Reducing Cost for PLC Analog Input Modules Using the MSP430™ MCU



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Introduction

A Programmable Logic Controller (PLC) is a robust process control unit widely used in factory automation, such as manufacturing lines and automated robotic machinery. A similar centralized control unit, sharing many requirements and functionality with a PLC, is used in building automation applications, such as smart homes or building security systems.

An Analog Input Module (AIN) as a subsystem of a PLC, is the front end between the physical real world parameters (that is, Temperature, Distance, Humidity) and the fully digital control unit. The AIN can be implemented into the main PLC unit, but extendable modular systems and remote AINs are also widely used. The AIN command input and typical sensory output are single- or bipolar-voltage (that is, 0 V to 10 V) or current (that is, 0 mA to 20 mA) signals.

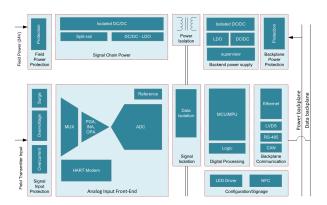


Figure 1. AIN AFE EE Diagram

The accelerating industrial 4.0 revolution keeps increasing the demand for scalable and extendable AIN modules and the reduction of development time, board size, and costs. A possible implementation for a cost-optimized AIN architecture based on a MSP430FR2355 MCU with integrated analog functionality is introduced.

MSP430FR2355 With Integrated Analog

The MSP430FR2355 is a low-power, low-cost MCU with integrated analog features. It integrates a 12-bit

SAR ADC and four configurable signal-chain modules called smart analog combos (SAC), each of which can be used as a 12-bit DAC or a configurable programmable-gain op amp to meet the specific needs of a system while reducing cost and PCB size.

Table 1. MSP430FR2355 Internal ADC

Channels	Up to 12 multiplexed channels
Resolution	10- or 12-bit mode
Sample frequency	Up to about 200 kHz
Voltage reference	Internal 1.5-V, 2.0-V, and 2.5-V rails or external

For a cost-sensitive entry or mid-performance AIN architecture an MCU-integrated ADC with low input voltage range in combination with an external front end is a good fit. Designing the input front-end with discrete components brings higher flexibility in meeting target requirements and reducing cost.

Voltage and Current Input Front End

The Input Front End converts the industries typical signal input range (that is, 0 V to 10 V, 0 mA to 20 mA) to the ADCs input voltage range as well as setting the input impedance and providing required overvoltage and overcurrent protection. Figure 2 and Figure 3 show a discrete unipolar architecture targeting the ADC 2.5-V rail with focus on achieving system accuracy below 1% and optimizing cost.

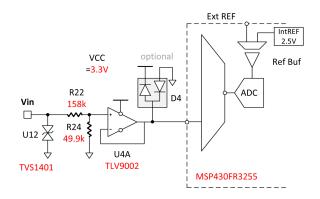


Figure 2. VIN Front End

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Table 2. VIN Device Selection

	Requirement	Specification
R1, R2 voltage divider	Trade off between accuracy and cost to achieve target	1%, 50ppm/°C
Op amp	system accuracy	TLV9002
Protection	Bidirectional surge protection, op-amp input protection	TVS1401, dual clamping diode

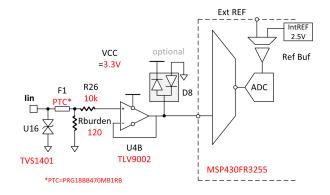


Figure 3. IIN Front End

Table 3. IIN Device Selection

	Requirement	Specification
Shunt Resistor	Trade off between accuracy and cost to achieve target	1%, 50ppm/°C, 1 W
Op amp	system accuracy	TLV9002
Protection	Bidirectional surge protection, shunt protection, optional op-amp input protection	TVS1401, PTC, dual clamping diode

The gain factor of the input stages is chosen so that the 10-V or 20-mA input signals result in an input voltage to the ADC a bit lower than the full-scale of 2.5 V to avoid saturation in consideration of offset, gain, and other errors. For the voltage input stage a gain of 0.24 V/V is used, that is, 2.4 V corresponds to the 10-V input. For the current input stage, a 20-mA input current over the $120-\Omega$ shunt resistor results in 2.4 V at the ADC input.

System Accuracy

The achievable system accuracy is derived from the overall system error. Offset, gain, and other static errors can be eliminated implementing calibration with the cost of lowering the dynamic range. The MSP430FR2355 supports calibration registers for the ADC and the internal reference, for implementing a factory calibration.

There is flexibility to meet individual accuracy specification, choosing front end components (that is, specification of resistors, op amp, ext. reference).

Table 4 lists various error specifications, derived from the data sheets of the system components

Table 4. Error Specification

	Resisto rs	Op Amp	int REF	REF Buffer	ADC	Unit
	1%, 50ppm	TLV900 2	MSP430FR2355			
Offset Error		1.5		16	1.5	mV
Offset Drift		0.004				mV/°C
Gain Error	1.25 (VIN) 1.00 (IIN)		1.5	0.64	0.0732	%FS
Gain Drift			50			ppm/°C
Noise			130			μV

These error values are referred to specific points in the signal chain. Consider the signal-chain gain when calculating the overall offset and gain errors at the input terminals. For example, a 1-mV offset at the input of the ADC corresponds to 1/2.4 mV = 4.167-mV offset at the input terminal.

Layout Size

To demonstrate the size of the proposed circuit, both voltage and current input front-ends including the protection devices are placed on a single-sided PCB. Both channels require only 10 mm \times 10 mm² as shown in Figure 4. The effective front-end area per channel is approximately 5×10 mm². Quad op-amp packages can be used instead of dual op amp and optional clamping diodes can be removed to further reduce the circuit area.

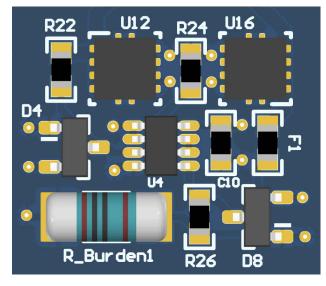


Figure 4. VIN and IIN Layout

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Measurement Procedure

For offset measurements the input is shorted to ground, while the FR2355 integrated ADC captures 1024 samples, from which the average value represents the offset values. If the offset is negative and cannot be measured directly from zero input, a slightly positive input of a few LSBs is added to determine the negative offset.

For measuring gain error the input signal is set to full-scale value. The ADC captures 1024 samples and the average value represents the full-scale error. The gain error is calculated by subtracting the offset error from the full-scale error.

Measurement Results

Table 5 details the voltage input.

To measure the signal-chain noise, DC voltage levels of 1.6 V, and 8.3 V are applied from battery cells, to minimize any added noise from the input source. The internal ADC captures 1024 samples, from which RMS noise is calculated. SNR and ENOBs are calculated with Equation 1 and Equation 2.

$$SNR = 20 \times log(input FS/ RMS noise)$$
 (1)

$$ENOB = (SNR - 1.76 dB) / 6.02$$
 (2)

Table 5. Voltage Input

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Parameter	Conditions ⁽¹⁾	At 20°C	At -25°C	At 55°C	Unit
DC offset		2.71 (6.89)	N/A ⁽²⁾	30.1 (76.5)	LSB (mV)
Gain error		7.27 (18.47)		21.4 (-54.5)	LSB (mV)
FS error		9.97 (25.37)	23 (58.5)	8.6 (22)	LSB (mV)
RMS noise	1.6-V DC input 8.3-V DC input	1.03 (2.61) 1.68 (4.27)	2.3 (5.8)	0.76 (1.9) 1.9 (4.9)	LSB (mV) LSB (mV)
SNR	1.6-V DC input 8.3-V DC input	71.7 67.4	64.7	74.2 66.1	
ENOB	1.6-V DC input 8.3-V DC input	11.6 10.9	10.5	12.0 10.7	Bits

- (1) ADC configuration: sample rate about 83 kSPS, 12-bit mode, 2.5-V internal ADC reference, LSB unit based on 10-V FS input
- (2) Negative offset: The measurement requires conducting the measurement with slightly positive input level (about 10 mV).

Table 6 lists the current input parameter.

Table 6. Current Input

Parameter	Conditions ⁽¹⁾	At 20°C	At -25°C	At 55°C	Unit
FS error		3.26 (16.57)	28.4 (144)	18.3 (93)	LSB (µA)
RMS noise	19.652-mA DC input	1.7 (8.5)	2 (10.4)	1.4 (7.2)	LSB (µA) LSB (µA)
SNR	19.652-mA DC input	67.5	65.7	69.0	
ENOB	19.652-mA DC input	10.9	10.6	11.1	Bits

(1) ADC configuration: sample rate approximately 83 kSPS, 12-bit mode, 2.5-V internal ADC reference, LSB unit based on 10-V FS input

The results for the current input stage are almost identical to these of the voltage input stage, with the burden resistors errors considered instead of the resistor divider.

The measurement results confirm the preliminary estimation of the offset error, gain error, and noise. The circuit maintains a minimum level of 10 b ENOBs over the temperature range, which can be further improved by running averaging over multiple samples. The offset and gain error can be reduced or eliminated by calibration. Single point calibration can be done in addition to correction based on the approximate temp measured by the MCU.

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References

- 1. Texas Instruments, MSP430FR4xx and MSP430FR2xx Family User's Guide
- 2. Texas Instruments, MSP430FR235x, MSP430FR215x Mixed-Signal Microcontrollers Data Sheet
- 3. Texas Instruments, Designing With the MSP430FR4xx and MSP430FR2xx ADC Application Report
- 4. Texas Instruments, *MSP430 ADC calibration* E2E[™] Forum

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