



TPS56x219A、4.5V～17V入力、2A、3A同期整流降压型レギュレータ 8ピンSOT-23パッケージ

1 特長

- TPS562219A: 133mΩおよび80mΩ FET内蔵2Aコンバータ
- TPS563219A: 68mΩおよび39mΩ FET内蔵3Aコンバータ
- D-CAP2™モード制御、スイッチング周波数 650kHz
- 入力電圧範囲: 4.5V～17V
- 出力電圧範囲: 0.76V～7V
- 650kHzのスイッチング周波数
- 低いシャットダウン電流: 10μA未満
- 帰還電圧精度: 1% (25°C)
- プリバイアス出力電圧からのスタートアップ
- サイクルごとの過電流制限
- ヒカップ・モード低電圧保護
- 非ラッチ型のOVP、UVLO、およびTSD保護
- 調整可能なソフト・スタート
- パワー・グッド出力

2 アプリケーション

- デジタル・テレビ用電源
- 高精細 Blu-ray Disc™ プレーヤー
- ネットワーク・ホーム・ターミナル
- デジタル・セットトップ・ボックス (STB)

3 概要

TPS562219AおよびTPS563219Aは、シンプルで使いやすい2A、3A同期整流降压型コンバータで、8ピンSOT-23パッケージで供給されます。

最小の外部部品点数で動作し、スタンバイ電流が小さくなるよう最適化されています。

これらのスイッチ・モード電源(SMPS)デバイスは、D-CAP2™モード制御を採用し、高速の過渡応答を実現します。また、特殊ポリマーなどの低ESR (等価直列抵抗)出力コンデンサと超低ESRセラミック・コンデンサの両方を外部補償部品なしでサポートします。

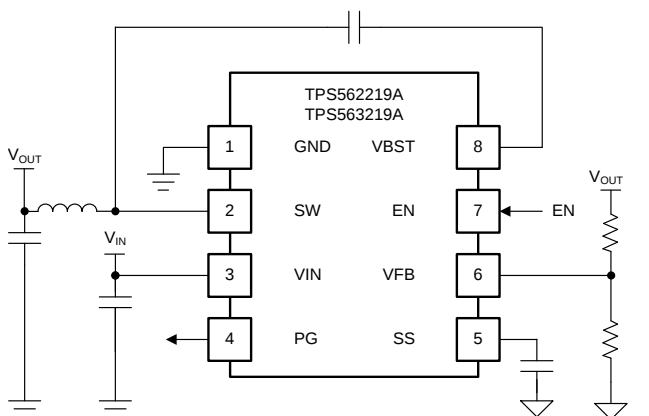
常に連続導通モードで動作するため、不連続導通モードと比較して軽負荷時の出力リップル電圧が軽減されます。TPS562219AおよびTPS563219Aは、8ピン1.6 × 2.9 (mm) SOT (DDF)パッケージで供給され、-40°C～85°Cの周囲温度範囲で仕様が規定されています。

製品情報⁽¹⁾

型番	パッケージ	本体サイズ(公称)
TPS562219A	SOT (8)	1.60mm×2.90mm
TPS563219A		

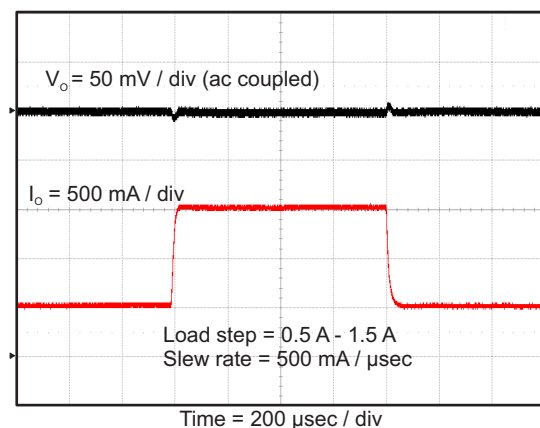
(1) 利用可能なすべてのパッケージについては、このデータシートの末尾にある注文情報を参照してください。

概略回路図



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TPS562219A 過渡応答



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4 改訂履歴

日付	改訂内容	注
2016年11月	*	初版

6 Specifications

6.1 Absolute Maximum Ratings

 $T_J = -40^{\circ}\text{C}$ to 150°C (unless otherwise noted) ⁽¹⁾

		MIN	MAX	UNIT
Input voltage range	VIN, EN	−0.3	19	V
	VBST	−0.3	25	V
	VBST (10 ns transient)	−0.3	27.5	V
	VBST (vs SW)	−0.3	6.5	V
	VFB, PG	−0.3	6.5	V
	SS	−0.3	5.5	V
	SW	−2	19	V
	SW (10 ns transient)	−3.5	21	V
Operating junction temperature, T_J		−40	150	$^{\circ}\text{C}$
Storage temperature, T_{stg}		−55	150	$^{\circ}\text{C}$

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

		VALUE	UNIT
$V_{\text{(ESD)}}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000
		Charged device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±500

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

 $T_J = -40^{\circ}\text{C}$ to 150°C (unless otherwise noted)

		MIN	MAX	UNIT
V_{IN}	Supply input voltage range	4.5	17	V
V_I	Input voltage range	VBST	−0.1	23
		VBST (10 ns transient)	−0.1	26
		VBST(vs SW)	−0.1	6
		EN	−0.1	17
		VFB, PG	−0.1	5.5
		SS	−0.1	5
		SW	−1.8	17
		SW (10 ns transient)	−3.5	20
T_A	Operating free-air temperature	−40	85	$^{\circ}\text{C}$

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		TPS562219A	TPS563219A	UNIT
		DDF (SOT)		
		8 PINS		
R _{θJA}	Junction-to-ambient thermal resistance	106.1	87.0	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	49.1	41.6	°C/W
R _{θJB}	Junction-to-board thermal resistance	10.9	14.6	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	8.6	4.7	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	10.8	14.6	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Electrical Characteristics

 $T_J = -40^{\circ}\text{C}$ to 150°C , $V_{IN} = 12\text{ V}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY CURRENT						
I _{VIN}	Operating – non-switching supply current	V _{IN} current, T _A = 25°C, EN = 5 V, V _{FB} = 0.8 V	650	900		μA
I _{VINSDN}	Shutdown supply current	V _{IN} current, T _A = 25°C, EN = 0 V	3	10		μA
LOGIC THRESHOLD						
V _{ENH}	EN high-level input voltage	EN	1.6			V
V _{ENL}	EN low-level input voltage	EN		0.6		V
R _{EN}	EN pin resistance to GND	V _{EN} = 12 V	225	450	900	kΩ
V _{FB} VOLTAGE AND DISCHARGE RESISTANCE						
V _{FBTH}	V _{FB} threshold voltage	T _A = 25°C, V _O = 1.05 V	757	765	773	mV
		T _A = 0°C to 85°C, V _O = 1.05 V ⁽¹⁾	753		777	
		T _A = -40°C to 85°C, V _O = 1.05 V ⁽¹⁾	751		779	
I _{VFB}	V _{FB} input current	V _{FB} = 0.8V, T _A = 25°C	0	±0.1		μA
MOSFET						
R _{DS(on)h}	High side switch resistance	T _A = 25°C, V _{BST} – SW = 5.5 V, TPS562219A	133			mΩ
		T _A = 25°C, V _{BST} – SW = 5.5 V, TPS563219A	68			
R _{DS(on)l}	Low side switch resistance	T _A = 25°C, TPS562219A	80			mΩ
		T _A = 25°C, TPS563219A	39			
CURRENT LIMIT						
I _{OCL}	Current limit ⁽¹⁾	DC current, V _{OUT} = 1.05 V, L1 = 2.2 μH, TPS562219A	2.5	3.2	4.3	A
		DC current, V _{OUT} = 1.05 V, L1 = 1.5 μH, TPS563219A	3.5	4.2	5.3	
THERMAL SHUTDOWN						
T _{SDN}	Thermal shutdown threshold ⁽¹⁾	Shutdown temperature	155			°C
		Hysteresis	35			
SOFT START						
I _{SS}	SS charge current	V _{SS} = 1.2 V	4.2	6	7.8	μA
POWER GOOD						
V _{THPG}	PG threshold	V _{FB} rising (Good)	85%	90%	95%	
		V _{FB} falling (Fault)		85%		
IPG	PG sink current	PG = 0.5 V	0.5	1		mA
OUTPUT UNDERVOLTAGE AND OVERVOLTAGE PROTECTION						
V _{OVP}	Output OVP threshold	OVP Detect	125% V _{fbth}			
V _{UVP}	Output UVP threshold	Hiccup detect	65% V _{fbth}			
t _{HiccupOn}	Hiccup Power On Time		1			cycle
t _{HiccupOff}	Hiccup Power Off Time		7			
UVLO						
UVLO	UVLO threshold	Wake up VIN voltage	3.45	3.75	4.05	V
		Hysteresis VIN voltage	0.13	0.32	0.55	

(1) Not production tested.

6.6 Timing Requirements

		MIN	TYP	MAX	UNIT
ON-TIME TIMER CONTROL					
t_{ON}	On time	$V_{IN} = 12\text{ V}$, $V_O = 1.05\text{ V}$	150		ns
$t_{OFF(MIN)}$	Minimum off time	$T_A = 25^{\circ}\text{C}$, $V_{FB} = 0.5\text{ V}$	260	310	ns

6.7 Typical Characteristics

6.7.1 TPS562219A Characteristics

$V_{IN} = 12V$ (unless otherwise noted)

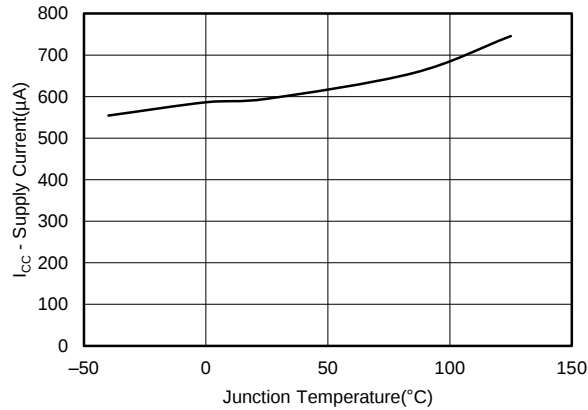


Figure 1. Supply Current vs Junction Temperature

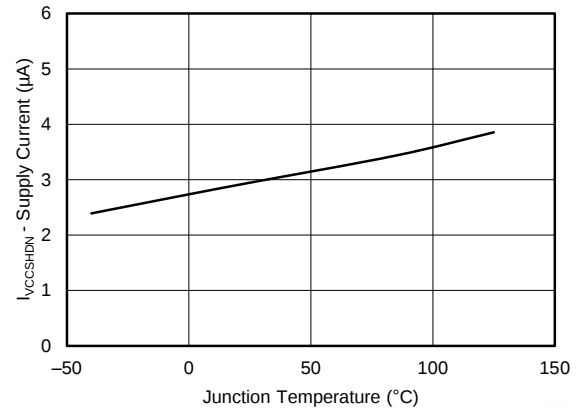


Figure 2. VIN Shutdown Current vs Junction Temperature

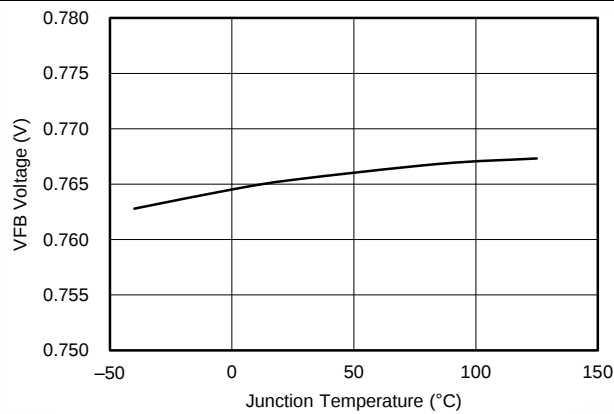


Figure 3. VFB Voltage vs Junction Temperature

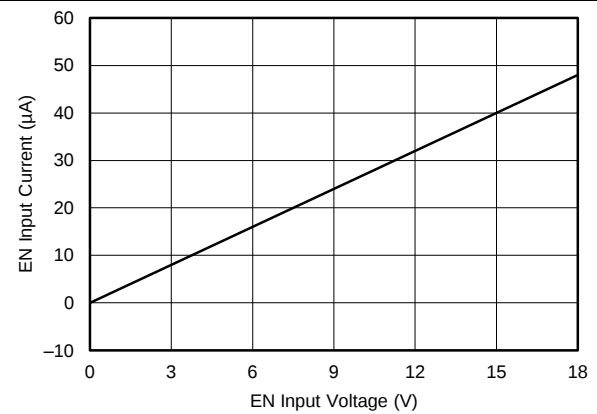


Figure 4. EN Current vs EN Voltage

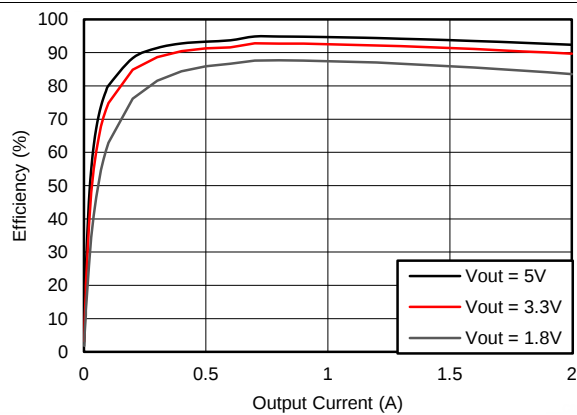


Figure 5. Efficiency vs Output Current

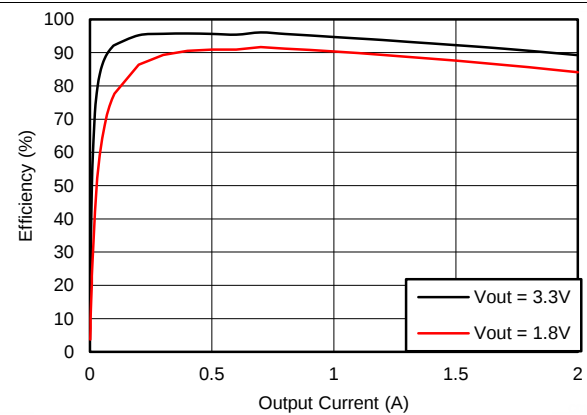
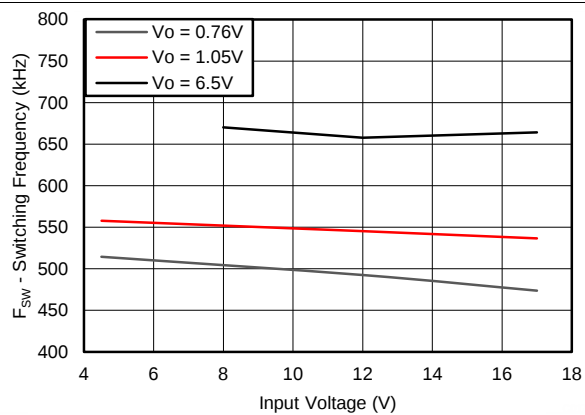
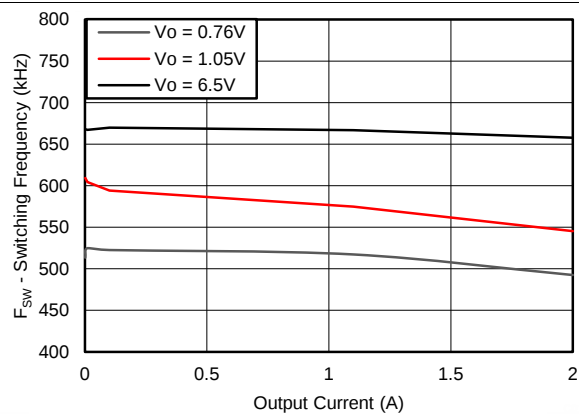


Figure 6. Efficiency vs Output Current ($V_{IN} = 5V$)

TPS562219A Characteristics (continued)



7. Switching Frequency vs Input Voltage



8. Switching Frequency vs Output Current

6.7.2 TPS563219A Characteristics

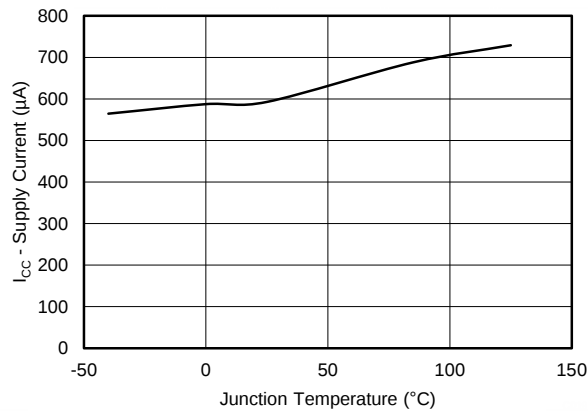


Figure 9. Supply Current vs Junction Temperature

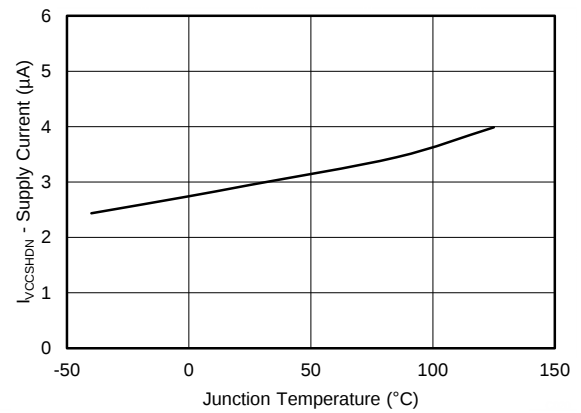


Figure 10. VIN Shutdown Current vs Junction Temperature

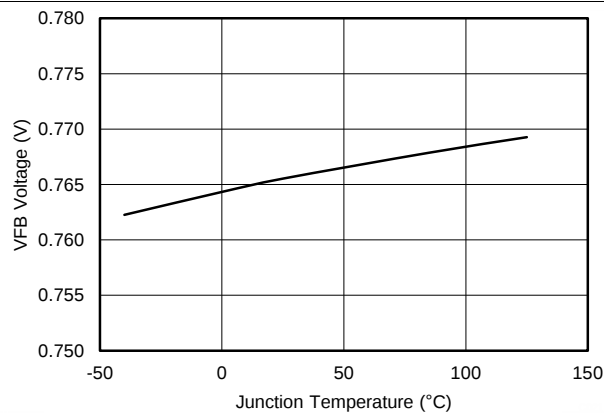


Figure 11. VFB Voltage vs Junction Temperature

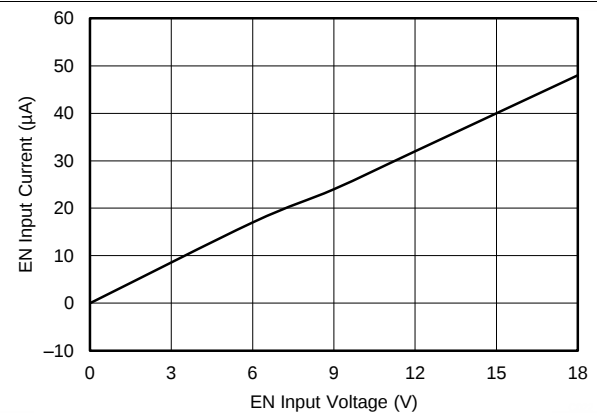


Figure 12. EN Current vs EN Voltage

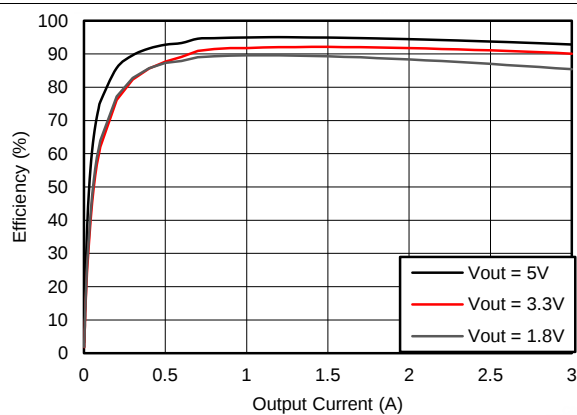


Figure 13. Efficiency vs Output Current

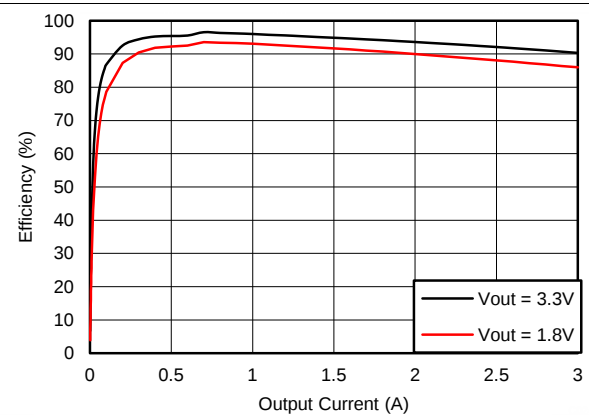
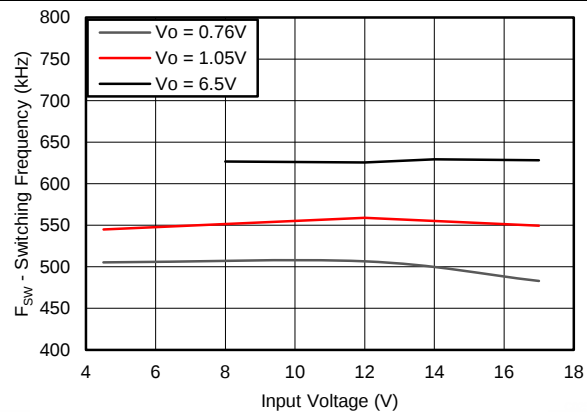
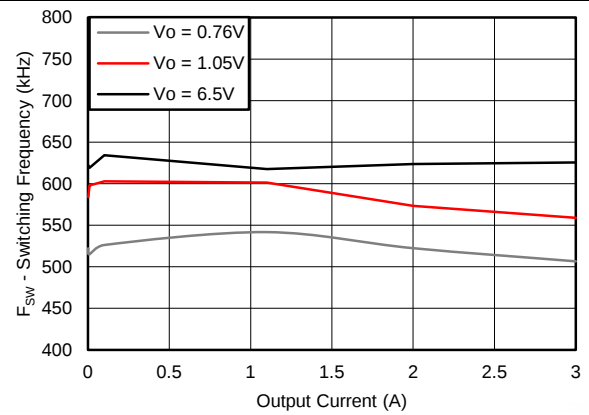


Figure 14. Efficiency vs Output Current ($V_i = 5V$)

TPS563219A Characteristics (continued)



ⓧ 15. Switching Frequency vs Input Voltage



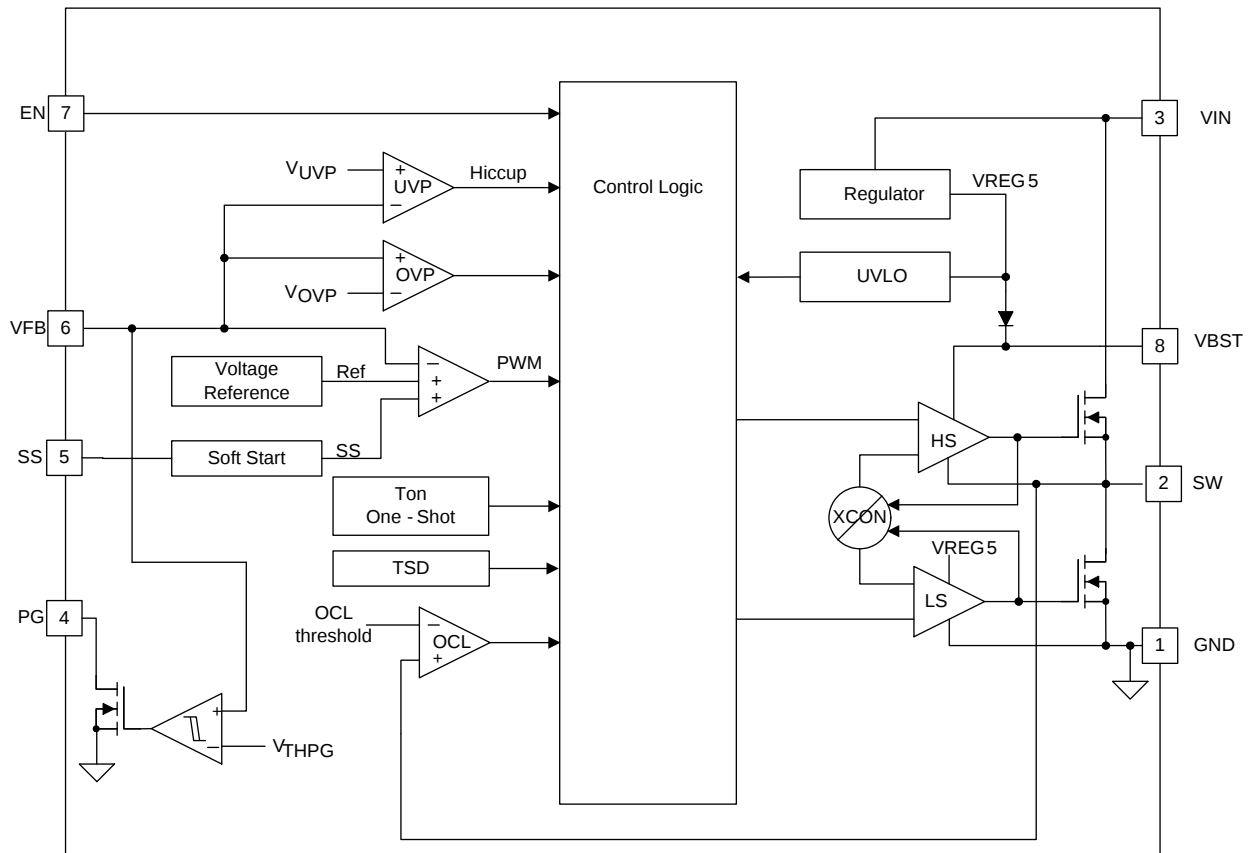
ⓧ 16. Switching Frequency vs Output Current

7 Detailed Description

7.1 Overview

The TPS562219A and TPS563219A are 2-A, 3-A synchronous step-down converters. The proprietary D-CAP2™ mode control supports low ESR output capacitors such as specialty polymer capacitors and multi-layer ceramic capacitors without complex external compensation circuits. The fast transient response of D-CAP2™ mode control can reduce the output capacitance required to meet a specific level of performance.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 The Adaptive On-Time Control and PWM Operation

The main control loop of the TPS562219A and TPS563219A are adaptive on-time pulse width modulation (PWM) controller that supports a proprietary D-CAP2™ mode control. The D-CAP2™ mode control combines adaptive on-time control with an internal compensation circuit for pseudo-fixed frequency and low external component count configuration with both low ESR and ceramic output capacitors. It is stable even with virtually no ripple at the output.

At the beginning of each cycle, the high-side MOSFET is turned on. This MOSFET is turned off after internal one shot timer expires. This one shot duration is set proportional to the converter input voltage, V_{IN} , and inversely proportional to the output voltage, V_O , to maintain a pseudo-fixed frequency over the input voltage range, hence it is called adaptive on-time control. The one-shot timer is reset and the high-side MOSFET is turned on again when the feedback voltage falls below the reference voltage. An internal ramp is added to reference voltage to simulate output ripple, eliminating the need for ESR induced output ripple from D-CAP2™ mode control.

Feature Description (continued)

7.3.2 Soft Start and Pre-Biased Soft Start

The TPS562219A and TPS563219A have adjustable soft-start. When the EN pin becomes high, the SS charge current (I_{SS}) begins charging the capacitor which is connected from the SS pin to GND (C_{SS}). Smooth control of the output voltage is maintained during start up. The equation for the soft start time, T_{SS} is shown in 式 1.

$$T_{SS}(ms) = \frac{C_{SS} \times V_{FBTH} \times 0.86}{I_{SS}} \quad (1)$$

where V_{FBTH} is 0.765 V and I_{SS} is 6 μ A.

If the output capacitor is pre-biased at startup, the devices initiate switching and start ramping up only after the internal reference voltage becomes greater than the feedback voltage V_{FB} . This scheme ensures that the converters ramp up smoothly into regulation point.

7.3.3 Power Good

The power good output, PG is an open drain output. The power good function becomes active after 1.7 times soft-start time. When the output voltage becomes within –10% of the target value, internal comparators detect power good state and the power good signal becomes high. If the feedback voltage goes under 15% of the target value, the power good signal becomes low.

7.3.4 Current Protection

The output over-current limit (OCL) is implemented using a cycle-by-cycle valley detect control circuit. The switch current is monitored during the OFF state by measuring the low-side FET drain to source voltage. This voltage is proportional to the switch current. To improve accuracy, the voltage sensing is temperature compensated.

During the on time of the high-side FET switch, the switch current increases at a linear rate determined by V_{IN} , V_{OUT} , the on-time and the output inductor value. During the on time of the low-side FET switch, this current decreases linearly. The average value of the switch current is the load current I_{OUT} . If the monitored current is above the OCL level, the converter maintains low-side FET on and delays the creation of a new set pulse, even the voltage feedback loop requires one, until the current level becomes OCL level or lower. In subsequent switching cycles, the on-time is set to a fixed value and the current is monitored in the same manner. If the over current condition exists consecutive switching cycles, the internal OCL threshold is set to a lower level, reducing the available output current. When a switching cycle occurs where the switch current is not above the lower OCL threshold, the counter is reset and the OCL threshold is returned to the higher value.

There are some important considerations for this type of over-current protection. The load current is higher than the over-current threshold by one half of the peak-to-peak inductor ripple current. Also, when the current is being limited, the output voltage tends to fall as the demanded load current may be higher than the current available from the converter. This may cause the output voltage to fall. When the V_{FB} voltage falls below the UVP threshold voltage, the UVP comparator detects it. And then, the device will shut down after the UVP delay time (typically 14 μ s) and re-start after the hiccup time.

When the over current condition is removed, the output voltage returns to the regulated value.

7.3.5 Over Voltage Protection

TPS562219A and TPS563219A detect over voltage condition by monitoring the feedback voltage (V_{FB}). When the feedback voltage becomes higher than 125% of the target voltage, the OVP comparator output goes high and the high-side MOSFET turns off. This function is non-latch operation.

7.3.6 UVLO Protection

Under voltage lock out protection (UVLO) monitors the internal regulator voltage. When the voltage is lower than UVLO threshold voltage, the device is shut off. This protection is non-latching.

7.3.7 Thermal Shutdown

The device monitors the temperature of itself. If the temperature exceeds the threshold value (typically 155°C), the device is shut off. This is a non-latch protection.

7.4 Device Functional Modes

7.4.1 Normal Operation

When the input voltage is above the UVLO threshold and the EN voltage is above the enable threshold, the TPS562219A and TPS563219A can operate in their normal switching modes. Normal continuous conduction mode (CCM) occurs when the minimum switch current is above 0 A. In CCM, the TPS562219A and TPS563219A operate at a quasi-fixed frequency of 650 kHz.

7.4.2 Forced CCM Operation

When the TPS562209 and TPS563209 are in the normal CCM operating mode and the switch current falls below 0 A, the TPS562219A and TPS563219A begin operating in forced CCM.

7.4.3 Standby Operation

When the TPS562219A and TPS563219A are operating in either normal CCM or forced CCM, they may be placed in standby by asserting the EN pin low.

8 Application and Implementation

注

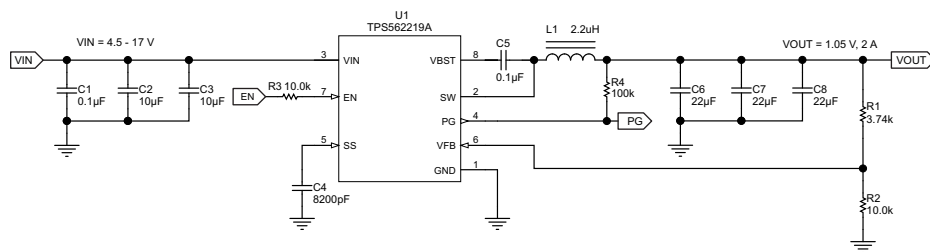
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The TPS562219A and TPS563219A are typically used as step down converters, which convert a voltage from 4.5 V - 17 V to a lower voltage. Webench software is available to aid in the design and analysis of circuits.

8.2 Typical Application

8.2.1 Typical Application, TPS562219A



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图 17. TPS562219A 1.05V/2A Reference Design

8.2.1.1 Design Requirements

For this design example, use the parameters shown in 表 1.

表 1. Design Parameters

PARAMETER	VALUES
Input voltage range	4.5 V to 17 V
Output voltage	1.05 V
Output current	2 A
Output voltage ripple	20 mVpp

8.2.1.2 Detailed Design Procedure

8.2.1.2.1 Output Voltage Resistors Selection

The output voltage is set with a resistor divider from the output node to the VFB pin. It is recommended to use 1% tolerance or better divider resistors. Start by using 式 2 to calculate V_{OUT} .

To improve efficiency at light loads consider using larger value resistors, too high of resistance are more susceptible to noise and voltage errors from the VFB input current are more noticeable.

$$V_{OUT} = 0.765 \times \left(1 + \frac{R1}{R2} \right) \quad (2)$$

8.2.1.2.2 Output Filter Selection

The LC filter used as the output filter has double pole at:

$$F_P = \frac{1}{2\pi\sqrt{L_{OUT} \times C_{OUT}}} \quad (3)$$

At low frequencies, the overall loop gain is set by the output set-point resistor divider network and the internal gain of the device. The low frequency phase is 180 degrees. At the output filter pole frequency, the gain rolls off at a –40 dB per decade rate and the phase drops rapidly. D-CAP2™ introduces a high frequency zero that reduces the gain roll off to –20 dB per decade and increases the phase to 90 degrees one decade above the zero frequency. The inductor and capacitor for the output filter must be selected so that the double pole of 式 3 is located below the high frequency zero but close enough that the phase boost provided by the high frequency zero provides adequate phase margin for a stable circuit. To meet this requirement use the values recommended in 表 2.

表 2. TPS562219A Recommended Component Values

Output Voltage (V)	R2 (kΩ)	R3 (kΩ)	L1(μH)			C6 + C7 + C8(μF)
			MIN	TYP	MAX	
1	3.09	10.0	1.5	2.2	4.7	20 - 68
1.05	3.74	10.0	1.5	2.2	4.7	20 - 68
1.2	5.76	10.0	1.5	2.2	4.7	20 - 68
1.5	9.53	10.0	1.5	2.2	4.7	20 - 68
1.8	13.7	10.0	1.5	2.2	4.7	20 - 68
2.5	22.6	10.0	2.2	3.3	4.7	20 - 68
3.3	33.2	10.0	2.2	3.3	4.7	20 - 68
5	54.9	10.0	3.3	4.7	4.7	20 - 68
6.5	75	10.0	3.3	4.7	4.7	20 - 68

The inductor peak-to-peak ripple current, peak current and RMS current are calculated using 式 4, 式 5 and 式 6. The inductor saturation current rating must be greater than the calculated peak current and the RMS or heating current rating must be greater than the calculated RMS current.

Use 650 kHz for f_{SW} . Make sure the chosen inductor is rated for the peak current of 式 5 and the RMS current of 式 6.

$$I_{P-P} = \frac{V_{OUT}}{V_{IN(MAX)}} \times \frac{V_{IN(MAX)} - V_{OUT}}{L_O \times f_{SW}} \quad (4)$$

$$I_{PEAK} = I_O + \frac{I_{P-P}}{2} \quad (5)$$

$$I_{LO(RMS)} = \sqrt{I_O^2 + \frac{1}{12} I_{P-P}^2} \quad (6)$$

For this design example, the calculated peak current is 2.34 A and the calculated RMS current is 2.01 A. The inductor used is a TDK CLF7045T-2R2N with a peak current rating of 5.5 A and an RMS current rating of 4.3 A.

The capacitor value and ESR determines the amount of output voltage ripple. The TPS562219A and TPS563219A are intended for use with ceramic or other low ESR capacitors. Recommended values range from 20 μF to 68 μF. Use 式 7 to determine the required RMS current rating for the output capacitor.

$$I_{CO(RMS)} = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{\sqrt{12} \times V_{IN} \times L_O \times f_{SW}} \quad (7)$$

For this design two TDK C3216X5R0J226M 22μF output capacitors are used. The typical ESR is 2 mΩ each. The calculated RMS current is 0.286A and each output capacitor is rated for 4A.

8.2.1.2.3 Input Capacitor Selection

The TPS562219A and TPS563219A require an input decoupling capacitor and a bulk capacitor is needed depending on the application. A ceramic capacitor over 10 μF is recommended for the decoupling capacitor. An additional 0.1 μF capacitor (C3) from pin 3 to ground is optional to provide additional high frequency filtering. The capacitor voltage rating needs to be greater than the maximum input voltage.

8.2.1.2.4 Bootstrap capacitor Selection

A 0.1 μ F ceramic capacitor must be connected between the VBST to SW pin for proper operation. It is recommended to use a ceramic capacitor.

8.2.1.3 Application Curves

The following application curves were generated using the application circuit of [Figure 17](#).

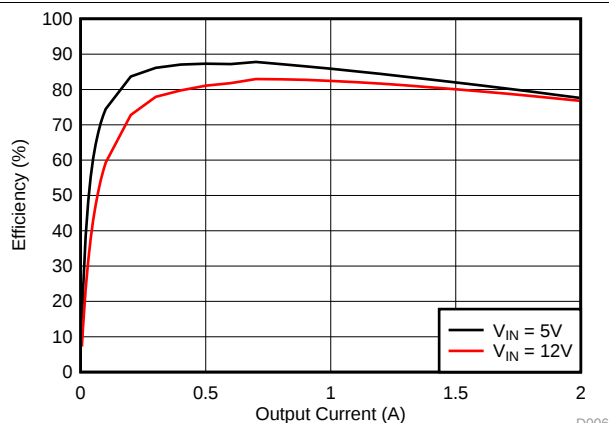


Figure 18. TPS562219A Efficiency

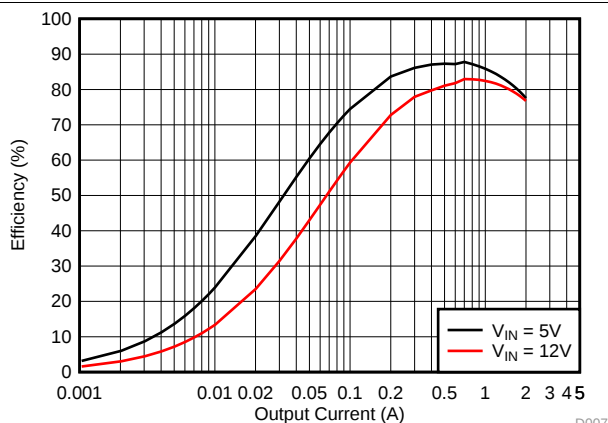


Figure 19. TPS562219A Light Load Efficiency

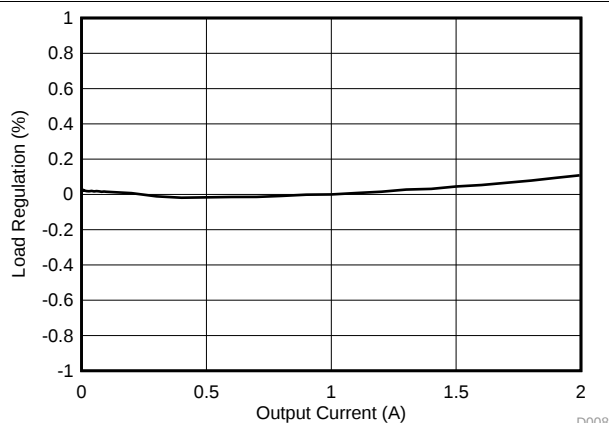


Figure 20. TPS562219A Load Regulation, $V_I = 5\text{ V}$

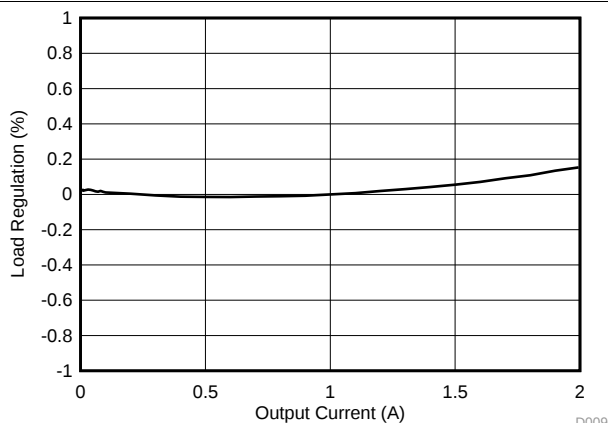


Figure 21. TPS562219A Load Regulation, $V_I = 12\text{ V}$

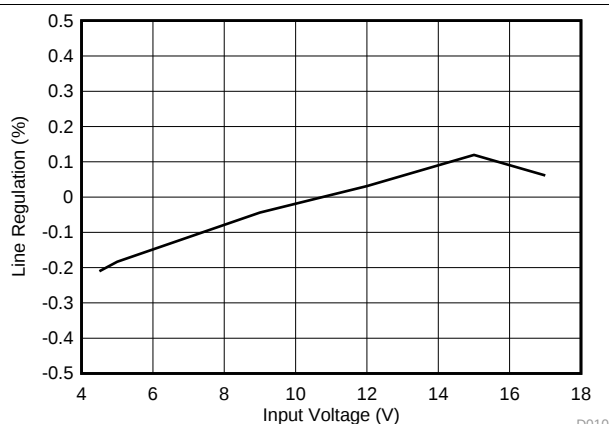


Figure 22. TPS562219A Line Regulation

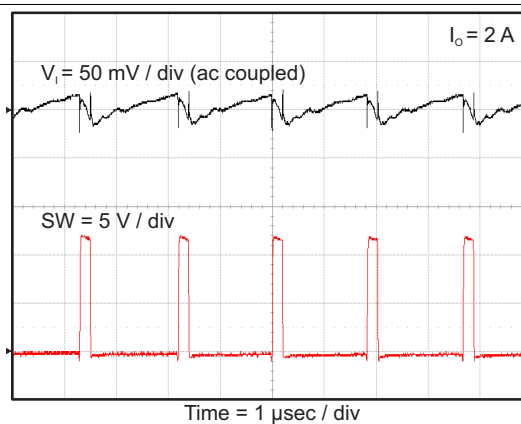
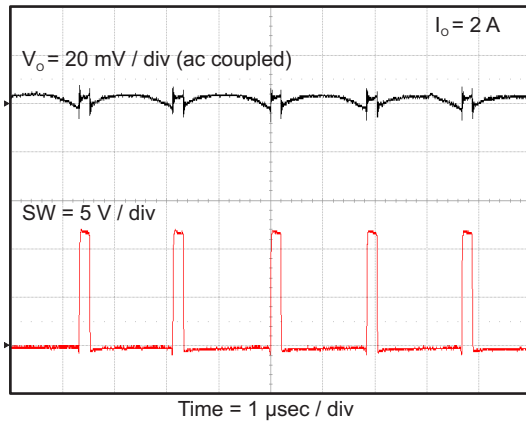
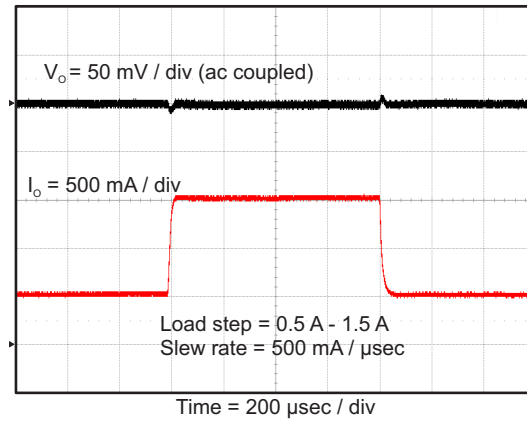


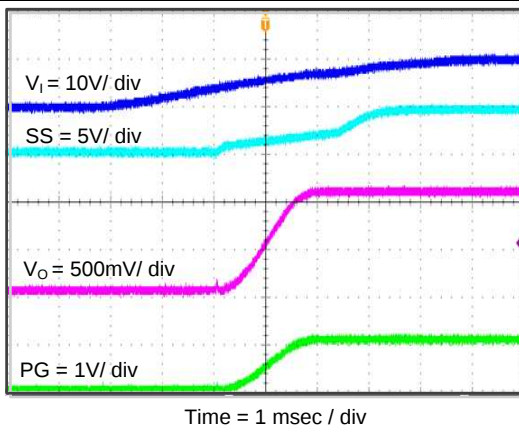
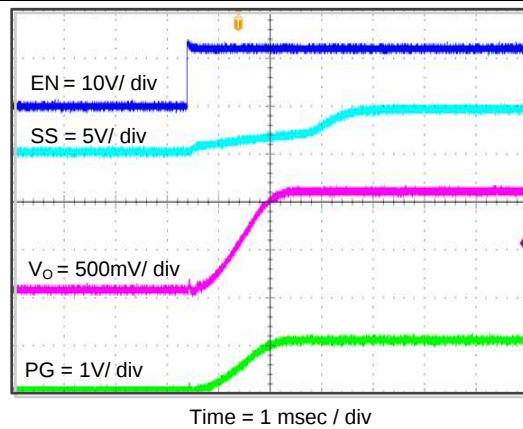
Figure 23. TPS562219A Input Voltage Ripple



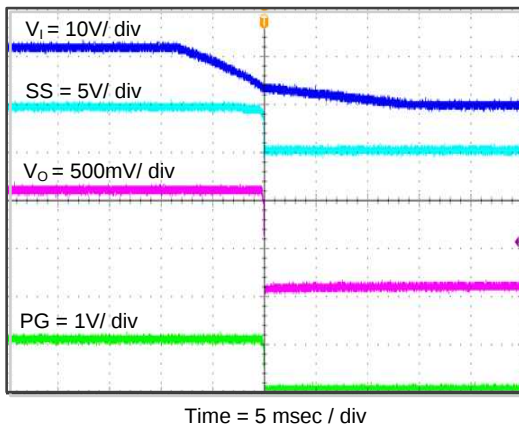
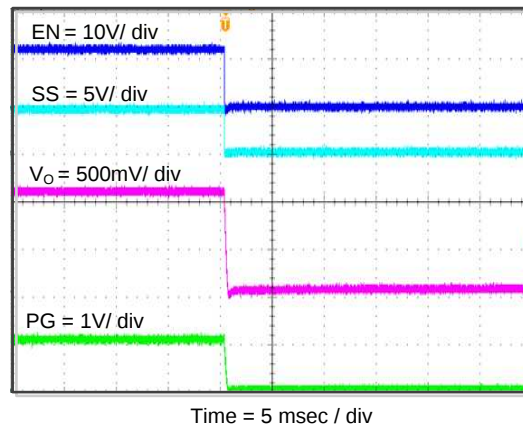
24. TPS562219A Output Voltage Ripple



25. TPS562219A Transient Response

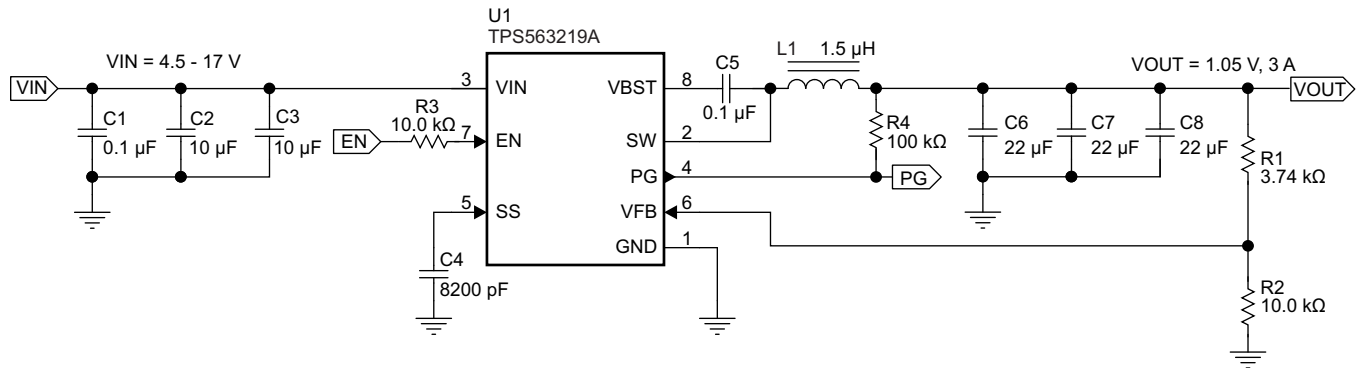

 26. TPS562219A Start Up Relative To V_I


27. TPS562219A Start Up Relative To EN


 28. TPS562219A Shut Down Relative To V_I


29. TPS562219A Shut Down Relative To EN

8.2.2 Typical Application, TPS563219A



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图 30. TPS563219A 1.05V/3A Reference Design

8.2.2.1 Design Requirements

For this design example, use the parameters shown in 表 3.

表 3. Design Parameters

PARAMETER	VALUE
Input voltage range	4.5 V to 17V
Output voltage	1.05V
Output current	3A
Output voltage ripple	20mVpp

8.2.2.2 Detailed Design Procedures

The detailed design procedure for TPS563219A is the same as for TPS562200 except for inductor selection.

8.2.2.2.1 Output Filter Selection

表 4. TPS563219A Recommended Component Values

Output Voltage (V)	R2 (kΩ)	R3 (kΩ)	L1 (μH)			C6 + C7 + C8 (μF)
			MIN	TYP	MAX	
1	3.09	10.0	1.0	1.5	4.7	20 - 68
1.05	3.74	10.0	1.0	1.5	4.7	20 - 68
1.2	5.76	10.0	1.0	1.5	4.7	20 - 68
1.5	9.53	10.0	1.0	1.5	4.7	20 - 68
1.8	13.7	10.0	1.5	2.2	4.7	20 - 68
2.5	22.6	10.0	1.5	2.2	4.7	20 - 68
3.3	33.2	10.0	1.5	2.2	4.7	20 - 68
5	54.9	10.0	2.2	3.3	4.7	20 - 68
6.5	75	10.0	2.2	3.3	4.7	20 - 68

The inductor peak-to-peak ripple current, peak current and RMS current are calculated using 式 8, 式 9 and 式 10. The inductor saturation current rating must be greater than the calculated peak current and the RMS or heating current rating must be greater than the calculated RMS current. Use 650 kHz for f_{SW} .

Use 650 kHz for f_{SW} . Make sure the chosen inductor is rated for the peak current of 式 9 and the RMS current of 式 10.

$$I_{P-P} = \frac{V_{OUT}}{V_{IN(MAX)}} \times \frac{V_{IN(MAX)} - V_{OUT}}{L_O \times f_{SW}} \quad (8)$$

$$I_{PEAK} = I_O + \frac{I_{P-P}}{2} \quad (9)$$

$$I_{LO(RMS)} = \sqrt{I_O^2 + \frac{1}{12} I_{P-P}^2} \quad (10)$$

For this design example, the calculated peak current is 3.505 A and the calculated RMS current is 3.014 A. The inductor used is a TDK CLF7045T-1R5N with a peak current rating of 7.3-A and an RMS current rating of 4.9-A.

The capacitor value and ESR determines the amount of output voltage ripple. The TPS563209 is intended for use with ceramic or other low ESR capacitors. Recommended values range from 20 μ F to 68 μ F. Use 式 6 to determine the required RMS current rating for the output capacitor. For this design, three TDK C3216X5R0J226M 22 μ F output capacitors are used. The typical ESR is 2 m Ω each. The calculated RMS current is 0.292 A and each output capacitor is rated for 4 A.

8.2.2.3 Application Curves

The following application curves were generated using the application circuit of 图 30.

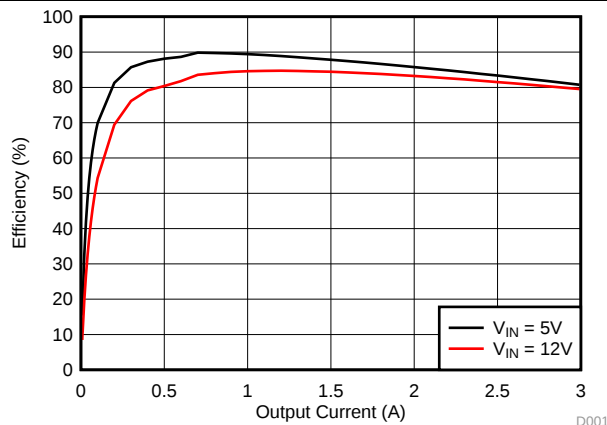


图 31. TPS563219A Efficiency

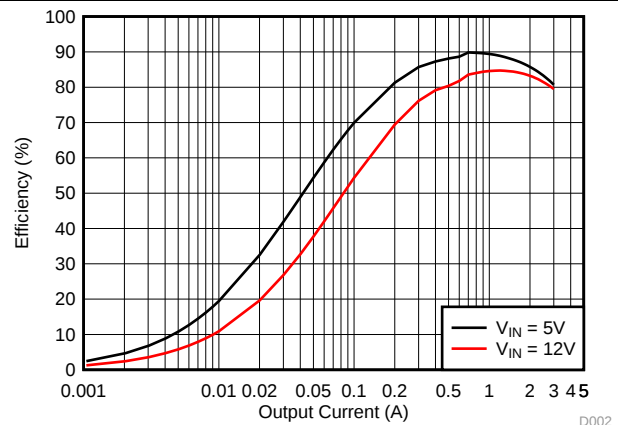


图 32. TPS563219A Light Load Efficiency

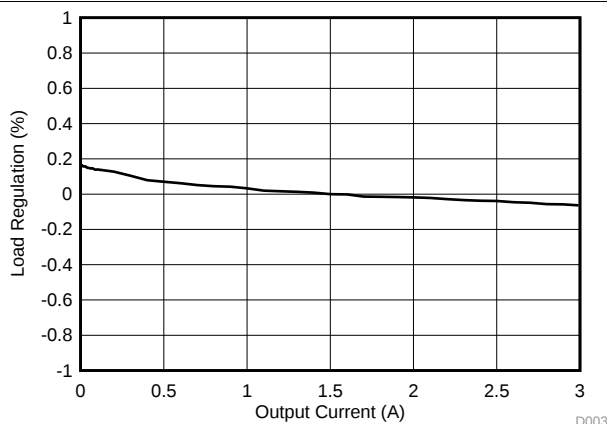


图 33. TPS563219A Load Regulation, $V_I = 5$ V

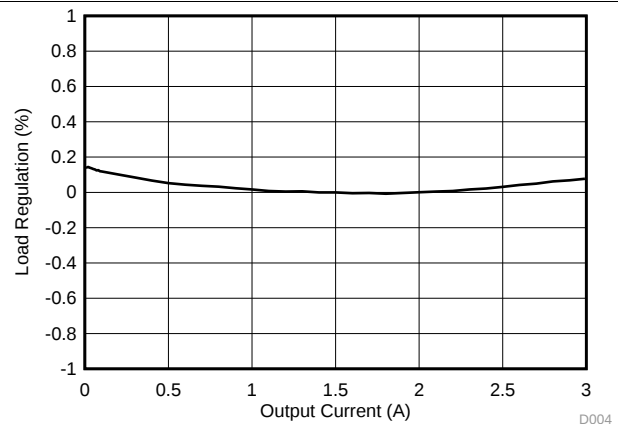


图 34. TPS563219A Load Regulation, $V_I = 12$ V

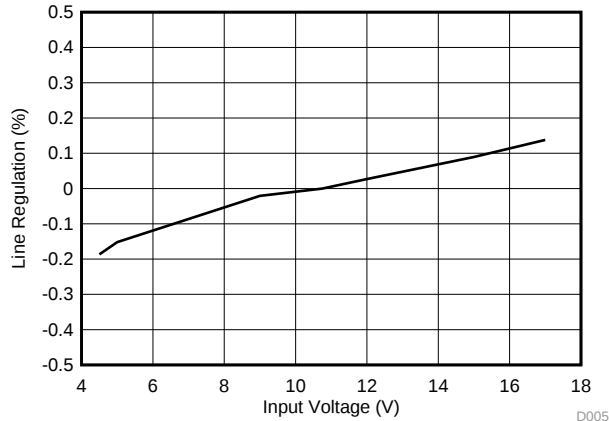


图 35. TPS563219A Line Regulation

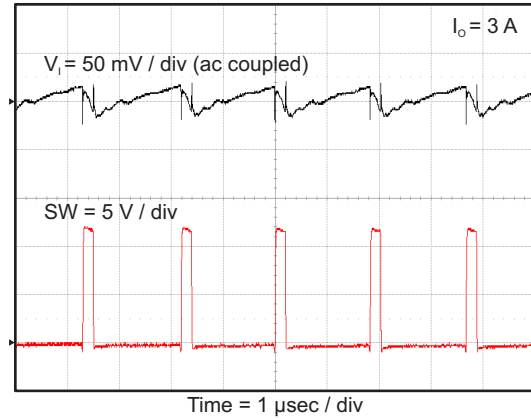


图 36. TPS563219A Input Voltage Ripple

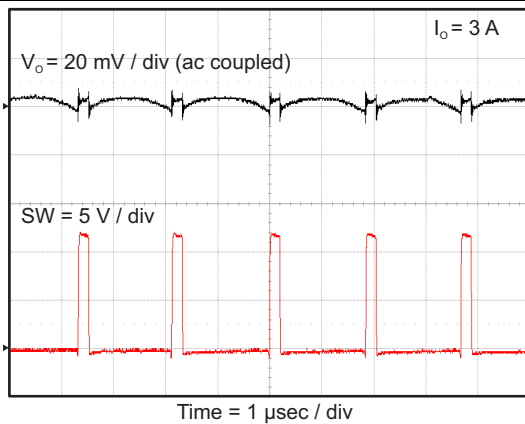


图 37. TPS563219A Output Voltage Ripple

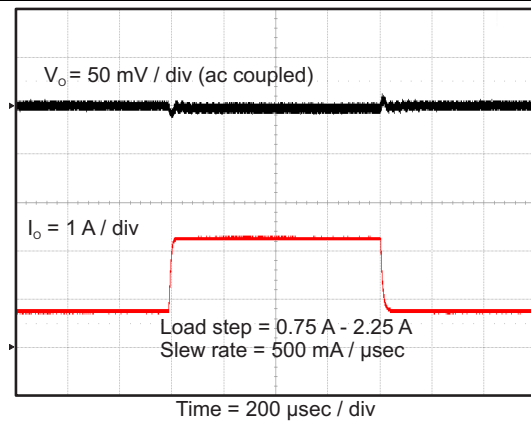


图 38. TPS563219A Transient Response

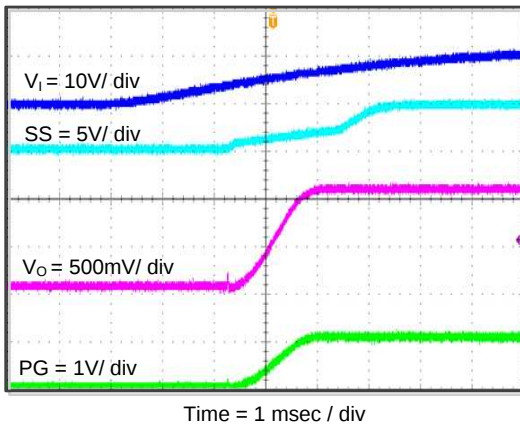


图 39. TPS563219A Start Up Relative To V_i

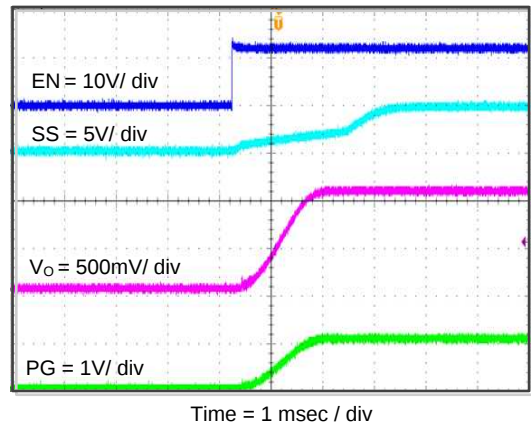


图 40. TPS563219A Start Up Relative To EN

TPS562219A, TPS563219A

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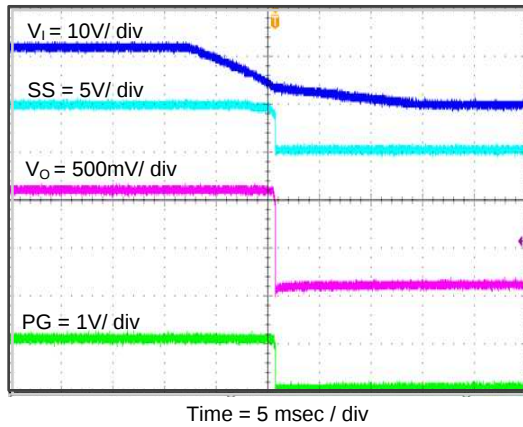


Figure 41. TPS563219A Shut Down Relative To V_I

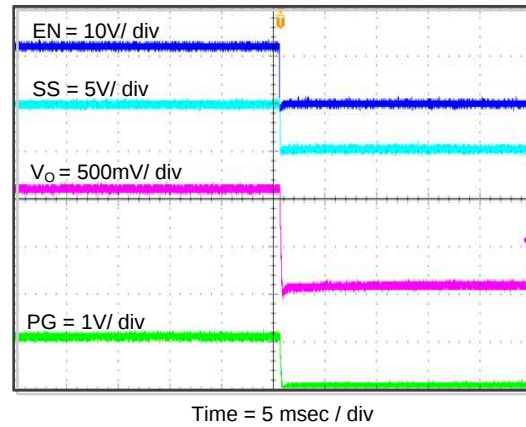


Figure 42. TPS563219A Shut Down Relative To EN

9 Power Supply Recommendations

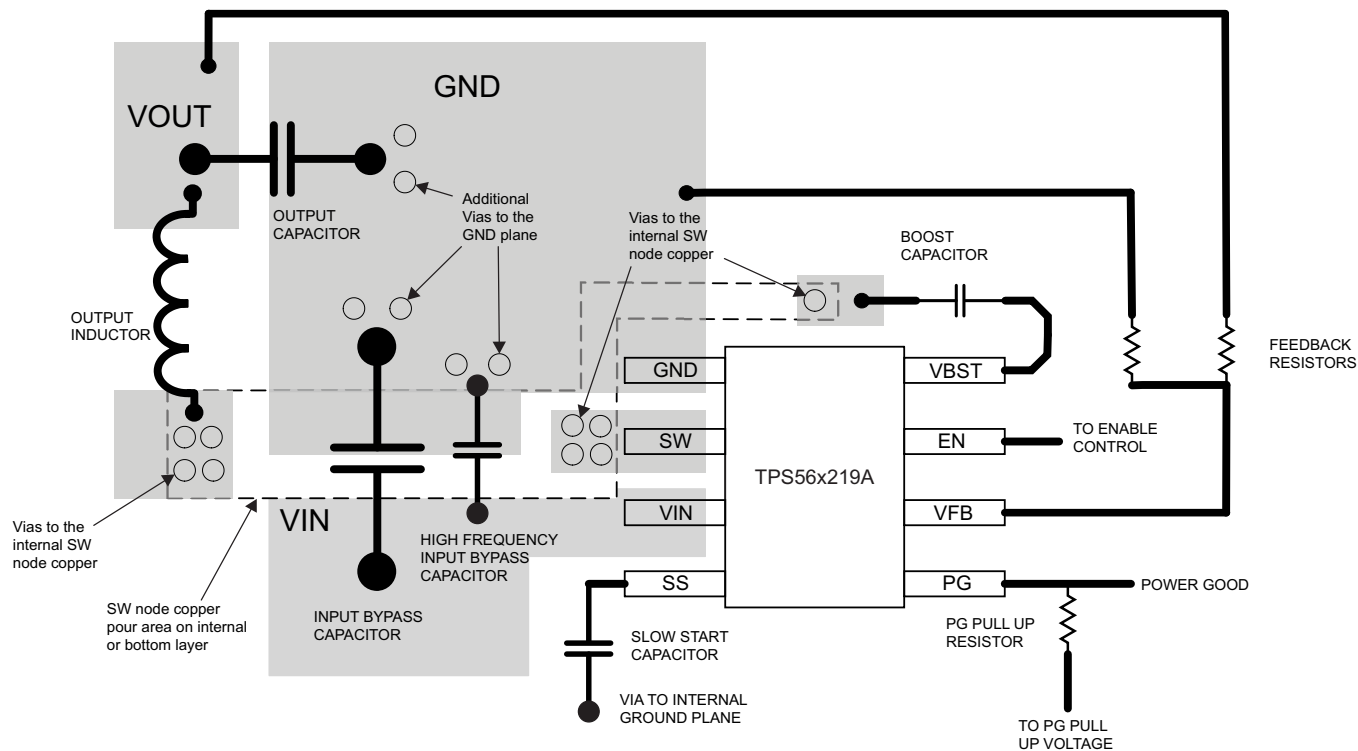
The TPS562209 and TPS563209 are designed to operate from input supply voltage in the range of 4.5V to 17V. Buck converters require the input voltage to be higher than the output voltage for proper operation. The maximum recommended operating duty cycle is 65%. Using that criteria, the minimum recommended input voltage is $V_O / 0.65$.

10 Layout

10.1 Layout Guidelines

1. VIN and GND traces should be as wide as possible to reduce trace impedance. The wide areas are also of advantage from the view point of heat dissipation.
2. The input capacitor and output capacitor should be placed as close to the device as possible to minimize trace impedance.
3. Provide sufficient vias for the input capacitor and output capacitor.
4. Keep the SW trace as physically short and wide as practical to minimize radiated emissions.
5. Do not allow switching current to flow under the device.
6. A separate VOUT path should be connected to the upper feedback resistor.
7. Make a Kelvin connection to the GND pin for the feedback path.
8. Voltage feedback loop should be placed away from the high-voltage switching trace, and preferably has ground shield.
9. The trace of the VFB node should be as small as possible to avoid noise coupling.
10. The GND trace between the output capacitor and the GND pin should be as wide as possible to minimize its trace impedance.

10.2 Layout Example



11 デバイスおよびドキュメントのサポート

11.1 デバイス・サポート

11.2 ドキュメントのサポート

11.3 関連リンク

次の表に、クイック・アクセス・リンクを示します。カテゴリには、技術資料、サポートおよびコミュニティ・リソース、ツールとソフトウェア、およびサンプル注文またはご購入へのクイック・アクセスが含まれます。

表 5. 関連リンク

製品	プロダクト・フォルダ	サンプルとご購入	技術資料	ツールとソフトウェア	サポートとコミュニティ
TPS562219A	ここをクリック	ここをクリック	ここをクリック	ここをクリック	ここをクリック
TPS563219A	ここをクリック	ここをクリック	ここをクリック	ここをクリック	ここをクリック

11.4 ドキュメントの更新通知を受け取る方法

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11.8 用語集

SLYZ022 — TI用語集.

この用語集には、用語や略語の一覧および定義が記載されています。

12 メカニカル、パッケージ、および注文情報

以降のページには、メカニカル、パッケージ、および注文に関する情報が記載されています。この情報は、そのデバイスについて利用可能な最新のデータです。このデータは予告なく変更されることがあり、ドキュメントが改訂される場合もあります。本データシートのブラウザ版を使用されている場合は、画面左側の説明をご覧ください。

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TPS562219ADDFR	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS562219ADDFR.A	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS562219ADDFR.B	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS562219ADDFT	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS562219ADDFT.A	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS562219ADDFT.B	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	2219A
TPS563219ADDFR	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A
TPS563219ADDFR.A	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A
TPS563219ADDFR.B	Active	Production	SOT-23-THIN (DDF) 8	3000 LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A
TPS563219ADDFT	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A
TPS563219ADDFT.A	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A
TPS563219ADDFT.B	Active	Production	SOT-23-THIN (DDF) 8	250 SMALL T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 85	3219A

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

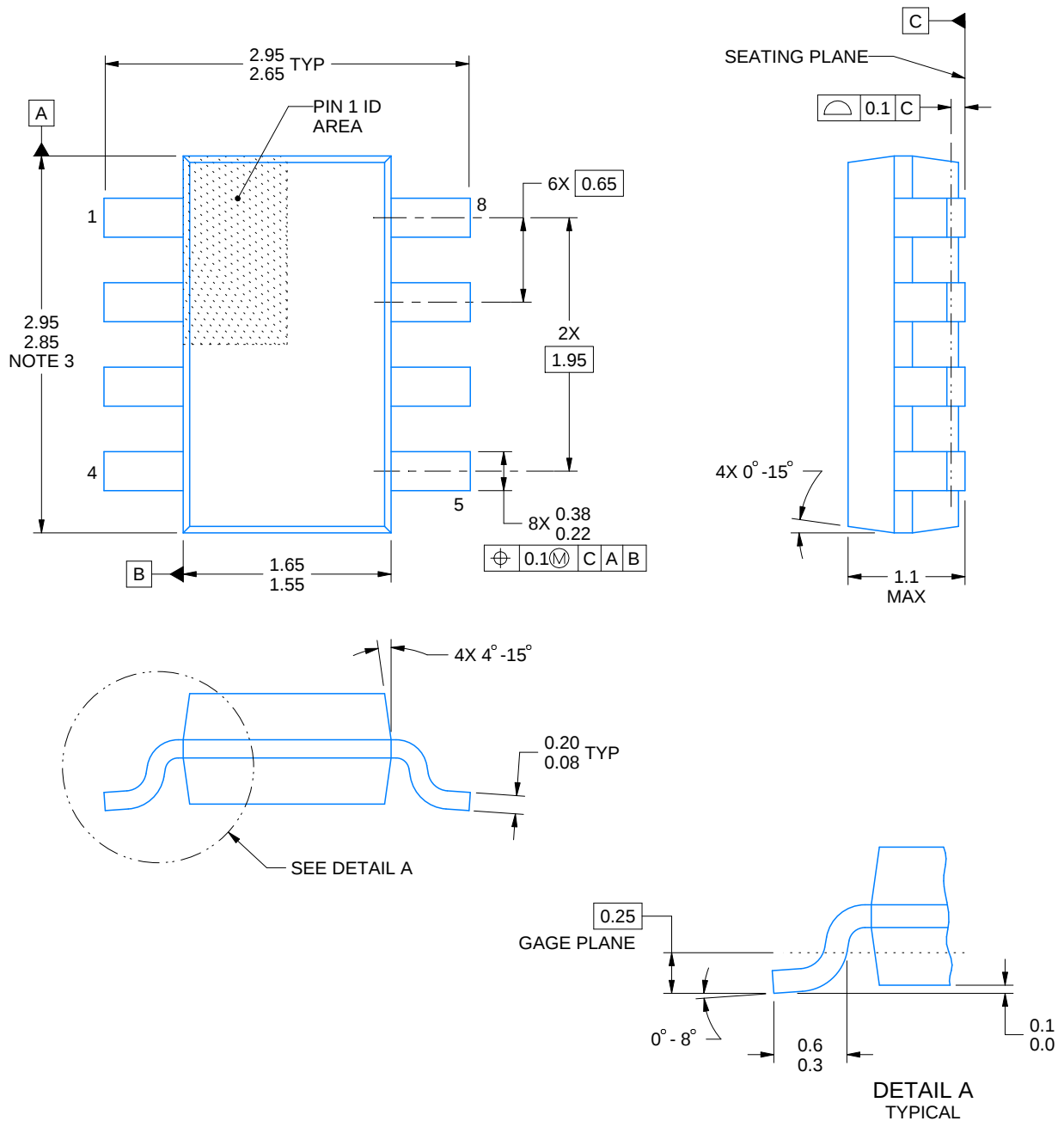
Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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DDF0008A**PACKAGE OUTLINE****SOT-23-THIN - 1.1 mm max height**

PLASTIC SMALL OUTLINE



4222047/E 07/2024

NOTES:

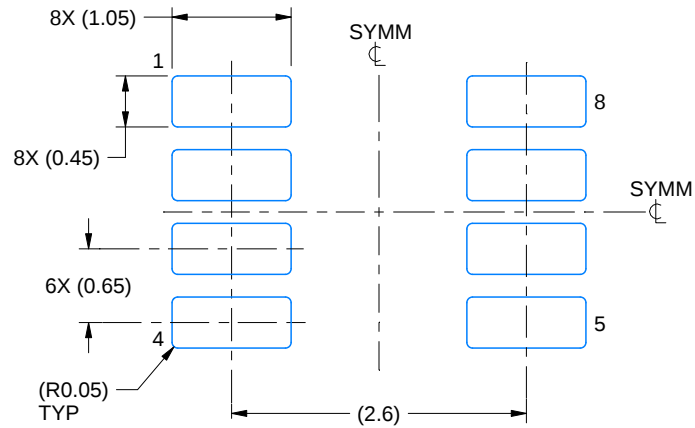
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.

EXAMPLE BOARD LAYOUT

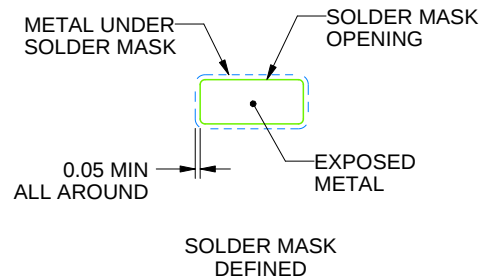
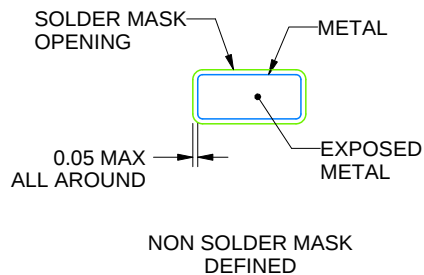
DDF0008A

SOT-23-THIN - 1.1 mm max height

PLASTIC SMALL OUTLINE



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X



SOLDER MASK DETAILS

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NOTES: (continued)

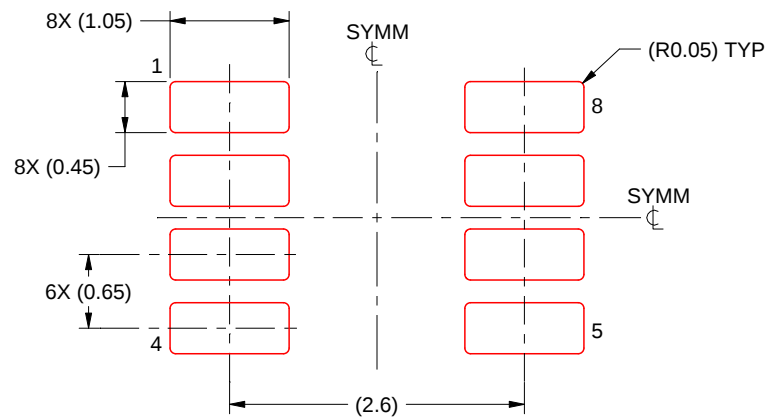
4. Publication IPC-7351 may have alternate designs.
5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

DDF0008A

SOT-23-THIN - 1.1 mm max height

PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:15X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
7. Board assembly site may have different recommendations for stencil design.

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