# デザイン・ガイド: TIDA-050027 最大限の省電力化のために柔軟なパーティションを採用したマルチ レール TV 電源のリファレンス・デザイン

# TEXAS INSTRUMENTS

#### 概要

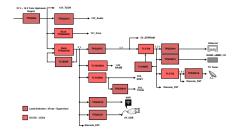
TIDA-050027 リファレンス・デザインは、テレビ (TV) プラッ トフォーム用のパワー・ディストリビューション・ソリューション です。この設計は、一般的な電圧レールとペリフェラルをエ ミュレートする構成により、LCD TV や OLED TV など、多 様なテレビ電源アーキテクチャのリファレンス・デザインとし て使用できます。入力では、高効率の降圧コンバータにより コア・レール、常時オンのレール、ペリフェラル・レールに必 要な電力を供給すると同時に、eFuse を使用して短絡保護 を実現します。降圧コンバータと LDO から電力供給される 5V レールには 2 つの電圧スーパーバイザがあり、レール を監視して低電圧状態になるとリセットします。また、Ioの小 さい低ドロップアウト・レギュレータ (LDO) により、ペリフェラ ル・レールにクリーンなデジタル電力を供給できます。これ らの電源部品に加えて、制御されたスルーレートで負荷ス イッチが電圧を制御することで、電源シーケンスの設定と消 費電力の節約が可能です。さらに、内蔵負荷スイッチは部 品点数 (BOM) を減らしソリューション・サイズを小型化する と同時に、自己保護機能も備えています。

#### リソース

TIDA-050027	デザイン・フォルダ
TPS22919、TPS22975	プロダクト・フォルダ
TPS2595、TPS22810	プロダクト・フォルダ
TLV755、TLV62568	プロダクト・フォルダ
TLV62569A、TLV809E	プロダクト・フォルダ
TPS564201、TPS566250	プロダクト・フォルダ



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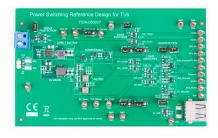


# 特長

- 一般的な TV レール用のフル機能電源ツリーにより、負荷スイッチ、降圧型コンバータ、LDO をシステム設計で適切に選択可能
- 高速過渡応答と、部品点数 (BOM) の削減によるソ リューションの小型化を実現する高効率降圧型コン バータ
- 内蔵保護機能によりシステムの信頼性を向上させ、返品を削減
- 未使用の負荷をオフにすることでシステムの効率を高め、スタンバイ電流を低減
- ディスクリート FET ソリューションと比較して、ソリューションのサイズと部品点数を低減する負荷スイッチ
- 高精度の電圧スーパーバイザにより電圧レールの状態 を適切に保ち、電圧レベルを維持

#### アプリケーション

- HDTV
- UHD 4K OLED TV
- UHD LCD TV





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# 1 System Description

This reference design showcases a power distribution solution for televisions (TVs). Since many TV architectures require similar core voltage and peripheral rails, this design can be used across multiple TV architectures including OLED and LCD TVs.

This design incorporates four buck converters as well as two LDOs to power the core rails, always-on rails, memory, and peripherals such as USB ports and TV tuner. Each buck converter is chosen to handle specific requirements for each of the rails. For the core rail, the requirements include steady output voltage with very good transient response to avoid affecting performance. The remainder of the rails require high efficiency with good transient response. The LDOs for the peripherals provide a  $I_Q$  power supply that also require a clean-power supply such as the TV Tuner.

This design also incorporates nine load switches for peripheral loads commonly found on most TVs including: Panel Power (TCON), Stand-by (STBY) rails, WIFI, NAND, USB, and TV Tuner blocks. All of the load switches have their ON pins connected to external jumpers for quick access to each peripheral rail. An eFuse at the input manages short-circuit and overvoltage events that could potentially damage downstream DC/DC converters and load switches. The reference design also includes common voltage rails found in most TV solutions, including 5 V Peripheral, 3.3 V Standby, 1.5 V DRAM, 3.3 V I/O, and 1.1 V Core rail. Finally, a USB-A header is included at the output of the 5 V USB switch to support USB loads. By using integrated devices, this reference design reduces BOM count, improves system reliability, and enables faster design cycles.

Each main voltage rail (5 V, 3.3 V, and 1.8 V) is accompanied by a discrete FET solution. This allows a direct performance comparison between the integrated load switch and a discrete FET solution. Test points are included at the output of each peripheral and voltage rail, and silkscreen traces outline the size of each solution.



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# 1.1 Key System Specifications

# 表 1. Key System Specifications

PARAMETER	SPECIFICATIONS	DETAILS
Input voltage range	V <sub>IN</sub>	12 V to 16 V

DEVICES	DEVICES	VIN	VOUT	IOUT
	TPS564201	12 V to 16 V	5 V	4 A
DC/DCS	TPS566250	12 V to 16 V	1 V	6 A
DC/DC3	TLV62569A	5 V	1.5 V	2 A
	TLV62568	5 V	3.3 V	1 A
LDOs	TLV759P	5 V	1.8 V	1 A
LDOS	TLV75518	3.3 V	1.8 V	400 mA

PARAMETER	DEVICES	VIN	RON	IOUT
eFuse	TPS2595	12 V to 16 V	34 mΩ	4 A
	TPS22810	12 V to 16 V	79 mΩ	2 A
Load Switch	TPS22975	5 V	16 mΩ	6 A
Zoda Omnon	TPS22919	5 V, 3.3 V, 1.8 V	89 mΩ at 5 V	1.5 A
USB Switch	TPS25221	5 V	70 mΩ	2 A

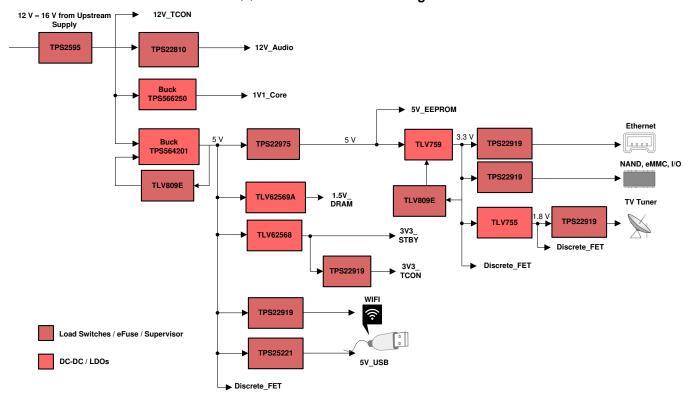


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# 2 System Overview

# 2.1 Block Diagram

# 図 1. TIDA-050027 Block Diagram



# 2.2 Design Considerations

#### 2.2.1 Total Solution Size

Referring to  $\boxtimes$  2, the design is laid out in three main stages: Front-end stage, Power Stage, and Peripherals. Using integrated devices simplifies the solution complexity while keeping the PCB space as compact and space-efficient as possible.



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# Power Switching Reference Design for TVs TIDA-050027 GND TIDA-050027 GND TIDA-050027 GND TIDA-050027 GND TIDA-050027 GND TOAD US: TOAD

#### 図 2. TIDA-050027 Reference Board

The front-end stage includes an eFuse at the input alongside two DC/DCs. The TPS2595 eFuse provides short-circuit and overvoltage protection, while the two DC/DCs emulate voltage rails commonly found on televisions: a 1.1 V core (1V1\_CORE) and a 5 V rail for downstream components. The core rail is an always-on, high current rail for the TV SoC, while the 5 V rail leads to downstream power components. The total solution size of the front end components is 111mm², with 7mm² for the eFuse and 104mm² for the DC/DC components.

The power stage components consist of additional DC/DCs to power a 3.3 V standby rail (3V3\_STBY), as well as a 1.5 V rail for DRAM or SoC Memory modules. The standby rail is useful when the TV enters a low-power mode, such as a sleep or rest state. The 1.5 V rail is another high current rail used for additional SoC modules, such a memory or DRAM. The total solution size of the DC/DC components is 56mm². There are also 5 V peripherals connected to the 5 V rail with integrated load switches, including WiFi (5V\_WIFI) and USB (5V\_USB). A discrete FET implementation is also included to compare performance and size against an integrated solution.

The Peripherals included on this TV reference design emulate common rails and outputs found on TV power designs. A TPS22975 load switch is included at the input of the 5 V peripheral rail, allowing easy access to power down all of the peripherals, for example during standby mode. An LDO converts the 5 V rail into a 3.3 V rail for peripherals including Ethernet, NAND, eMMC, and I/O. Another LDO branches off the 3.3 V down to 1.8 V, providing access to 1.8 V loads such as TV tuners and digital voltage rails required by the SoC. Finally, load switches are included on the front end 12 V line for 12 V rails such as audio and TCON.

表 2. Solution Area Size

Name of Box	Included Components	Total Solution Size
Input Protection	TPS2595	7mm <sup>2</sup>
Core / SoC Rail	TPS566250	57mm <sup>2</sup>
5 V Main	TPS564201	47mm <sup>2</sup>



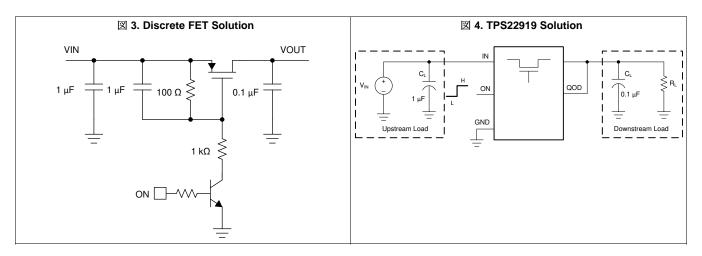
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Name of Box	Included Components	Total Solution Size
DC/DCs	TLV62568, TLV62569A	56mm²
Peripherals	TPS22975, TLV757	14.18mm <sup>2</sup>

# 2.2.2 Power Switching Complexity and Size

As televisions push for higher resolution and greater complexity, their power designs are also getting larger and more complex. As the number of power rails and peripherals increases, additional devices and passive components need to be selected and integrated into the system. As the BOM complexity increases, the number of design considerations increase: including design and layout cycles, PCB space, and points of failure/visual inspection.

A discrete switching circuit contains several components to control the turnon and turnoff behavior of a discrete power MOSFET. 🗵 3 shows a common discrete FET implementation, which is used on the reference design. In comparison, 🗵 4 shows a TPS22919 integrated load switch that can also be used to turn on and or off the corresponding load. Compared to a discrete solution, integrated load switches reduce the BOM count, which helps to reduce the total solution complexity and size.



The TPS22919 also contains additional features that are difficult or not possible to implement discretely. Features such as self-protection and recovery from short circuits, controlled slew rate, thermal shutdown, and quick output discharge (QOD), would require additional components or oversized FETs to be done discretely. With an integrated load switch, this can be accomplished using a single device. 3.2.2 will demonstrate and compare the discrete circuit and the TPS22919 in-depth, providing power on and or off waveforms, thermal images, and transient event waveforms. For more information regarding the drawbacks and limitations of a discrete switching solution, please refer to Integrated Load Switches versus Discrete MOSFETs app note.

表 3. Discrete FET vs. TPS22919 Load Switch Comparison

	Discrete Solution	TPS22919
BOM Count	6 components	1 component
Solution Size	17.08mm <sup>2</sup>	4.1mm <sup>2</sup>
Rise Time	RC Based	Linear, ~1 ms
Thermal Shutdown		✓
Short Circuit Protection		✓



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# 表 3. Discrete FET vs. TPS22919 Load Switch Comparison (continued)

	Discrete Solution	TPS22919
Quick Output Discharge (QOD)	Implemented using External Components	<b>~</b>

☑ 5 compares the solution size of both switching solutions. The three discrete FETs, (1V8\_DIS, 3V3\_DIS, 5V\_DIS), consists of 18 components with a total solution size of 275mm². In comparison, the three integrated rails consists of three TPS22919 load switches, with a 76 percent reduction in solution size.



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# 表 4. Discrete vs. Integrated Solution Size

Solution	Included Components	Number of Components	Total Solution Size
3 Load Switches	3* TPS22919 (1V8_TV, 3V3_ETH, 5V_WIFI)	3	12.3mm <sup>2</sup>
3 Discrete FETs	3* PMOS FET 3* BJT 9*Resistors 3*Gate Capacitor	18	51.24mm <sup>2</sup>

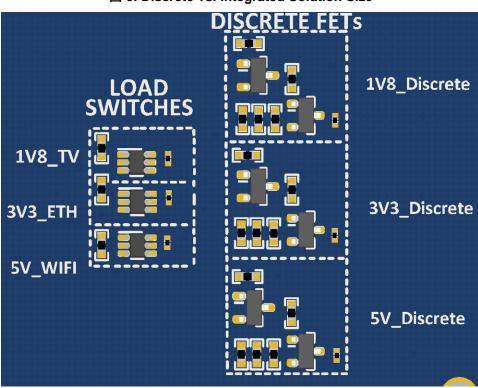


図 5. Discrete vs. Integrated Solution Size

# 2.2.3 Protection Features

Protection features allows for safer and more reliable systems, reducing system failures and customer returns. In television power architectures, features such as inrush current protection, short circuit protection, thermal shutdown, and power sequencing ensure that the system remains robust, while preventing damage to sensitive components.

During startup, inrush current is controlled to protect the MOSFET, PCB traces, and prevent voltage dip that could leave the system in an undesired state. When power is initially applied to the system, capacitor charging can result in a spike of inrush current, which can exceed the nominal load current. The TPS2595 eFuse at the input manages inrush current to the downstream voltage rails, while the TPS22919 manages inrush current to the various peripheral loads. By increasing the rise time of each voltage rail using slew rate control, this minimizes the inrush current while protecting the downstream load. To implement this configuration using discrete components, additional passive components such as a gate resistor and capacitors need to be included, which increases BOM count, solution size, and timing complexity.



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The eFuse and integrated load switches also provide short circuit protection and thermal shutdown. The eFuse at the input prevents damage to downstream components in case of an overvoltage/overcurrent or fault condition, while the TPS22919 offers self protection, meaning that it will protect itself from short circuit events on the peripheral rails. The TPS22919 also has thermal shutdown to prevent damage to the device from overheating, which could occur during startup into a capacitive load or a high-current fault condition.

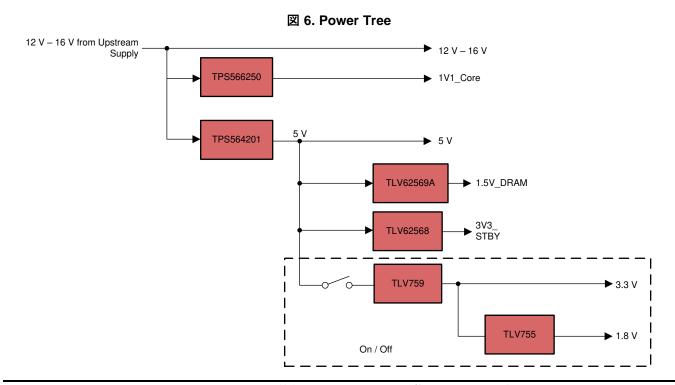
The voltage supervisor allows for monitoring of the buck converter and LDO for undervoltage conditions and turn off appropriately when they reach below the voltage thresholds for each device. When the voltage at the output of these devices falls below the set voltage thresholds for the voltage supervisors, they will output an active-low signal from the RESET pin to disable the devices.

# 2.2.4 Power Consumption

To reduce standby power for the system, this reference design reduces power consumption by choosing high efficiency and low  $I_Q$  power components and by disconnecting loads and peripherals from DC/DC converters when not in use, which is common practice in standby mode.

In this design, load switches are used to emulate various peripheral rails, which can be turned on or off when not in use. If the TV requires all the peripherals to be off in a certain mode, such as standby mode, the TPS22975 can also be used to turn off all the various peripheral rails. The switches can also be used for power sequencing. Power Sequencing is critical for voltage rails that must turn on in a specific order to ensure operation safety and reliability. Sequencing the rails also staggers the inrush current during startup, which reduces system stress and prevents unexpected reverse bias conditions. On this TV power design, the load switches can be sequenced to ensure that the core rails (1V1\_CORE, DRAM, Panel), are up before the peripheral rails are operational.

Televisions also contain always-on rails to power critical system components such as core SoC rails, standby rails, and DRAM power. These rails must remain on throughout all operation modes. Therefore, all of the four DC/DC converters remain enabled to emulate powering always-on loads.





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# 2.3 Highlighted Products

#### 2.3.1 TPS22919

The TPS22919 device is a small, single channel load switch with controlled slew rate. When power is first applied, a Smart Pull Down is used to keep the ON pin from floating until system sequencing is complete. When the pin is deliberately driven High (>VIH), the Smart Pull Down is disconnected to prevent unnecessary power loss. The TPS22919 load switch is also self-protected, meaning that it will protect itself from short circuit events on the output of the device. It also has thermal shutdown to prevent any damage from overheating.

# 2.3.2 TPS22810

The TPS22810 is a single channel load switch with configurable rise time and with an integrated quick output discharge (QOD). In addition, the device features thermal shutdown to protect the device against high junction temperature. Because of this, safe operating area of the device is inherently ensured. The device contains an N-channel MOSFET that can operate over an input voltage range of 2.7 V to 18 V.

#### 2.3.3 TPS22975

The TPS22975 is a single-channel load switch that provides a configurable rise time to minimize inrush current. TPS22975 has an optional  $230-\Omega$  on-chip load resistor for quick output discharge when switch is turned off. The TPS22975 is available in a small, space-saving 2-mm × 2-mm 8-pin SON package (DSG) with integrated thermal pad allowing for high power dissipation.

# 2.3.4 TPS2595

The TPS2595 eFuse (integrated FET hot swap devices) is a highly integrated circuit protection and power management solution in a small package. The device provides multiple protection modes using very few external components and are a robust defense against overloads, short circuits, voltage surges, and excessive inrush current.

#### 2.3.5 TPS566260

The TPS566250 is a 18 V, 6A synchronous step-down (buck) converter with dynamic voltage scaling (DVS) capability through  $I^2c$ . It operates using D-CAP2<sup>TM</sup> control mode which provides very fast transient response, and reduces the required output capacitance required to meet a specific level of performance. This device also would not require external compensation which reduces the total size and cost of implementation. The output voltage of the device can be set by either FB with resistor divider or it can be dynamically set from 0.6 V  $\sim$  1.87 V through  $I^2C$  compatible interface.

#### 2.3.6 TPS564201A

The TPS564201 is a simple, easy-to-use, 4-A synchronous step-down converter in SOT-23 package. The device is optimized to operate with minimum external component count and also optimized to achieve high efficiency at both normal and light load conditions. This solution is a part of a pin-to-pin compatible family of devices (TPS56x201) which would support from 1A ~5A current loads.



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#### 2.3.7 TLV65268

The TLV62568 device is a synchronous step-down buck DC-DC converter optimized for high efficiency and compact solution size. The device integrates switches capable of delivering an output current up to 1A. In order to provide the lowest power consumption for an always on rail, the device enters power save mode at light loads. If lower ripple is required, the TLV62568A can be used instead. In shutdown, the current consumption is reduced to less than  $2~\mu\text{A}$ .

#### 2.3.8 TLV65269A

The TLV62569A device is a synchronous step-down buck DC-DC converter optimized for high efficiency and compact solution size. The device integrates switches capable of delivering an output current up to 2 A. In order to provide the lowest ripple and best transient response for powering the DRAM rail, the device operates only in pulse width modulation (PWM) mode without a power saving mode. If light load efficiency is important, the TLV62569 can be used instead. In shutdown, the current consumption is reduced to less than 2  $\mu$ A.

#### 2.3.9 TLV755/7P

The TLV755P is an ultra-small, low quiescent current, low-dropout regulator (LDO) that sources 500mA and 1 A for the TLV757P, with good line and load transient performance. The TLV755P is optimized for a wider variety of applications by supporting an input voltage range from 1.45 V to 5.5 V. To minimize cost and solution size, the device is offered in fixed output voltages ranging from 0.6 V to 5 V to support the lower core voltages of modern microcontrollers (MCUs). Additionally, the TLV755P has a low I<sub>Q</sub> with enable functionality to minimize stand-by power. When shutdown, the device actively pulls down the output to quickly discharge the outputs and ensure a known start-up state.

#### 2.3.10 TLV758/9P

The TLV759P low-dropout regulator (LDO) is an adjustable output, ultra-small, low quiescent current LDO that sources 1 A and 500mA for the TLV758P, with good line and load transient performance. The TLV759P is optimized for a wide variety of applications by supporting an input voltage range from 1.45 V to 5.5 V. Additionally, the TLV759P has a low  $I_Q$  with enable functionality to minimize standby power. When shutdown, the device actively pulls down the output to quickly discharge the outputs and ensure a known start-up state. Furthermore, the TLV759P supports adjustable output via resistor divider for voltage options not supported by the TLV755/7.

#### 2.3.11 TLV809E

The TLV809E voltage supervisor is a low current (250 nA typical, 2  $\mu$ A max) circuit (reset IC) that monitors a  $V_{DD}$  voltage level. This device initiates a reset signal whenever supply voltage drops below the factory programmed threshold voltage,  $V_{IT}$ . The reset output remains low for a fixed reset time delay  $t_D$  after the  $V_{DD}$  voltage rises above the threshold voltage and hysteresis.



# 3 Hardware, Software, Testing Requirements, and Test Results

# 3.1 Required Hardware

# 3.1.1 System Overview

The TIDA-050027 TV Reference Design power rails can be sequenced depending on the downstream loads and peripherals. 表 5 provides an overview of the jumpers and test points on the TI Design.

表 5. Jumper/Connector Summary

Stage	Connector	Label	Description
	J2	-	Input Power, connector for 12 - 16 V
Input	JP9	-	Enable for TPS2595 (U12)
	TP21	12V_UP	Test Point for eFuse output
CORE / SoC Rail	JP11, TP23, TP24	SDA / SCL	Test Points for TPS566250 communication
	TP22	1V1_Core	Test Point for DC/DC output
5 V Main	TP25	5V_Down	Test Point for DC/DC output
DC/DCs	TP19	1V5_DRAM	Test Point for DC/DC output
DC/DC\$	TP20	3V3_STBY	Test Point for DC/DC output
Peripherals	JP5	Peripheral_EN	Enable for TPS22975
Periprierais	TP14	5V_Periph	Output of 5 V Peripheral Rail
	JP10	12V_EN	Enable for TPS22810
	JP2	3V3_EN	Enable for TPS22919
	JP1	1V8_EN	Enable for TPS22919
	JP7	5V_EN	Enable for TPS22919
	TP8	TV	Output of 1.8 V Peripheral Rail
Load Switches	TP1	TCON	Output 12 V Rail
	TP4	Audio	Output 12 V Rail
	TP9	Ethernet	Output of 3.3 V Peripheral Rail
	TP10	NAND	Output of 3.3 V Peripheral Rail
	TP11	3V3_Panel	Output of 3.3 V Standby Rail
	TP12	WIFI	Output of 5 V Peripheral Rail
	JP3	1V8_D_EN	Enable for Discrete FET
	JP4	3V3_D_EN	Enable for Discrete FET
Discrete FETs	JP8	5V_D_EN	Enable for Discrete FET
DISCIPLE FETS	TP5	1V8_Dis	Output of 1.8 V Discrete Rail
	TP6	3V3_Dis	Output of 3.3 V Discrete Rail
	TP7	5V_Dis	Output of 5 V Discrete Rail



# 3.2 Testing and Results

# 3.2.1 Test Setup

To configure the reference design for operation, connect the jumpers as shown in 🗵 7.

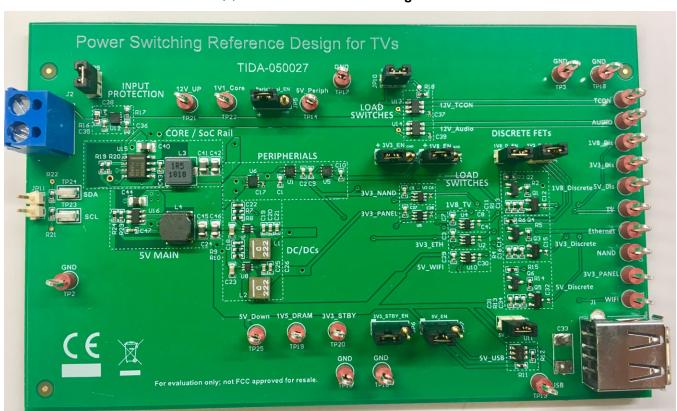


図 7. TIDA-050027 Test Configuration

表 6. TV Configuration Setup

Description	Jumper / Connector	Value
Input Power, connector for 12 - 16 V	J2	Connect pins 1 and 2
12 V Load Switch enable	JP5, JP10	Connect pins 1 and 2
5 V, 3.3 V, 1.8 V Load Switch enable	JP1, JP2, JP6, JP7	Connect pins 1 and 2
Discrete FET Enable	JP3, JP4, JP8	Connect pins 1 and 2

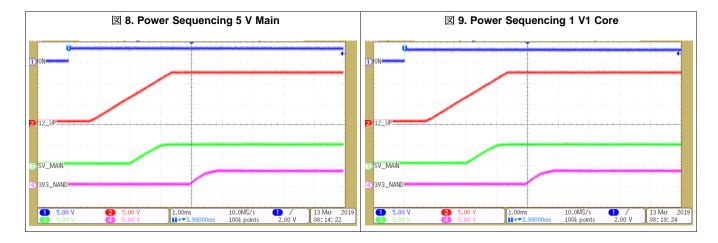
#### 3.2.2 Test Results

# 3.2.2.1 Power Sequencing Demonstration

Power Sequencing is critical on voltage rails that must turn on in a specific order to ensure operational safety and reliability. For example, core rails for downstream SoCs and microcontrollers need to be powered before WIFI and other peripherals are turned on. Sequencing the rails also helps stagger the inrush current during power-up, which reduces system stress and prevents unexpected reverse bias conditions.



The TV Reference Design's voltage rails are sequenced to emulate the power-on of a television system. After the system is initially turned on, the 12 -16 V input passes downstream and turns on the two critical rails, 1V1\_Core and 5 V Main. Once both of these rails are high, the 5 V Peripherals are turned on followed by the 3.3 V and 1.8 V Peripherals.  $\boxtimes$  8 and  $\boxtimes$  9 demonstrate the power-on sequence.



# 3.2.2.2 Standby Power Demonstration

The system standby current measurements were taken in various on/off modes. In standby mode, the measured input current was 1.256mA. This includes all of the necessary "always-on" core rails, including 1V1\_Core and 3V3\_STBY. With all of the peripherals and load switches enabled, the design was drawing roughly 1.5mA of quiescent current. This accounts for all of the LDOs and peripheral load switches, which drew an additional 220  $\mu$ A current when enabled. In comparison, three discrete FETs drew roughly 500 $\mu$ A when enabled.

Mode of Operation	Devices Enabled	Measured Input Current
Standby Mode	TPS2595, TPS566250, TPS564201, TLV62569A, TLV62568	1.256mA
Everything on except Discrete FETs	TPS2595, TPS566250, TPS564201, TLV62569A, TLV62568, TPS22975, TLV757, TPS22919	1.478mA
All Devices Enabled	TPS2595, TPS566250, TPS564201, TLV62569A, TLV62568, TPS22975, TLV757, TPS22919, 1V8_Dis, 3V3_Dis, 5V_Dis	1.923mA

表 7. Measured Input Current

#### 3.2.2.3 Discrete FET vs. Integrated Load Switch Comparison

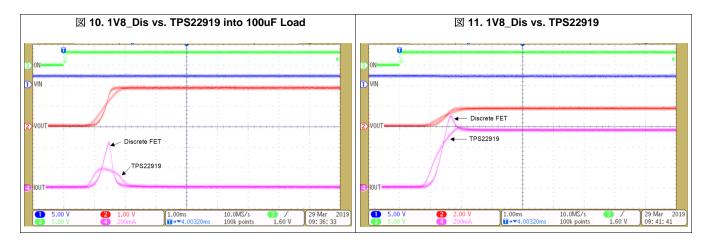
表	8.	Turn	On/Off	Summary	/ Table

Voltage	Parameter	Description	Section
1.8 V	1.8 V Capacitive Load	100uF	図 10
	1.8 V Load	100uF, 3.6Ω (500mA)	図 11
3.3 V	3.3 V Capacitive Load	100uF	図 12
	3.3 V Load	100uF, 5.5Ω (600mA)	図 13
5 V	5 V Capacitive Load	100uF	図 14
5 V	5 V Load	100uF, 1Ω (1A)	図 15



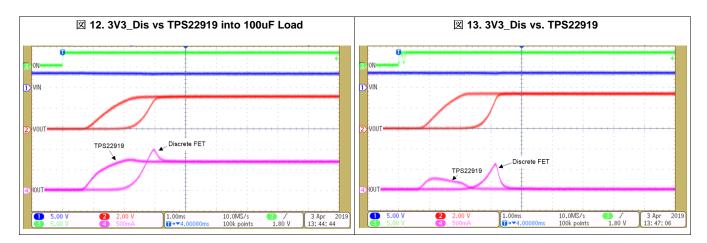
#### 3.2.2.3.1 1.8V Load Demonstration

As shown in 🗵 4 the discrete solution contains an NPN BJT is connected to the gate of the PMOS "pass FET." When the NPN BJT is enabled, it pulls down the gate of the PMOS and allows for power to flow from VIN to VOUT. However, compared to a TPS22919 solution, the discrete solution comes with a set of limitations. During power on and or off sequences, the discrete solution will have a nonlinear (RC) turn-on behavior, which can lead to a large spike of inrush current. The following waveforms will compare/contrast the turnon behaviors overlaid on one another. In the first waveform, 🗵 10, both switching solutions are starting into a 100uF capacitive load. The TPS22919 exhibits a linear turn on behavior, limiting the inrush current to roughly 200mA. In comparison, the discrete solution experiences a peak of over 400mA of inrush current. In 🗵 11, both solutions are starting into a 500mA / 100uF load. The TPS22919 helps to manage the inrush current with the controlled slew rate, while the discrete solution peaks at 700mA of inrush current.



# 3.2.2.3.2 3.3V Load Demonstration

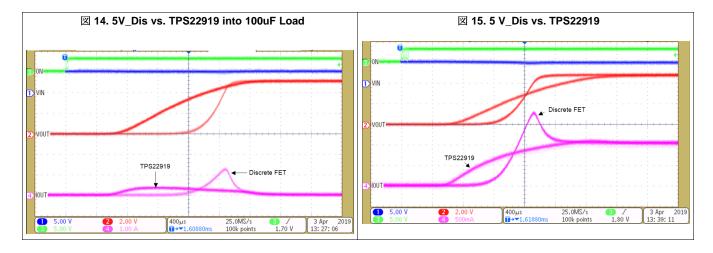
☑ 12 and ☑ 13 provide a comparison between the discrete solution and the TPS22919 at 3.3 V. When starting into a capacitive load, the TPS22919 limits the inrush current to 250mA while the inrush current on the discrete solution jumps to 700mA. When starting into a 1A load with 100uF of capacitance, the TPS22919 limits the inrush current while the discrete FET peaks to 1A.





#### 3.2.2.3.3 5 V Load Demonstration

☑ 14 and ☑ 15 provide a comparison between the discrete solution and the TPS22919 at 5 V. When starting into a capacitive load, the TPS22919 limits the inrush current to 250mA while the inrush current on the discrete solution jumps to 1.3A. When starting into a 1A load with 100uF of capacitance, the TPS22919 limits the inrush current while the discrete FET peaks at over 1.75A.

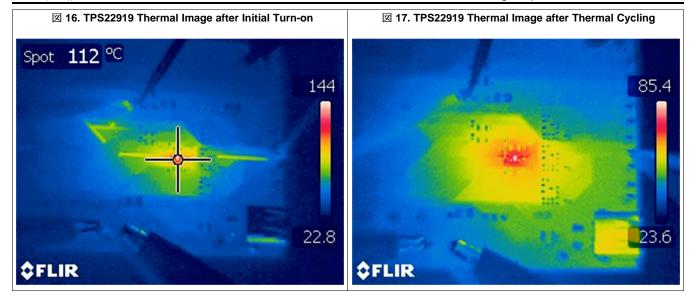


#### 3.2.2.3.4 Thermal Protection

Thermal shutdown protects the device from failure when the device junction temperature exceeds its safe limit. During operation, the junction temperature can rise due to many factors, including a high current load, large inrush current during startup, or during a fault condition. Thermal protection turns off the switch to protect itself by reducing the power dissipation, also potentially avoiding damage to upstream and downstream components. In comparison, a discrete FET cannot protect itself if it exceeds it's thermal junction temperature. In this case, the FET could potentially break, causing a short and damaging the downstream load or peripheral.

☑ 16 and ☑ 17 provide a thermal image of the TPS22919 with a high current load at 25C. In this scenario, a 3A load is connected to 3V3\_Eth, causing the thermal junction temperature to rise. As the temperature increases past 144C, the TPS22919 turns off and thermal cycles, bringing the board temperature back down to roughly 85C.





In comparison, 🗵 18 shows the package temperature of the discrete FET at 25C under the same 3A load. The 3.3 V discrete FET will continue to power the downstream load until it fails. Unlike the TPS22919, the discrete FET does not have thermal shutdown, and will continue to be thermally stressed until the load is removed. If the FET is stressed beyond its absolute maximum rating, the device could also potentially be damaged. At higher ambient temperatures, the discrete FET will continue to be stressed beyond 143C, whereas the TPS22919 will continue to protect itself and thermally cycle.

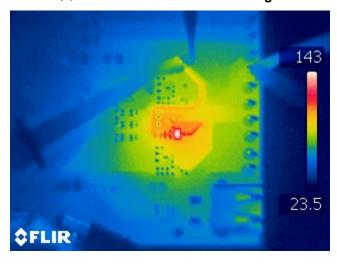


図 18. Discrete FET Thermal Image

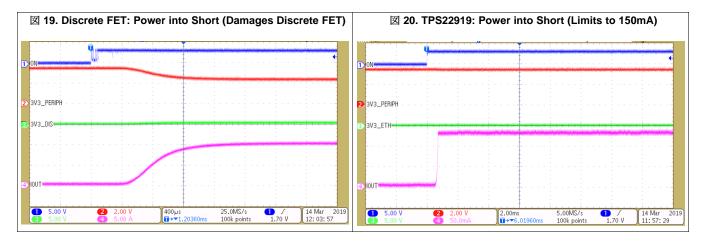
#### 3.2.2.3.5 Transient Protection Features

TV Power Architectures can be stressed during startup and fault conditions. During these conditions, discrete FETs can dissipate more power that exceeds the physical and thermal limits of the MOSFET. If proper tests aren't taken into consideration, the system can fail due to overstressing these components. Additionally, some of these fault conditions, such as short circuit protection and hot plug protection, are often difficult to implement discretely and require additional components and design complexity. By using integrated devices on the TIDA-050027, these additional protection features can make the design more robust while simplifying the design.



The first major transient event that could severely damage the system is *Power Into Short*. In this condition, the output of the FET / IC is shorted to ground before the switch is turned on. As the FET is turned on, VOUT remains grounded at 0V. Therefore, the stress across the FET is equal to VIN. The power dissipated into the FET is equal to VIN \* IOUT. Done discretely, as shown in 🗵 19, the FET doesn't limit the current in the transient event and passes 10A of current downstream. Since the junction temperature of the FET exceeds its thermal limit during this event, the FET is damaged and sends 10A of continuous current to the downstream component.

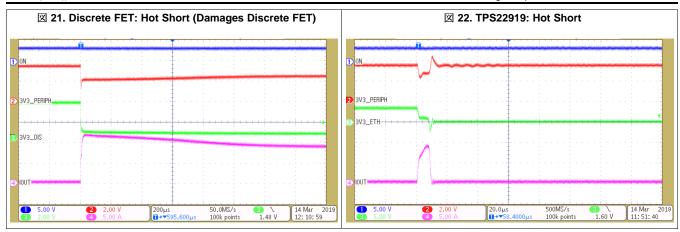
In comparison, the TPS22919 integrated load switch is self-protecting and prevents damaging itself during the power into short event. When the short occurs, the large VIN to VOUT voltage drop causes the switch to limit the output current (Isc). Since the output voltage starts at 0 V, a lower limit is used to minimize the power dissipation until the fault is removed. In this case, as shown in  $\boxtimes$  20, the TPS22919 limits the output current to 0.5A while protecting itself from damage.



The other major transient event that could severely damage the system is *Hot Short*. In this condition, the output of the FET / IC is shorted to ground when the device is already turned on. This event is considered the more stressful event since the switch usually starts out hotter due to the pre-existing load current. The power dissipated into the switch is equal to VIN \* IOUT.  $\boxtimes$  21 demonstrates a discrete implementation. Since the discrete FET doesn't limit the current in the transient, more than 10A passes downstream and the FET fails.

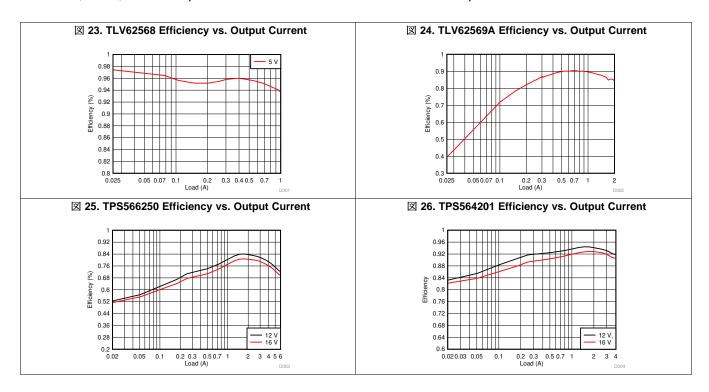
In comparison, the TPS22919 integrated load switch measures the voltage drop across the device and limits the output current within  $t_{\rm sc}$ . The device continues to limit the current until it reaches the thermal shutdown temperature. At this point, the device turns off until the junction temperature has lowered. As shown in  $\boxtimes$  22 the TPS22919 limits the output current and turns off, dropping the output voltage to 0 V and protecting itself from damage.



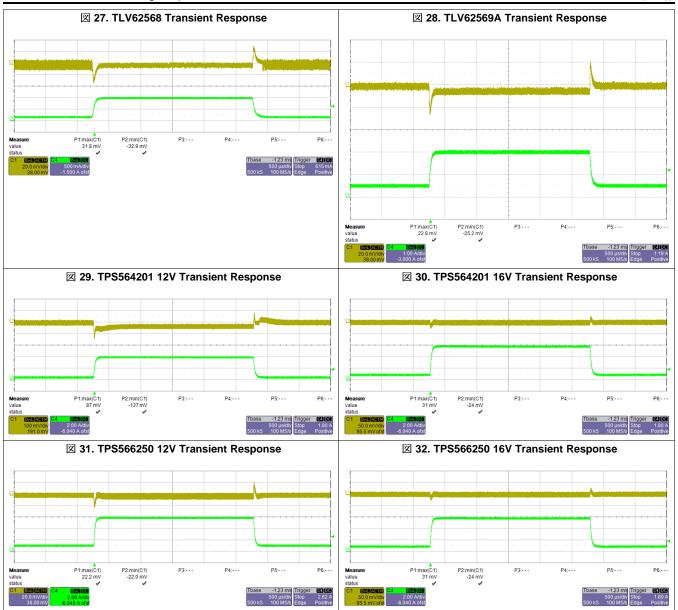


# 3.2.2.4 DC/DC Efficiency and Transient Testing Results

図 23, 図 24, 図 25, and 図 26 provides the TIDA-050027 DC/DC efficiency results. 図 27, 図 28, 図 29, 図 30, 図 31, and 図 32 provides the TIDA-050027 DC/DC transient responses.









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# 4 Design Files

# 4.1 Altium Project

To download the Altium Designer® project files, see the design files at TIDA-050027.

#### 4.2 Schematics

To download the schematics, see the design files at TIDA-050027.

#### 4.3 Bill of Materials

To download the bill of materials (BOM), see the design files at TIDA-050027.

# 4.4 PCB Layout Recommendations

# 4.4.1 Layout Prints

To download the layer plots, see the design files at TIDA-050027.

# 4.5 Gerber Files

To download the Gerber files, see the design files at TIDA-050027.

# 4.6 Assembly Drawings

To download the assembly drawings, see the design files at TIDA-050027.

#### 5 Software Files

To download the software files, see the design files at TIDA-050027.

# 6 Related Documentation

- 1. Integrated Load Switches versus Discrete MOSFETs app report
- 2. Basics of Power Switches app report

#### 6.1 商標

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改訂履歴 www.tij.co.jp

# 改訂履歴

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Revision A (May 2019) から Revision B に変更	Page
<ul> <li>降圧コンバータと LDO から電力供給される 5V レールには 2 つの電圧スーパーバイザがあり、レとリセットします。 追加</li> <li>プロダクト・フォルダ 追加</li> </ul>	
• 電圧スーパーバイザに関する情報を 追加	1
<ul><li>TLV809E を入れて図を 変更</li><li>TLV809E to block diagram. 追加</li></ul>	4
<ul><li>section on voltage supervisors. 追加</li><li>new section on TLV809E. 追加</li></ul>	
2019年5月発行のものから更新	Page
タイトルを「TV 用の電源スイッチングのリファレンス・デザイン」から「最大限の省電力化のために柔ルチレール TV 電源のリファレンス・デザイン」に 変更	<i>軟なパーティションを採用したマ</i> 1

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