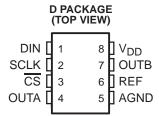
features

- **Controlled Baseline**
 - One Assembly/Test Site, One Fabrication
- **Extended Temperature Performance of** -55°C to 125°C
- **Enhanced Diminishing Manufacturing** Sources (DMS) Support
- **Enhanced Product-Change Notification**
- Qualification Pedigree[†]
- **Dual 12-Bit Voltage Output DAC**
- **Programmable Settling Time**
 - 3 μs in Fast Mode
 - 10 us in Slow Mode
- † Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- Compatible With TMS320 and SPI Serial **Ports**
- Differential Nonlinearity < 0.5 LSB Typ
- **Monotonic Over Temperature**
- **Direct Replacement for TLC5618A**

applications

- **Digital Servo Control Loops**
- **Digital Offset and Gain Adjustment**
- **Industrial Process Control**
- **Machine and Motion Control Devices**
- Mass Storage Devices



description

The TLV5618A is a dual 12-bit voltage output DAC with a flexible 3-wire serial interface. The serial interface is compatible with TMS320, SPI™, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing 4 control and 12 data bits.

The resistor string output voltage is buffered by an x2 gain rail-to-rail output buffer. The buffer features a Class-AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed versus power dissipation.

Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package.

The TLV5618AM is characterized for operation from -55°C to 125°C.

ORDERING INFORMATION

| TA | PACKA | AGE [‡] | ORDERABLE PART NUMBER | TOP-SIDE MARKING |
|----------------|----------|------------------|--------------------------|---------------------|
| –55°C to 125°C | SOIC (D) | Tape and reel | TLV5618AMDREP | 5618ME |

[‡] Package drawings, standard packing quantities, thermal data, symbolization, and PCB design quidelines are available at www.ti.com/sc/package.



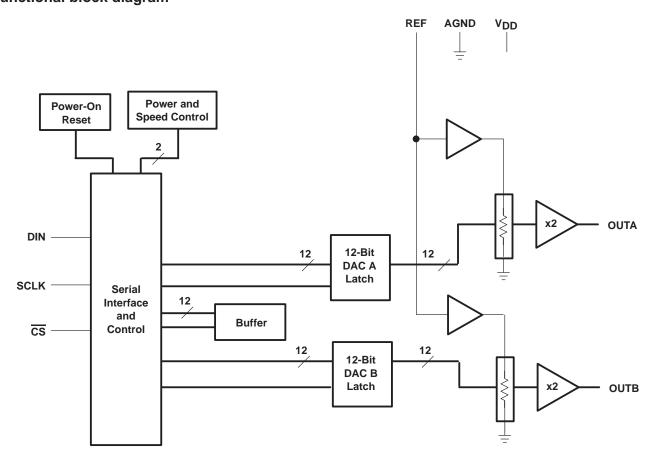
Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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Microwire is a trademark of National Semiconductor Corporation.



functional block diagram



Terminal Functions

| TERM | INAL | 1/0/7 | DECORIDE |
|----------|------|-------|---|
| NAME | NO. | I/O/P | DESCRIPTION |
| AGND | 5 | Р | Ground |
| CS | 3 | I | Chip select. Digital input active low, used to enable/disable inputs. |
| DIN | 1 | I | Digital serial data input |
| OUTA | 4 | 0 | DAC A analog voltage output |
| OUTB | 7 | 0 | DAC B analog voltage output |
| REF | 6 | I | Analog reference voltage input |
| SCLK | 2 | I | Digital serial clock input |
| V_{DD} | 8 | Р | Positive power supply |



TLV5618A-EP 2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN

SGLS214 - OCTOBER 2003

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

| Supply voltage (V _{DD} to AGND) | 7 V |
|--|------------------------------------|
| Reference input voltage range | – 0.3 V to V _{DD} + 0.3 V |
| Digital input voltage range | |
| Operating free-air temperature range [‡] , T _A : TLV5618AM | –55°C to 125°C |
| Storage temperature range [‡] , T _{stg} | –65°C to 150°C |
| Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds | 260°C |
| Package thermal impedance, Rθ _{JA} : D package | |

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

| | | | MIN | NOM | MAX | UNIT |
|---|------------------------------------|--|------|-------|----------------------|------|
| Own houself and M | V _{DD} = 5 V | | 4.5 | 5 | 5.5 | |
| Supply voltage, V _{DD} | V _{DD} = 3 V | | 2.7 | 3 | 3.3 | V |
| Power on reset | | | 0.55 | | 2 | V |
| I Pak Javal Pakal Sanatasakana M | V _{DD} = 2.7 V | | 2 | | | |
| High-level digital input voltage, V _{IH} | V _{DD} = 5.5 V | 4.5 5 5.5 V 2.7 3 3.3 V | | | | |
| | V _{DD} = 2.7 V | | | | 0.6 | ., |
| Low-level digital input voltage, V _{IL} | V _{DD} = 5.5 V | 4.5 5 5.5 2.7 3 3.3 0.55 2 2 2.4 0.6 1 1) AGND 2.048 V _{DD} -1.5 AGND 1.024 V _{DD} -1.5 2 100 p | V | | | |
| Defendance V to DEF to miles! | V _{DD} = 5 V (see Note 1) | | AGND | 2.048 | V _{DD} -1.5 | ., |
| Reference voltage, V _{ref} to REF terminal | V _{DD} = 3 V (see Note 1) | | AGND | 1.024 | V _{DD} -1.5 | V |
| Load resistance, R _L | | | 2 | | | kΩ |
| Load capacitance, C _L | | | | | 100 | pF |
| Clock frequency, f(CLK) | | | | | 20 | MHz |
| Operating free-air temperature, TA | TLV5618AM | | -55 | | 125 | °C |

NOTE 1: Due to the x2 output buffer, a reference input voltage \geq (V_{DD}-0.4 V)/2 causes clipping of the transfer function.



[‡] Long term high-temperature storage and/or extended use at maximum recommended operating conditions may result in a reduction of overall device life. See http://www.ti.com/ep_quality for additional information on enhanced plastic packaging.

TLV5618A-EP 2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH POWER DOWN

SGLS214 - OCTOBER 2003

electrical characteristics over recommended operating conditions (unless otherwise noted)

power supply

| | PARAMETER | TEST C | CONDITIONS | MIN | TYP | MAX | UNIT | |
|------------------------------|------------------------|--|--|------|-----|-----|------|-----|
| I _{DD} Power supply | Downer ounnly ourrent | No load, All inputs = AGND or | V 27V to F F V | Fast | | 1.8 | 2.3 | A |
| | Power supply current | V _{DD} , DAC latch = All ones | $V_{DD} = 2.7 \text{ V to } 5.5 \text{ V}$ | Slow | | 0.8 | 1 | mA |
| | Power down supply cu | | | | 1 | | μΑ | |
| DODD | Barrer armate action | | Zero scale, See Note 2 | | | -65 | | -ID |
| PSRR P | Power supply rejection | ratio | Full scale, See Note 3 | | -65 | | dB | |

NOTES: 2. Power supply rejection ratio at zero scale is measured by varying V_{DD} and is given by: $PSRR = 20 log [(E_{ZS}(V_{DD}max) - E_{ZS}(V_{DD}min)/V_{DD}max]]$

3. Power supply rejection ratio at full scale is measured by varying V_{DD} and is given by: PSRR = 20 log [(E_G(V_{DD}max) – E_G(V_{DD}min)/V_{DD}max]

static DAC specifications

| | PARAMETER | TES | T CONDITIONS | MIN | TYP | MAX | UNIT |
|----------------------|---|------------|---------------------------------|------|-----|------|----------------|
| | Resolution | | | 12 | | | bits |
| INL | Integral nonlinearity | See Note 4 | | | ±2 | ±4 | LSB |
| DNL | Differential nonlinearity | See Note 5 | | ±0.5 | ±1 | LSB | |
| EZS | Zero-scale error (offset error at zero scale) | See Note 6 | | | ±12 | mV | |
| E _{ZS} (TC) | Zero-scale-error temperature coefficient | See Note 7 | | | 3 | | ppm/°C |
| EG | Gain error | See Note 8 | V _{DD} = 2.7 V – 5.5 V | | | ±0.6 | % full scale V |
| E _G (TC) | Gain-error temperature coefficient | See Note 9 | | | 1 | | ppm/°C |

- NOTES: 4. The relative accuracy of integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale, excluding the effects of zero-code and full-scale errors.
 - 5. The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1-LSB amplitude change of any two adjacent codes.
 - 6. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.
 - 7. Zero-scale-error temperature coefficient is given by: E_{ZS} TC = $[E_{ZS}$ (T_{max}) E_{ZS} (T_{min})]/2 V_{ref} × 10⁶/(T_{max} T_{min}).
 - 8. Gain error is the deviation from the ideal output ($2V_{ref} 1$ LSB) with an output load of 10 k Ω .
 - 9. Gain temperature coefficient is given by: $E_G T_C = [E_G (T_{max}) E_g (T_{min})]/2V_{ref} \times 10^6/(T_{max} T_{min})$.

output specifications

| | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|----|---------------------------------|---|-----|-----|----------------------|------|
| VO | Output voltage range | $R_L = 10 \text{ k}\Omega$ | 0 | | V _{DD} -0.4 | ٧ |
| | Output load regulation accuracy | V_O = 4.096 V, 2.048 V, R _L = 2 kΩ to 10 kΩ | | | ±0.29 | % FS |



electrical characteristics over recommended operating conditions (unless otherwise noted) (continued)

reference input

| | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT | |
|---------|---------------------------|---|------|-----|-----|---------------------|-----------|
| ٧ı | Input voltage range | | | 0 | | V _{DD-1.5} | V |
| RĮ | Input resistance | | | | 10 | | $M\Omega$ |
| C_{I} | Input capacitance | | | | 5 | | pF |
| | Deference input handwidth | DEE 0.2 V + 4.024 V do | Fast | | 1.3 | | MHz |
| | Reference input bandwidth | REF = $0.2 \text{ V}_{pp} + 1.024 \text{ V dc}$ | Slow | | 525 | | kHz |
| | Reference feedthrough | REF = 1 V _{pp} at 1 kHz + 1.024 V dc (see Note | | -80 | | dB | |

NOTE 10: Reference feedthrough is measured at the DAC output with an input code = 0x000.

digital inputs

| | PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|-----------------|----------------------------------|----------------------|-----|-----|-----|------|
| lн | High-level digital input current | $V_I = V_{DD}$ | | | 1 | μΑ |
| I _{IL} | Low-level digital input current | V _I = 0 V | -1 | | | μΑ |
| Ci | Input capacitance | | | 8 | | pF |

analog output dynamic performance

| | PARAMETER | | TEST CONDITIONS | | | | | UNIT |
|--------|------------------------------------|------------------------------|--|------|--|-----|----|------|
| | 0 | D 4010 | 0 400 F 0 N 44 | Fast | | 1 | 3 | |
| ts(FS) | Output settling time, full scale | $R_L = 10 \text{ k}\Omega$, | $C_L = 100 \text{ pF}, \text{ See Note 11}$ | Slow | | 3 | 10 | μs |
| | | 5 4010 | 0 400 5 0 11 40 | Fast | | 1 | | |
| ts(CC) | Output settling time, code to code | $R_L = 10 \text{ k}\Omega$, | $_{-}$ = 10 kΩ, C_{L} = 100 pF, See Note 12 | | | 2 | | μs |
| | | | 0 100 F 0 N 1 10 | Fast | | 3 | | |
| SR | Slew rate | $R_L = 10 \text{ k}\Omega$, | $C_L = 100 \text{ pF}$, See Note 13 | Slow | | 0.5 | | V/μs |
| | Glitch energy | DIN = 0 to 1, | DIN = 0 to 1, FCLK = 100 kHz, $\overline{CS} = V_{DD}$ | | | | | nV-s |
| SNR | Signal-to-noise ratio | | | | | 76 | | |
| SINAD | Signal-to-noise + distortion | f _S = 102 kSPS, | $f_{out} = 1 \text{ kHz}, R_L = 10 \text{ k}\Omega,$ | | | 68 | | |
| THD | Total harmonic distortion | C _L = 100 pF | _ | | | -68 | | dB |
| SFDR | Spurious free dynamic range | | | | | 72 | | |

- NOTES: 11. Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0xFDF and 0xFDF to 0x020 respectively. Not tested, assured by design.
 - 12. Settling time is the time for the output signal to remain within \pm 0.5 LSB of the final measured value for a digital input code change of one count. Not tested, assured by design.
 - 13. Slew rate determines the time it takes for a change of the DAC output from 10% to 90% of full-scale voltage.



digital input timing requirements

| | | | MIN | NOM | MAX | UNIT |
|-------------------------|---|-----------------------|-----|-----|-----|------|
| t _{su(CS-CK)} | Setup time, CS low before first negative SCLK edge | | 10 | | | ns |
| t _{su(C16-CS)} | Setup time, 16 th negative SCLK edge before CS rising edge | | 10 | | | ns |
| t _{w(H)} | SCLK pulse width high | | 25 | | | ns |
| t _{w(L)} | SCLK pulse width low | 25 | | | ns | |
| t _{su(D)} | Setup time, data ready before SCLK falling edge | | 8 | | | ns |
| ^t h(D) | Hold time, data held valid after SCLK falling edge | | 10 | | | ns |
| ^t h(CSH) | | V _{DD} = 5 V | 25 | | | |
| | Hold time, CS high between cycles | V _{DD} = 3 V | 50 | | | ns |

timing requirements

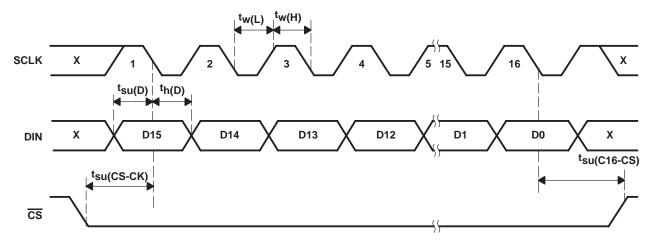


Figure 1. Timing Diagram



TYPICAL CHARACTERISTICS

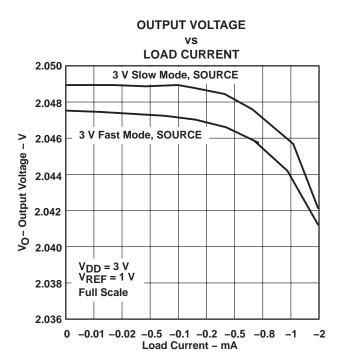
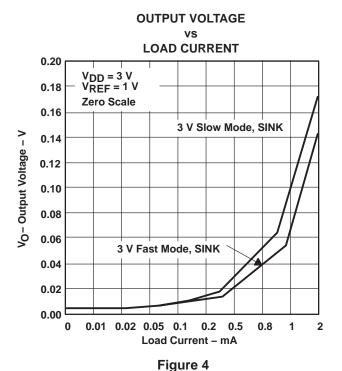


Figure 2



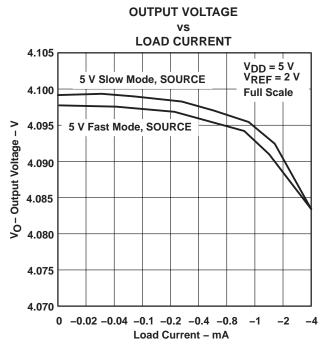


Figure 3

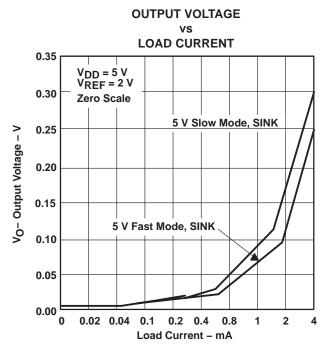
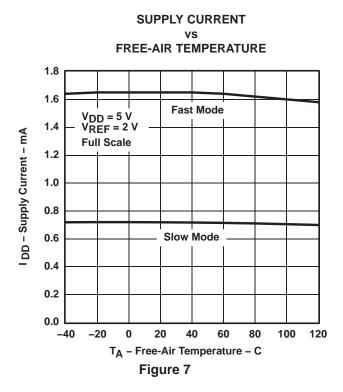
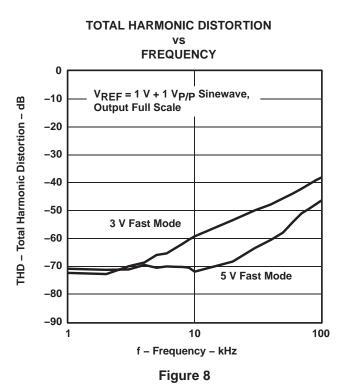


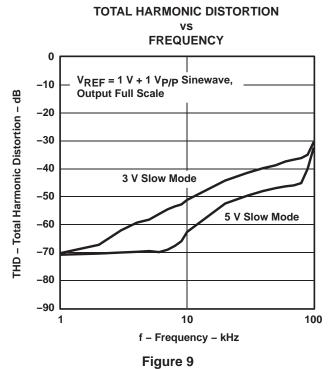
Figure 5

TYPICAL CHARACTERISTICS

SUPPLY CURRENT FREE-AIR TEMPERATURE 1.8 V_{DD} = 3 V V_{REF} = 1 V 1.6 **Full Scale Fast Mode** 1.4 I DD - Supply Current - mA 1.2 1.0 8.0 **Slow Mode** 0.6 0.4 0.2 0.0 -40 20 40 60 80 100 120 -20 T_A - Free-Air Temperature - C Figure 6







TYPICAL CHARACTERISTICS

UNTEGRAL NONLINEARITY ERROR VS DIGITAL CODE 4.0 2.5 2.0 1.5 1.5 1.0 0.5 0.0 -1.5 -2.0 -1.5 -3.0 -3.5 -3.0 Digital Code

Figure 10

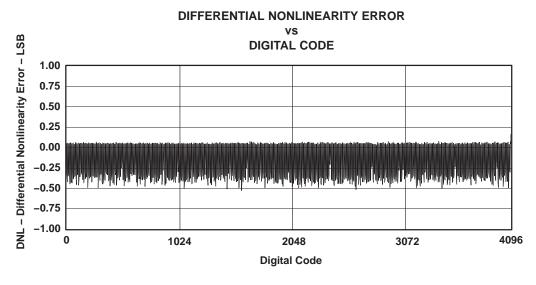


Figure 11

APPLICATION INFORMATION

general function

The TLV5618A is a dual 12-bit, single-supply DAC, based on a resistor-string architecture. It consists of a serial interface, a speed and power down control logic, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by the reference) is given by:

$$2 REF \frac{CODE}{2^n} [V]$$

Where REF is the reference voltage and CODE is the digital input value within the range of 0₁₀ to 2ⁿ–1, where n=12 (bits). The 16-bit data word, consisting of control bits and the new DAC value, is illustrated in the data format section. A power-on reset initially resets the internal latches to a defined state (all bits zero).

serial interface

A falling edge of $\overline{\text{CS}}$ starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or \overline{CS} rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 12 shows examples of how to connect the TLV5618A to TMS320, SPI, and Microwire.

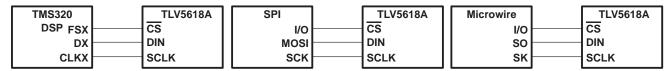


Figure 12. Three-Wire Interface

Notes on SPI and Microwire: Before the controller starts the data transfer, the software has to generate a falling edge on the pin connected to CS. If the word width is 8 bits (SPI and Microwire) two write operations must be performed to program the TLV5618A. After the write operation(s), the holding registers or the control register are updated automatically on the next positive clock edge following the 16th falling clock edge.

serial clock frequency and update rate

The maximum serial clock frequency is given by:

$$f_{sclkmax} = \frac{1}{t_{whmin} + t_{wlmin}} = 20 \text{ MHz}$$

The maximum update rate is:

$$f_{updatemax} = \frac{1}{16(t_{whmin} + t_{wlmin})} = 1.25 \text{ MHz}$$

Note that the maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5618A should also be considered.



APPLICATION INFORMATION

data format

The 16-bit data word for the TLV5618A consists of two parts:

• Program bits (D15..D12)

New data (D11..D0)

| D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|-----|-----|-----|-----|-----|-----|--------------|----|----|----|----|----|----|----|-----|----|
| R1 | SPD | PWR | R0 | MSB | | 12 Data bits | | | | | | | | LSB | |

SPD: Speed control bit $1 \rightarrow$ fast mode $0 \rightarrow$ slow mode PWR: Power control bit $1 \rightarrow$ power down $0 \rightarrow$ normal operation On power up, SPD and PWD are reset to 0 (slow mode and normal operation)

The following table lists all possible combinations of register-select bits:

register-select bits

| R1 | R0 | REGISTER |
|----|----|--|
| 0 | 0 | Write data to DAC B and BUFFER |
| 0 | 1 | Write data to BUFFER |
| 1 | 0 | Write data to DAC A and update DAC B with BUFFER content |
| 1 | 1 | Reserved |

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

examples of operation

Set DAC A output, select fast mode:

Write new DAC A value and update DAC A output:

| | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|-----|-----|-----|-----|-----|-----|----|----|-----|---------|-----------|------|----|----|----|----|
| ſ | 1 | 1 | 0 | 0 | | | | | Nev | V DAC A | output va | alue | | | | |

The DAC A output is updated on the rising clock edge after D0 is sampled.

Set DAC B output, select fast mode:

Write new DAC B value to BUFFER and update DAC B output:

| D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|-----|-----|-----|-----|-----|---|----|----|----|----|----|----|----|----|----|----|
| 0 | 1 | 0 | 0 | | New BUFFER content and DAC B output value | | | | | | | | | | |

The DAC A output is updated on the rising clock edge after D0 is sampled.

- Set DAC A value, set DAC B value, update both simultaneously, select slow mode:
 - 1. Write data for DAC B to BUFFER:

| I | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|-----|-----|-----|-----|-----|-----------------|----|----|----|----|----|----|----|----|----|----|
| ı | 0 | 0 | 0 | 1 | | New DAC B value | | | | | | | | | | |

2. Write new DAC A value and update DAC A and B simultaneously:

| D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|-----|-----|-----|-----|-----|-----------------|----|----|----|----|----|----|----|----|----|----|
| 1 | 0 | 0 | 0 | | New DAC A value | | | | | | | | | | |



APPLICATION INFORMATION

examples of operation (continued)

Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

Set power-down mode:

| | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|---|-----|-----|-----|-----|-----|-----|----|----|----|----|----|----|----|----|----|----|
| I | Χ | Х | 1 | Х | Х | Х | Х | Х | Х | Х | Х | Х | Х | Х | Х | Х |

X = Don't care

linearity, offset, and gain error using single ended supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 13.

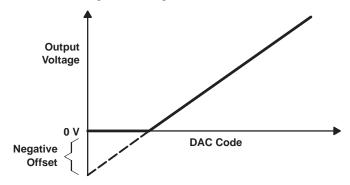


Figure 13. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.



APPLICATION INFORMATION

definitions of specifications and terminology

integral nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

differential nonlinearity (DNL)

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

zero-scale error (E_{ZS})

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

gain error (E_G)

Gain error is the error in slope of the DAC transfer function.

total harmonic distortion (THD)

THD is the ratio of the rms value of the first six harmonic components to the value of the fundamental signal. The value for THD is expressed in decibels.

signal-to-noise ratio + distortion (S/N+D)

S/N+D is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

spurious free dynamic range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the largest spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.



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PACKAGING INFORMATION

| Orderable part number | Status | Material type | Package Pins | Package qty Carrier | RoHS (3) | Lead finish/ Ball material | MSL rating/ Peak reflow | Op temp (°C) | Part marking (6) |
|-----------------------|--------|---------------|----------------|-----------------------|-----------------|-------------------------------|----------------------------|--------------|------------------|
| TLV5618AMDREP | Active | Production | SOIC (D) 8 | 2500 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -55 to 125 | 5618ME |
| TLV5618AMDREP.A | Active | Production | SOIC (D) 8 | 2500 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -55 to 125 | 5618ME |
| V62/04646-01XE | Active | Production | SOIC (D) 8 | 2500 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -55 to 125 | 5618ME |

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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OTHER QUALIFIED VERSIONS OF TLV5618A-EP:

Catalog: TLV5618A

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

PACKAGE OPTION ADDENDUM

www.ti.com 11-Nov-2025

Military : TLV5618AM

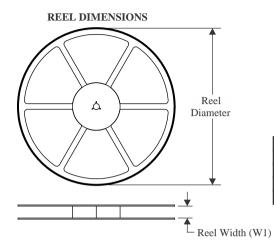
NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Military QML certified for Military and Defense Applications

PACKAGE MATERIALS INFORMATION

www.ti.com 5-Dec-2023

TAPE AND REEL INFORMATION





| A0 | Dimension designed to accommodate the component width |
|----|---|
| В0 | Dimension designed to accommodate the component length |
| K0 | Dimension designed to accommodate the component thickness |
| W | Overall width of the carrier tape |
| P1 | Pitch between successive cavity centers |

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

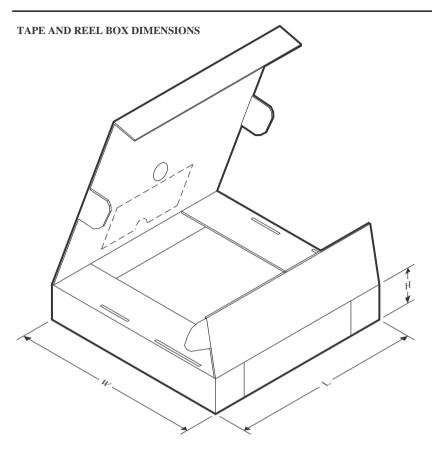


*All dimensions are nominal

| Device | Package Type | Package Drawing | | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|---------------|-----------------|--------------------|---|------|--------------------------|--------------------------|------------|------------|------------|------------|-----------|------------------|
| TLV5618AMDREP | SOIC | D | 8 | 2500 | 330.0 | 12.4 | 6.4 | 5.2 | 2.1 | 8.0 | 12.0 | Q1 |

PACKAGE MATERIALS INFORMATION

www.ti.com 5-Dec-2023



*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|---------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TLV5618AMDREP | SOIC | D | 8 | 2500 | 350.0 | 350.0 | 43.0 |

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