

# LSF0102-Q1 車載用 2 チャネル、自動双方向マルチ電圧レベル・トランスレータ

## 1 特長

- 車載アプリケーション向けに AEC-Q100 認定済み
  - 温度グレード 1:  $-40^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$
  - デバイス HBM ESD 分類レベル 2
  - CDM ESD 分類レベル C6
- 方向ピンを必要としない双方向電圧レベル変換を実現
- I<sup>2</sup>C、SPI、UART、MDIO、SDIO、GPIO などオープン・ドレインおよびプッシュプルアプリケーションをサポート
- 30pF 以下の容量性負荷で 100MHz までの昇圧変換および 100MHz を超える降圧変換をサポートし、50pF の容量性負荷で 40MHz までの昇圧または降圧変換をサポート
- 以下の各電圧間で双方向電圧レベル変換が可能
  - 0.95V ↔ 1.8/2.5/3.3/5V
  - 1.2V ↔ 1.8/2.5/3.3/5V
  - 1.8V ↔ 2.5/3.3/5V
  - 2.5V ↔ 3.3/5V
  - 3.3V ↔ 5V
- 低いスタンバイ電流
- 5V 許容の I/O ポートにより TTL 電圧レベルに対応
- 低い R<sub>ON</sub> により信号歪みを低減
- EN = Low のとき高インピーダンスとなる I/O ピン
- フロースルー・ピン配置により PCB 配線を簡素化
- JESD 78、Class II 準拠で 100mA 超のラッチアップ性能

## 2 アプリケーション

- インフォテインメント用ヘッド・ユニット
- グラフィカル・クラスター
- ADAS フェュージョン
- ADAS フロント・カメラ
- HEV バッテリー管理システム

## 3 概要

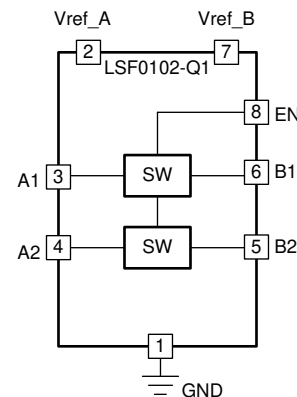
LSF0102-Q1 デバイスは、自動双方向電圧トランスレータで、方向ピンの必要なしに広い範囲の電源電圧間の変換を行います。LSF0102-Q1 は、30pF 以下の容量性負荷で 100MHz までの昇圧変換および 100MHz を超える降圧変換をサポートしています。さらに、LSF0102-Q1 は、50pF の容量性負荷で最高 40MHz の昇圧および降圧変換をサポートしているため、I<sup>2</sup>C、SPI、GPIO、SDIO、UART、MDIO など車載用アプリケーションで一般的に見られる広範な標準インターフェイスをサポートできます。

LSF0102-Q1 デバイスのデータ入力は 5V 許容です。このため、このデバイスは TTL 出力レベルと互換性があります。さらに、LSF0102-Q1 は混在モードの電圧変換をサポートしており、デバイスはチャンネルごとに異なる電圧レベルの昇圧および降圧変換を行うことができます。

### パッケージ情報 (1)

部品番号	パッケージ	本体サイズ (公称)
LSF0102-Q1	DCU (VSSOP, 8)	2.30mm × 2.00mm

- (1) 利用可能なすべてのパッケージについては、このデータシートの末尾にある注文情報を参照してください。



機能ブロック図



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## 4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision A (April 2021) to Revision B (May 2023)	Page
• Updated the <i>Recommended Operating Conditions</i> table to reflect maximum of 5.5 V.....	4
• Updated the <i>Thermal Information</i> table.....	4
• Changed all <i>Switching Characteristic Test Conditions</i> .....	6
• Added the <i>Output Enable</i> section.....	9
• Added the <i>Up and Down Translation</i> sections.....	11
• Changed <i>pull up resistor</i> to <i>bias resistor</i> in <i>Enable, Disable, and Reference Voltage Guidelines</i> section.....	12
• Added the <i>Bias Circuitry</i> section.....	13
• Updated the current values in the table titled <i>Pull-up Resistor Values</i> .....	13
• Added image to the <i>Mixed-Mode Voltage Translation</i> section.....	14
• Added the <i>Single Supply Translation</i> section.....	15
• Added section <i>Voltage Translation for Vref_B &lt; Vref_A + 0.8 V</i> .....	17

Changes from Revision * (May 2018) to Revision A (April 2021)	Page
• ドキュメント全体にわたって表、図、相互参照の採番方法を更新.....	1
• Updated the <i>Bidirectional Translation</i> section to include inclusive terminology.....	13

## 5 Pin Configuration and Functions

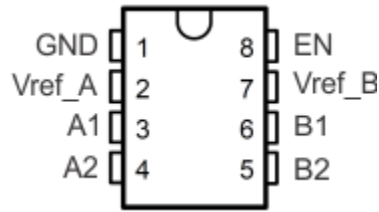


図 5-1. LSF0102-Q1 DCU Package, 8-Pin VSSOP (Top View)

表 5-1. Pin Functions

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NAME	NO.		
A1	3	I/O	Input/Output A port for Channel 1
A2	4	I/O	Input/Output A port for Channel 2
B1	6	I/O	Input/Output B port for Channel 1
B2	5	I/O	Input/Output B port for Channel 2
EN	8	I	I/O enable input; see 図 9-1 for typical setup. Should be tied directly to V <sub>ref_B</sub> to be enabled or pulled LOW to disable all I/O pins.
GND	1	—	Ground
Vref_A	2	—	A side reference supply voltage; see セクション 9 for setup and supply voltage range.
Vref_B	7	—	B side reference supply voltage. Must be connected to supply through 200 kΩ; see セクション 9 for setup and supply voltage range.

(1) I = input, O = output

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature (unless otherwise noted)<sup>(1)</sup>

	MIN	MAX	UNIT
Input voltage <sup>(2)</sup> , $V_I$	-0.5	7	V
Input/output voltage <sup>(2)</sup> , $V_{I/O}$	-0.5	7	V
Continuous channel current		128	mA
Input clamp current, $I_{IK}$	$V_I < 0$	-50	mA
Storage temperature range, $T_{stg}$	-65	150	°C
Operating junction temperature, $T_J$		150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) The input and input/output negative-voltage ratings may be exceeded if the input and input/output clamp-current ratings are observed.

### 6.2 ESD Ratings

	VALUE	UNIT
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2000
	Charged-device model (CDM), per AEC Q100-011	±1000

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
$V_{I/O}$ Input/output voltage		5.5	V
$V_{ref\_A/B/EN}$ Reference voltage		5.5	V
$I_{PASS}$ Pass transistor current		64	mA
$T_A$ Operating free-air temperature	-40	125	°C

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>	LSF0102-Q1	UNIT
	DCU (US8)	
	8 PINS	
$R_{\theta JA}$ Junction-to-ambient thermal resistance	279.7	°C/W
$R_{\theta JC(top)}$ Junction-to-case (top) thermal resistance	129.9	°C/W
$R_{\theta JB}$ Junction-to-board thermal resistance	191.3	°C/W
$\Psi_{JT}$ Junction-to-top characterization parameter	66.3	°C/W
$\Psi_{JB}$ Junction-to-board characterization parameter	190.1	°C/W

- (1) For more information about traditional and new thermal metrics, see the [IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics

over recommended operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
$V_{IK}$	Input clamp voltage	$I_I = -18 \text{ mA}, V_{EN} = 0$			-1.2	V
$I_{IH}$	I/O input high leakage	$V_I = 5 \text{ V}, V_{EN} = 0$			5.0	$\mu\text{A}$
$I_{CCBA}$	$V_{ref\_B}$ to $V_{ref\_A}$ leakage	$V_{ref\_B} = V_{EN} = 5.5 \text{ V}, V_{ref\_A} = 4.5 \text{ V}, I_O = 0, V_I = V_{CC}$ or GND		1		$\mu\text{A}$
$C_{I(ref\_A/B/EN)}$	Input capacitance	$V_I = 3 \text{ V}$ or 0		11		pF
$C_{io(off)}$	I/O pin off-state capacitance	$V_O = 3 \text{ V}$ or 0, $V_{EN} = 0$		4.0	6.0	pF
$C_{io(on)}$	I/O Pin on-state capacitance	$V_O = 3 \text{ V}$ or 0, $V_{EN} = 3 \text{ V}$		10.5	12.5	pF
$r_{on}$ <sup>(2)</sup>	On-state resistance	$V_I = 0, I_O = 64 \text{ mA}$	$V_{ref\_A} = 3.3 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	8.0		$\Omega$
			$V_{ref\_A} = 1.8 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	9.0		
			$V_{ref\_A} = 1.0 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	10		
		$V_I = 0, I_O = 32 \text{ mA}$	$V_{ref\_A} = 1.8 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	10		$\Omega$
			$V_{ref\_A} = 2.5 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	15		
		$V_I = 1.8 \text{ V}, I_O = 15 \text{ mA}$	$V_{ref\_A} = 3.3 \text{ V}; V_{ref\_B} = V_{EN} = 5 \text{ V}$	9.0		$\Omega$
		$V_I = 1.0 \text{ V}, I_O = 10 \text{ mA}$	$V_{ref\_A} = 1.8 \text{ V}; V_{ref\_B} = V_{EN} = 3.3 \text{ V}$	18		$\Omega$
		$V_I = 0 \text{ V}, I_O = 10 \text{ mA}$	$V_{ref\_A} = 1.0 \text{ V}; V_{ref\_B} = V_{EN} = 3.3 \text{ V}$	20		$\Omega$
$V_I = 0 \text{ V}, I_O = 10 \text{ mA}$	$V_{ref\_A} = 1.0 \text{ V}; V_{ref\_B} = V_{EN} = 1.8 \text{ V}$	30		$\Omega$		

(1) All typical values are at  $T_A = 25^\circ\text{C}$ .

(2) Measured by the voltage drop between the A and B pins at the indicated current through the switch. On-state resistance is determined by the lowest voltage of the two (A or B) pins.

### 6.6 Switching Characteristics (Translating Down): $V_{CCB} = 3.3\text{ V}$

over recommended operating free-air temperature range,  $V_{CCB} = 3.3\text{ V}$ ,  $V_{CCB} = V_{IH} = V_{ref\_A} + 1$ ,  $V_{IL} = 0$ , and  $V_M = 0.5V_{ref\_A}$  (unless otherwise noted) (see [Parameter Measurement Information](#))

PARAMETER	TEST CONDITIONS	$C_L = 50\text{ pF}$			$C_L = 30\text{ pF}$			$C_L = 15\text{ pF}$			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$	Propagation delay time, low-to-high output From (input) A or B to (output) B or A		1.1			0.7			0.3		ns
$t_{PHL}$	Propagation delay time, high-to-low output From (input) A or B to (output) B or A		1.2			0.8			0.4		

### 6.7 Switching Characteristics (Translating Down): $V_{CCB} = 2.5\text{ V}$

over recommended operating free-air temperature range,  $V_{CCB} = 2.5\text{ V}$ ,  $V_{CCB} = V_{IH} = V_{ref\_A} + 1$ ,  $V_{IL} = 0$ , and  $V_M = 0.5V_{ref\_A}$  (unless otherwise noted) (see [Parameter Measurement Information](#))

PARAMETER	TEST CONDITIONS	$C_L = 50\text{ pF}$			$C_L = 30\text{ pF}$			$C_L = 15\text{ pF}$			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$	Propagation delay time, low-to-high output From (input) A or B to (output) B or A		1.2			0.8			0.35		ns
$t_{PHL}$	Propagation delay time, high-to-low output From (input) A or B to (output) B or A		1.3			1			0.5		

### 6.8 Switching Characteristics Translating Up): $V_{CCB} = 3.3\text{ V}$

over recommended operating free-air temperature range,  $V_{CCB} = 3.3\text{ V}$ ,  $V_{CCB} = V_T = V_{ref\_A} + 1$ ,  $V_{ref\_A} = V_{IH}$ ,  $V_{IL} = 0$ ,  $V_M = 0.5V_{ref\_A}$  and  $R_L = 300$  (unless otherwise noted) (see [Parameter Measurement Information](#))

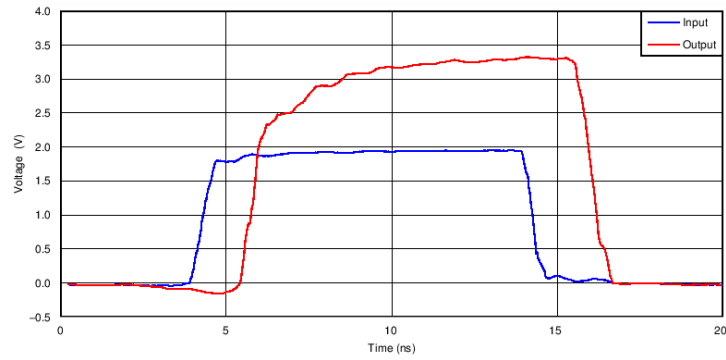
PARAMETER	TEST CONDITIONS	$C_L = 50\text{ pF}$			$C_L = 30\text{ pF}$			$C_L = 15\text{ pF}$			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$	Propagation delay time, low-to-high output From (input) A or B to (output) B or A		1			0.8			0.4		ns
$t_{PHL}$	Propagation delay time, high-to-low output From (input) A or B to (output) B or A		1			0.9			0.4		

### 6.9 Switching Characteristics (Translating Up): $V_{CCB} = 2.5\text{ V}$

over recommended operating free-air temperature range,  $V_{CCB} = 2.5\text{ V}$ ,  $V_{CCB} = V_T = V_{ref\_A} + 1$ ,  $V_{ref\_A} = V_{IH}$ ,  $V_{IL} = 0$ ,  $V_M = 0.5V_{ref\_A}$  and  $R_L = 300$  (unless otherwise noted) (see [Parameter Measurement Information](#))

PARAMETER	TEST CONDITIONS	$C_L = 50\text{ pF}$			$C_L = 30\text{ pF}$			$C_L = 15\text{ pF}$			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
$t_{PLH}$	Propagation delay time, low-to-high output From (input) A or B to (output) B or A		1.1			0.9			0.45		ns
$t_{PHL}$	Propagation delay time, high-to-low output From (input) A or B to (output) B or A		1.3			1.1			0.6		

## 6.10 Typical Characteristics

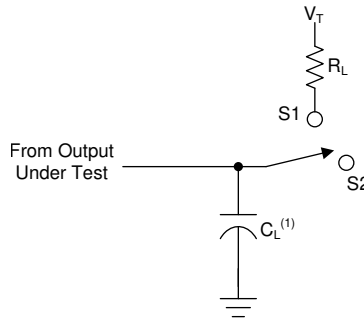


**6-1. Signal Integrity (1.8 to 3.3 V Translation Up at 50 MHz)**

## 7 Parameter Measurement Information

The outputs are measured one at a time, with one transition per measurement. All input pulses are supplied by generators that have the following characteristics:

- $PRR \leq 10 \text{ MHz}$
- $Z_O = 50 \Omega$
- $t_r \leq 2 \text{ ns}$
- $t_f \leq 2 \text{ ns}$

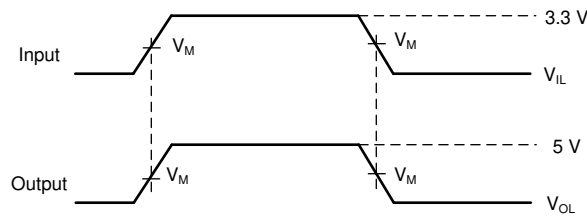


A.  $C_L$  includes probe and jig capacitance.

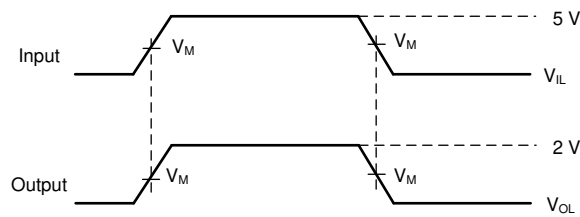
**7-1. Load Circuit**

USAGE	SWITCH
Translating Up	S1
Translating Down	S2

**7-2. Translating Up and Down Table**



**7-3. Translating Up**



**7-4. Translating Down**

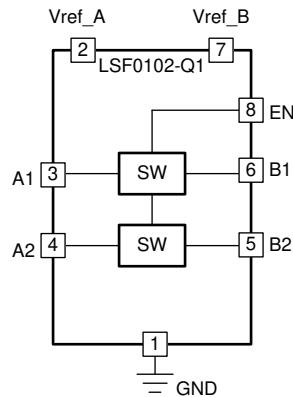


## 8 Detailed Description

### 8.1 Overview

The LSF0102-Q1 device can be used in level translation applications for interfacing devices or systems operating at different interface voltages. The LSF0102-Q1 device is ideal for use in applications where an open-drain driver is connected to the data I/Os. With appropriate pull-up resistors and layout, the LSF0102-Q1 device can achieve 100 MHz. The LSF0102-Q1 can also be used in applications where a push-pull driver is connected to the data I/Os.

### 8.2 Functional Block Diagram



### 8.3 Feature Description

#### 8.3.1 Auto Bidirectional Voltage Translation

The LSF0102-Q1 device is an auto bidirectional voltage level translator that operates from 0.95 to 5.5 V on  $V_{ref\_A}$  and 1.8 to 5.5 V on  $V_{ref\_B}$ . This allows bidirectional voltage translation between 0.95 V and 5.5 V without the need for a direction pin in open-drain or push-pull applications. The LSF0102-Q1 device supports level translation applications with transmission speeds greater than 100 Mbps for open-drain systems using a 250- $\Omega$  pull-up resistor with a 30-pF capacitive load.

#### 8.3.2 Output Enable

To enable the I/O pins, the EN input should be tied directly to  $V_{ref\_B}$  during operation and both pins must be pulled up to the HIGH side ( $V_{CCB}$ ) through a bias resistor (typically 200 k $\Omega$ ). To ensure the high impedance state during power-up, power-down, or during operation, the EN pin must be LOW. The EN pin should always be tied directly to the  $V_{ref\_B}$  pin and is recommended to be disabled by an open-drain driver without a pullup resistor. This allows  $V_{ref\_B}$  to regulate the EN input and bias the channels for proper translation. A filter capacitor on  $V_{ref\_B}$  is recommended for a stable supply at the device.

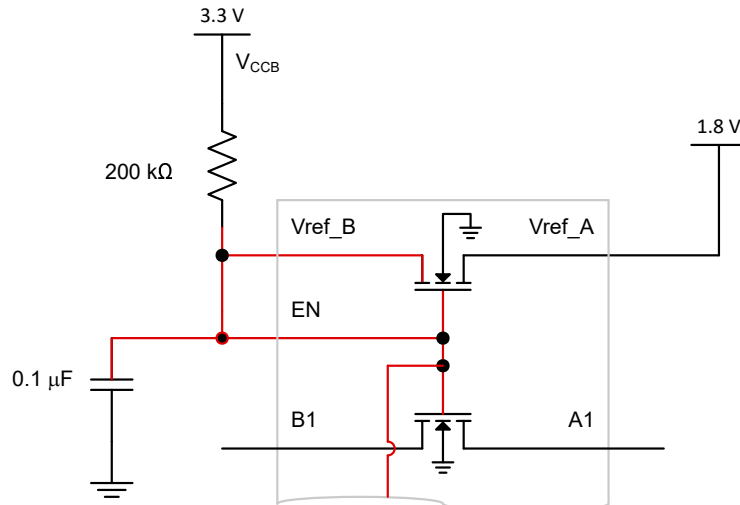


图 8-1. EN Pin Tied to  $V_{ref\_B}$  Directly and to  $V_{CCB}$  Through a Pull-Up Resistor

The supply voltage of open drain I/O devices can be completely different from the supplies used for the LSF and has no impact on the operation. For additional details on how to use the enable pin, see the [Using the Enable Pin with the LSF Family video](#).

表 8-1. EN Pin Function Table

INPUT EN <sup>(1)</sup> PIN	Data Port State
Tied directly to $V_{ref\_B}$	An = Bn
L	Hi-Z

(1) EN is controlled by  $V_{ref\_B}$  logic levels.

### 8.3.3 Device Functional Modes

For each channel (n), when either the An or Bn port is LOW, the switch provides a low impedance path between the An and Bn ports; the corresponding Bn or An port will be pulled LOW. The low  $R_{ON}$  of the switch allows connections to be made with minimal propagation delay and signal distortion.

表 8-1 provides a summary of device operation. For additional details on the functional operation of the LSF family of devices, see the [Down Translation with the LSF Family](#) and [Up Translation with the LSF Family](#) videos.

表 8-2. Device Functionality

Signal Direction <sup>(1)</sup>	Input State	Switch State	Functionality
B to A (Down Translation)	B = LOW	ON (Low Impedance)	A-side voltage is pulled low through the switch to the B-side voltage
	B = HIGH	OFF (High Impedance)	A-side voltage is clamped at $V_{ref\_A}$ <sup>(2)</sup>
A to B (Up Translation)	A = LOW	ON (Low Impedance)	B-side voltage is pulled low through the switch to the A-side voltage
	A = HIGH	OFF (High Impedance)	B-side voltage is clamped at $V_{ref\_A}$ and then pulled up to the $V_{PU}$ supply voltage

(1) The downstream channel should not be actively driven through a low impedance driver, or else bus contention may occur.

(2) The A-side can have a pullup to  $V_{ref\_A}$  for additional current drive capability or may also be pulled above  $V_{ref\_A}$  with a pullup resistor. Specifications in the [Recommended Operating Conditions](#) section should always be followed.

### 8.3.3.1 Up and Down Translation

#### 8.3.3.1.1 Up Translation

When the signal is being driven from A to B and the An port is HIGH, the switch will be OFF and the Bn port will then be driven to a voltage higher than  $V_{ref\_A}$  by the pull-up resistor that is connected to the pull-up supply voltage ( $V_{PU}$ ). This functionality allows seamless translation between higher and lower voltages selected by the user, without the need for directional control. Pull-up resistors are always required on the high side, and pull-ups are only required on the low side, if the low side of the device's output is open drain or its input has a leakage greater than  $1\ \mu\text{A}$ .

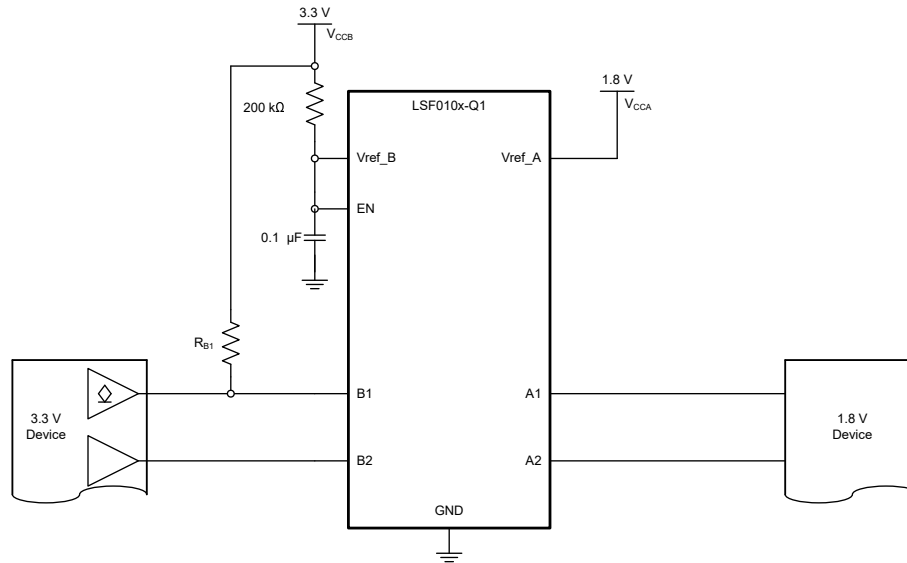


Figure 8-2. Up Translation Example Schematic with Push-Pull and Open Drain Configuration

Up translation with the LSF requires attention to two important factors: maximum data rate and sink current. Maximum data rate is directly related to the rising edge of the output signal. Sink current depends on supply values and the chosen pull-up resistor values. 式 1 shows the maximum data rate formula and 式 2 shows the maximum sink current formula, both of which are estimations. A low RC value is needed to reach high speeds, which also require strong drivers. Please see the [Up Translation with the LSF Family](#) video for estimated data rate and sink current calculations based on circuit components.

$$\frac{1}{3 \times 2R_{B1}C_{B1}} = \frac{1}{6R_{B1}C_{B1}} \left( \frac{\text{bits}}{\text{second}} \right) \quad (1)$$

$$I_{OL} \cong \frac{V_{CCA}}{R_{A1}} + \frac{V_{CCB}}{R_{B1}} \quad (A) \quad (2)$$

#### 8.3.3.1.2 Down Translation

When the signal is being driven HIGH from the Bn port to An port, the switch will be OFF, clamping the voltage on the An port to the voltage set by  $V_{ref\_A}$ . A pull-up resistor can be added on either side of the device. There are special circumstances that allow the removal of one or both of the pull-up resistors. If the signal is always going to be down translated from a push-pull transmitter, then the resistor on the B-side can be removed. If the leakage current into the receiver on the A-side is less than  $1\ \mu\text{A}$ , then the resistor on the A-side can also be removed. This arrangement with no external pull-up resistors can be used when down translating from a push-pull output to a low-leakage input. For an open drain transmitter, the pull-up resistor on the B-side is necessary because an open drain output can't drive high by itself. For a summary of device operation, refer to [セクション 8.3.3](#). For additional details on the functional operation of the LSF family of devices, see the [Up Translation with the LSF Family](#) and [Down Translation with the LSF Family](#) videos.

## 9 Application and Implementation

### 注

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### 9.1 Application Information

The LSF0102-Q1 device is able to perform voltage translation for open-drain or push-pull interfaces such as I<sup>2</sup>C, SPI, UART, MDIO, SDIO, and GPIO.

### 9.2 Typical Application

#### 9.2.1 Bidirectional Translation

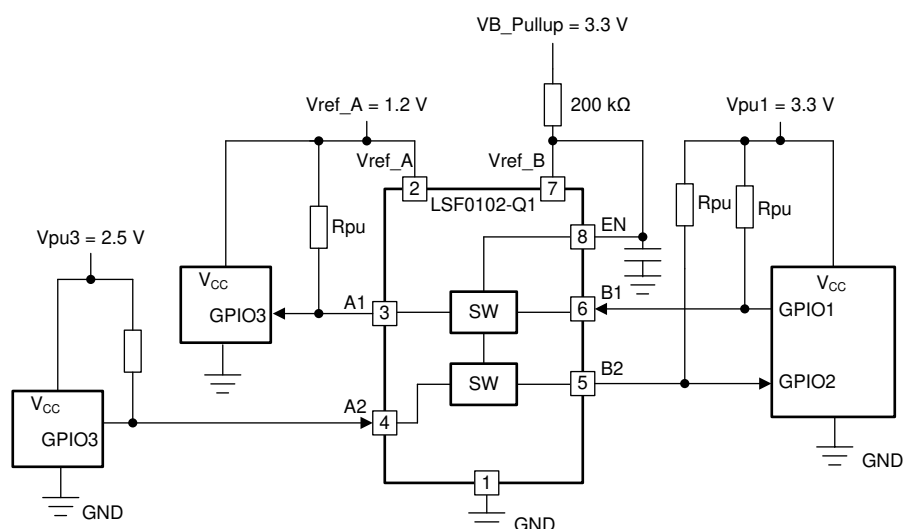


図 9-1. Bidirectional Translation to Multiple Voltage Levels

#### 9.2.1.1 Design Requirements

##### 9.2.1.1.1 Enable, Disable, and Reference Voltage Guidelines

The LSF0102-Q1 device has an EN input that is used to disable the device by setting EN LOW, which places all I/Os in the high-impedance state. Since LSF family is switch-type voltage translator, the power consumption is very low. It is recommended to always enable LSF0102-Q1 device for bidirectional applications by connecting the EN pin to the  $V_{ref\_B}$  pin, as shown in 図 9-1. For additional details on setting up the  $V_{ref\_A}$ ,  $V_{ref\_B}$ , and EN pins, see the [Understanding the Bias Circuit for the LSF Family](#) video.

表 9-1. Application Operating Condition

PARAMETER		MIN	TYP	MAX	UNIT
$V_{ref\_A}$ <sup>(1)</sup>	reference voltage (A)	0.95		5.5	V
$V_{ref\_B}$	reference voltage (B)	$V_{ref\_A} + 0.8$		5.5	V
$V_{I(EN)}$	input voltage on EN pin	$V_{ref\_A} + 0.8$	$V_{ref\_B}$	5.5	V
$V_{PU}$	pull-up supply voltage	0		$V_{ref\_B}$	V

(1)  $V_{ref\_A}$  is required to be the lowest voltage level across all inputs and outputs.

The 200 kΩ, bias resistor is required to allow  $V_{ref\_B}$  to regulate the EN input. A filter capacitor on  $V_{ref\_B}$  is recommended. Also  $V_{ref\_B}$  and  $V_{I(EN)}$  are recommended to be 1.0 V higher than  $V_{ref\_A}$  for best signal integrity.

### 9.2.1.1.2 Bias Circuitry

For proper operation,  $V_{CCA}$  must always be at least 0.8 V less than  $V_{CCB}$  ( $V_{CCA} + 0.8 \leq V_{CCB}$ ). The 200 k $\Omega$  bias resistor is required to allow  $V_{ref\_B}$  to regulate the EN input and properly bias the device for translation. A 0.1  $\mu$ F capacitor is recommended for providing a path from  $V_{ref\_B}$  to ground for high frequency noise.  $V_{ref\_B}$  and  $V_{I(EN)}$  are recommended to be 1.0 V higher than  $V_{ref\_A}$  for best signal integrity.

Attempting to drive the EN pin directly with a push-pull output device is a very common design error with the LSF0102-Q1 series of devices. It is also very important to note that current does flow into the A-side voltage supply during normal operation. Not all voltage sources can sink current, so be sure that applicable designs can handle this current. For more design details, see the [Understanding the Bias Circuit for the LSF Family](#) video.

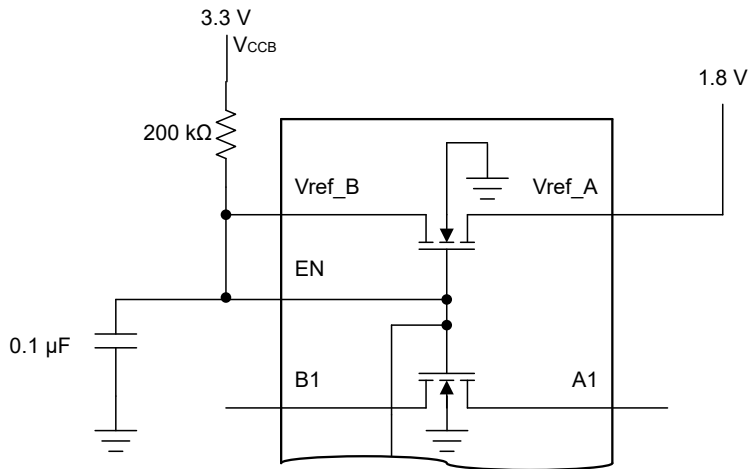


Figure 9-2. Bias Circuitry Inside the LSF010x-Q1 Device

## 9.2.1.2 Detailed Design Procedure

### 9.2.1.2.1 Bidirectional Translation

For the bidirectional clamping configuration (higher voltage to lower voltage or lower voltage to higher voltage), the EN input must be connected to  $V_{ref\_B}$  and both pins pulled to HIGH side  $V_{CCB}$  through a bias resistor (typically 200 k $\Omega$ ), as shown in Figure 9-1. This allows  $V_{ref\_B}$  to regulate the EN input. A filter capacitor on  $V_{ref\_B}$  is recommended. The controller output driver can be push-pull or open-drain (pull-up resistors may be required) and the peripheral device output can be push-pull or open-drain (pull-up resistors are required to pull the Bn outputs to  $V_{pu}$ ).

If either output is push-pull, data must be unidirectional or the outputs must be tri-state and be controlled by some direction-control mechanism to prevent HIGH-to-LOW contention in either direction. If both outputs are open-drain, no direction control is needed.

In Figure 9-1, the reference supply voltage  $V_{ref\_A}$  is connected to the processor core power supply voltage.  $V_{ref\_B}$  is connected through a 200 k $\Omega$  resistor to a 3.3 V  $V_{B\_Pullup}$  power supply and  $V_{ref\_A}$  is set to 1.2 V. The output of A1 has a maximum output voltage equal to  $V_{ref\_A}$ , and the bidirectional interface on channel 2 has a maximum output voltage equal to  $V_{PU1}$ .

### 9.2.1.2.2 Pull-Up Resistor Sizing

To maintain an appropriate output low voltage, the pull-up resistor value should limit the current through the pass transistor when it is in the ON state to less than 15 mA. This ensures a pass voltage of 260 mV to 350 mV. To set the current through each pass transistor at 15 mA, the pull-up resistor value can be calculated using the following equation:

$$R_{pu} = \frac{(V_{pu} - 0.35 V)}{0.015 A} \quad (3)$$

The appropriate pull up resistor will depend on the current requirements of the application. 表 9-2 provides resistor values, reference voltages, and currents at 15 mA, 10 mA, and 3 mA. The resistor value shown in the +10% column (or a larger value) should be used to ensure that the pass voltage of the transistor is 350 mV or less. The external driver must be able to sink the total current from the resistors on both sides of the LSF0102-Q1 device at 0.175 V, although the 15 mA applies only to current flowing through the LSF0102-Q1.

表 9-2. Pull-up Resistor Values

$V_{PU}$	8 mA		5 mA		3 mA	
	NOMINAL ( $\Omega$ )	+10% <sup>(1)</sup> ( $\Omega$ )	NOMINAL ( $\Omega$ )	+10% <sup>(1)</sup> ( $\Omega$ )	NOMINAL ( $\Omega$ )	+10% <sup>(1)</sup> ( $\Omega$ )
5 V	581	639	930	1023	1550	1705
3.3 V	369	406	590	649	983	1082
2.5 V	269	296	430	473	717	788
1.8 V	181	199	290	319	483	532
1.5 V	144	158	230	253	383	422
1.2 V	106	117	170	187	283	312

(1) +10% to compensate for  $V_{DD}$  range and resistor tolerance

### 9.2.1.2.3 Application Curve

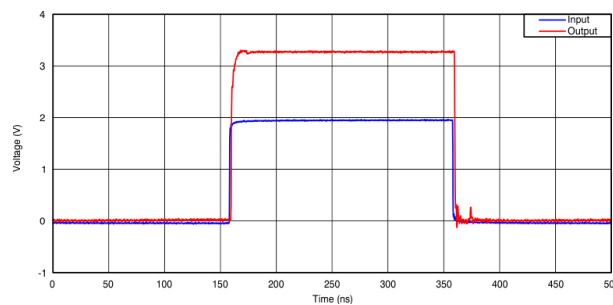


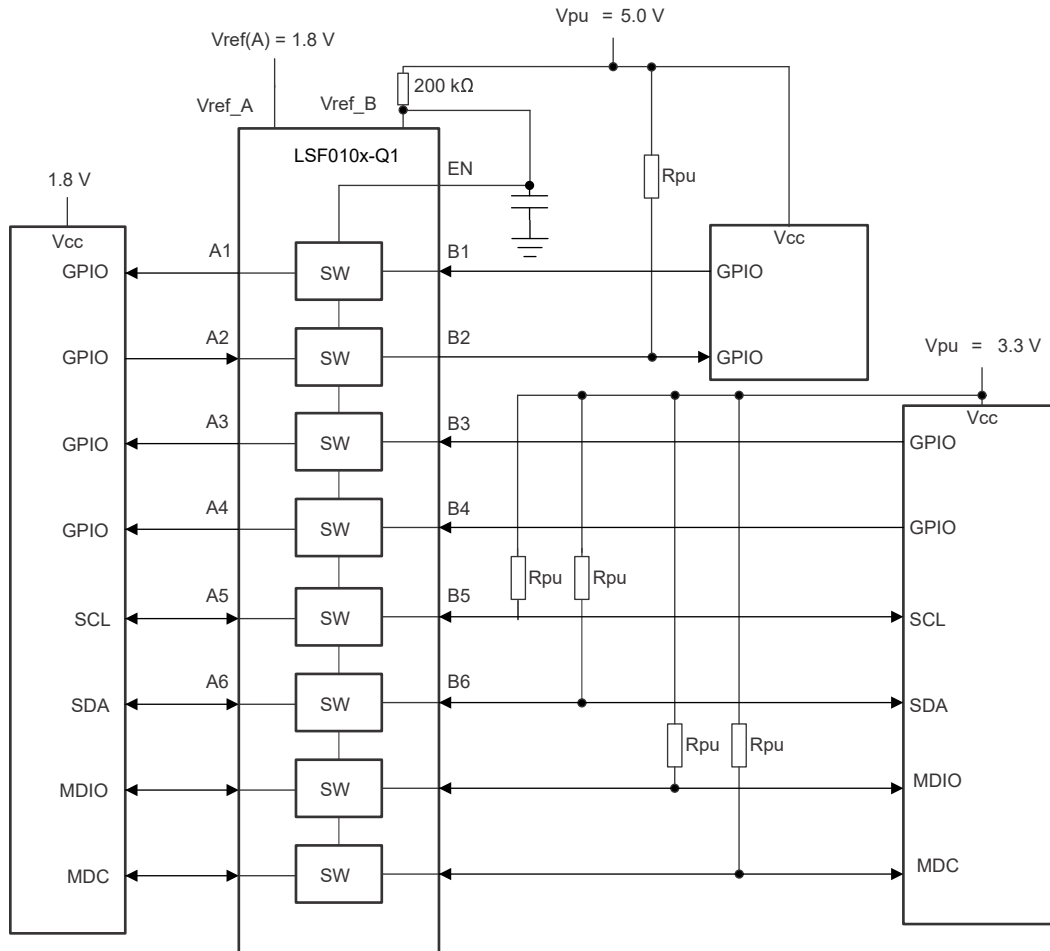
图 9-3. Captured Waveform From Above I<sup>2</sup>C Set-Up (1.8 V to 3.3 V at 2.5 MHz)

### 9.2.1.2.4 Mixed-Mode Voltage Translation

The supply voltage ( $V_{PU}$ ) for each channel can be individually set with a pull-up resistor. 图 9-4 shows an example of this mixed-mode multi-voltage translation. For additional details on multi-voltage translation, see the [Multi-voltage Translation with the LSF Family](#) video.

With the  $V_{ref\_B}$  pulled up to 5 V and  $V_{ref\_A}$  connected to 1.8 V, all channels will be clamped to 1.8 V at which point a pullup can be used to define the high level voltage for a given channel.

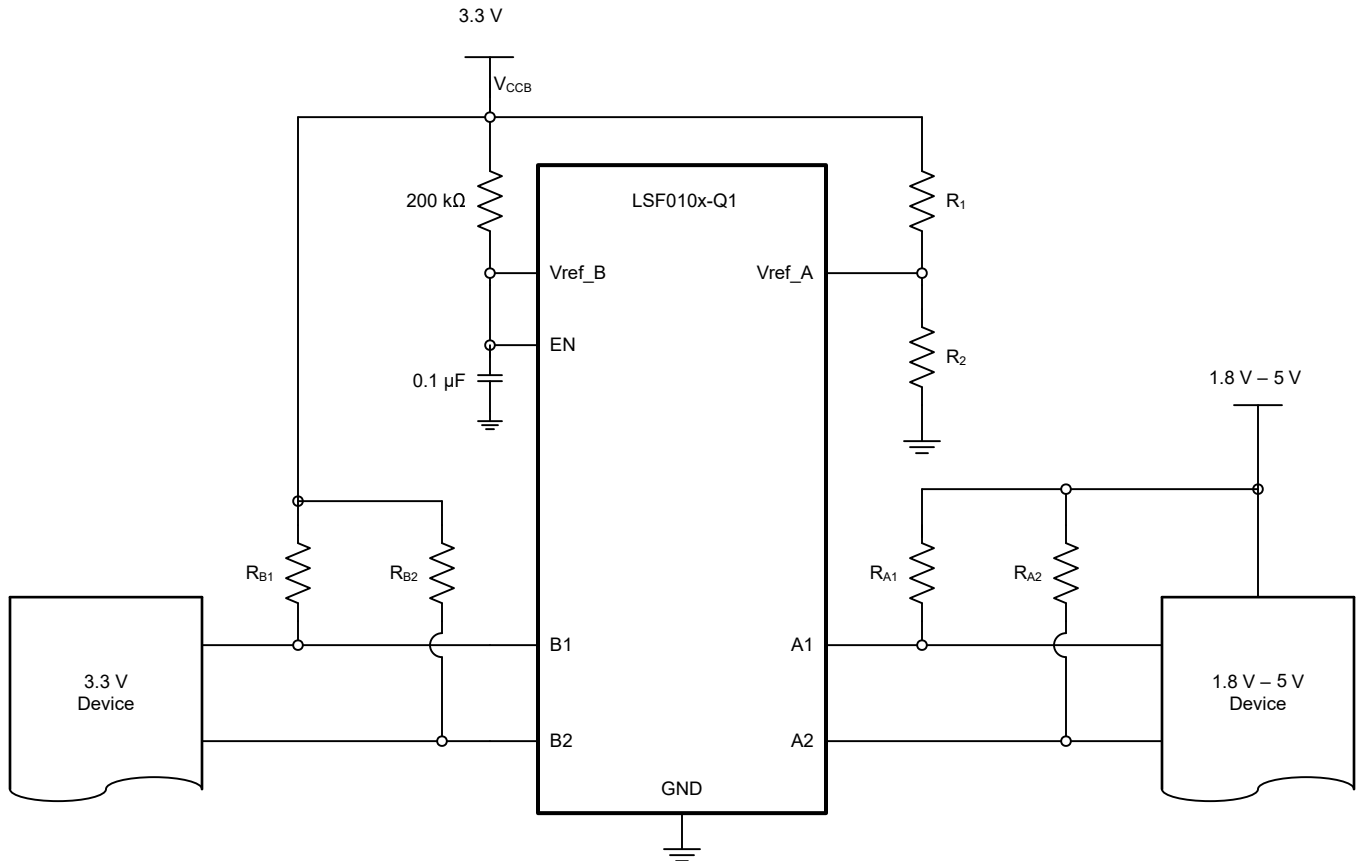
- **Push-Pull Down Translation (5 V to 1.8 V):** Channel 1 is an example of this setup. When B1 is 5 V, A1 is clamped to 1.8 V, and when B1 is LOW, A1 is driven LOW through the switch.
- **Push-Pull Up Translation (1.8 V to 5 V):** Channel 2 is an example of this setup. When A2 is 1.8 V, the switch is high impedance and the B2 channel is pulled up to 5 V. When A2 is LOW, B2 is driven LOW through the switch.
- **Push-Pull Down Translation (3.3 V to 1.8 V):** Channels 3 and 4 are examples of this setup. When either B3 or B4 are driven to 3.3 V, A3 or A4 are clamped to 1.8 V, and when either B3 or B4 are LOW, A3 or A4 are driven LOW through the switch.
- **Open-Drain Bidirectional Translation (3.3 V ↔ 1.8 V):** Channels 5 through 8 are examples of this setup. These channels are for bidirectional operation for I<sup>2</sup>C and MDIO to translate between 1.8 V and 3.3 V with open-drain drivers.



**9-4. Multi-Voltage Translation with the LSF010x-Q1**

#### 9.2.1.2.5 Single Supply Translation

Sometimes, an external device will have an unknown voltage that could be above or below the desired translation voltage, preventing a normal connection of the LSF. Resistors are added on the A side in place of the second supply in this case – this is an example of when LSF single supply operation is utilized, shown in [Figure 9-5](#). In the following figure, a single 3.3 V supply is used to translate between a 3.3 V device and a device that can change between 1.8 V and 5.0 V. R1 and R2 are added in place of the second supply. Note that due to some current coming out of the  $V_{ref\_A}$  pin, this cannot be treated as a simple voltage divider.



**9-5. Single Supply Translation with 3.3 V Supply**

The steps to select the resistor values for R1 and R2 are as follows:

1. Select a value for R1. Typically, 1 MΩ is used to reduce current consumption.
2. Plug in values for your system into the following equation. Note that  $V_{ref\_A}$  is the lowest voltage in the system.  $V_{CCB}$  is the primary supply and R1 is the selected value from step 1.

$$R_2 = \frac{200(10^3) \times R_1 \times V_{REFA}}{(200(10^3) + R_1)(V_{CCB} - V_{REFA}) - 0.85 \times R_1} \quad (4)$$

The single supply used must be at least 0.8 V larger than the lowest desired translation voltage. The voltage at  $V_{ref\_A}$  must be selected as the lowest voltage to be used in the system. The LSF evaluation module (LSF-EVM) contains unpopulated pads to place R1 and R2 for single supply operation testing. For an example single supply translation schematic and details, see the [Single Supply Translation with the LSF Family](#) video.



### 9.2.1.2.6 Voltage Translation for $V_{ref\_B} < V_{ref\_A} + 0.8\text{ V}$

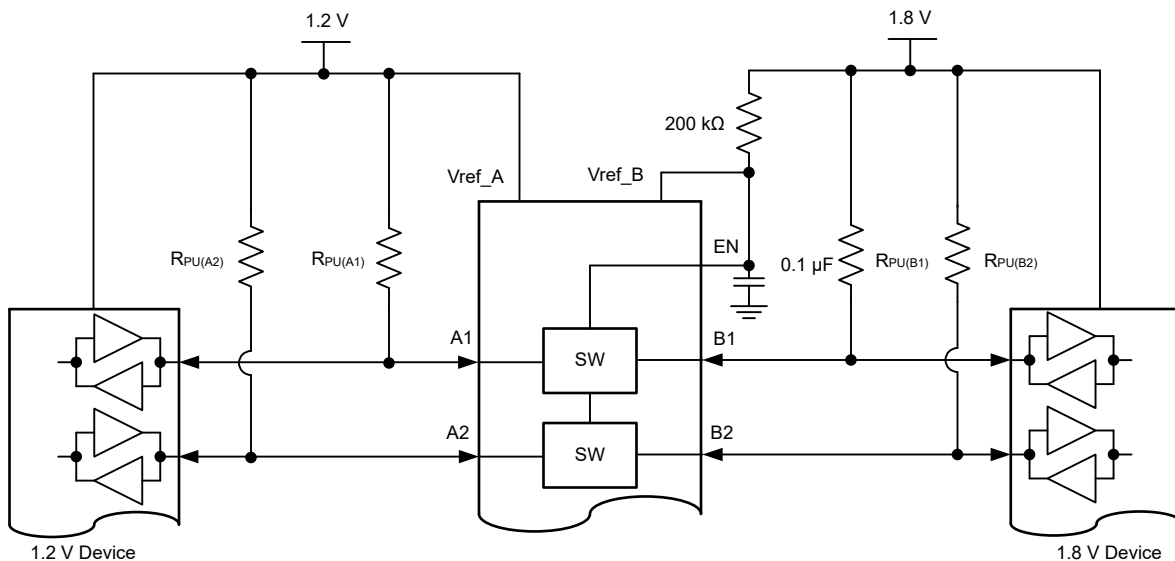
As described in the *Enable, Disable, and Reference Voltage Guidelines* section, it is generally recommended that  $V_{ref\_B} > V_{ref\_A} + 0.8\text{ V}$ ; however, the device can still operate in the condition where  $V_{ref\_B} < V_{ref\_A} + 0.8\text{ V}$  as long as additional considerations are made for the design.

**Typical Operation ( $V_{ref\_B} > V_{ref\_A} + 0.8\text{ V}$ ):** in this scenario, pullup resistors are not required on the A-side for proper down-translation as is shown for channels 1 and 2 of [Figure 9-4](#). The typical operating mode of the device ensures that when down translating from B to A, the A-side I/O ports will clamp at  $V_{ref\_A}$  to provide proper voltage translation. For further explanation of device operation, see the [Down Translation with the LSF Family](#) video.

**Requirements for  $V_{ref\_B} < V_{ref\_A} + 0.8\text{ V}$  Operation:** in this scenario, there is not a large enough voltage difference between  $V_{ref\_A}$  and  $V_{ref\_B}$  to ensure that the A side I/O ports will be clamped at  $V_{ref\_A}$ , but rather at a voltage approximately equal to  $V_{ref\_B} - 0.8\text{ V}$ . For example, if  $V_{ref\_B} = 1.8\text{ V}$  and  $V_{ref\_A} = 1.2\text{ V}$ , the A-side I/Os will clamp to a voltage around  $1.0\text{ V}$ . Therefore, to operate in such a condition, the following additional design considerations must be met:

- $V_{ref\_B}$  must be greater than  $V_{ref\_A}$  during operation ( $V_{ref\_B} > V_{ref\_A}$ )
- Pullup resistors should be populated on A-side I/O ports to ensure the line will be fully pulled up to the desired voltage.

[Figure 9-6](#) shows an example of this setup, where  $1.2\text{ V} \leftrightarrow 1.8\text{ V}$  translation is achieved with the LSF0102-Q1. This type of setup also applies for other voltage nodes such as  $1.8\text{ V} \leftrightarrow 2.5\text{ V}$ ,  $1.05\text{ V} \leftrightarrow 1.5\text{ V}$ , and others as long as the [Recommended Operating Conditions](#) table is followed.



**Figure 9-6. 1.2 V to 1.8 V Level Translation with LSF010x**

## 9.3 Power Supply Recommendations

There are no power sequence requirements for the LSF family. [Table 9-3](#) provides the recommended operating voltages for all supply and input pins.

**Table 9-3. Recommended Operating Voltages**

PARAMETER		MIN	TYP	MAX	UNIT
$V_{ref\_A}$ <sup>(1)</sup>	reference voltage (A)	0.95		5.5	V
$V_{ref\_B}$	reference voltage (B)	$V_{ref\_A} + 0.8$		5.5	V
$V_{I(EN)}$	input voltage on EN pin	$V_{ref\_A} + 0.8$		5.5	V

**表 9-3. Recommended Operating Voltages (continued)**

PARAMETER	MIN	TYP	MAX	UNIT
V <sub>PU</sub> pull-up supply voltage	0		V <sub>ref_B</sub>	V

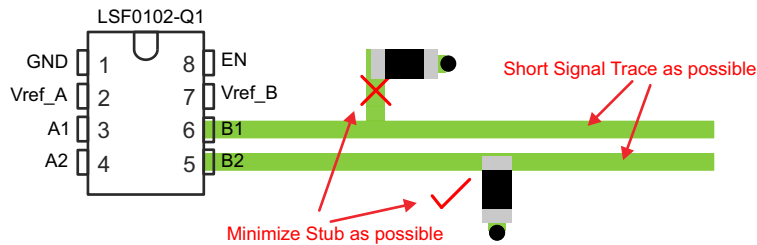
## 9.4 Layout

### 9.4.1 Layout Guidelines

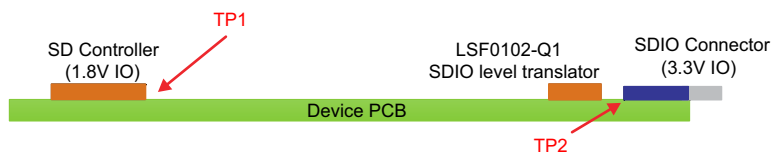
Because the LSF0102-Q1 device is a switch-type level translator, the signal integrity is dependent upon the pull-up resistor value and PCB board parasitics. Consider the following recommendations when designing with the LSF0102-Q1.

- Minimize the signal trace length to reduce capacitance
- Avoid using stubs in the signal path to reduce parasitics.
- Place the LSF0102-Q1 device near the high voltage side.
- Select the appropriate pull-up resistor that applies to translation levels and driving capability of transmitter.

### 9.4.2 Layout Example



**图 9-7. Short Trace Layout**



**图 9-8. Device Placement**

## 10 Device and Documentation Support

### 10.1 Documentation Support

#### 10.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, [TI Logic Minute: Introduction – Voltage Level Translation with the LSF Family](#) video
- Texas Instruments, [Voltage-Level Translation with the LSF Family](#) application report

### 10.2 ドキュメントの更新通知を受け取る方法

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[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

## 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable part number	Status (1)	Material type (2)	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
<a href="#">LSF0102QDCURQ1</a>	Active	Production	VSSOP (DCU)   8	3000   LARGE T&R	Yes	NIPDAU   NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	(2SMT, NG2SQ)
LSF0102QDCURQ1.A	Active	Production	VSSOP (DCU)   8	3000   LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	(2SMT, NG2SQ)

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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**OTHER QUALIFIED VERSIONS OF LSF0102-Q1 :**

- Catalog : [LSF0102](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product

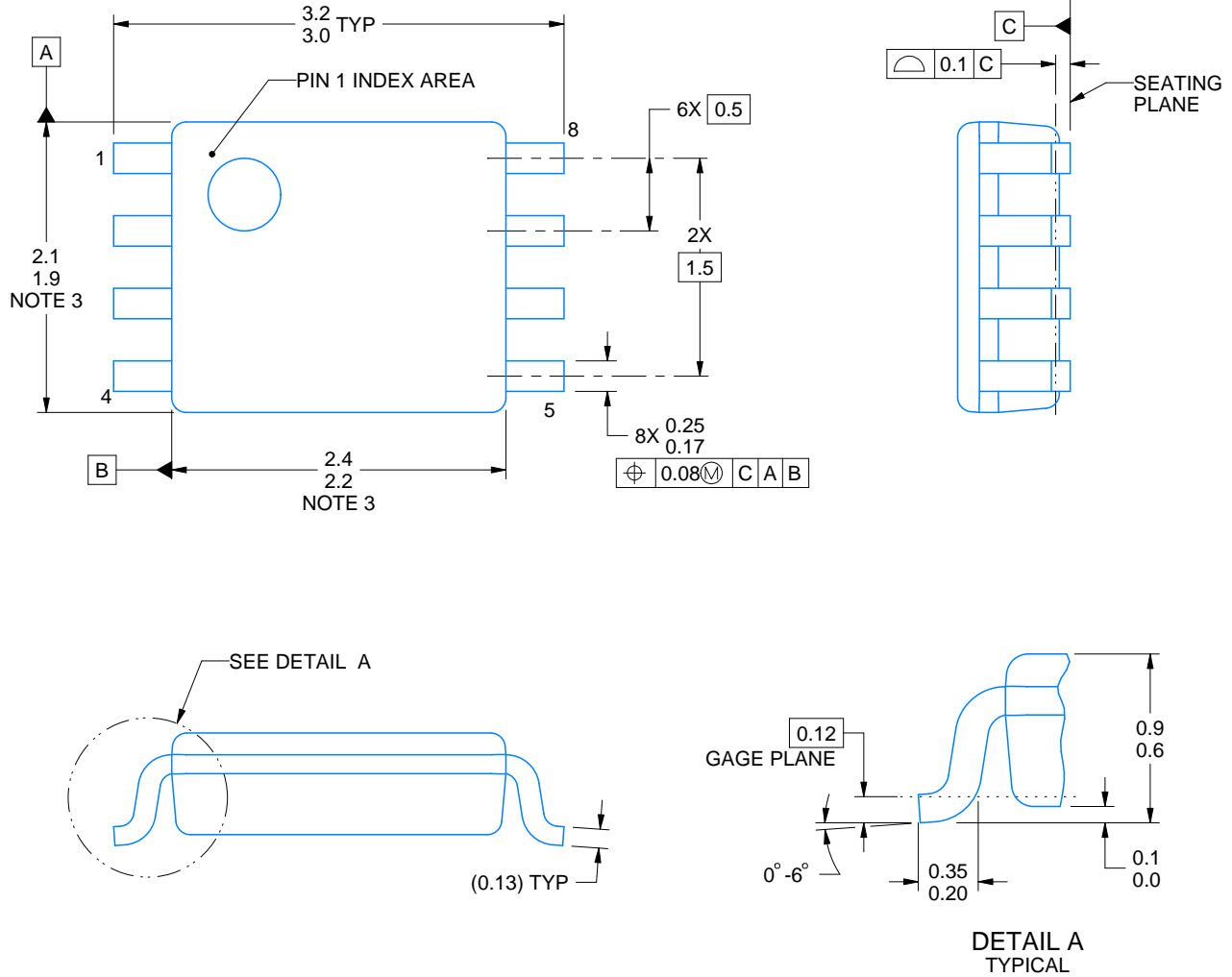
DCU0008A



# PACKAGE OUTLINE

VSSOP - 0.9 mm max height

SMALL OUTLINE PACKAGE



4225266/A 09/2014

NOTES:

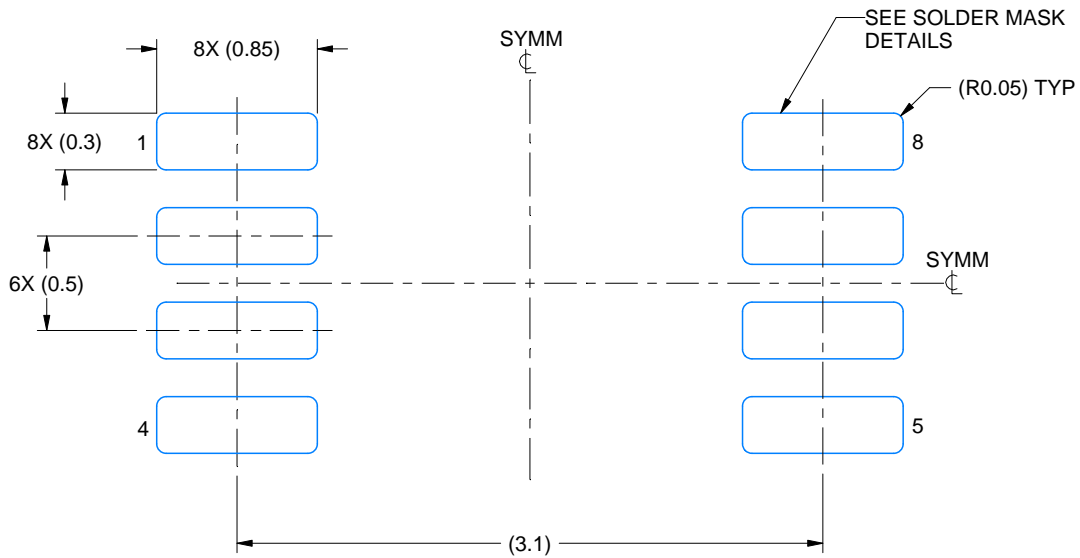
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. Reference JEDEC registration MO-187 variation CA.

# EXAMPLE BOARD LAYOUT

DCU0008A

VSSOP - 0.9 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 25X



4225266/A 09/2014

NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DCU0008A

VSSOP - 0.9 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE: 25X

4225266/A 09/2014

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.



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