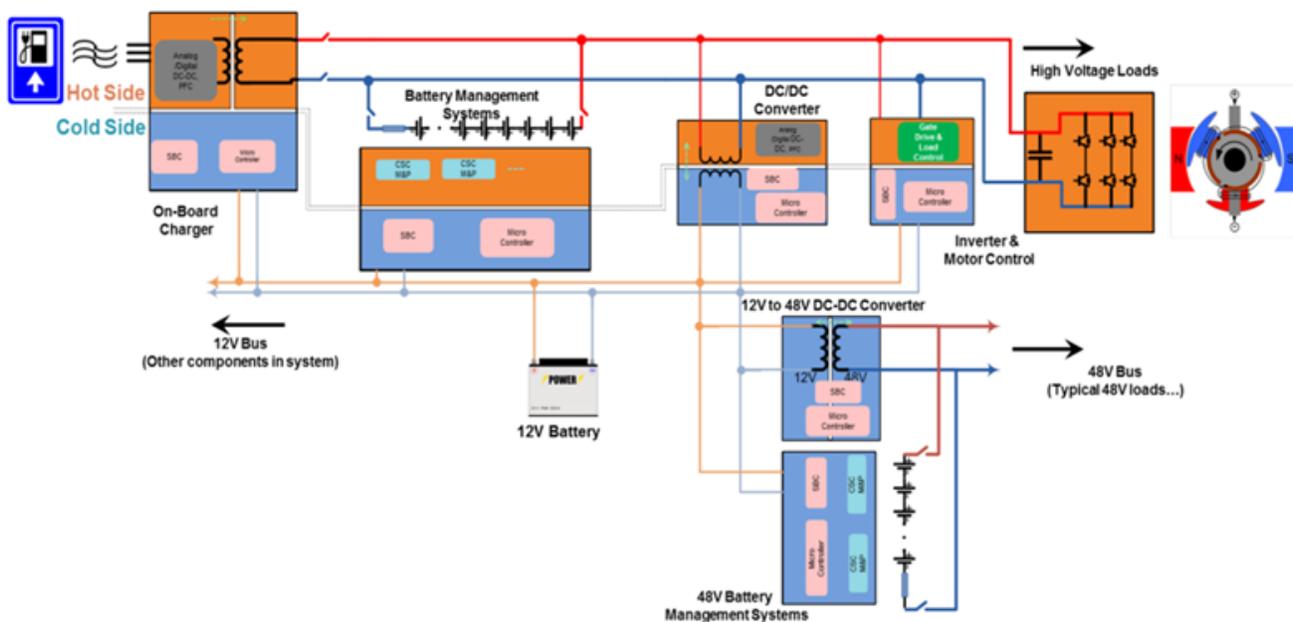
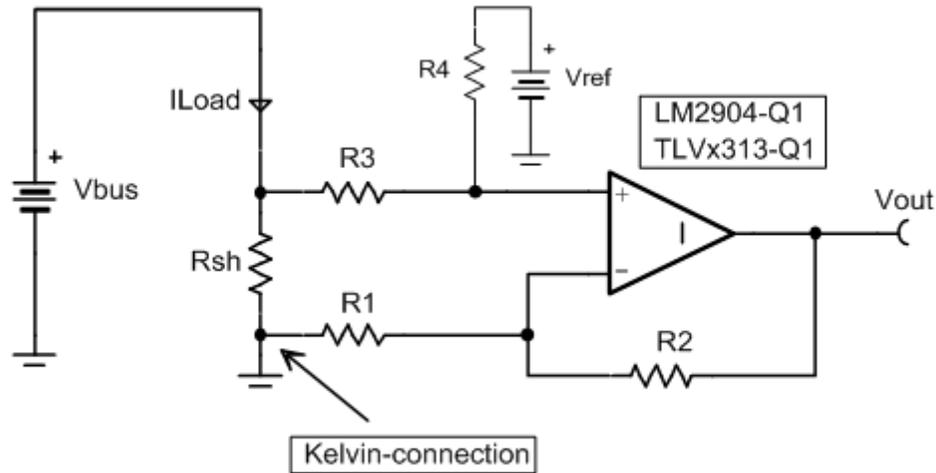


Figure 1. Typical HEV/EV System

While we won't cover the operation of the subsystems in this post, designers sometimes use discrete DAs for low-side current sensing (as shown in [Figure 2](#)), particularly when the load current is bidirectional or they need a small reference voltage to pedestal the output to a known value greater than the amplifier's minimum output swing. For the printed circuit board (PCB) layout, you should Kelvin-connect the shunt resistor. It is important to use a device with a common-mode voltage range that includes ground, such as the [LM2904-Q1](#) or [TLV313-Q1](#).

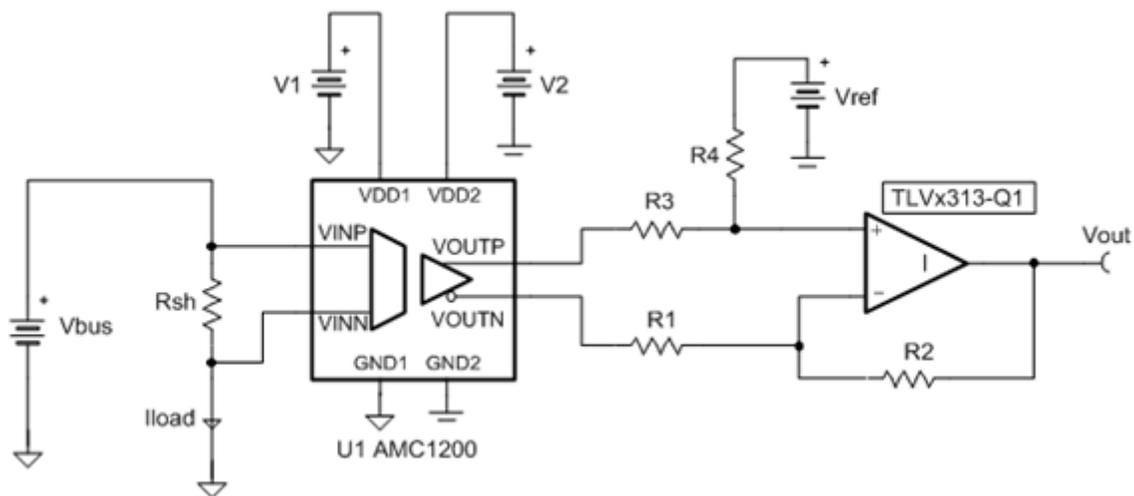


**Figure 2. Difference Amplifier for Low-side Current Sensing**

Equation 1 gives the transfer function for this circuit, assuming  $R_4 = R_2$  and  $R_3 = R_1$ . The amplifier for this application usually has a common-mode range that includes ground. Depending on the situation, however, the common-mode voltage may not be near ground because it actually depends on the voltage divider created by voltage across the shunt resistor ( $I_{Load} \times R_{sh}$ ),  $V_{ref}$  and the ratio of  $R_3$  to  $R_4$ . Make sure that the gain ( $R_2/R_1$ ) is such that the output-voltage swing does not saturate.

$$V_{out} = \left( (I_{Load} \times R_{sh}) \times \frac{R_2}{R_1} \right) + V_{ref} \quad (1)$$

Figure 3 depicts a high-side, in-line isolated current-sensing solution. The AMC1200 is a differential-in, differential-out isolated amplifier. Isolated amplifiers help protect low-voltage circuitry from high common-mode voltages. The DA converts the differential signal to single-ended, which an analog-to-digital converter (ADC) can then digitize.

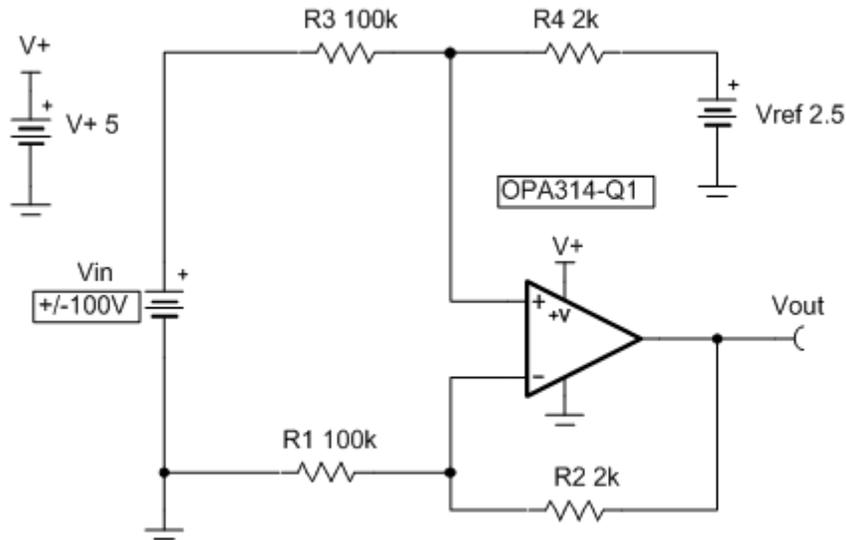


**Figure 3. Difference Amplifier Used for Differential to Single-ended Conversion, Isolated In-line Sensing**

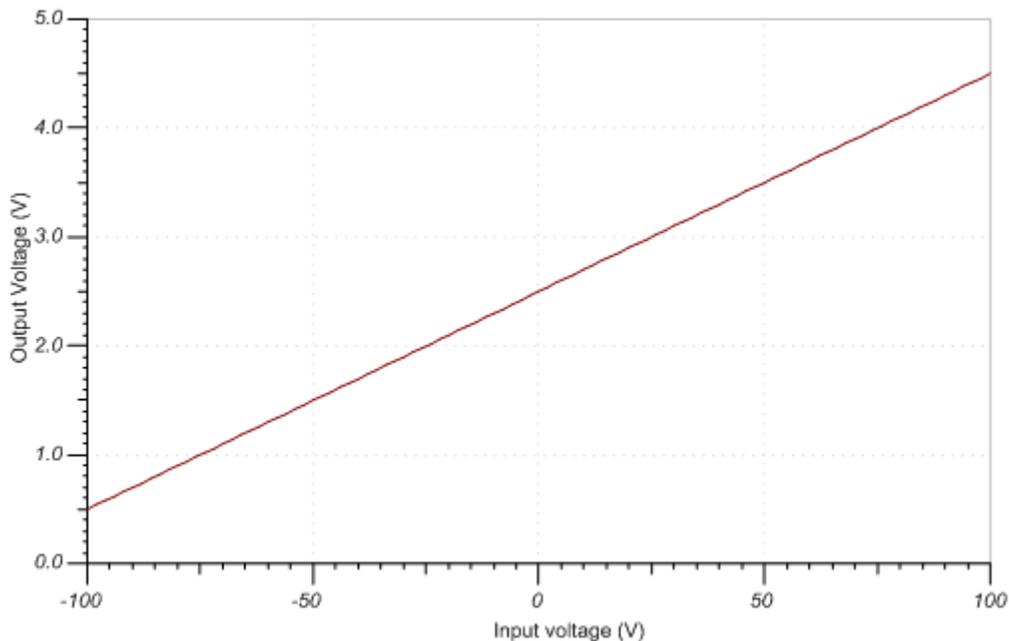
Equation 2 gives the transfer function for the circuit shown in Figure 3, assuming  $R_4 = R_2$  and  $R_3 = R_1$ . The gain of the AMC1200 is 8V/V.  $V_{ref}$  typically connects to  $V_2/2$  to bias the output to mid supply to support bidirectional current. A rail-to-rail input/output (RRIO) amplifier such as the TLV313-Q1 is a good fit for this application.

$$V_{out} = \left( \left( (I_{Load} \times R_{sh}) \times 8 \frac{V}{V} \right) \times \frac{R_2}{R_1} \right) + V_{ref}$$

(2) Figure 4 and Figure 5 depict an attenuating DA and the corresponding TINA-TI™ simulation. In this circuit, the DA resistors divide the input voltage ( $\pm 100V$ ) down. Since the input voltage is bipolar, a reference voltage set to mid supply biases the output. When selecting the amplifier, be sure to comply with the device's input common-mode and output-voltage swing specifications. This example utilizes the OPA314-Q1 because of its wide common-mode range.



**Figure 4. Attenuating Difference Amplifier**



**Figure 5. TINA-TI Software Simulation of Attenuating Difference Amplifier**

Equation 3 gives the transfer function for the circuit shown in [Figure 4](#), assuming  $R_4 = R_2$  and  $R_3 = R_1$ . Notice that the gain is less than unity.

$$V_{out} = \left( V_{in} \times \frac{R_2}{R_1} \right) + V_{ref} = \left( V_{in} \times 0.02 \frac{V}{V} \right) + 2.5V \quad (3)$$

In this blog post, I presented three common circuits that appear in automotive subsystems such as the OBC, BMS, DC/DC converters and inverters. Additionally, I highlighted that HEV/EV applications use voltage and current sensing to monitor critical system functions, helping to maximize vehicle lifetimes. To learn more about the key operational amplifier specifications mentioned in this blog (for example, common-mode voltage) check out [TI Precision Labs](#).

### Additional Resources

- For additional information on PCB layout, read these blog posts:
  - [“The basics: How to layout a PCB for an op amp.”](#)
  - [“How to layout a PCB for an instrumentation amplifier.”](#)
  - Watch the video, [“TI Precision Labs – Op Amps: Input and Output Limitations.”](#)
  - Find common analog design formulas in the [“Analog Engineer’s Pocket Reference.”](#)

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