

LMH6640 TFT-LCD Single, 16V Rail-to-Rail High Output Operational Amplifier

Check for Samples: LMH6640

FEATURES

- (V_S = 16V, R_L= 2 kΩ to V⁺/2, 25°C, Typical Values Unless Specified)
- Supply current (no load) 4 mA
- Output resistance (closed loop 1 MHz) 0.35Ω
- -3 dB BW (A_V = 1) 190 MHz
- Settling time (±0.1%, 2 V_{PP}) 35 ns
- Input common mode voltage −0.3V to 15.1V
- Output voltage swing 100 mV from rails
- Linear output current ±100 mA
- Total harmonic distortion (2 V_{PP}, 5 MHz) -64 dBc
- Fully characterized for: 5V & 16V
- No output phase reversal with CMVR exceeded
- Differential gain ($R_1 = 150\Omega$) 0.12%
- Differential phase (R_L = 150Ω) 0.12°

APPLICATIONS

- TFT panel V_{COM} buffer amplifier
- Active filters
- CD/DVD ROM
- ADC buffer amplifier
- Portable video
- Current sense buffer

DESCRIPTION

The LMH[™]6640 is a voltage feedback operational amplifier with a rail-to-rail output drive capability of 100 mA. Employing TI's patented VIP10 process, the LMH6640 delivers a bandwidth of 190 MHz at a current consumption of only 4mA. An input common mode voltage range extending to 0.3V below the V-and to within 0.9V of V⁺, makes the LMH6640 a true single supply op-amp. The output voltage range extends to within 100 mV of either supply rail providing the user with a dynamic range that is especially desirable in low voltage applications.

The LMH6640 offers a slew rate of 170 V/µs resulting in a full power bandwidth of approximately 28 MHz with 5V single supply (2 V_{PP}, -1 dB). Careful attention has been paid to ensure device stability under all operating voltages and modes. The result is a very well behaved frequency response characteristic for any gain setting including +1, and excellent specifications for driving video cables including total harmonic distortion of -64 dBc @ 5 MHz, differential gain of 0.12% and differential phase of 0.12°.

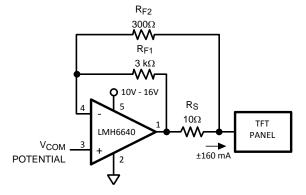


Figure 1. Typical Application as a TFT Panel V_{COM} Driver

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings (1)

2 KV
200V
±2.5V
±10 mA
18V
V ⁺ +0.8V, V [−] −0.8V
−65°C to +150°C
+150°C
235°C
260°C

- (1) Absolute maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For specifications and the test conditions, see the Electrical Characteristics.
- (2) Human body model, 1.5 k Ω in series with 100 pF. Machine Model, 0Ω in series with 200 pF.
- (3) The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} T_A) / \theta_{JA}$. All numbers apply for packages soldered directly onto a PC board.

Operating Ratings (1)

<u> </u>	
Supply Voltage (V ⁺ – V ⁻)	4.5V to 16V
Operating Temperature Range (2)	-40°C to +85°C
Package Thermal Resistance (2)	
5-Pin SOT-23	265°C/W

- (1) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150 °C Short circuit test is a momentary test. Output short circuit duration is infinite for V_S < 6V at room temperature and below. For V_S > 6V, allowable short circuit duration is 1.5 ms.
- infinite for V_S < 6V at room temperature and below. For V_S > 6V, allowable short circuit duration is 1.5 ms.

 (2) The maximum power dissipation is a function of T_{J(MAX)}, θ_{JA}, and T_A. The maximum allowable power dissipation at any ambient temperature is P_D = (T_{J(MAX)}-T_A) / θ_{JA}. All numbers apply for packages soldered directly onto a PC board.



5V Electrical Characteristics

Unless otherwise specified, All limits specified for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_O = V_{CM} = V^+/2$ and $R_L = 2 \text{ k}\Omega$ to $V^+/2$. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter Conditions				Typ ⁽³⁾	Max ⁽²⁾	Units	
BW	-3 dB Bandwidth	$A_V = +1 \ (R_L = 100\Omega)$		150		N 41 1		
		$A_V = -1 \ (R_L = 100\Omega)$		58		MHz		
BW _{0.1 dB}	0.1 dB Gain Flatness	A _V = −3		18		MHz		
FPBW	Full Power Bandwidth	A _V = +1, V _{OUT} = 2 V _{PP} , −1 dB					MHz	
LSBW	-3 dB Bandwidth	$A_V = +1$, $V_O = 2 V_{PP} (R_L = 100)$	Ω)		32		MHz	
GBW	Gain Bandwidth Product	$A_V = +1$, $(R_L = 100\Omega)$					MHz	
SR	Slew Rate (4)	A _V = −1			170		V/µs	
e _n	Input Referred Voltage Noise		f = 10 kHz		23		->4//	
			f = 1 MHz		15		nV/√Hz	
i _n	Input Referred Current Noise		f = 10 kHz		1.1		- A / /II	
			f = 1 MHz		0.7		pA/√Hz	
THD	Total Harmonic Distortion	$f = 5 \text{ MHz}, V_O = 2 V_{PP}, A_V = +1 R_L = 1 k\Omega \text{ to } V^+/2$	2		-65		dBc	
t _s	Settling Time	$V_O = 2 V_{PP}, \pm 0.1\%, A_V = -1$			35		ns	
V _{OS}	Input Offset Voltage				1	5 7	mV	
I _B	Input Bias Current (5)				-1.2	-2.6 -3.25	μА	
I _{OS}	Input Offset Current			34	800 1400	nA		
CMVR	Common Mode Input Voltage Range	,				-0.2 -0.1		
				4.0 3.6	4.1		V	
CMRR	Common Mode Rejection Ratio	$V^{-} \le V_{CM} \le V^{+} -1.5V$		72	90		dB	
A _{VOL}	Large Signal Voltage Gain	$V_{O} = 4 V_{PP}, R_{L} = 2 k\Omega \text{ to } V^{+}/2$		86 82	95		40	
		$V_{O} = 3.75 V_{PP}, R_{L} = 150\Omega \text{ to } V_{C}$	/+/2	74 70	78		dB	
Vo	Output Swing High	$R_L = 2 k\Omega \text{ to } V^+/2$		4.90	4.94			
		$R_L = 150\Omega \text{ to V}^{+}/2$		4.75	4.80		.,	
	Output Swing Low	$R_L = 2 k\Omega \text{ to } V^+/2$			0.06	0.10	V	
		$R_L = 150\Omega \text{ to V}^+/2$			0.20	0.25		
I _{SC}	Output Short Circuit Current (6)			100 75	130			
		Sinking from V ⁺ /2		100 70	130		mA mA	
I _{OUT}	Output Current	V _O = 0.5V from either Supply		+75/-90		mA		
PSRR	Power Supply Rejection Ratio	4V ≤ V ⁺ ≤ 6V	72	80		dB		
Is	Supply Current	No Load			3.7	5.5 8.0	mA	
R _{IN}	Common Mode Input Resistance	$A_V = +1$, $f = 1$ kHz, $R_S = 1$ M Ω			15		ΜΩ	

⁽¹⁾ Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T_J = T_A. Parametric performance is indicated in the electrical tables under conditions of internal self-heating where T_J > T_A.

⁽²⁾ All limits are specified by testing or statistical analysis.

⁽³⁾ Typical Values represent the most likely parametric norm.

⁽⁴⁾ Slew rate is the average of the rising and falling slew rates

⁽⁵⁾ Positive current corresponds to current flowing into the device.

⁽⁶⁾ Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150 °C Short circuit test is a momentary test. Output short circuit duration is infinite for V_S < 6V at room temperature and below. For V_S > 6V, allowable short circuit duration is 1.5 ms.



5V Electrical Characteristics (continued)

Unless otherwise specified, All limits specified for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_O = V_{CM} = V^+/2$ and $R_L = 2 \text{ k}\Omega$ to $V^+/2$. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
C _{IN}	Common Mode Input Capacitance	$A_V = +1, R_S = 100 \text{ k}\Omega$		1.7		рF
R _{OUT}	Output Resistance Closed Loop	$R_F = 10 \text{ k}\Omega, f = 1 \text{ kHz}, A_V = -1$		0.1		Ω
		$R_F = 10 \text{ k}\Omega, f = 1 \text{ MHz}, A_V = -1$		0.4		12
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.13		%
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.10		deg



16V Electrical Characteristics

Unless otherwise specified, All limits specified for $T_J = 25^{\circ}C$, $V^+ = 16V$, $V^- = 0V$, $V_O = V_{CM} = V^+/2$ and $R_L = 2 \text{ k}\Omega$ to $V^+/2$. **Boldface** limits apply at temperature extremes. (1)

Symbol Parameter		Parameter Conditions		Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
BW	-3 dB Bandwidth	$A_V = +1 (R_L = 100\Omega)$			190		
		$A_V = -1 \ (R_L = 100\Omega)$		60		MHz	
BW _{0.1 dB}	0.1 dB Gain Flatness	A _V = −2.7		20		MHz	
LSBW	-3 dB Bandwidth	$A_V = +1$, $V_O = 2 V_{PP} (R_L = 10)$		35		MHz	
GBW	Gain Bandwidth Product	$A_V = +1$, $(R_L = 100\Omega)$			62		MHz
SR	Slew Rate (4)	A _V = −1			170		V/µs
e _n	Input Referred Voltage Noise		f = 10 kHz		23		\ , , , /
			f = 1 MHz		15		nV/√Hz
i _n	Input Referred Current Noise		f = 10 kHz		1.1		A / /III
			f = 1 MHz		0.7		pA/√Hz
THD	Total Harmonic Distortion	$f = 5 \text{ MHz}, V_O = 2 V_{PP}, A_V = R_L = 1 k\Omega \text{ to } V^+/2$	+2		-64		dBc
t _s	Settling Time	$V_O = 2 V_{PP}, \pm 0.1\%, A_V = -1$			35		ns
V _{OS}	Input Offset Voltage				1	5 7	mV
I _B	Input Bias Current (5)				-1	-2.6 -3.5	μА
los	Input Offset Current			34	800 1800	nA	
CMVR	Common Mode Input Voltage Range	CMRR ≥ 50 dB		-0.3	-0.2 -0.1	V	
				15.0 14.6	15.1		V
CMRR	Common Mode Rejection Ratio	$V^- \le V_{CM} \le V^+ -1.5V$	72	90		dB	
A _{VOL}	Large Signal Voltage Gain	$V_O = 15 \text{ V}_{PP}, R_L = 2 \text{ k}\Omega \text{ to } V^T$	⁺ /2	86 82	95		٩D
		$V_{O} = 14 \text{ V}_{PP}, R_{L} = 150\Omega \text{ to V}$	/+/2	74 70	78		dB
Vo	Output Swing High	$R_L = 2 k\Omega \text{ to } V^+/2$		15.85	15.90		
		$R_L = 150\Omega \text{ to } V^+/2$		15.45	15.78		V
	Output Swing Low	$R_L = 2 k\Omega$ to $V^+/2$			0.10	0.15	V
		$R_L = 150\Omega \text{ to } V^+/2$			0.21	0.55	
I _{SC}	Output Short Circuit Current (6)	Sourcing to V ⁺ /2		60 30	95		A
		Sinking from V ⁺ /2		50 15	75		mA
I _{OUT}	Output Current	V _O = 0.5V from either Supply		±100		mA	
PSRR	Power Supply Rejection Ratio	15V ≤ V ⁺ ≤ 17V		72	80		dB
Is	Supply Current	No Load			4	6.5 7.8	mA
R _{IN}	Common Mode Input Resistance	$A_V = +1$, $f = 1$ kHz, $R_S = 1$ M	Ω		32		ΜΩ
C _{IN}	Common Mode Input Capacitance	$A_V = +1, R_S = 100 \text{ k}\Omega$			1.7		pF

⁽¹⁾ Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T_J = T_A. Parametric performance is indicated in the electrical tables under conditions of internal self-heating where T_J > T_A.

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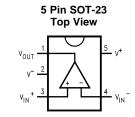


16V Electrical Characteristics (continued)

Unless otherwise specified, All limits specified for $T_J = 25^{\circ}C$, $V^+ = 16V$, $V^- = 0V$, $V_O = V_{CM} = V^+/2$ and $R_L = 2 \text{ k}\Omega$ to $V^+/2$. **Boldface** limits apply at temperature extremes.

Symbol	Parameter	Conditions	Min ⁽²⁾	Typ ⁽³⁾	Max ⁽²⁾	Units
R _{OUT}	Output Resistance Closed Loop	$R_F = 10 \text{ k}\Omega, f = 1 \text{ kHz}, A_V = -1$		0.1		Ω
		$R_F = 10 \text{ k}\Omega$, $f = 1 \text{ MHz}$, $A_V = -1$		0.3		Ω
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to V ⁺ /2		0.12		%
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.12		deg

CONNECTION DIAGRAM



See Package Number DBV0005A



Typical Performance Characteristics

At $T_J = 25^{\circ}C$, $V^+ = 16$ V, $V^- = 0$ V, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1$ k Ω for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified.

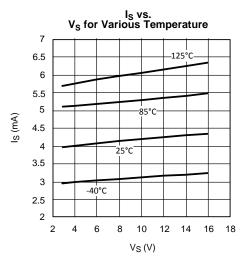


Figure 2.

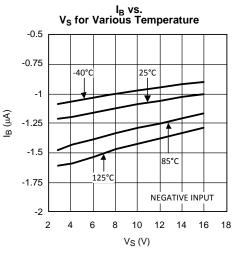
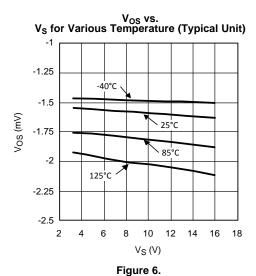


Figure 4.



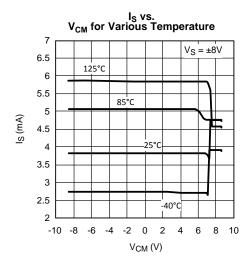


Figure 3.

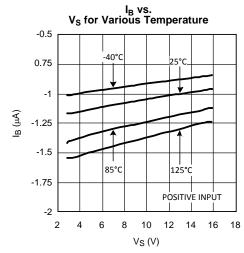


Figure 5.

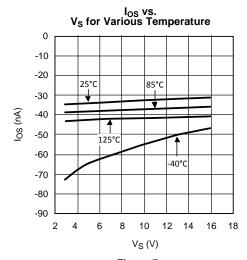


Figure 7.



At $T_J = 25^{\circ}C$, $V^+ = 16$ V, $V^- = 0$ V, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1$ k Ω for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified.

Positive Output Saturation Voltage vs. V_S for Various Temperature

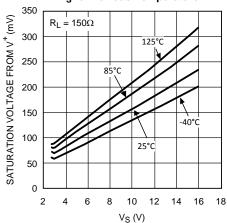


Figure 8.

Output Sinking Saturation Voltage vs. I_{SINKING} for Various Temperature

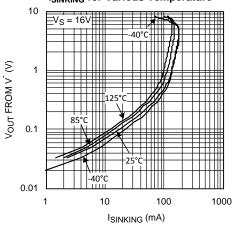
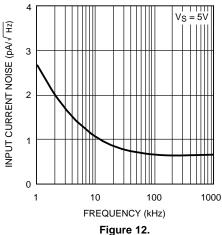


Figure 10.

Input Current Noise vs. Frequency



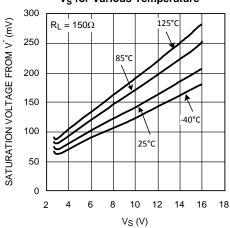


Figure 9.

Output Sourcing Saturation Voltage vs. I_{SOURCING} for Various Temperature

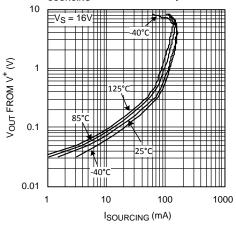


Figure 11.

Input Voltage Noise vs. Frequency

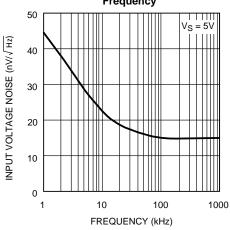


Figure 13.



At $T_J = 25^{\circ}C$, $V^+ = 16$ V, $V^- = 0$ V, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1$ k Ω for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified.

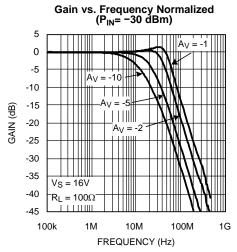
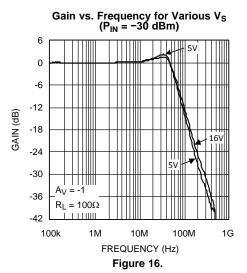
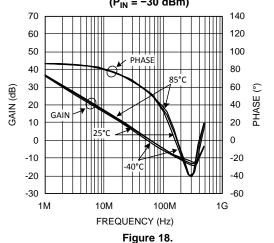


Figure 14.



Open Loop Gain & Phase vs. Frequency for Various Temperature (P_{IN} = −30 dBm)



Gain vs. Frequency Normalized (P_{IN}=-30dBm)

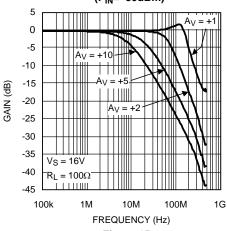
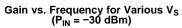


Figure 15.



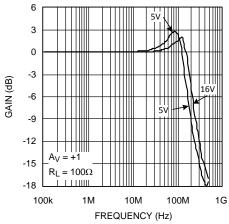


Figure 17.

Relative Gain vs. Frequency for Various Temperature ($P_{IN} = -10 \text{ dBm}$)

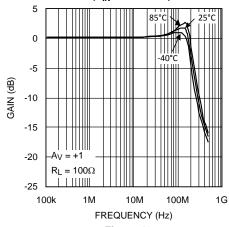


Figure 19.



At $T_J = 25^{\circ}C$, $V^+ = 16$ V, $V^- = 0$ V, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1$ k Ω for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified.

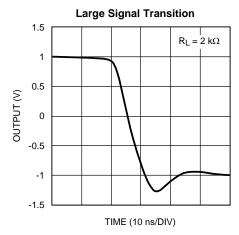


Figure 20.

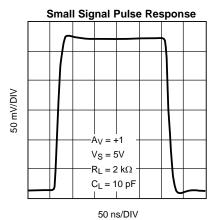


Figure 22.

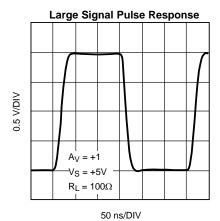


Figure 24.

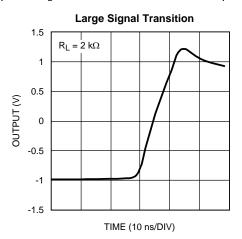


Figure 21.

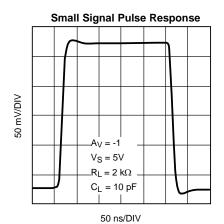


Figure 23.

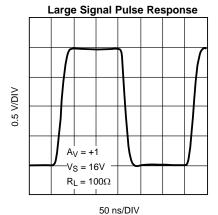


Figure 25.



At $T_J = 25^{\circ}C$, $V^+ = 16$ V, $V^- = 0$ V, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1$ k Ω for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified.

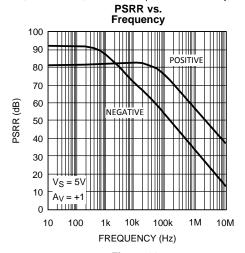


Figure 26.

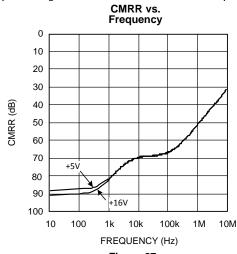


Figure 27.



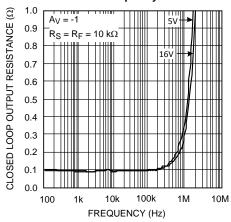


Figure 28.

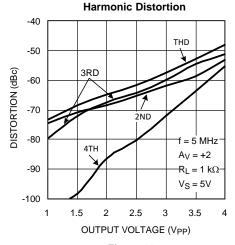
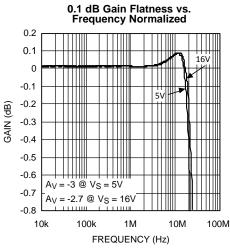


Figure 29.





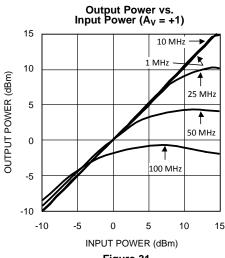


Figure 31.



At $T_J = 25^{\circ}\text{C}$, $V^+ = 16 \text{ V}$, $V^- = 0\text{V}$, $R_F = 330\Omega$ for $A_V = +2$, $R_F = 1 \text{ k}\Omega$ for $A_V = -1$. R_L tied to $V^+/2$. Unless otherwise specified. Differential Gain/Phase vs. IRE

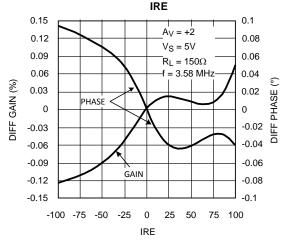


Figure 32.



APPLICATION INFORMATION

Application Notes

With its high output current and speed, one of the major applications for the LMH6640 is the V_{COM} driver in a TFT panel. This application is a specially taxing one because of the demands it places on the operational amplifier's output to drive a large amount of bi-directional current into a heavy capacitive load while operating under unity gain condition, which is a difficult challenge due to loop stability reasons. For a more detailed explanation of what a TFT panel is and what its amplifier requirements are, please see the Application Notes section of the LM6584 found on the web at: http://www.ti.com/lit/pdf/snosb08

Because of the complexity of the TFT V_{COM} waveform and the wide variation in characteristics between different TFT panels, it is difficult to decipher the results of circuit testing in an actual panel. The ability to make simplifying assumptions about the load in order to test the amplifier on the bench allows testing using standard equipment and provides familiar results which could be interpreted using standard loop analysis techniques. This is what has been done in this application note with regard to the LMH6640's performance when subjected to the conditions found in a TFT V_{COM} application.

Figure 33, shows a typical simplified V_{COM} application with the LMH6640 buffering the V_{COM} potential (which is usually around ½ of panel supply voltage) and looking into the simplified model of the load. The load represents the cumulative effect of all stray capacitances between the V_{COM} node and both row and column lines. Associated with the capacitances shown, is the distributed resistance of the lines to each individual transistor switch. The other end of this R-C ladder is driven by the column driver in an actual panel and here is driven with a low impedance MOSFET driver (labeled "High Current Driver") for the purposes of this bench test to simulate the effect that the column driver exerts on the V_{COM} load.

The modeled TFT V_{COM} load, shown in Figure 33, is based on the following simplifying assumptions in order to allow for easy bench testing and yet allow good matching results obtained in the actual application:

- The sum of all the capacitors and resistors in the R-C ladder is the total V_{COM} capacitance and resistance respectively. This total varies from panel to panel; capacitance could range from 50 nF-200 nF and the resistance could be anywhere from $20\Omega-100\Omega$.
- The number of ladder sections has been reduced to a number (4 sections in this case) which can easily be
 put together in the lab and which behaves reasonably close to the actual load.

In this example, the LMH6640 was tested under the simulated conditions of total 209 nF capacitance and 54Ω as shown in Figure 33.

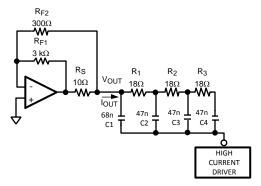
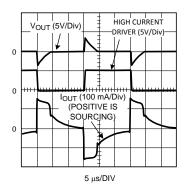


Figure 33. LMH6640 in a V_{COM} Buffer Application with Simulated TFT Load

 R_S is sometimes used in the panel to provide additional isolation from the load while R_{F2} provides a more direct feedback from the V_{COM} . R_{F1} , R_{F2} , and R_S are trimmed in the actual circuit with settling time and stability trade-offs considered and evaluated. When tested under simulated load conditions of Figure 33, here are the resultant voltage and current waveforms at the LMH6640 output:





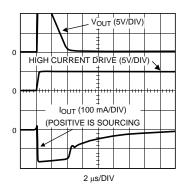


Figure 34. V_{COM} Output, High Current Drive Waveform, & LMH6640 Output Current Waveforms

Figure 35. Expanded View of Figure 34 Waveforms showing LMH6640 Current Sinking ½ Cycle

As can be seen, the LMH6640 is capable of supplying up to 160 mA of output current and can settle the output in $4.4 \mu s$.

The LMH6640 is a cost effective amplifier for use in the TFT V_{COM} application and is made even more attractive by its large supply voltage range and high output current. The combination of all these features is not readily available in the market, especially in the space saving SOT-23 5 pin package. All this performance is achieved at the low power consumption of 65 mW which is of utmost importance in today's battery driven TFT panels.





REVISION HISTORY

Cł	Changes from Revision A (March 2013) to Revision B					
•	Changed layout of National Data Sheet to TI format		14			

www.ti.com 11-Nov-2025

PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
LMH6640MF/NOPB	Obsolete	Production	SOT-23 (DBV) 5	-	-	Call TI	Call TI	-	AH1A

⁽¹⁾ Status: For more details on status, see our product life cycle.

- (3) RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.
- (4) Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- (5) MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- (6) Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

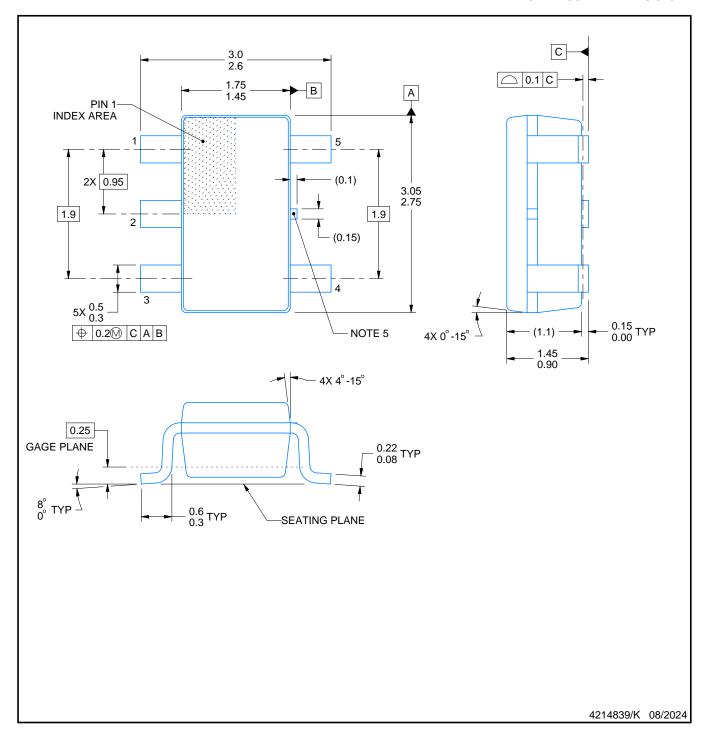
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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.



SMALL OUTLINE TRANSISTOR



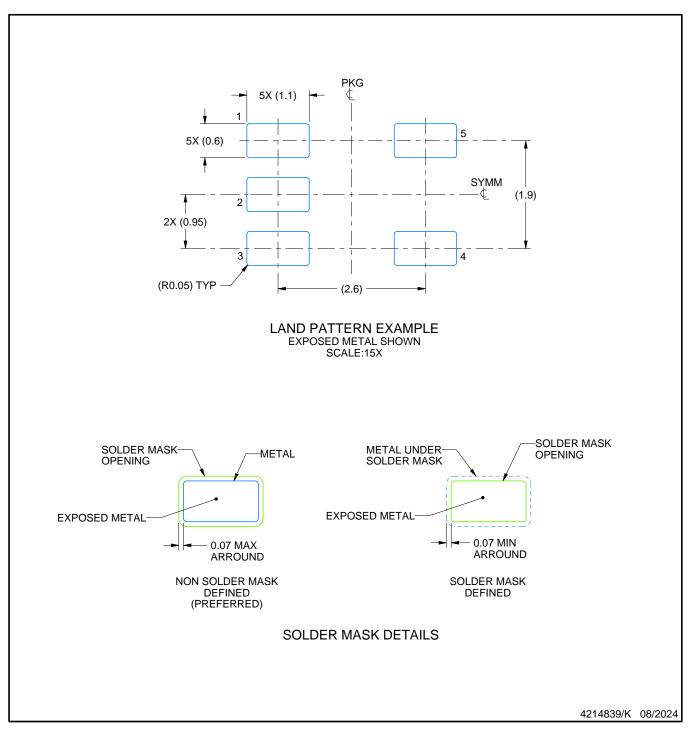
NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
 3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
- 5. Support pin may differ or may not be present.



SMALL OUTLINE TRANSISTOR



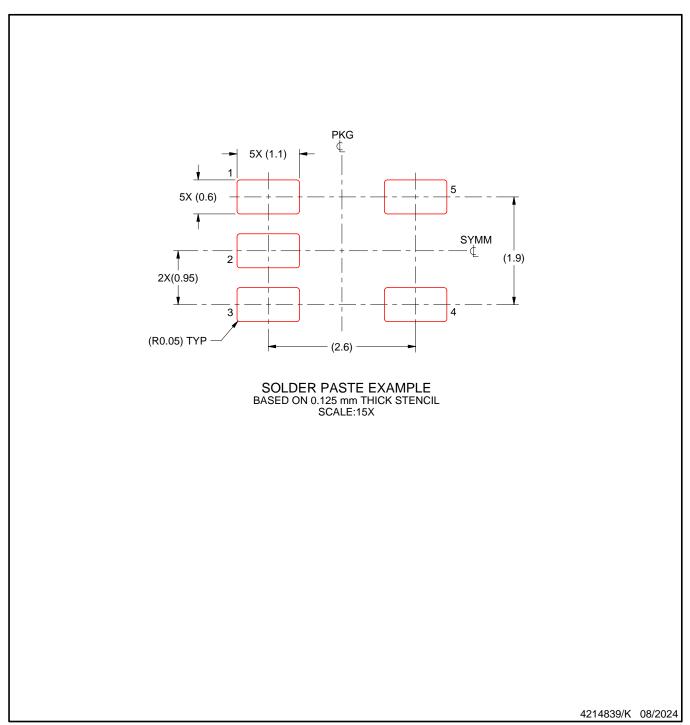
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE TRANSISTOR



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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