

TMUX721x 44V、低 RON、1:1 (SPST)、4 チャンネル高精度スイッチ、ラッチアップフリー、1.8V ロジック

1 特長

- ラッチアップフリー
- 両電源電圧範囲: $\pm 4.5V \sim \pm 22V$
- 単電源電圧範囲: $4.5V \sim 44V$
- 低いオン抵抗: 2Ω
- 大電流のサポート: 220mA (最大値) (TSSOP)
- $-40^{\circ}C \sim +125^{\circ}C$ の動作温度範囲
- ロジックピンにプルダウン抵抗を内蔵
- 1.8V ロジック互換
- フェイルセーフロジック
- レールツーレール動作
- 双方向動作

2 アプリケーション

- EV 充電ステーション向け電源モジュール
- 先進運転支援システム (ADAS)
- 車載ゲートウェイ
- アナログおよびデジタルの多重化 / 多重分離
- 車載ヘッド・ユニット
- テレマティクス制御ユニット
- 緊急通話 (eCall)
- インフォテインメント
- 車体制御モジュール (BCM)
- ボディ・エレクトロニクス / 照明
- バッテリー管理システム (BMS)
- HVAC コントローラ・モジュール
- ADAS ドメイン・コントローラ

3 概要

TMUX7212 は、独立して選択できる 4 つの 1:1 単極単投 (SPST) スイッチ チャンネルを備えた相補型金属酸化膜半導体 (CMOS) スイッチです。このデバイスは単一電源 ($4.5V \sim 44V$)、デュアル電源 ($\pm 4.5V \sim \pm 22V$)、非対称電源 ($V_{DD} = 12V$, $V_{SS} = -5V$ など) で動作します。TMUX721x は、ソース (Sx) およびドレイン (Dx) ピンで、 V_{SS} から V_{DD} までの範囲の双方向アナログおよびデジタル信号をサポートします。

TMUX721x のスイッチは、SELx ピンの適切なロジック制御入力で制御されます。TMUX721x は高精度スイッチおよびマルチプレクサファミリの製品であり、オンおよびオフ時のリーク電流が非常に小さいため、高精度の測定用途に使用できます。

TMUX721x ファミリはラッチアップフリーであるため、過電圧イベントによってよく発生するデバイス内の寄生構造間の好ましくない大電流イベントを防止できます。ラッチアップ状態は通常、電源レールがオフにされるまで継続するため、デバイスの障害の原因となる場合があります。このラッチアップフリーという特長により、TMUX721x は過酷な環境でも使用できます。

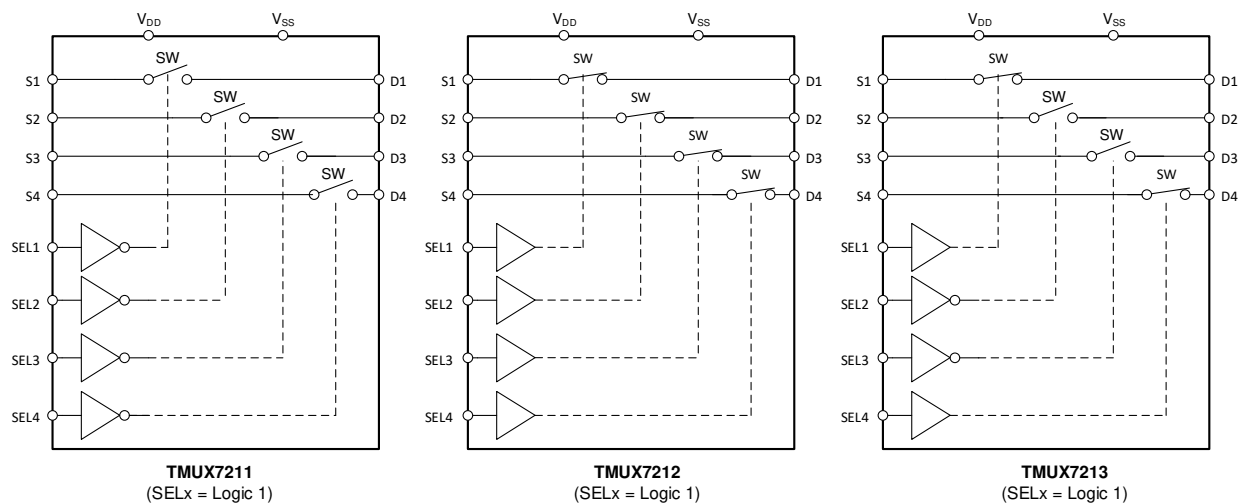
表 3-1. 製品情報

部品番号 ⁽¹⁾	アクティブ ロジック	パッケージ ⁽²⁾
TMUX7211	ロジック Low	PW (TSSOP, 16)
TMUX7212	ロジック High	RUM (WQFN, 16)
TMUX7213	ロジック Low + ロジック High	

(1) 製品比較表を参照してください。

(2) 詳細については、セクション 12 を参照してください。





TMUX721x 機能ブロック図

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4 Device Comparison Table

PRODUCT	DESCRIPTION
TMUX7211	Low-Leakage-Current, Precision, 4-Channel, 1:1 (SPST) Switches (Logic Low)
TMUX7212	Low-Leakage-Current, Precision, 4-Channel, 1:1 (SPST) Switches (Logic High)
TMUX7213	Low-Leakage-Current, Precision, 4-Channel, 1:1 (SPST) Switches (Logic Low + Logic High)

5 Pin Configuration and Functions

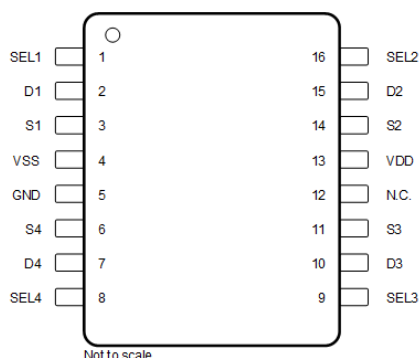


図 5-1. PW Package, 16-Pin TSSOP (Top View)

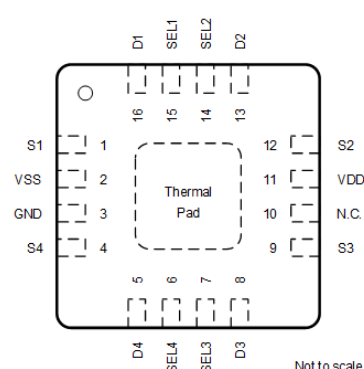


図 5-2. RUM Package, 16-Pin WQFN (Top View)

表 5-1. Pin Functions

PIN			TYPE ⁽¹⁾	DESCRIPTION ⁽²⁾
NAME	TSSOP	WQFN		
D1	2	16	I/O	Drain pin 1. Can be an input or output.
D2	15	13	I/O	Drain pin 2. Can be an input or output.
D3	10	8	I/O	Drain pin 3. Can be an input or output.
D4	7	5	I/O	Drain pin 4. Can be an input or output.
GND	5	3	P	Ground (0 V) reference
N.C.	12	10	—	No internal connection. Can be shorted to GND or left floating.
S1	3	1	I/O	Source pin 1. Can be an input or output.
S2	14	12	I/O	Source pin 2. Can be an input or output.
S3	11	9	I/O	Source pin 3. Can be an input or output.
S4	6	4	I/O	Source pin 4. Can be an input or output.
SEL1	1	15	I	Logic control input 1, has internal 4 MΩ pull-down resistor. Controls channel 1 state as shown in セクション 8.5 .
SEL2	16	14	I	Logic control input 2, has internal 4 MΩ pull-down resistor. Controls channel 2 state as shown in セクション 8.5 .
SEL3	9	7	I	Logic control input 3, has internal 4 MΩ pull-down resistor. Controls channel 3 state as shown in セクション 8.5 .
SEL4	8	6	I	Logic control input 4, has internal 4 MΩ pull-down resistor. Controls channel 4 state as shown in セクション 8.5 .
VDD	13	11	P	Positive power supply. This pin is the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from 0.1 μF to 10 μF between V _{DD} and GND.
VSS	4	2	P	Negative power supply. This pin is the most negative power-supply potential. In single-supply applications, this pin can be connected to ground. For reliable operation, connect a decoupling capacitor ranging from 0.1 μF to 10 μF between V _{SS} and GND.
Thermal Pad			—	The thermal pad is not connected internally. No requirement to solder this pad, if connected it is recommended that the pad be left floating or tied to GND

(1) I = input, O = output, I/O = input and output, P = power.

(2) Refer to [セクション 8.4](#) for what to do with unused pins.

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)^{(1) (2)}

		MIN	MAX	UNIT
$V_{DD} - V_{SS}$	Supply voltage		48	V
V_{DD}		−0.5	48	V
V_{SS}		−48	0.5	V
V_{SEL} or V_{EN}	Logic control input pin voltage (SELx)	−0.5	48	V
I_{SEL} or I_{EN}	Logic control input pin current (SELx)	−30	30	mA
V_S or V_D	Source or drain voltage (Sx, Dx)	$V_{SS} - 0.5$	$V_{DD} + 0.5$	V
I_{IK}	Diode clamp current ⁽³⁾	−30	30	mA
I_S or $I_D (CONT)$	Source or drain continuous current (Sx, Dx)		$I_{DC} + 10\%$ ⁽⁴⁾	mA
T_A	Ambient temperature	−55	150	°C
T_{stg}	Storage temperature	−65	150	°C
T_J	Junction temperature		150	°C
P_{tot}	Total power dissipation (QFN) ⁽⁵⁾		1650	mW
	Total power dissipation (TSSOP) ⁽⁵⁾		700	mW

- (1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltages are with respect to ground, unless otherwise specified.
- (3) Pins are diode-clamped to the power-supply rails. Over voltage signals must be voltage and current limited to maximum ratings.
- (4) Refer to *Source or Drain Continuous Current* table for I_{DC} specifications.
- (5) For QFN package: P_{tot} derates linearly above $T_A = 70^\circ\text{C}$ by $24.2\text{mW}/^\circ\text{C}$.
For TSSOP package: $P_{tot} = 700\text{ mW (max)}$ and derates linearly above $T_A = 70^\circ\text{C}$ by $10.7\text{mW}/^\circ\text{C}$.

6.2 ESD Ratings

			VALUE	UNIT
TMUX721x in PW package				
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ⁽¹⁾	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±500	
TMUX721x in RUM package				
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ⁽¹⁾	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±500	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Thermal Information

THERMAL METRIC ⁽¹⁾		TMUX721x		UNIT
		PW (TSSOP)	RUM (WQFN)	
		16 PINS	16 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	94.5	41.5	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	25.5	25.1	°C/W
R _{θJB}	Junction-to-board thermal resistance	41.1	16.5	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	1.1	0.3	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	40.4	16.4	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	N/A	2.9	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V _{DD} – V _{SS} ⁽¹⁾	Power supply voltage differential	4.5		44	V
V _{DD}	Positive power supply voltage	4.5		44	V
V _S or V _D	Signal path input/output voltage (source or drain pin) (Sx, D)	V _{SS}		V _{DD}	V
V _{SEL} or V _{EN}	Address or enable pin voltage	0		44	V
I _S or I _D (CONT)	Source or drain continuous current (Sx, D)			I _{DC} ⁽²⁾	mA
T _A	Ambient temperature	–40		125	°C

(1) V_{DD} and V_{SS} can be any value as long as 4.5 V ≤ (V_{DD} – V_{SS}) ≤ 44 V, and the minimum V_{DD} is met.

(2) Refer to *Source or Drain Continuous Current* table for I_{DC} specifications.

6.5 Source or Drain Continuous Current

at supply voltage of V_{DD} ± 10%, V_{SS} ± 10 % (unless otherwise noted)

CONTINUOUS CURRENT PER CHANNEL (I _{DC}) ⁽²⁾		T _A = 25°C	T _A = 85°C	T _A = 125°C	UNIT
PACKAGE	TEST CONDITIONS				
PW (TSSOP)	+44 V Dual Supply ⁽¹⁾	220	160	100	mA
	±15 V Dual Supply	220	160	100	mA
	+12 V Single Supply	190	130	90	mA
	±5 V Dual Supply	170	120	80	mA
	+5 V Single Supply	130	90	60	mA
RUM (WQFN)	+44 V Single Supply ⁽¹⁾	330	220	120	mA
	±15 V Dual Supply	330	220	120	mA
	+12 V Single Supply	260	180	110	mA
	±5 V Dual Supply	240	160	100	mA
	+5 V Single Supply	180	120	80	mA

(1) Specified for nominal supply voltage only.

(2) Refer to Total power dissipation (P_{tot}) limits in *Absolute Maximum Ratings* table that must be followed with max continuous current specification.

6.6 ±15 V Dual Supply: Electrical Characteristics

$V_{DD} = +15\text{ V} \pm 10\%$, $V_{SS} = -15\text{ V} \pm 10\%$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +15\text{ V}$, $V_{SS} = -15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG SWITCH							
R _{ON}	On-resistance	V _S = −10 V to +10 V I _D = −10 mA Refer to On-Resistance	25°C		2	2.7	Ω
			−40°C to +85°C			3.4	Ω
			−40°C to +125°C			4	Ω
ΔR _{ON}	On-resistance mismatch between channels	V _S = −10 V to +10 V I _D = −10 mA Refer to On-Resistance	25°C		0.1	0.18	Ω
			−40°C to +85°C			0.19	Ω
			−40°C to +125°C			0.21	Ω
R _{ON FLAT}	On-resistance flatness	V _S = −10 V to +10 V I _S = −10 mA Refer to On-Resistance	25°C		0.2	0.46	Ω
			−40°C to +85°C			0.65	Ω
			−40°C to +125°C			0.7	Ω
R _{ON DRIFT}	On-resistance drift	V _S = 0 V, I _S = −10 mA Refer to On-Resistance	−40°C to +125°C		0.008		Ω/°C
I _{S(OFF)}	Source off leakage current ⁽¹⁾	V _{DD} = 16.5 V, V _{SS} = −16.5 V Switch state is off V _S = +10 V / −10 V V _D = −10 V / +10 V Refer to Off-Leakage Current	25°C	−0.25	0.05	0.25	nA
			−40°C to +85°C	−3		3	nA
			−40°C to +125°C	−20		20	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	V _{DD} = 16.5 V, V _{SS} = −16.5 V Switch state is off V _S = +10 V / −10 V V _D = −10 V / +10 V Refer to Off-Leakage Current	25°C	−0.25	0.05	0.25	nA
			−40°C to +85°C	−3		3	nA
			−40°C to +125°C	−20		20	nA
I _{S(ON)} I _{D(ON)}	Channel on leakage current ⁽²⁾	V _{DD} = 16.5 V, V _{SS} = −16.5 V Switch state is on V _S = V _D = ±10 V Refer to On-Leakage Current	25°C	−0.4	0.1	0.4	nA
			−40°C to +85°C	−1		1	nA
			−40°C to +125°C	−3		3	nA
LOGIC INPUTS (SEL / EN pins)							
V _{IH}	Logic voltage high		−40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		−40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		−40°C to +125°C		0.4	2	μA
I _{IL}	Input leakage current		−40°C to +125°C	−0.1	−0.005		μA
C _{IN}	Logic input capacitance		−40°C to +125°C		3.5		pF
POWER SUPPLY							
I _{DD}	V _{DD} supply current	V _{DD} = 16.5 V, V _{SS} = −16.5 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		35	56	μA
			−40°C to +85°C			65	μA
			−40°C to +125°C			80	μA
I _{SS}	V _{SS} supply current	V _{DD} = 16.5 V, V _{SS} = −16.5 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		5	20	μA
			−40°C to +85°C			24	μA
			−40°C to +125°C			35	μA

(1) When V_S is positive, V_D is negative, or when V_S is negative, V_D is positive.

(2) When V_S is at a voltage potential, V_D is floating, or when V_D is at a voltage potential, V_S is floating.

6.7 ±15 V Dual Supply: Switching Characteristics

$V_{DD} = +15\text{ V} \pm 10\%$, $V_{SS} = -15\text{ V} \pm 10\%$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +15\text{ V}$, $V_{SS} = -15\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
t_{ON}	Turn-on time from control input	$V_S = 10\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		100	175	ns
			-40°C to $+85^\circ\text{C}$			205	ns
			-40°C to $+125^\circ\text{C}$			225	ns
t_{OFF}	Turn-off time from control input	$V_S = 10\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		80	205	ns
			-40°C to $+85^\circ\text{C}$			225	ns
			-40°C to $+125^\circ\text{C}$			240	ns
t_{BBM}	Break-before-make time delay (TMUX7213 Only)	$V_S = 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$	25°C		27		ns
			-40°C to $+85^\circ\text{C}$	5			ns
			-40°C to $+125^\circ\text{C}$	5			ns
$t_{ON(VDD)}$	Device turn on time (V_{DD} to output)	V_{DD} rise time = $1\ \mu\text{s}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on (VDD) Time	25°C		0.17		ms
			-40°C to $+85^\circ\text{C}$		0.18		ms
			-40°C to $+125^\circ\text{C}$		0.18		ms
t_{PD}	Propagation delay	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ Refer to Propagation Delay	25°C		260		ps
Q_{INJ}	Charge injection	$V_S = 0\text{ V}$, $C_L = 100\text{ pF}$ Refer to Charge Injection	25°C		60		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 100\text{ kHz}$ Refer to Off Isolation	25°C		-70		dB
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$ Refer to Off Isolation	25°C		-50		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 100\text{ kHz}$ Refer to Crosstalk	25°C		-114		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$ Refer to Crosstalk	25°C		-93		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$ Refer to Bandwidth	25°C		56		MHz
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		-0.15		dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP} = 0.62\text{ V}$ on V_{DD} and V_{SS} $R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$ Refer to ACPSRR	25°C		-68		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 15\text{ V}$, $V_{BIAS} = 0\text{ V}$ $R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $f = 20\text{ Hz}$ to 20 kHz Refer to THD + Noise	25°C		0.0004		%
$C_{S(OFF)}$	Source off capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		28		pF
$C_{D(OFF)}$	Drain off capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		45		pF
$C_{S(ON)}$, $C_{D(ON)}$	On capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		145		pF

6.8 ±20 V Dual Supply: Electrical Characteristics

$V_{DD} = +20\text{ V} \pm 10\%$, $V_{SS} = -20\text{ V} \pm 10\%$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +20\text{ V}$, $V_{SS} = -20\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG SWITCH							
R _{ON}	On-resistance	V _S = −15 V to +15 V I _D = −10 mA Refer to On-Resistance	25°C		1.7	2.5	Ω
			−40°C to +85°C			3	Ω
			−40°C to +125°C			3.6	Ω
ΔR _{ON}	On-resistance mismatch between channels	V _S = −15 V to +15 V I _D = −10 mA Refer to On-Resistance	25°C		0.1	0.18	Ω
			−40°C to +85°C			0.19	Ω
			−40°C to +125°C			0.21	Ω
R _{ON FLAT}	On-resistance flatness	V _S = −15 V to +15 V I _S = −10 mA Refer to On-Resistance	25°C		0.3	0.6	Ω
			−40°C to +85°C			0.8	Ω
			−40°C to +125°C			0.95	Ω
R _{ON DRIFT}	On-resistance drift	V _S = 0 V, I _S = −10 mA Refer to On-Resistance	−40°C to +125°C		0.008		Ω/°C
I _{S(OFF)}	Source off leakage current ⁽¹⁾	V _{DD} = 22 V, V _{SS} = −22 V Switch state is off V _S = +15 V / −15 V V _D = −15 V / +15 V Refer to Off-Leakage Current	25°C	−1	0.05	1	nA
			−40°C to +85°C	−4.5		4.5	nA
			−40°C to +125°C	−33		33	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	V _{DD} = 22 V, V _{SS} = −22 V Switch state is off V _S = +15 V / −15 V V _D = −15 V / +15 V Refer to Off-Leakage Current	25°C	−1	0.1	1	nA
			−40°C to +85°C	−4.5		4.5	nA
			−40°C to +125°C	−33		33	nA
I _{S(ON)} I _{D(ON)}	Channel on leakage current ⁽²⁾	V _{DD} = 22 V, V _{SS} = −22 V Switch state is on V _S = V _D = ±15 V Refer to On-Leakage Current	25°C	−1	0.1	1	nA
			−40°C to +85°C	−1.5		1.5	nA
			−40°C to +125°C	−8		8	nA
LOGIC INPUTS (SEL / EN pins)							
V _{IH}	Logic voltage high		−40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		−40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		−40°C to +125°C		0.4	2	μA
I _{IL}	Input leakage current		−40°C to +125°C	−0.1	−0.005		μA
C _{IN}	Logic input capacitance		−40°C to +125°C		3.5		pF
POWER SUPPLY							
I _{DD}	V _{DD} supply current	V _{DD} = 22 V, V _{SS} = −22 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		33	65	μA
			−40°C to +85°C			74	μA
			−40°C to +125°C			90	μA
I _{SS}	V _{SS} supply current	V _{DD} = 22 V, V _{SS} = −22 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		7	26	μA
			−40°C to +85°C			30	μA
			−40°C to +125°C			45	μA

(1) When V_S is positive, V_D is negative, or when V_S is negative, V_D is positive.

(2) When V_S is at a voltage potential, V_D is floating, or when V_D is at a voltage potential, V_S is floating.

6.9 ±20 V Dual Supply: Switching Characteristics

$V_{DD} = +20\text{ V} \pm 10\%$, $V_{SS} = -20\text{ V} \pm 10\%$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +20\text{ V}$, $V_{SS} = -20\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
t_{ON}	Turn-on time from control input	$V_S = 10\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		100	185	ns
			-40°C to $+85^\circ\text{C}$			210	ns
			-40°C to $+125^\circ\text{C}$			230	ns
t_{OFF}	Turn-off time from control input	$V_S = 10\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		90	210	ns
			-40°C to $+85^\circ\text{C}$			225	ns
			-40°C to $+125^\circ\text{C}$			235	ns
t_{BBM}	Break-before-make time delay (TMUX7213 Only)	$V_S = 10\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$	25°C		28		ns
			-40°C to $+85^\circ\text{C}$	10			ns
			-40°C to $+125^\circ\text{C}$	10			ns
$t_{ON(VDD)}$	Device turn on time (V_{DD} to output)	V_{DD} rise time = $1\ \mu\text{s}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on (VDD) Time	25°C		0.17		ms
			-40°C to $+85^\circ\text{C}$		0.18		ms
			-40°C to $+125^\circ\text{C}$		0.18		ms
t_{PD}	Propagation delay	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ Refer to Propagation Delay	25°C		260		ps
Q_{INJ}	Charge injection	$V_S = 0\text{ V}$, $C_L = 100\text{ pF}$ Refer to Charge Injection	25°C		92		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 100\text{ kHz}$ Refer to Off Isolation	25°C		-70		dB
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$ Refer to Off Isolation	25°C		-50		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 100\text{ kHz}$ Refer to Crosstalk	25°C		-112		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$ Refer to Crosstalk	25°C		-93		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$ Refer to Bandwidth	25°C		48		MHz
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		-0.14		dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP} = 0.62\text{ V}$ on V_{DD} and V_{SS} $R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$ Refer to ACPSRR	25°C		-68		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 20\text{ V}$, $V_{BIAS} = 0\text{ V}$ $R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $f = 20\text{ Hz}$ to 20 kHz Refer to THD + Noise	25°C		0.0003		%
$C_{S(OFF)}$	Source off capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		28		pF
$C_{D(OFF)}$	Drain off capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		45		pF
$C_{S(ON)}$, $C_{D(ON)}$	On capacitance	$V_S = 0\text{ V}$, $f = 1\text{ MHz}$	25°C		145		pF

6.10 44 V Single Supply: Electrical Characteristics

$V_{DD} = +44\text{ V}$, $V_{SS} = 0\text{ V}$, GND = 0 V (unless otherwise noted)

Typical at $V_{DD} = +44\text{ V}$, $V_{SS} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG SWITCH							
R _{ON}	On-resistance	V _S = 0 V to 40 V I _D = −10 mA Refer to On-Resistance	25°C		2	2.4	Ω
			−40°C to +85°C			3.2	Ω
			−40°C to +125°C			3.8	Ω
ΔR _{ON}	On-resistance mismatch between channels	V _S = 0 V to 40 V I _D = −10 mA Refer to On-Resistance	25°C		0.1	0.18	Ω
			−40°C to +85°C			0.19	Ω
			−40°C to +125°C			0.21	Ω
R _{ON FLAT}	On-resistance flatness	V _S = 0 V to 40 V I _D = −10 mA Refer to On-Resistance	25°C		0.65	0.8	Ω
			−40°C to +85°C			1.1	Ω
			−40°C to +125°C			1.2	Ω
R _{ON DRIFT}	On-resistance drift	V _S = 22 V, I _S = −10 mA Refer to On-Resistance	−40°C to +125°C		0.007		Ω/°C
I _{S(OFF)}	Source off leakage current ⁽¹⁾	V _{DD} = 44 V, V _{SS} = 0 V Switch state is off V _S = 40 V / 1 V V _D = 1 V / 40 V Refer to Off-Leakage Current	25°C	−1	0.05	1	nA
			−40°C to +85°C	−7		7	nA
			−40°C to +125°C	−50		50	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	V _{DD} = 44 V, V _{SS} = 0 V Switch state is off V _S = 40 V / 1 V V _D = 1 V / 40 V Refer to Off-Leakage Current	25°C	−1	0.05	1	nA
			−40°C to +85°C	−7		7	nA
			−40°C to +125°C	−50		50	nA
I _{S(ON)} I _{D(ON)}	Channel on leakage current ⁽²⁾	V _{DD} = 44 V, V _{SS} = 0 V Switch state is on V _S = V _D = 40 V or 1 V Refer to On-Leakage Current	25°C	−1	0.05	1	nA
			−40°C to +85°C	−3.5		3.5	nA
			−40°C to +125°C	−5		5	nA
LOGIC INPUTS (SEL / EN pins)							
V _{IH}	Logic voltage high		−40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		−40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		−40°C to +125°C		0.6	2	μA
I _{IL}	Input leakage current		−40°C to +125°C	−0.1	−0.005		μA
C _{IN}	Logic input capacitance		−40°C to +125°C		3.5		pF
POWER SUPPLY							
I _{DD}	V _{DD} supply current	V _{DD} = 44 V, V _{SS} = 0 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		44	79	μA
			−40°C to +85°C			88	μA
			−40°C to +125°C			105	μA

(1) When V_S is positive, V_D is negative, or when V_S is negative, V_D is positive.

(2) When V_S is at a voltage potential, V_D is floating, or when V_D is at a voltage potential, V_S is floating.

6.11 44 V Single Supply: Switching Characteristics

$V_{DD} = +44\text{ V}$, $V_{SS} = 0\text{ V}$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +44\text{ V}$, $V_{SS} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
t_{ON}	Turn-on time from control input	$V_S = 18\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		80	185	ns
			-40°C to $+85^\circ\text{C}$			205	ns
			-40°C to $+125^\circ\text{C}$			225	ns
t_{OFF}	Turn-off time from control input	$V_S = 18\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		90	205	ns
			-40°C to $+85^\circ\text{C}$			220	ns
			-40°C to $+125^\circ\text{C}$			228	ns
t_{BBM}	Break-before-make time delay (TMUX7213 Only)	$V_S = 18\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$	25°C		27		ns
			-40°C to $+85^\circ\text{C}$	10			ns
			-40°C to $+125^\circ\text{C}$	10			ns
$t_{ON(VDD)}$	Device turn on time (V_{DD} to output)	V_{DD} rise time = $1\ \mu\text{s}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on (VDD) Time	25°C		0.14		ms
			-40°C to $+85^\circ\text{C}$		0.15		ms
			-40°C to $+125^\circ\text{C}$		0.15		ms
t_{PD}	Propagation delay	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ Refer to Propagation Delay	25°C		270		ps
Q_{INJ}	Charge injection	$V_S = 22\text{ V}$, $C_L = 100\text{ pF}$ Refer to Charge Injection	25°C		104		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 100\text{ kHz}$ Refer to Off Isolation	25°C		-70		dB
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$ Refer to Off Isolation	25°C		-50		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 100\text{ kHz}$ Refer to Crosstalk	25°C		-112		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$ Refer to Crosstalk	25°C		-93		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$ Refer to Bandwidth	25°C		46		MHz
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$	25°C		-0.15		dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP} = 0.62\text{ V}$ on V_{DD} and V_{SS} $R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$ Refer to ACPSRR	25°C		-66		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 22\text{ V}$, $V_{BIAS} = 22\text{ V}$ $R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $f = 20\text{ Hz}$ to 20 kHz Refer to THD + Noise	25°C		0.0003		%
$C_{S(OFF)}$	Source off capacitance	$V_S = 22\text{ V}$, $f = 1\text{ MHz}$	25°C		28		pF
$C_{D(OFF)}$	Drain off capacitance	$V_S = 22\text{ V}$, $f = 1\text{ MHz}$	25°C		45		pF
$C_{S(ON)}$, $C_{D(ON)}$	On capacitance	$V_S = 22\text{ V}$, $f = 1\text{ MHz}$	25°C		145		pF

6.12 12 V Single Supply: Electrical Characteristics

$V_{DD} = +12\text{ V} \pm 10\%$, $V_{SS} = 0\text{ V}$, $GND = 0\text{ V}$ (unless otherwise noted)

Typical at $V_{DD} = +12\text{ V}$, $V_{SS} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG SWITCH							
R _{ON}	On-resistance	V _S = 0 V to 10 V I _D = −10 mA Refer to On-Resistance	25°C		2.8	5.4	Ω
			−40°C to +85°C			6.8	Ω
			−40°C to +125°C			7.4	Ω
ΔR _{ON}	On-resistance mismatch between channels	V _S = 0 V to 10 V I _D = −10 mA Refer to On-Resistance	25°C		0.13	0.21	Ω
			−40°C to +85°C			0.23	Ω
			−40°C to +125°C			0.25	Ω
R _{ON FLAT}	On-resistance flatness	V _S = 0 V to 10 V I _S = −10 mA Refer to On-Resistance	25°C		1	1.7	Ω
			−40°C to +85°C			1.9	Ω
			−40°C to +125°C			2	Ω
R _{ON DRIFT}	On-resistance drift	V _S = 6 V, I _S = −10 mA Refer to On-Resistance	−40°C to +125°C		0.015		Ω/°C
I _{S(OFF)}	Source off leakage current ⁽¹⁾	V _{DD} = 13.2 V, V _{SS} = 0 V Switch state is off V _S = 10 V / 1 V V _D = 1 V / 10 V Refer to Off-Leakage Current	25°C	−0.25	0.01	0.25	nA
			−40°C to +85°C	−2		2	nA
			−40°C to +125°C	−16		16	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	V _{DD} = 13.2 V, V _{SS} = 0 V Switch state is off V _S = 10 V / 1 V V _D = 1 V / 10 V Refer to Off-Leakage Current	25°C	−0.25	0.05	0.25	nA
			−40°C to +85°C	−2		2	nA
			−40°C to +125°C	−16		16	nA
I _{S(ON)} I _{D(ON)}	Channel on leakage current ⁽²⁾	V _{DD} = 13.2 V, V _{SS} = 0 V Switch state is on V _S = V _D = 10 V or 1 V Refer to On-Leakage Current	25°C	−0.5	0.05	0.5	nA
			−40°C to +85°C	−1		1	nA
			−40°C to +125°C	−3		3	nA
LOGIC INPUTS (SEL / EN pins)							
V _{IH}	Logic voltage high		−40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		−40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		−40°C to +125°C		0.4	2	μA
I _{IL}	Input leakage current		−40°C to +125°C	−0.1	−0.005		μA
C _{IN}	Logic input capacitance		−40°C to +125°C		3.5		pF
POWER SUPPLY							
I _{DD}	V _{DD} supply current	V _{DD} = 13.2 V, V _{SS} = 0 V Logic inputs = 0 V, 5 V, or V _{DD}	25°C		30	44	μA
			−40°C to +85°C			52	μA
			−40°C to +125°C			62	μA

(1) When V_S is positive, V_D is negative, or when V_S is negative, V_D is positive.

(2) When V_S is at a voltage potential, V_D is floating, or when V_D is at a voltage potential, V_S is floating.

6.13 12 V Single Supply: Switching Characteristics

$V_{DD} = +12\text{ V} \pm 10\%$, $V_{SS} = 0\text{ V}$, $GND = 0\text{ V}$ (unless otherwise noted)
Typical at $V_{DD} = +12\text{ V}$, $V_{SS} = 0\text{ V}$, $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T_A	MIN	TYP	MAX	UNIT
t_{ON}	Turn-on time from control input	$V_S = 8\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		170	225	ns
			-40°C to $+85^\circ\text{C}$			276	ns
			-40°C to $+125^\circ\text{C}$			315	ns
t_{OFF}	Turn-off time from control input	$V_S = 8\text{ V}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on and Turn-off Time	25°C		75	248	ns
			-40°C to $+85^\circ\text{C}$			285	ns
			-40°C to $+125^\circ\text{C}$			310	ns
t_{BBM}	Break-before-make time delay (TMUX7213 Only)	$V_S = 8\text{ V}$, $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$	25°C		30		ns
			-40°C to $+85^\circ\text{C}$	13			ns
			-40°C to $+125^\circ\text{C}$	13			ns
$t_{ON(VDD)}$	Device turn on time (V_{DD} to output)	V_{DD} rise time = $1\ \mu\text{s}$ $R_L = 300\ \Omega$, $C_L = 35\text{ pF}$ Refer to Turn-on (VDD) Time	25°C		0.17		ms
			-40°C to $+85^\circ\text{C}$		0.18		ms
			-40°C to $+125^\circ\text{C}$		0.18		ms
t_{PD}	Propagation delay	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ Refer to Propagation Delay	25°C		270		ps
Q_{INJ}	Charge injection	$V_S = 6\text{ V}$, $C_L = 100\text{ pF}$ Refer to Charge Injection	25°C		12		pC
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 100\text{ kHz}$ Refer to Off Isolation	25°C		-70		dB
O_{ISO}	Off-isolation	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$ Refer to Off Isolation	25°C		-50		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 100\text{ kHz}$ Refer to Crosstalk	25°C		-112		dB
X_{TALK}	Crosstalk	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$ Refer to Crosstalk	25°C		-93		dB
BW	-3dB Bandwidth	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$ Refer to Bandwidth	25°C		125		MHz
I_L	Insertion loss	$R_L = 50\ \Omega$, $C_L = 5\text{ pF}$ $V_S = 6\text{ V}$, $f = 1\text{ MHz}$	25°C		-0.25		dB
ACPSRR	AC Power Supply Rejection Ratio	$V_{PP} = 0.62\text{ V}$ on V_{DD} and V_{SS} $R_L = 50\ \Omega$, $C_L = 5\text{ pF}$, $f = 1\text{ MHz}$ Refer to ACPSRR	25°C		-70		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 6\text{ V}$, $V_{BIAS} = 6\text{ V}$ $R_L = 10\text{ k}\Omega$, $C_L = 5\text{ pF}$, $f = 20\text{ Hz}$ to 20 kHz Refer to THD + Noise	25°C		0.001		%
$C_{S(OFF)}$	Source off capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$	25°C		35		pF
$C_{D(OFF)}$	Drain off capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$	25°C		50		pF
$C_{S(ON)}$, $C_{D(ON)}$	On capacitance	$V_S = 6\text{ V}$, $f = 1\text{ MHz}$	25°C		145		pF

6.14 Typical Characteristics

at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

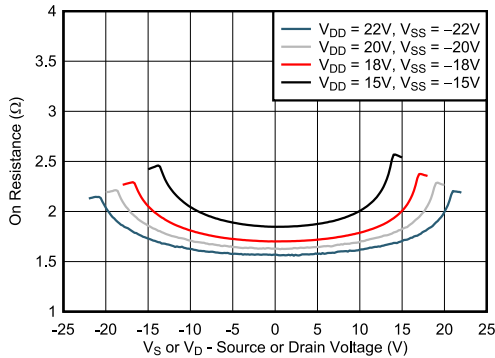


Figure 6-1. On-Resistance vs. Source or Drain Voltage – Dual Supply

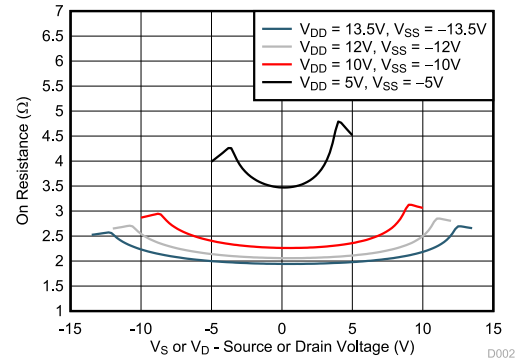


Figure 6-2. On-Resistance vs. Source or Drain Voltage – Dual Supply

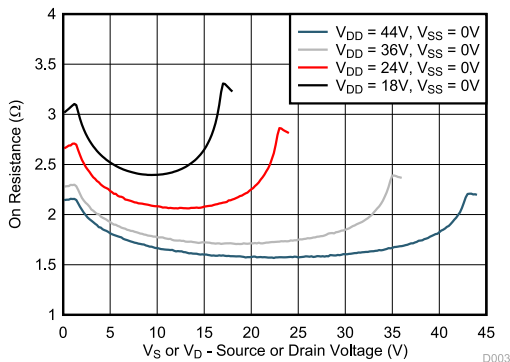


Figure 6-3. On-Resistance vs. Source or Drain Voltage – Single Supply

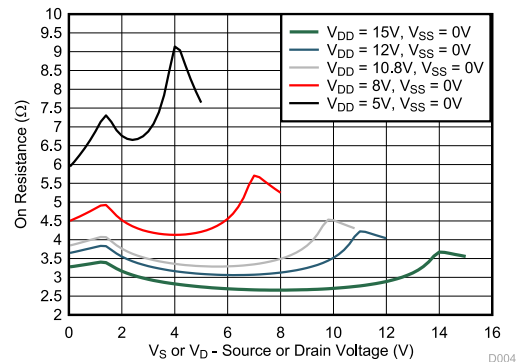


Figure 6-4. On-Resistance vs. Source or Drain Voltage – Single Supply

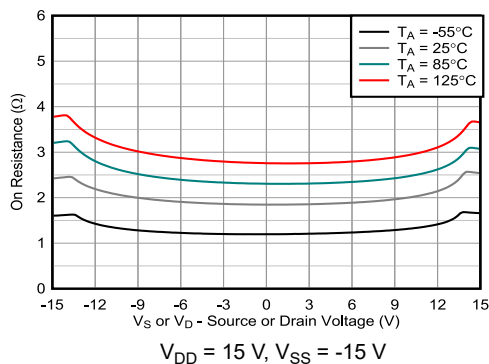


Figure 6-5. On-Resistance vs Temperature

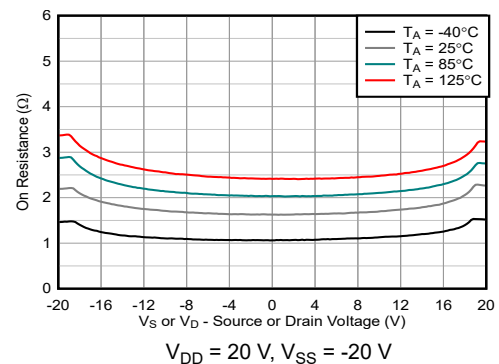


Figure 6-6. On-Resistance vs Temperature

6.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

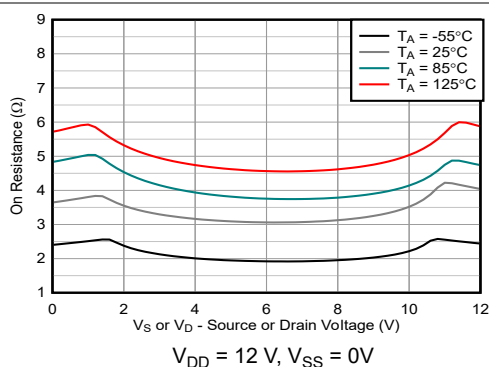


Figure 6-7. On-Resistance vs Temperature

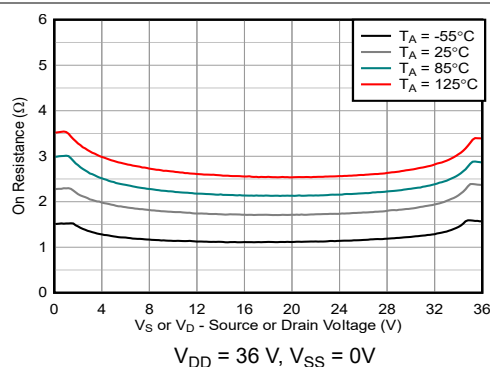


Figure 6-8. On-Resistance vs Temperature

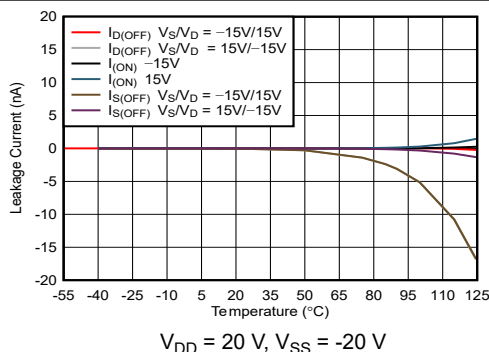


Figure 6-9. Leakage Current vs Temperature

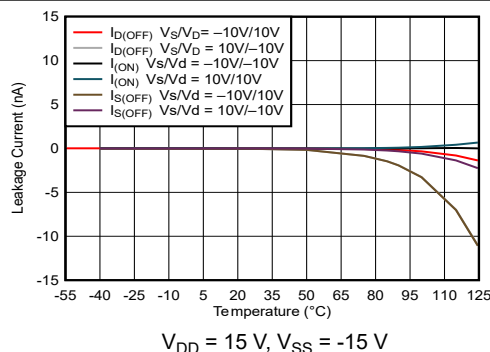


Figure 6-10. Leakage Current vs Temperature

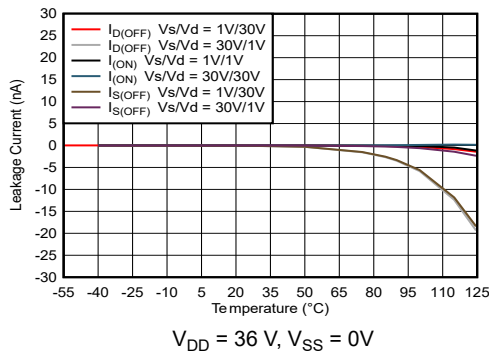


Figure 6-11. Leakage Current vs Temperature

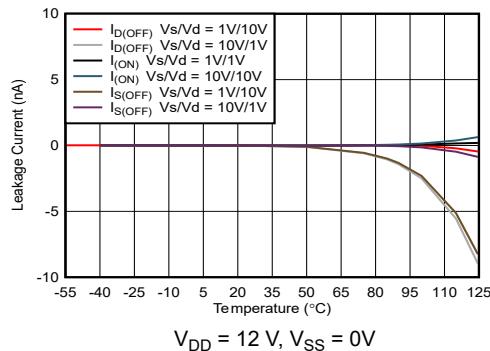


Figure 6-12. Leakage Current vs Temperature

6.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

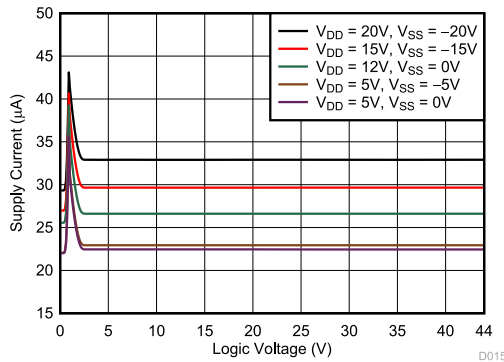


Figure 6-13. Supply Current vs. Logic Voltage

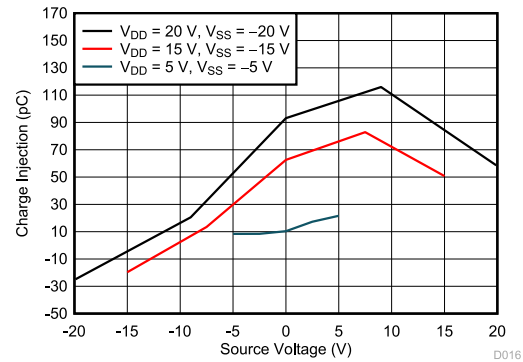


Figure 6-14. Charge Injection vs. Source Voltage – Dual Supply

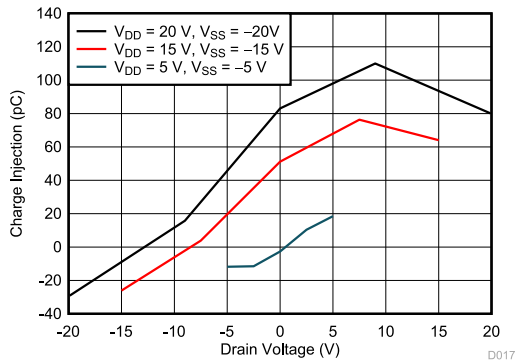


Figure 6-15. Charge Injection vs. Drain Voltage – Dual Supply

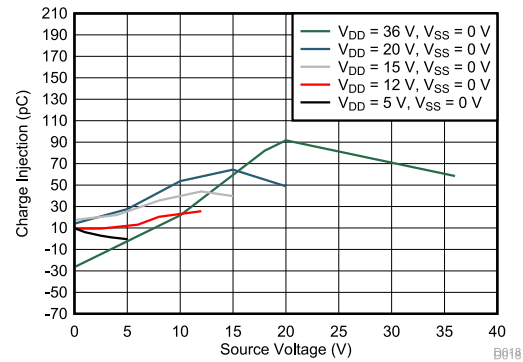


Figure 6-16. Charge Injection vs. Source Voltage – Single Supply

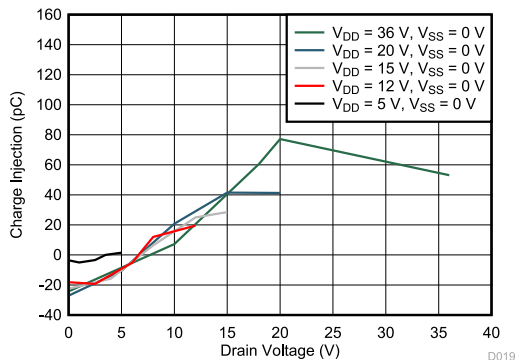


Figure 6-17. Charge Injection vs. Drain Voltage – Single Supply

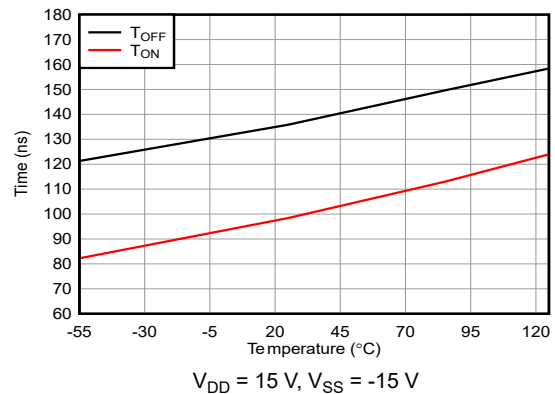


Figure 6-18. T_{ON} and T_{OFF} vs. Temperature

6.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

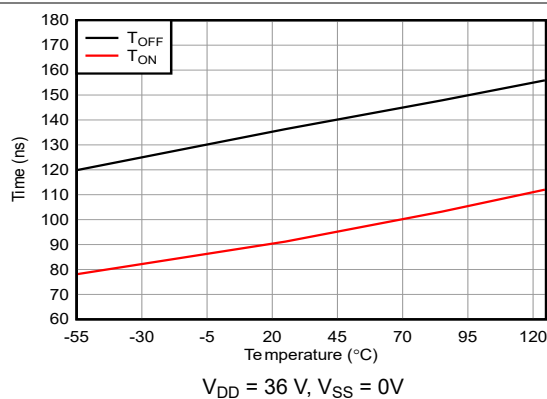


Figure 6-19. T_{ON} and T_{OFF} vs. Temperature

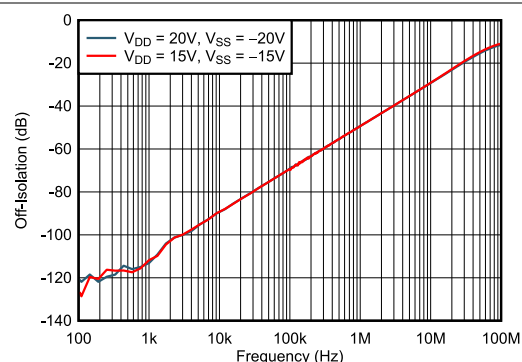


Figure 6-20. Off-Isolation vs Frequency

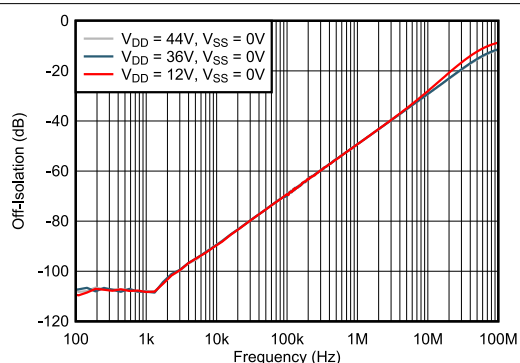


Figure 6-21. Off-Isolation vs Frequency

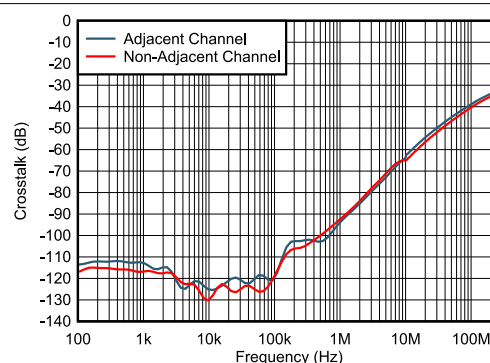


Figure 6-22. Crosstalk vs Frequency

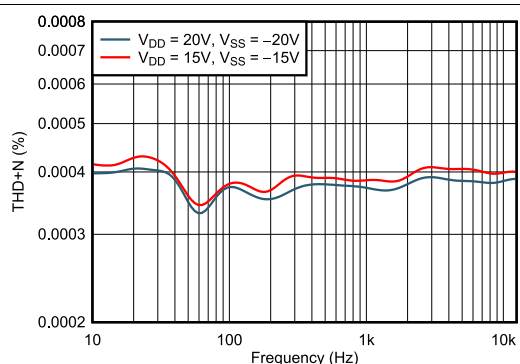


Figure 6-23. THD+N vs. Frequency (Dual Supplies)

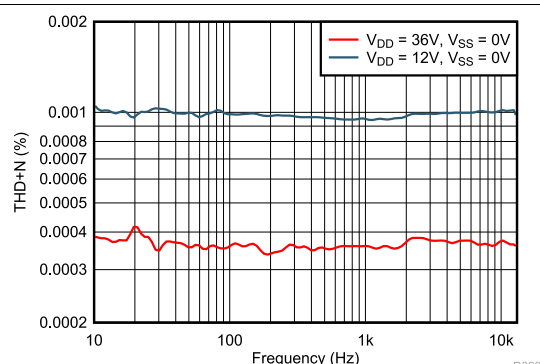


Figure 6-24. THD+N vs. Frequency (Single Supplies)

6.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

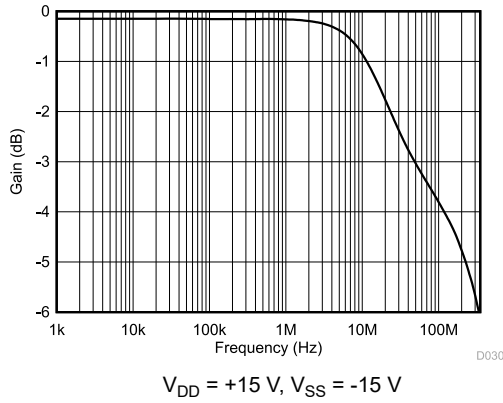


図 6-25. On Response vs Frequency

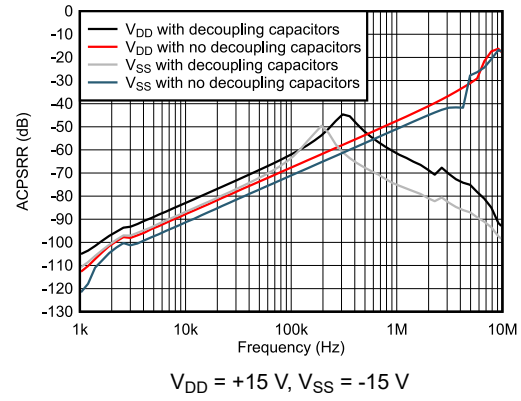


図 6-26. ACPSRR vs. Frequency

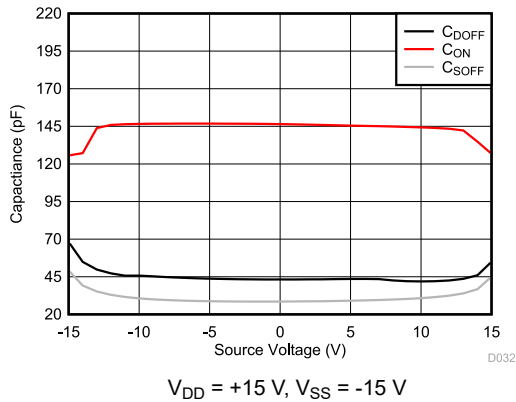


図 6-27. Capacitance vs. Source Voltage or Drain Voltage

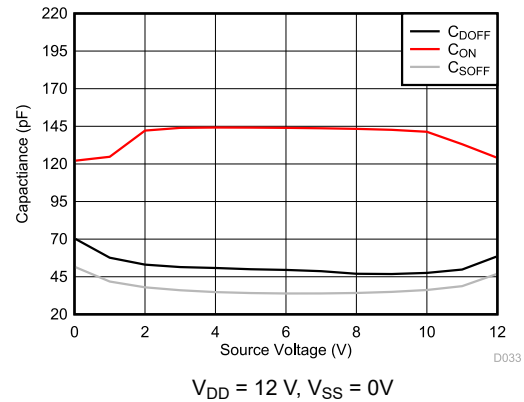
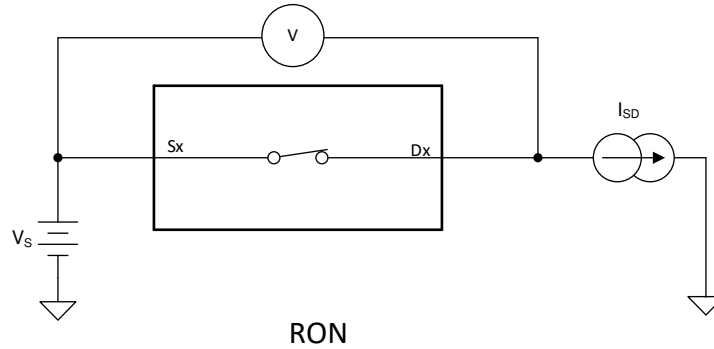


図 6-28. Capacitance vs. Source Voltage or Drain Voltage

7 Parameter Measurement Information

7.1 On-Resistance

The on-resistance of a device is the ohmic resistance between the source (Sx) and drain (Dx) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol R_{ON} is used to denote on-resistance. 7-1 shows the measurement setup used to measure R_{ON} . Voltage (V) and current (I_{SD}) are measured using this setup, and R_{ON} is computed with $R_{ON} = V / I_{SD}$:



7-1. On-Resistance Measurement Setup

7.2 Off-Leakage Current

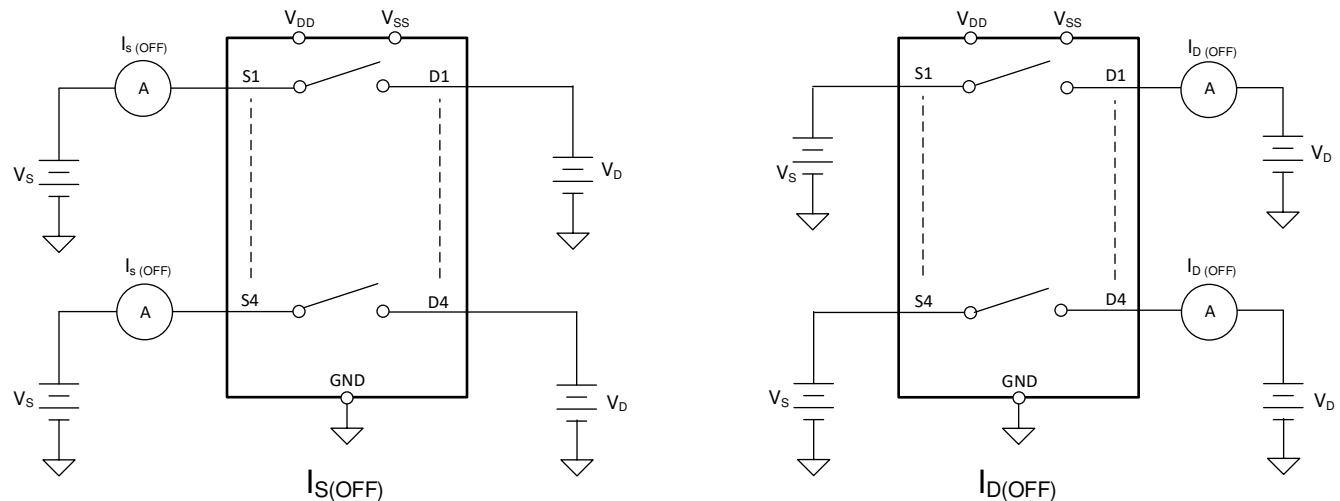
There are two types of leakage currents associated with a switch during the off state:

1. Source off-leakage current.
2. Drain off-leakage current.

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol $I_{S(OFF)}$.

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol $I_{D(OFF)}$.

7-2 shows the setup used to measure both off-leakage currents.



7-2. Off-Leakage Measurement Setup

7.3 On-Leakage Current

Source on-leakage current is defined as the leakage current flowing into or out of the source pin when the switch is on. This current is denoted by the symbol $I_{S(ON)}$.

Drain on-leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is on. This current is denoted by the symbol $I_{D(ON)}$.

Either the source pin or drain pin is left floating during the measurement. Figure 7-3 shows the circuit used for measuring the on-leakage current, denoted by $I_{S(ON)}$ or $I_{D(ON)}$.

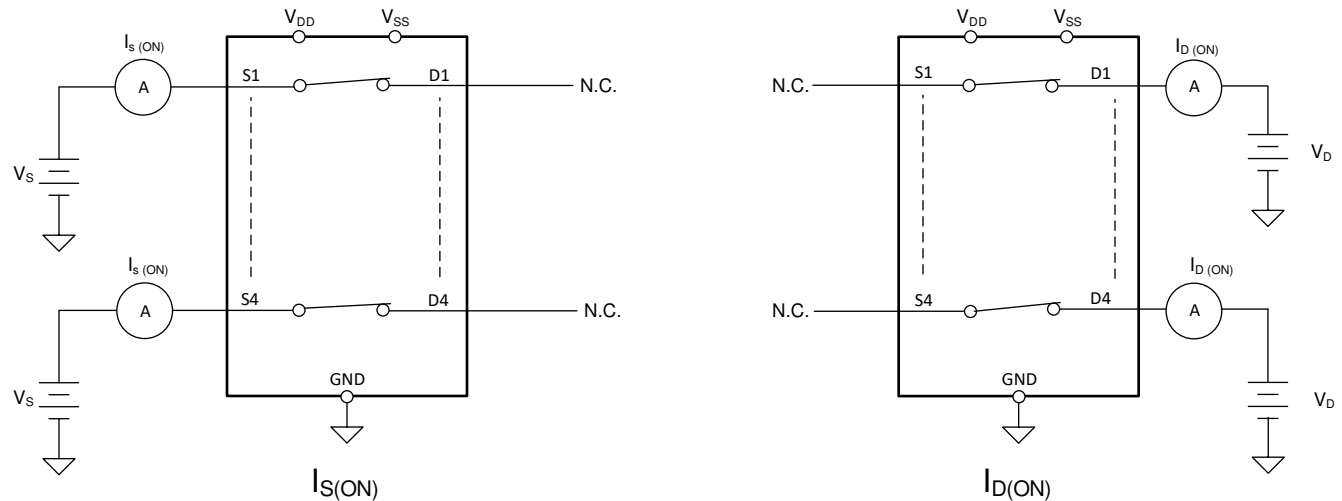


Figure 7-3. On-Leakage Measurement Setup

7.4 t_{ON} and t_{OFF} Time

Turn-on time is defined as the time taken by the output of the device to rise to 90% after the enable has risen past the logic threshold. The 90% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. Figure 7-4 shows the setup used to measure turn-on time, denoted by the symbol t_{ON} .

Turn-off time is defined as the time taken by the output of the device to fall to 10% after the enable has fallen past the logic threshold. The 10% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. Figure 7-4 shows the setup used to measure turn-off time, denoted by the symbol t_{OFF} .

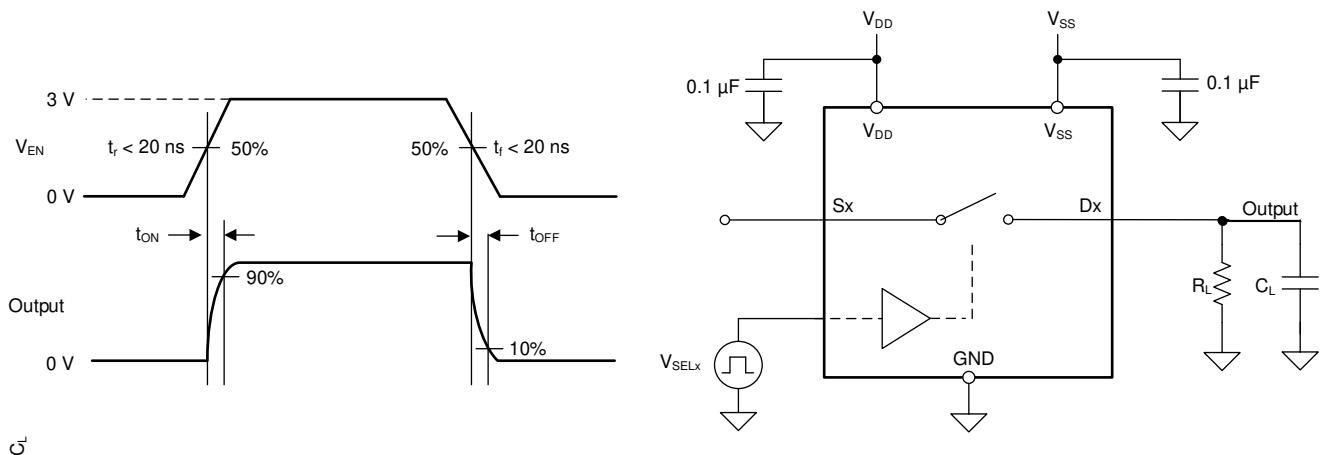
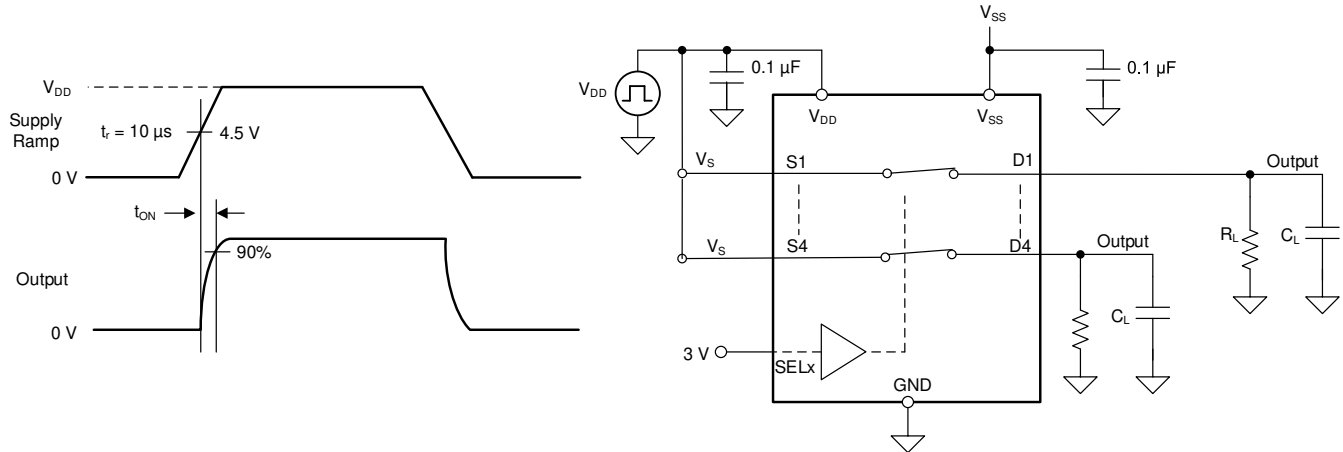


Figure 7-4. Turn-On and Turn-Off Time Measurement Setup

7.5 $t_{ON(VDD)}$ Time

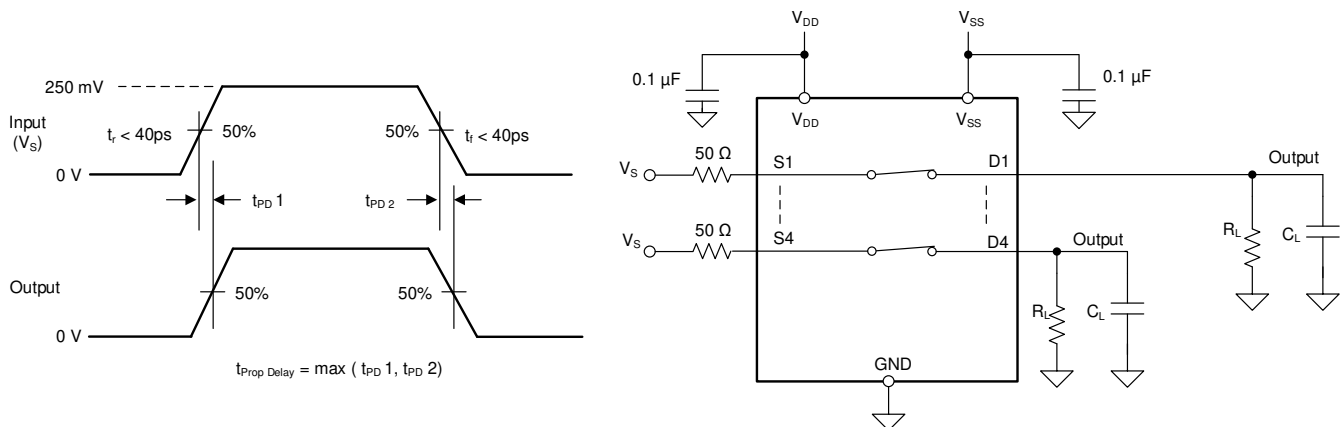
The $t_{ON(VDD)}$ time is defined as the time taken by the output of the device to rise to 90% after the supply has risen past the supply threshold. The 90% measurement is used to provide the timing of the device turning on in the system. 7-5 shows the setup used to measure turn on time, denoted by the symbol $t_{ON(VDD)}$.



7-5. $t_{ON(VDD)}$ Time Measurement Setup

7.6 Propagation Delay

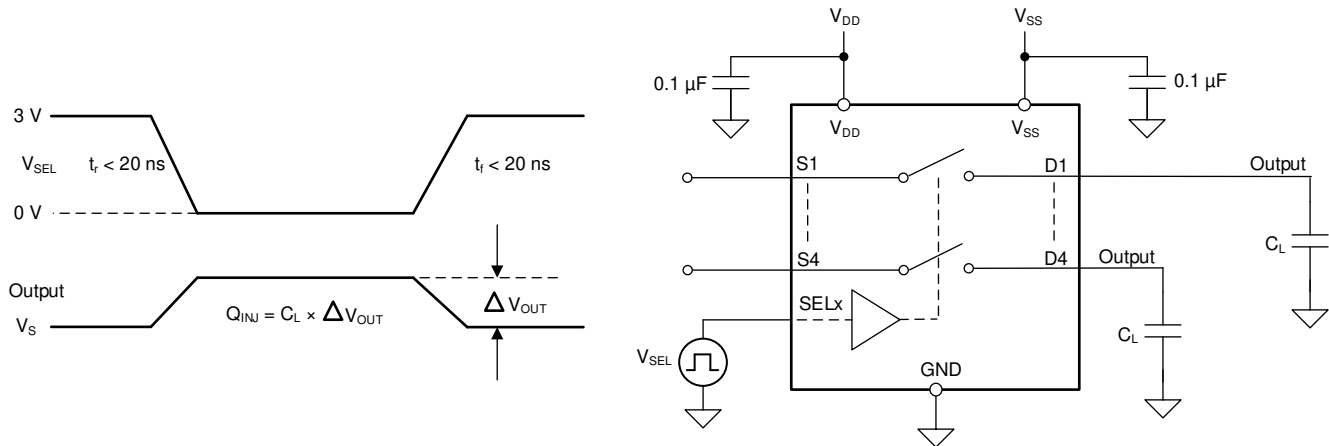
Propagation delay is defined as the time taken by the output of the device to rise or fall 50% after the input signal has risen or fallen past the 50% threshold. 7-6 shows the setup used to measure propagation delay, denoted by the symbol t_{PD} .



7-6. Propagation Delay Measurement Setup

7.7 Charge Injection

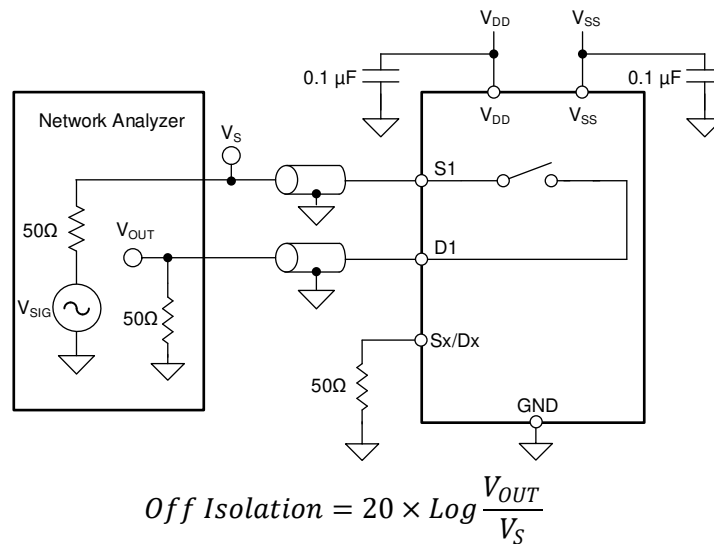
The TMUX721x devices have a transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol Q_C . 7-7 shows the setup used to measure charge injection from source (Sx) to drain (Dx).



7-7. Charge-Injection Measurement Setup

7.8 Off Isolation

Off isolation is defined as the ratio of the signal at the drain pin (Dx) of the device when a signal is applied to the source pin (Sx) of an off-channel. The characteristic impedance, Z_0 , for the measurement is 50 Ω . 7-8 shows the setup used to measure off isolation. Use off isolation equation to compute off isolation.



$$\text{Off Isolation} = 20 \times \text{Log} \frac{V_{OUT}}{V_S}$$

7-8. Off Isolation Measurement Setup

7.9 Channel-to-Channel Crosstalk

Crosstalk is defined as the ratio of the signal at the drain pin (Dx) of a different channel, when a signal is applied at the source pin (Sx) of an on-channel. The characteristic impedance, Z_0 , for the measurement is 50 Ω . 図 7-9 shows the setup used to measure, and the equation used to compute crosstalk.

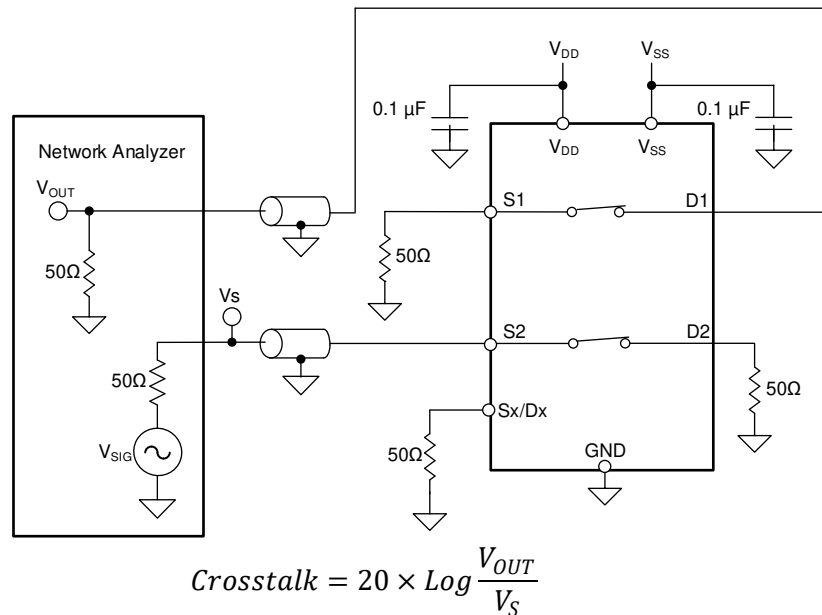


図 7-9. Channel-to-Channel Crosstalk Measurement Setup

7.10 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by less than 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (Dx) of the device. The characteristic impedance, Z_0 , for the measurement is 50 Ω . 図 7-10 shows the setup used to measure bandwidth.

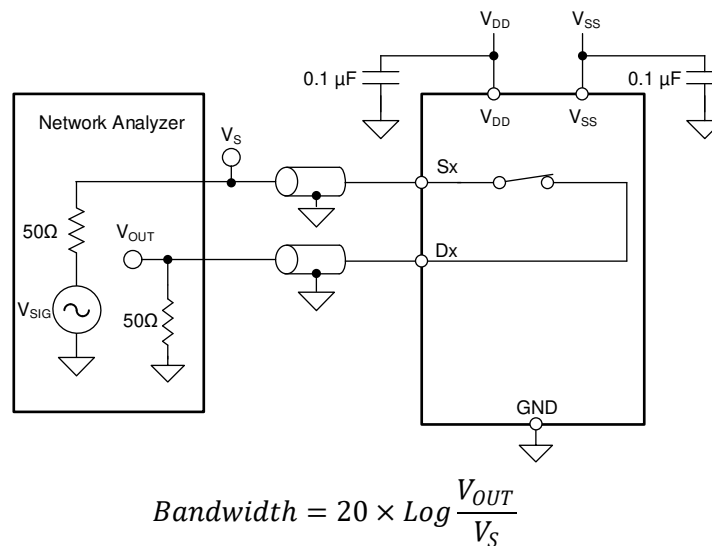
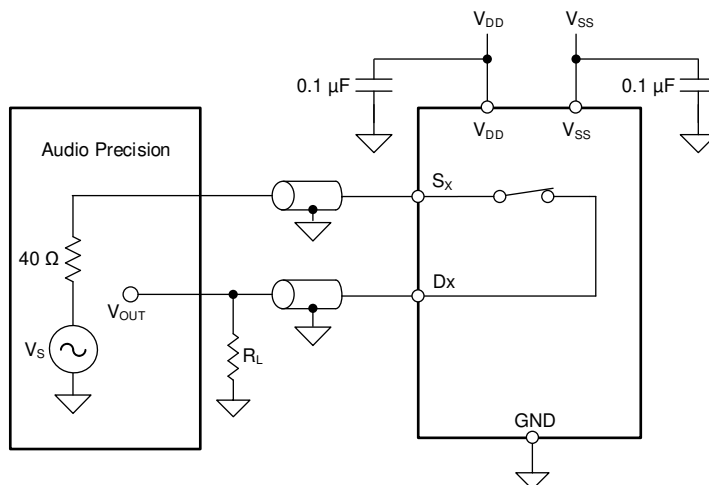


図 7-10. Bandwidth Measurement Setup

7.11 THD + Noise

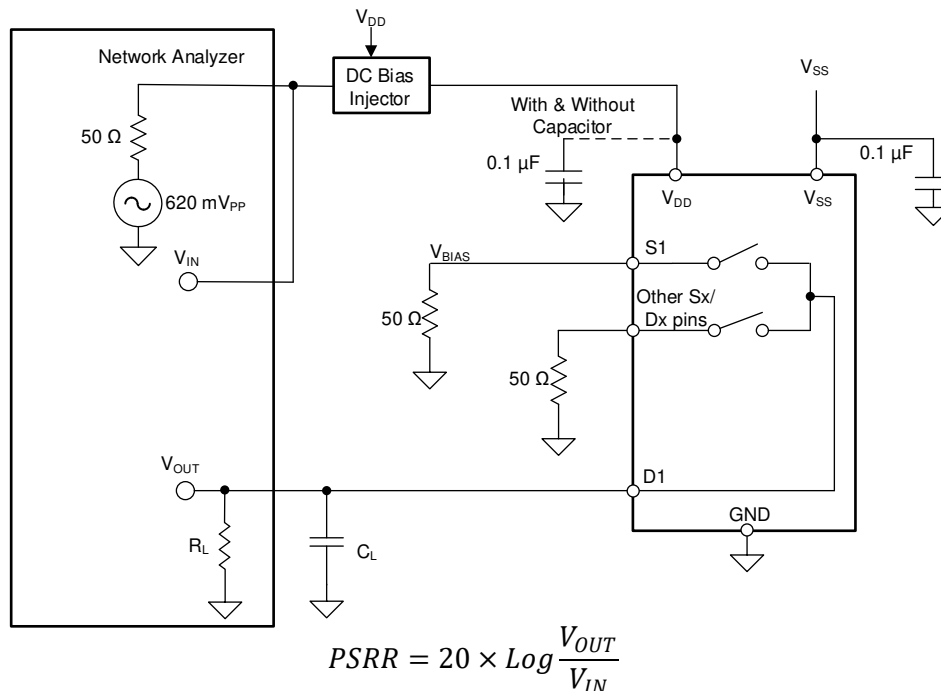
The total harmonic distortion (THD) of a signal is a measurement of the harmonic distortion, and is defined as the ratio of the sum of the powers of all harmonic components to the power of the fundamental frequency at the mux output. The on-resistance of the device varies with the amplitude of the input signal and results in distortion when the drain pin is connected to a low-impedance load. Total harmonic distortion plus noise is denoted as THD + N.



7-11. THD + N Measurement Setup

7.12 Power Supply Rejection Ratio (PSRR)

PSRR measures the ability of a device to prevent noise and spurious signals that appear on the supply voltage pin from coupling to the output of the switch. The DC voltage on the device supply is modulated by a sine wave of 100 mV_{PP}. The ratio of the amplitude of signal on the output to the amplitude of the modulated signal is the AC PSRR.



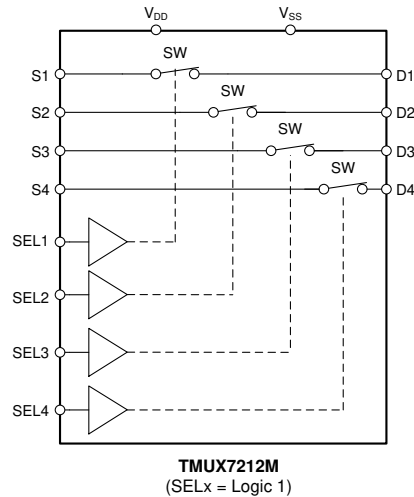
7-12. AC PSRR Measurement Setup

8 Detailed Description

8.1 Overview

The TMUX7212 is a 1:1 (SPST), 4-channel switch. The device has four independently selectable single-pole, single-throw switches that are turned-on or turned-off based on the state of the corresponding select pin.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Bidirectional Operation

The TMUX721x conducts equally well from source (Sx) to drain (Dx) or from drain (Dx) to source (Sx). Each channel has similar characteristics in both directions and supports both analog and digital signals.

8.3.2 Rail-to-Rail Operation

The valid signal path input and output voltage for TMUX721x ranges from V_{SS} to V_{DD} .

8.3.3 1.8V Logic Compatible Inputs

The TMUX721x device has 1.8V logic compatible control for all logic control inputs. 1.8V logic level inputs allows the TMUX721x to interface with processors that have lower logic I/O rails and eliminates the need for an external translator, which saves both space and bill of material cost. For more information on 1.8V logic implementations refer to [Simplifying Design with 1.8V logic Muxes and Switches](#).

8.3.4 Integrated Pull-Down Resistor on Logic Pins

The TMUX721x has internal weak pull-down resistors to GND to ensure the logic pins are not left floating. The value of this pull-down resistor is approximately 4M Ω , but is clamped to about 1 μ A at higher voltages. This feature integrates up to four external components and reduces system size and cost.

8.3.5 Fail-Safe Logic

The TMUX721x supports Fail-Safe Logic on the control input pins (SEL1, SEL2, SEL3, and SEL4) allowing for operation up to 44V, regardless of the state of the supply pin. This feature allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage. Fail-Safe Logic minimizes system complexity by removing the need for power supply sequencing on the logic control pins. For example, the Fail-Safe Logic feature allows the select pins of the TMUX721x to be ramped to 44V while V_{DD} and V_{SS} = 0V. The logic control inputs are protected against positive faults of up to 44V in powered-off condition, but do not offer protection against negative overvoltage conditions.

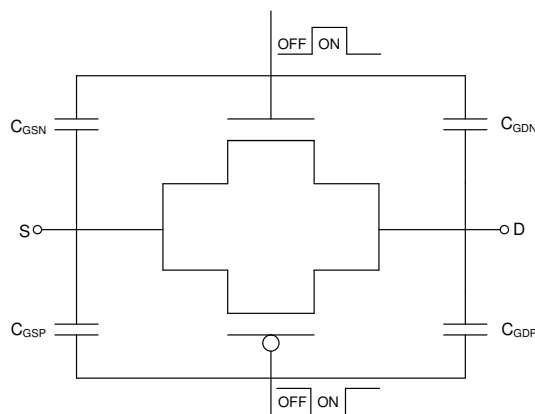
8.3.6 Latch-Up Immune

Latch-up is a condition where a low impedance path is created between a supply pin and ground. This condition is caused by a trigger (current injection or overvoltage), but once activated, the low impedance path remains even after the trigger is no longer present. This low impedance path may cause system upset or catastrophic damage due to excessive current levels. The latch-up condition typically requires a power cycle to eliminate the low impedance path.

The TMUX721x family of devices are constructed on silicon on insulator (SOI) based process where an oxide layer is added between the PMOS and NMOS transistor of each CMOS switch to prevent parasitic structures from forming. The oxide layer is also known as an insulating trench and prevents triggering of latch up events due to overvoltage or current injections. The latch-up immunity feature allows the TMUX721x family of switches and multiplexers to be used in harsh environments. For more information on latch-up immunity refer to [Using Latch Up Immune Multiplexers to Help Improve System Reliability](#).

8.3.7 Ultra-Low Charge Injection

✎ 8-1 shows how the TMUX721x devices have a transmission gate topology. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed.



✎ 8-1. Transmission Gate Topology

The TMUX721x contains specialized architecture to reduce charge injection on the Drain (Dx). To further reduce charge injection in a sensitive application, a compensation capacitor (C_p) can be added on the Source (Sx). This will ensure that excess charge from the switch transition will be pushed into the compensation capacitor on the Source (Sx) instead of the Drain (Dx). As a general rule, C_p should be 20x larger than the equivalent load capacitance on the Drain (Dx). ✎ 8-2 shows charge injection variation with different compensation capacitors on the Source side. This plot was captured on the TMUX7219M as part of the TMUX72xx family with a 100 pF load capacitance.

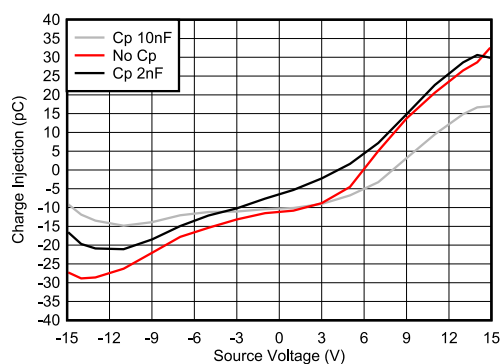


図 8-2. Charge Injection Compensation

8.4 Device Functional Modes

The TMUX721x device has four independently selectable single-pole, single-throw switches that are turned-on or turned-off based on the state of the corresponding select pin. The control pins can be as high as 44V.

The TMUX721x devices can operate without any external components except for the supply decoupling capacitors. The SELx pins have internal pull-down resistors of 4MΩ. If unused, then the SELx pin must be tied to GND for the device to not consume additional current as highlighted in [Implications of Slow or Floating CMOS Inputs](#). Unused signal path inputs (Sx or Dx) should be connected to GND.

8.5 Truth Tables

表 8-1 shows the truth tables for the TMUX7212.

表 8-1. TMUX7212 Truth Table

SEL x ⁽¹⁾	CHANNEL x
0	Channel x OFF
1	Channel x ON

(1) x denotes 1, 2, 3, or 4 for the corresponding channel.

9 Application and Implementation

注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくことになります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

9.1 Application Information

The TMUX721x is part of the precision switches and multiplexers family of devices. These devices operate with dual supplies (± 4.5 V to ± 22 V), a single supply (4.5 V to 44 V), or asymmetric supplies (such as $V_{DD} = 12$ V, $V_{SS} = -5$ V), and offer true rail-to-rail input and output. The TMUX721x offers low R_{ON} , low on and off leakage currents and ultra-low charge injection performance. These features makes the TMUX721x a family of precision, robust, high-performance analog multiplexer for high-voltage, industrial applications.

9.2 Typical Application

One example to take advantage of TMUX721x precision performance is the implementation of parametric measurement unit (PMU) in the semiconductor automatic test equipment (ATE) application.

In Automated Test Equipment (ATE) systems, the Parametric Measurement Unit (PMU) is tasked to measure device (DUT) parametric information in terms of voltage and current. When measuring voltage, current is applied at the DUT pin, and current range adjustment can be done through changing the value of the internal sense resistor. There is sometimes a need, depending on the DUT, to use even higher testing current than natively supported by the system. A 4 channel SPST switch, together with external higher current amplifier and resistor, can be used to achieve the flexibility. The PMU operating voltage is typically in mid voltage (up to 20 V). An appropriate switch like the TMUX721x with low leakage current (0.05 nA typical) works well in these applications to ensure measurement accuracy and low R_{ON} and flat $R_{ON_FLATNESS}$ allows the current range to be controlled more precisely. 図 9-1 shows simplified diagram of such implementations in memory and semiconductor test equipment.

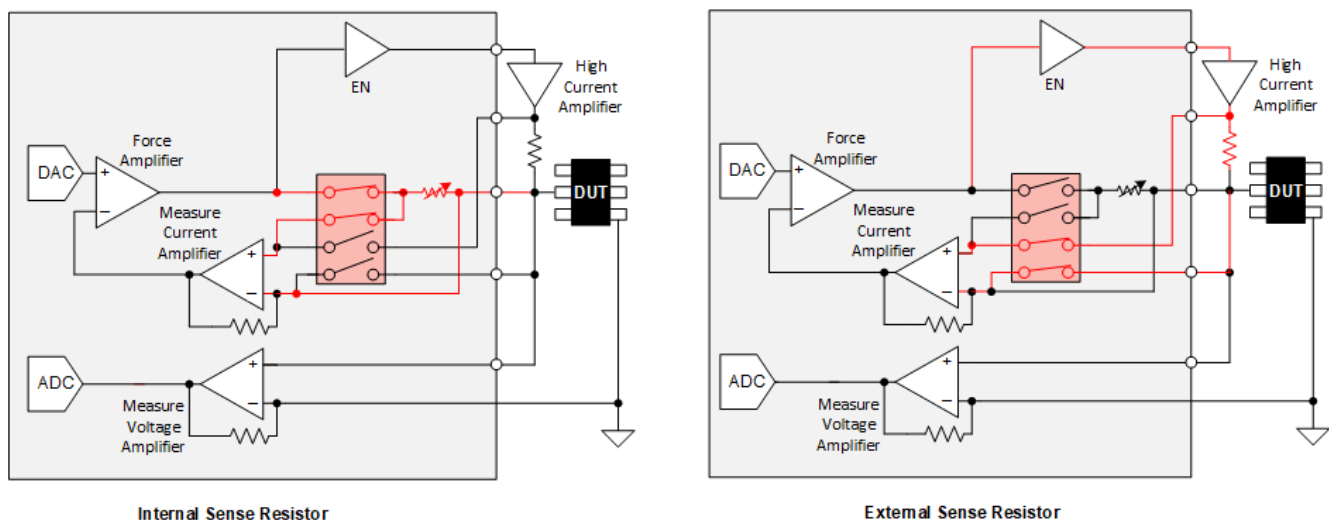


図 9-1. High Current Range Selection Using External Resistor

9.2.1 Design Requirements

For this design example, use the parameters listed in 表 9-1.

表 9-1. Design Parameters

PARAMETERS	VALUES
Supply (V_{DD})	20 V
Supply (V_{SS})	-10 V
Input / Output signal range	-10 V to 20 V (Rail-to-Rail)
Control logic thresholds	1.8 V compatible

9.2.2 Detailed Design Procedure

Figure 10-1 demonstrates how the TMUX721x can be used in semiconductor test equipment for high-precision, high-voltage, multi-channel measurement applications. The TMUX721x can support 1.8 V logic signals on the control input, allowing the device to interface with low logic controls of an FPGA or MCU. The TMUX721x can be operated without any external components except for the supply decoupling capacitors. The select pins have an internal pull-down resistor to prevent floating input logic. All inputs to the switch must fall within the recommend operating conditions of the TMUX721x including signal range and continuous current. For this design with a positive supply of 20 V on V_{DD} , and negative supply of -10 V on V_{SS} , the signal range can be 20 V to -10 V. The max continuous current (I_{DC}) can be up to 330 mA as shown in the *Recommended Operating Conditions* table for wide-range current measurement.

9.2.3 Application Curve

The TMUX721x have excellent charge injection performance and ultra-low leakage current, making them ideal choices to minimize sampling error for the sample and hold application.

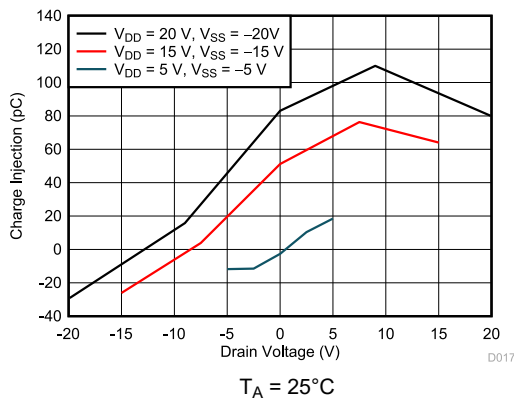


図 9-2. Charge Injection vs. Drain Voltage

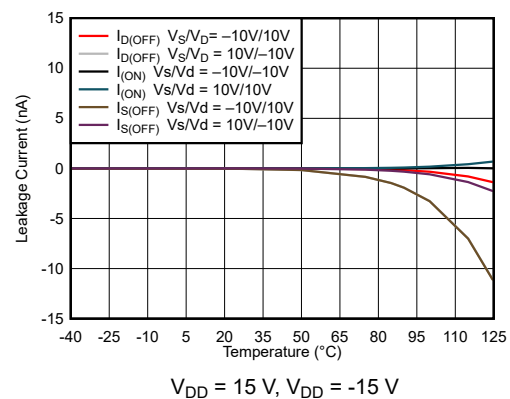


図 9-3. On-Leakage vs. Source or Drain Voltage

9.3 Power Supply Recommendations

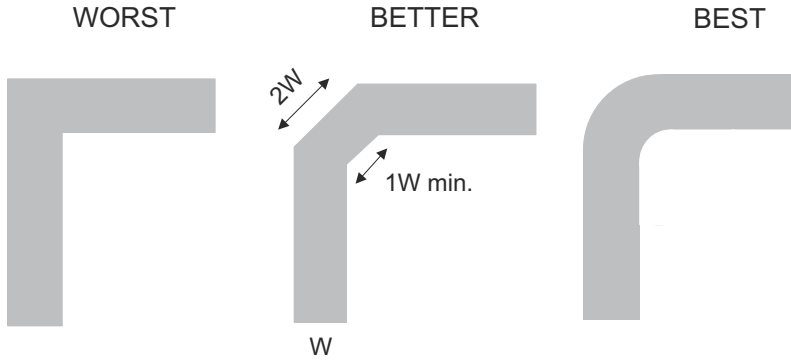
The TMUX721x device operates across a wide supply range of $\pm 4.5V$ to $\pm 22V$ (4.5V to 44V in single-supply mode). The device also perform well with asymmetrical supplies such as $V_{DD} = 12 V$ and $V_{SS} = -5 V$.

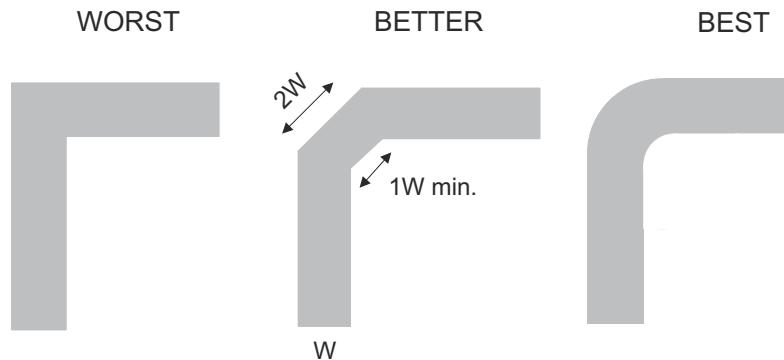
Power-supply bypassing improves noise margin and prevents switching noise propagation from the supply rails to other components. Good power-supply decoupling is important to achieve optimum performance. For improved supply noise immunity, use a supply decoupling capacitor ranging from 0.1 μF to 10 μF at both the V_{DD} and V_{SS} pins to ground. Place the bypass capacitors as close to the power supply pins of the device as possible using low-impedance connections. TI recommends using multi-layer ceramic chip capacitors (MLCCs) that offer low equivalent series resistance (ESR) and inductance (ESL) characteristics for power-supply decoupling purposes. For very sensitive systems, or for systems in harsh noise environments, avoiding the use of vias for connecting the capacitors to the device pins may offer superior noise immunity. The use of multiple vias in

parallel lowers the overall inductance and is beneficial for connections to ground and power planes. Always ensure the ground (GND) connection is established before supplies are ramped.

9.4 Layout

9.4.1 Layout Guidelines

When a PCB trace turns a corner at a 90° angle, a reflection can occur. A reflection occurs primarily because of the change of width of the trace. At the apex of the turn, the trace width increases to 1.414 times the width. This increase upsets the transmission-line characteristics, especially the distributed capacitance and self-inductance of the trace which results in the reflection. Not all PCB traces can be straight and therefore some traces must turn corners.  9-4 shows progressively better techniques of rounding corners. Only the last example (BEST) maintains constant trace width and minimizes reflections.



 9-4. Trace Example

Route high-speed signals using a minimum of vias and corners which reduces signal reflections and impedance changes. When a via must be used, increase the clearance size around it to minimize its capacitance. Each via introduces discontinuities in the signal's transmission line and increases the chance of picking up interference from the other layers of the board. Be careful when designing test points, through-hole pins are not recommended at high frequencies.

Some key considerations are:

- For reliable operation, connect a decoupling capacitor ranging from 0.1µF to 10µF between VDD/VSS and GND. We recommend a 0.1µF and 1µF capacitor, placing the lowest value capacitor as close to the pin as possible. Make sure that the capacitor voltage rating is sufficient for the supply voltage.
- Keep the input lines as short as possible.
- Use a solid ground plane to help reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.
- Using multiple vias in parallel will lower the overall inductance and is beneficial for connection to ground planes.

9.4.2 Layout Example

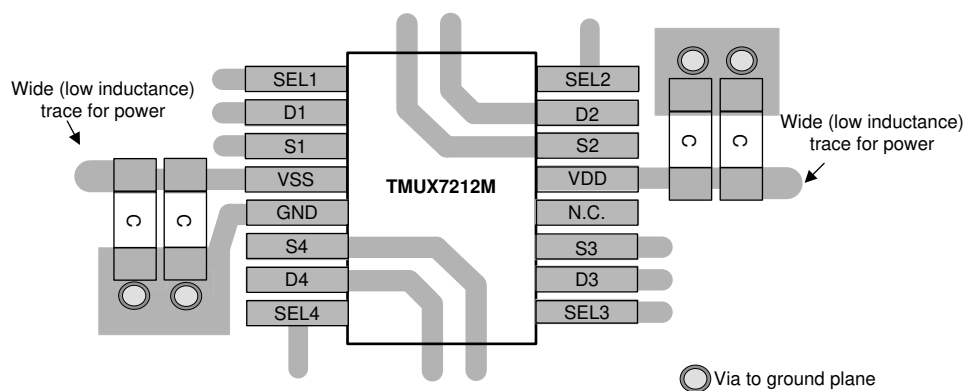


図 9-5. TMUX721x Layout Example

10 Device and Documentation Support

10.1 Documentation Support

10.1.1 Related Documentation

- Texas Instruments, [Using Latch Up Immune Multiplexers to Help Improve System Reliability](#) application note
- Texas Instruments, [Improve Stability Issues with Low CON Multiplexers](#) application brief
- Texas Instruments, [Improving Signal Measurement Accuracy in Automated Test Equipment](#) application brief
- Texas Instruments, [Sample & Hold Glitch Reduction for Precision Outputs Reference Design](#) reference guide
- Texas Instruments, [Simplifying Design with 1.8V logic Muxes and Switches](#) application brief
- Texas Instruments, [System-Level Protection for High-Voltage Analog Multiplexers](#) application note
- Texas Instruments, [True Differential, 4 x 2 MUX, Analog Front End, Simultaneous-Sampling ADC Circuit](#) application note
- Texas Instruments, [Quad Flatpack No-Lead Logic Packages](#) application note

10.2 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、www.tij.co.jp のデバイス製品フォルダを開いてください。[通知] をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取ることができます。変更の詳細については、改訂されたドキュメントに含まれている改訂履歴をご覧ください。

10.3 サポート・リソース

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10.4 Trademarks

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すべての商標は、それぞれの所有者に帰属します。

10.5 静電気放電に関する注意事項



この IC は、ESD によって破損する可能性があります。テキサス・インスツルメンツは、IC を取り扱う際には常に適切な注意を払うことを推奨します。正しい取り扱いおよび設置手順に従わない場合、デバイスを破損するおそれがあります。

ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

10.6 用語集

[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

11 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision C (August 2021) to Revision D (July 2024)	Page
• Update HBM ESD value on PW and RUM package.....	5
• Updated IIH max specification.....	7

Changes from Revision B (April 2021) to Revision C (August 2021)	Page
• TMUX7211, TMUX7212, TMUX7213 の QFN パッケージを追加.....	1
• Added ESD detail for RUM package.....	5
• Changed THD+N typical for 12V supply.....	14
• Added the <i>Integrated Pull-Down Resistor on Logic Pins</i> section.....	26
• Updated the <i>Ultra-Low Charge Injection</i> section.....	27
• Updated the <i>TMUX721x Layout Example</i> figure in the <i>Layout Example</i> section.....	33

Changes from Revision A (March 2021) to Revision B (April 2021)	Page
• Included Break-before-make time delay for TMUX7213	8
• Updated the <i>Charge Injection Compensation</i> figure.....	27

Changes from Revision * (December 2020) to Revision A (March 2021)	Page
• 「特長」セクションに WQFN の大電流サポートを追加.....	1
• Added thermal information for QFN package.....	6
• Updated I _{DC} specs for TSSOP package in <i>Source or Drain Continuous Current</i> table	6
• Added I _{DC} specs for QFN package in <i>Source or Drain Continuous Current</i> table	6
• Updated V _{DD} rise time value from 100ns to 1μs in T _{ON(VDD)} test condition.....	8
• Updated C _L value from 1nF to 100pF in Charge Injection test condition.....	8
• Updated Figure 7-10 <i>Leakage Current vs. Temperature</i>	15

12 Mechanical, Packaging, and Orderable Information

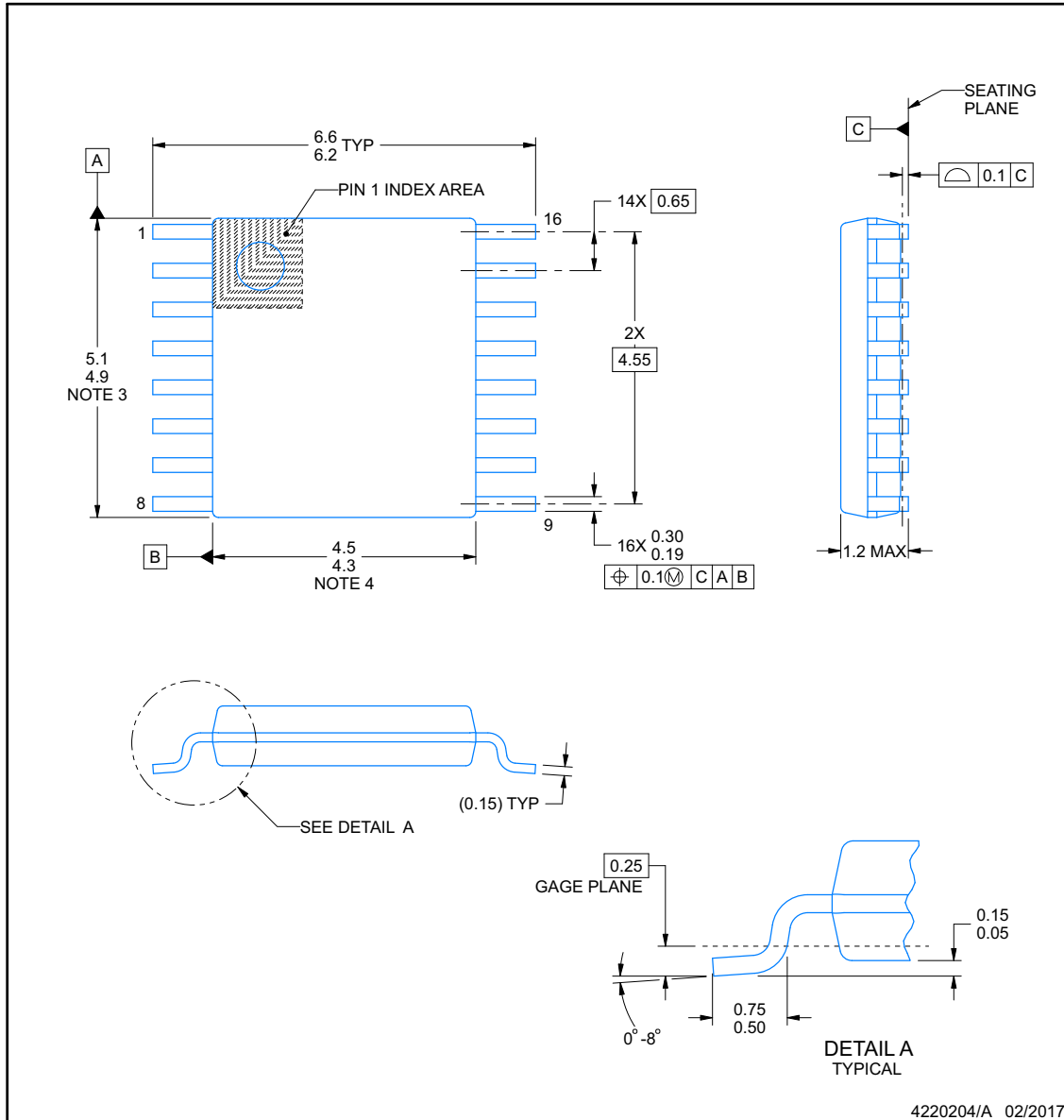
The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PW0016A



PACKAGE OUTLINE TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES:

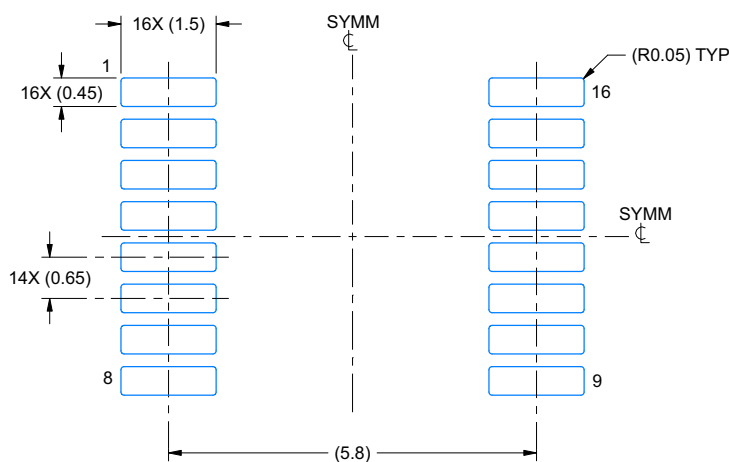
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153.

EXAMPLE BOARD LAYOUT

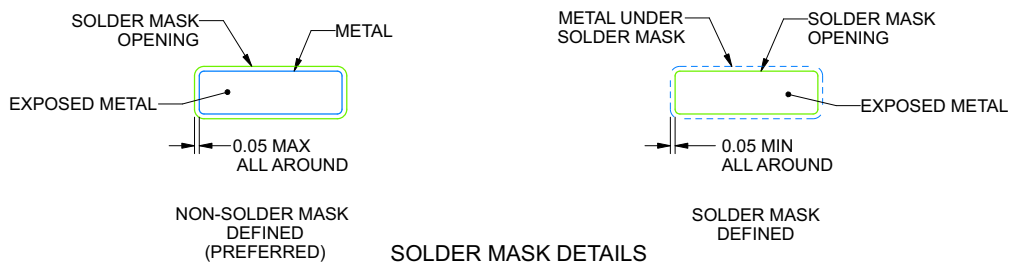
PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



4220204/A 02/2017

NOTES: (continued)

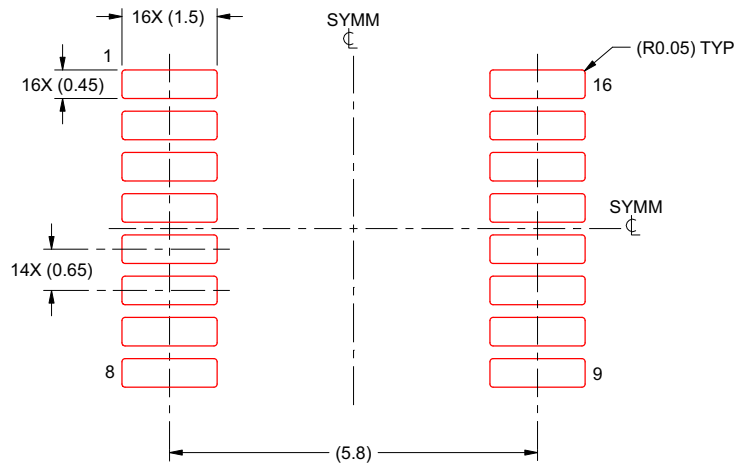
6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

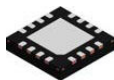


SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4220204/A 02/2017

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

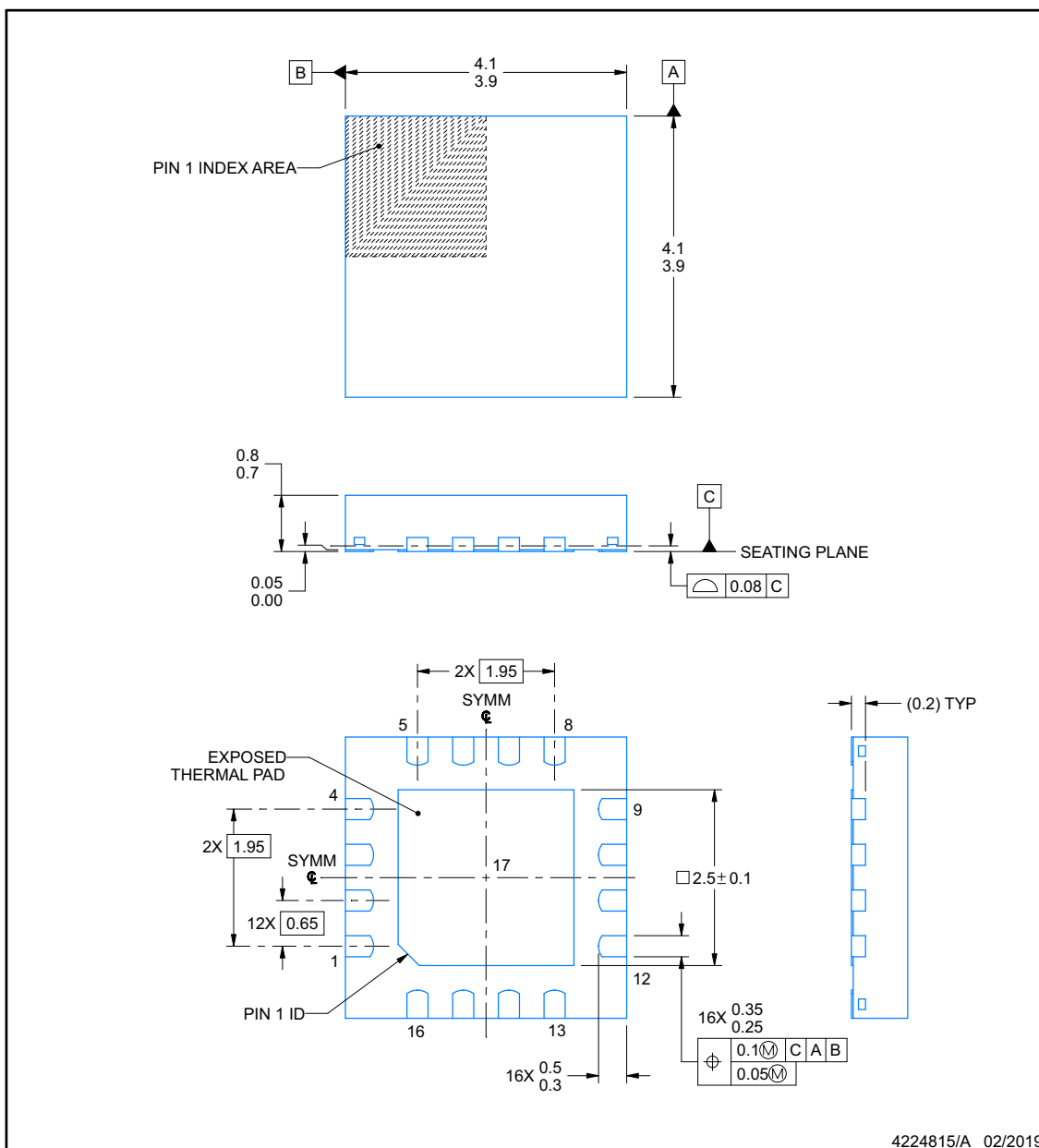


RUM0016E

PACKAGE OUTLINE

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

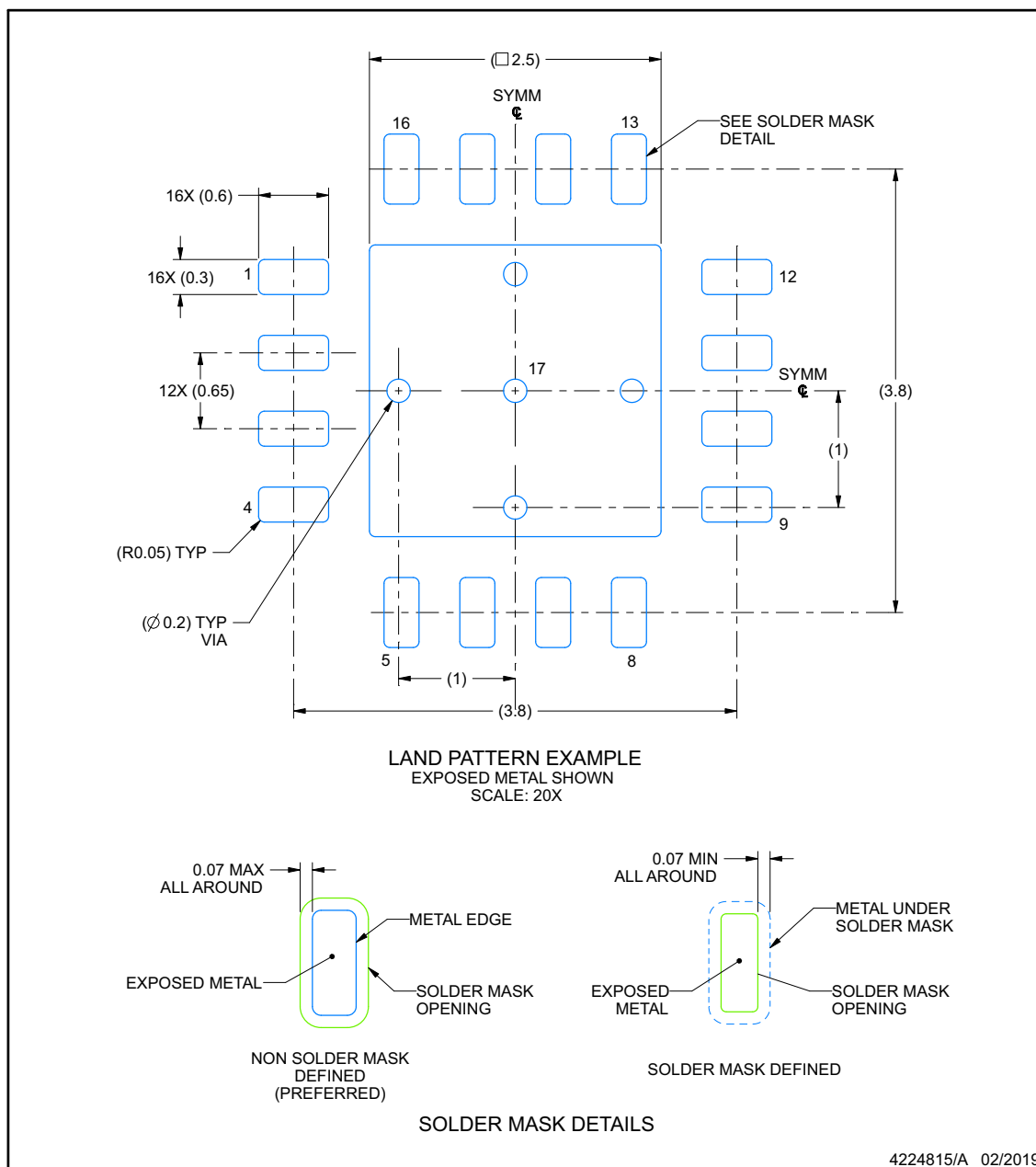
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

RUM0016E

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

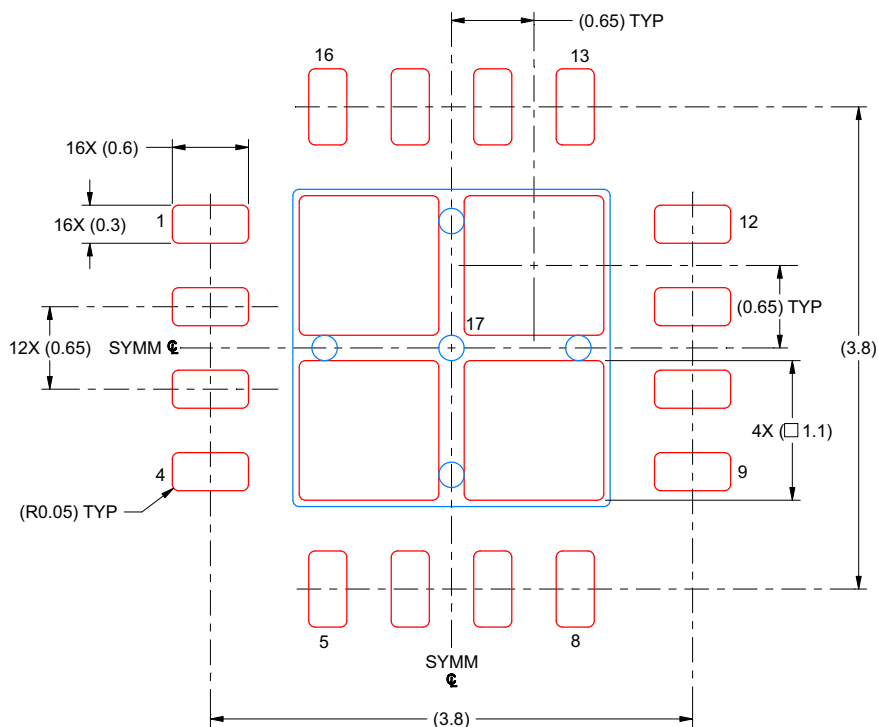
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/sluea271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RUM0016E

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 MM THICK STENCIL
SCALE: 20X
EXPOSED PAD 17
77% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

4224815/A 02/2019

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TMUX7211PWR	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X211
TMUX7211PWR.B	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X211
TMUX7211RUMR	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X211
TMUX7211RUMR.B	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X211
TMUX7211RUMRG4.B	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X211
TMUX7212PWR	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X212
TMUX7212PWR.B	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X212
TMUX7212RUMR	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X212
TMUX7212RUMR.B	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X212
TMUX7212RUMRG4.B	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X212
TMUX7213PWR	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X213
TMUX7213PWR.B	Active	Production	TSSOP (PW) 16	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	X213
TMUX7213RUMR	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X213
TMUX7213RUMR.B	Active	Production	WQFN (RUM) 16	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	TMUX X213

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TMUX7211PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX7211RUMR	WQFN	RUM	16	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
TMUX7212PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX7212RUMR	WQFN	RUM	16	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
TMUX7213PWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
TMUX7213RUMR	WQFN	RUM	16	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

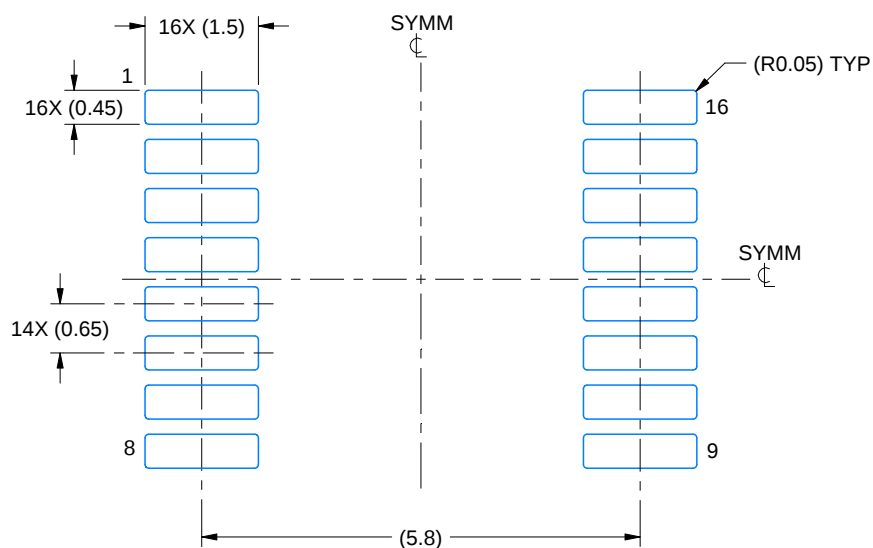
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMUX7211PWR	TSSOP	PW	16	2000	356.0	356.0	35.0
TMUX7211RUMR	WQFN	RUM	16	3000	367.0	367.0	35.0
TMUX7212PWR	TSSOP	PW	16	2000	356.0	356.0	35.0
TMUX7212RUMR	WQFN	RUM	16	3000	367.0	367.0	35.0
TMUX7213PWR	TSSOP	PW	16	2000	356.0	356.0	35.0
TMUX7213RUMR	WQFN	RUM	16	3000	367.0	367.0	35.0

EXAMPLE BOARD LAYOUT

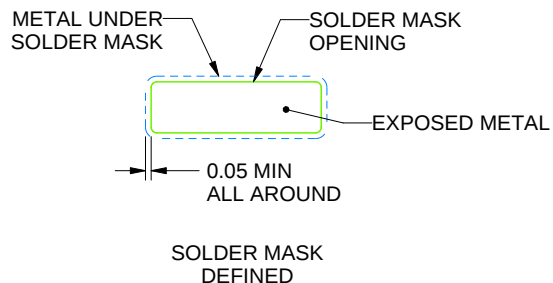
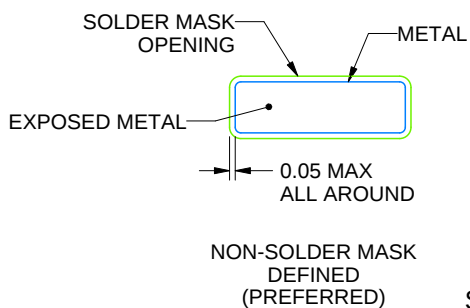
PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 10X



SOLDER MASK DETAILS

4220204/A 02/2017

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

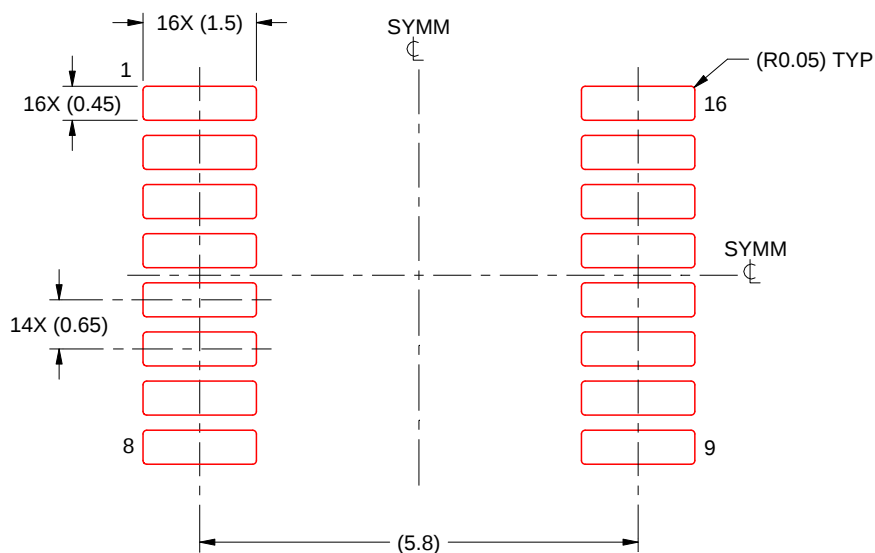
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE: 10X

4220204/A 02/2017

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

GENERIC PACKAGE VIEW

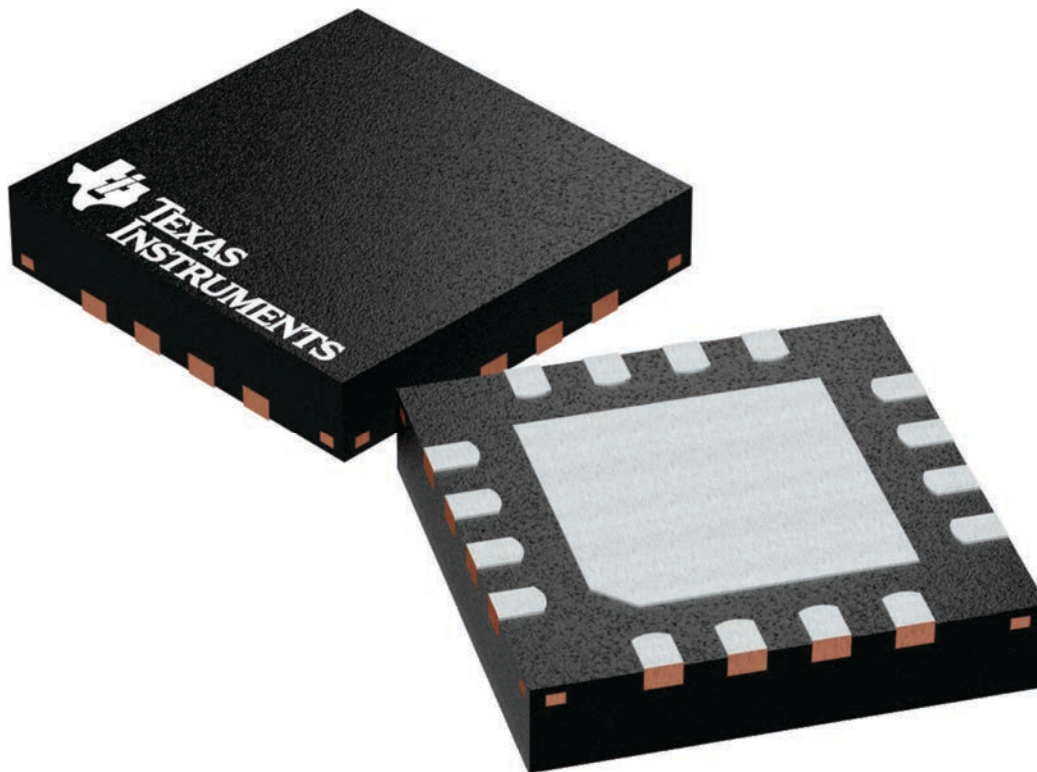
RUM 16

WQFN - 0.8 mm max height

4 x 4, 0.65 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224843/A

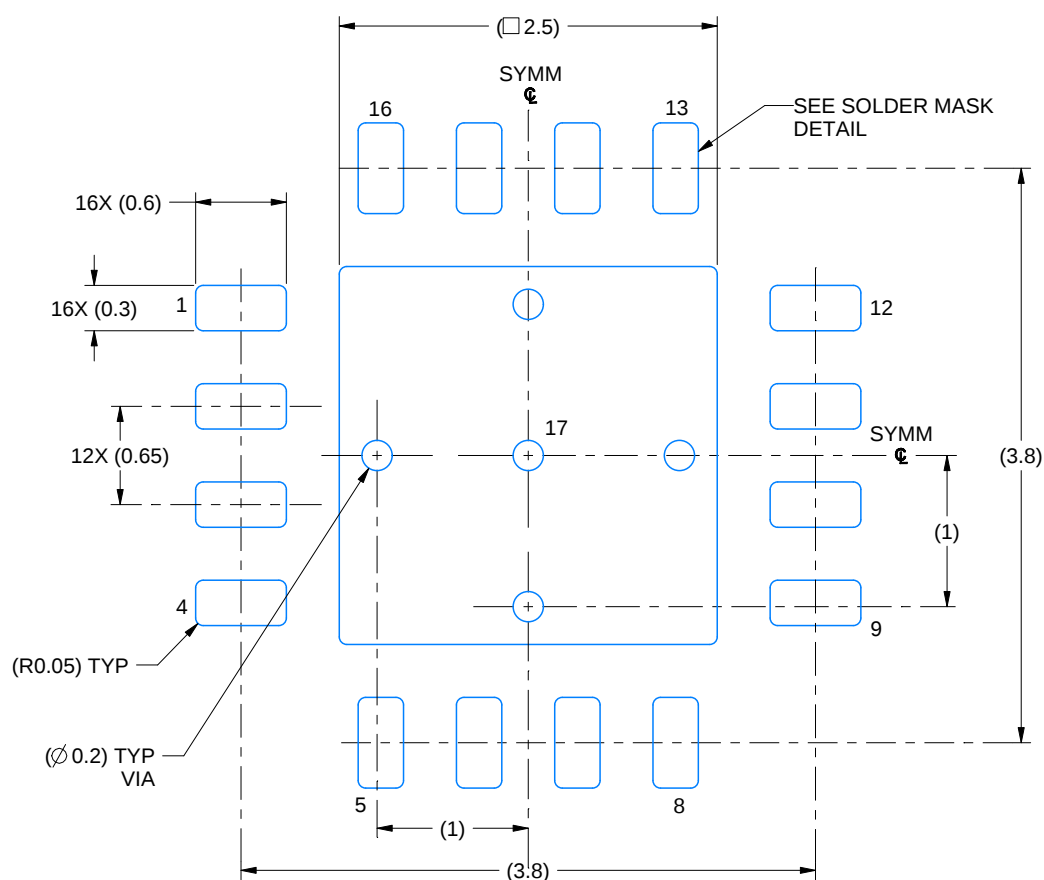


1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

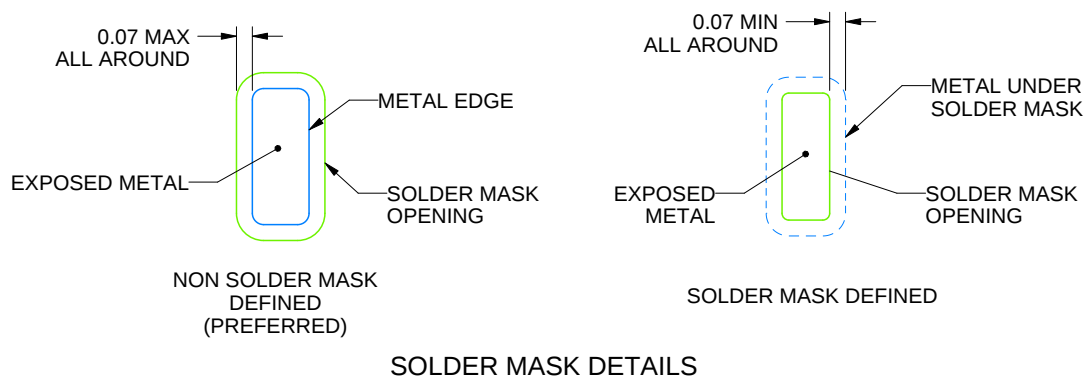
RUM0016E

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 20X



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NOTES: (continued)

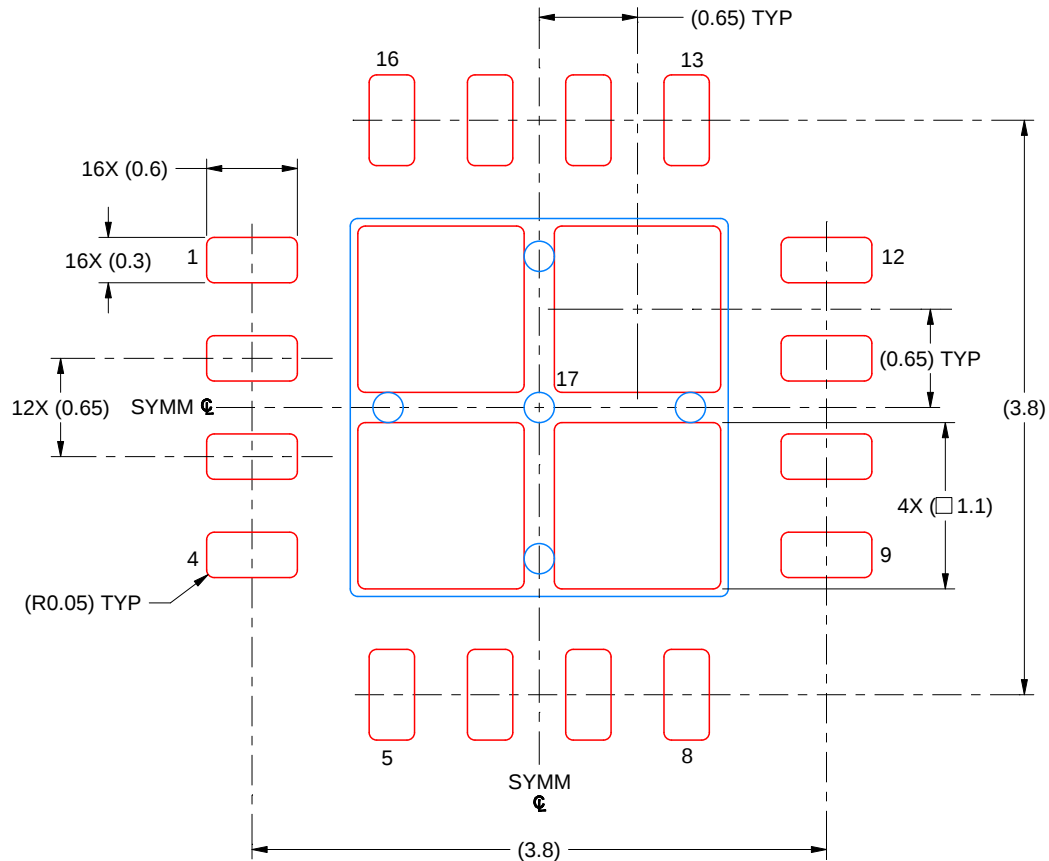
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

EXAMPLE STENCIL DESIGN

RUM0016E

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
 BASED ON 0.125 MM THICK STENCIL
 SCALE: 20X

EXPOSED PAD 17
 77% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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