

SNAS438A - NOVEMBER 2008 - REVISED APRIL 2013

# LME49724 High Performance, High Fidelity, Fully-Differential Audio Operational Amplifier

Check for Samples: LME49724

## FEATURES

- Drives 600Ω Loads with Full Output Signal Swing
- Optimized for Superior Audio Signal Fidelity
- Output Short Circuit Protection
- PSRR and CMRR Exceed 100dB (typ)
- Available in SO PowerPad Package

# APPLICATIONS

- Ultra High Quality Audio Amplification
- High Fidelity Preamplifiers and Active Filters
- Simple Single-Ended to Differential Conversion
- State of the Art D-to-A Converters
- State of the Art A-to-D input Amplifiers
- Professional Audio
- High Fidelity Equalization and Crossover Networks
- High Performance Line Drivers and Receivers

# DESCRIPTION

The LME49724 is an ultra-low distortion, low noise, high slew rate fully-differential operational amplifier optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state of the art circuit design, the LME49724 fully-differential audio operational amplifier delivers superior audio signal amplification for outstanding audio performance. The LME49724 combines extremely low voltage noise density (2.1nV/vHz) with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LME49724 has a high slew rate of ±18V/µs and an output current capability of ±80mA. Further, dynamic range is maximized by an output stage that drives 600 $\Omega$  loads to 52V<sub>P-P</sub> while operating on a ±15V supply voltage.

The LME49724's outstanding CMRR (102dB), PSRR (125dB), and  $V_{OS}$  (0.2mV) results in excellent operational amplifier DC performance.

The LME49724 has a wide supply range of  $\pm 2.5V$  to  $\pm 18V$ . Over this supply range the LME49724's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LME49724 is unity gain stable. This Fully-Differential Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with capacitive values as high as 100pF.

#### Table 1. Key Specifications

Power Supply Voltage Range		±2.5V to ±18V
THD+N (A <sub>V</sub> = 1, V <sub>OUT</sub> = $3V_{RMS}$ , f <sub>IN</sub> = 1kHz)	$R_L = 2k\Omega$	0.00003% (typ)
	$R_L = 600\Omega$	0.00003% (typ)
Input Noise Density		2.1nV/√Hz (typ)
Slew Rate		±18V/µs (typ)
Gain Bandwidth Product		50 MHz (typ)
Open Loop Gain ( $R_L = 600\Omega$ )		125 dB (typ)
Input Bias Current		60nA (typ)
Input Offset Voltage		0.2mV (typ)
DC Gain Linearity Error		0.00009%

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## **Typical Application**

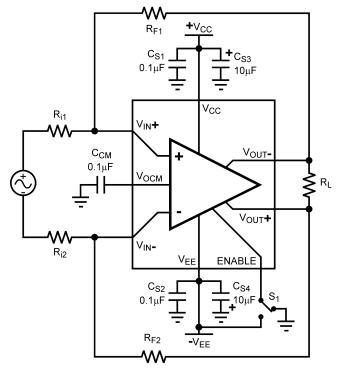
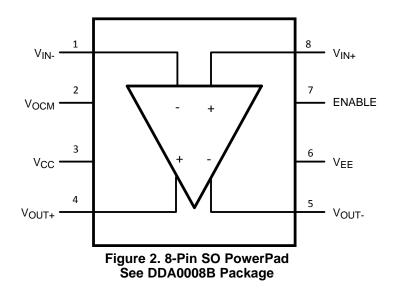


Figure 1. Typical Application Circuit

## **Connection Diagram**



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Pin	n Name Pin Function			
1	V <sub>IN-</sub>	Input pin	Analog Input	
2	V <sub>OCM</sub>	Sets the output DC voltage. Internally set by a resistor divider to the midpoint of the voltages on the V <sub>CC</sub> and V <sub>EE</sub> pins. Can be forced externally to a different voltage ( $50k\Omega$ input impedance).	Analog Input	
3	V <sub>CC</sub>	Positive power supply pin.	Power Supply	
4	V <sub>OUT+</sub>	Output pin. Signal is inverted relative to $V_{IN}$ where the feedback loop is connected.	Analog Output	
5	V <sub>OUT-</sub>	Output pin. Signal is inverted relative to $V_{IN+}$ where the feedback loop is connected.	Analog Output	
6	V <sub>EE</sub>	Negative power supply pin or ground for a single supply configuration.	Power Supply	
7	ENABLE	Enables the LME49724 when the voltage is greater than 2.35V above the voltage on the V <sub>EE</sub> pin. Disable the LME49724 by connecting to the same voltage as on the V <sub>EE</sub> pin which will reduce current consumption to less than 0.3mA (typ).	Analog Input	
8	V <sub>IN+</sub>	Input pin	Analog Input	
xposed Pad		Exposed pad for improved thermal performance. Connect to the same potential as the $V_{\text{EE}}$ pin or electrically isolate.		

PIN DESCRIPTIONS



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### Absolute Maximum Ratings (1)(2)(3)

Power Supply Voltage	$(V_{S} = V_{CC} +  V_{EE} )$	38V	
Storage Temperature		−65°C to 150°C	
Input Voltage		$(V_{EE}) - 0.7V$ to $(V_{CC}) + 0.7V$	
Output Short Circuit	Continuous		
Power Dissipation <sup>(4)</sup>		Internally Limited	
ESD Rating <sup>(5)</sup>	D Rating <sup>(5)</sup>		
ESD Rating <sup>(6)</sup>		200V	
Junction Temperature (T <sub>JMAX</sub> )		150°C	
Soldering Information	Vapor Phase (60sec.)	215°C	
	Infrared (60sec.)	220°C	
Thermal Resistance	θ <sub>JA</sub> (MR)	49.6°C/W	

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified.

(2) The Electrical Characteristics tables list specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not ensured.

(3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.

(4) The maximum power dissipation must be derated at elevated temperatures and is dictated by  $T_{JMAX}$ ,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . The maximum allowable power dissipation is  $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$  or the number given in *Absolute Maximum Ratings*, whichever is lower.

(5) Human body model, applicable std. JESD22-A114C.

(6) Machine model, applicable std. JESD22-A115-A.

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## Operating Ratings (1)(2)

Temperature Range	$T_{MIN} \le T_A \le T_{MAX}$	-40°C ≤ T <sub>A</sub> ≤ +85°C
Supply Voltage Range		$\pm 2.5V \le V_S \le \pm 18V$

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified.

(2) The Electrical Characteristics tables list specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not ensured.

### Electrical Characteristics (1)(2)

The following specifications apply for  $V_S = \pm 15V$ ,  $R_L = 2k\Omega$ ,  $f_{IN} = 1kHz$ , and  $T_A = 25^{\circ}C$ , unless otherwise specified.

0	Bananatan	O and litilaria	LME4	LME49724		
Symbol	Parameter	Conditions	Typical <sup>(3)</sup>	Limit <sup>(4)</sup>	(Limits)	
POWER SU	PPLY					
Vs	Operating Power Supply			±2.5V ±18V	V (min) V (max)	
I <sub>CCQ</sub>	Total Quiescent Current	$V_O = 0V, I_O = 0mA$ Enable = GND Enable = $V_{EE}$	10 0.3	15 0.5	mA (max) mA (max)	
PSRR	Power Supply Rejection Ratio	$V_{\rm S} = \pm 5 V$ to $\pm 15 V$ <sup>(5)</sup>	125	95	dB (min)	
V <sub>ENIH</sub>	Enable High Input Voltage	Device active, $T_A = 25^{\circ}C^{(6)}$	V <sub>EE</sub> + 2.35		V	
V <sub>ENIL</sub>	Enable Low Input Voltage	Device disabled, $T_A = 25^{\circ}C^{(6)}$	V <sub>EE</sub> + 1.75		V	
DYNAMIC P	ERFORMANCE					
THD+N	Total Harmonic Distortion + Noise		0.00003 0.00003	0.00009	% % (max)	
IMD	Intermodulation Distortion	$A_V = 1$ , $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1	0.0005		%	
GBWP	Gain Bandwidth Product		50	35	MHz (min)	
FPBW	Full Power Bandwidth	$V_{OUT} = 1V_{P-P}$ , -3dB referenced to output magnitude at f = 1kHz	13		MHz	
SR	Sew Rate	$R_L = 2k\Omega$	±18	±13	V/µs (min)	
t <sub>S</sub>	Settling time	$A_V = -1$ , 10V step, $C_L = 100 pF$ settling time to 0.1%			μs	
		$-10V < V_{OUT} < 10V, R_{L} = 600\Omega$	125	100	dB (min)	
A <sub>VOL</sub>	Open-Loop Voltage Gain	$-10V < V_{OUT} < 10V, R_L = 2k\Omega$	125		dB	
		$-10V < V_{OUT} < 10V, R_{L} = 10k\Omega$	125		dB	

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified.

(2) The Electrical Characteristics tables list specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not ensured.

(3) Typical values represent most likely parametric norms at T<sub>A</sub> = +25°C, and at the *Recommended Operation Conditions* at the time of product characterization and are not ensured.

(4) Datasheet min/max specification limits are specified by test or statistical analysis.

(5) PSRR is measured as follows:  $V_{OS}$  is measured at two supply voltages, ±5V and ±15V. PSRR = |  $20\log(\Delta V_{OS}/\Delta V_S)$  |.

(6) The ENABLE threshold voltage is determined by V<sub>BE</sub> voltages and will therefore vary with temperature. The typical values represent the most likely parametric norms at T<sub>A</sub> = +25°C.



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# Electrical Characteristics <sup>(1)(2)</sup> (continued)

The following specifications apply for  $V_S = \pm 15V$ ,  $R_L = 2k\Omega$ ,  $f_{IN} = 1kHz$ , and  $T_A = 25^{\circ}C$ , unless otherwise specified.

Symbol	Beremeter	Conditions	LME	LME49724			
Symbol	Parameter	Conditions	Typical <sup>(3)</sup>	Limit <sup>(4)</sup>	(Limits)		
NOISE	•	•		+	*		
_	Equivalent Input Noise Voltage	$f_{BW} = 20$ Hz to 20kHz	0.30	0.64	μV <sub>RMS</sub> (max)		
e <sub>N</sub>	Equivalent Input Noise Density	f = 1kHz f = 10Hz	2.1 3.7		nV <b>/</b> √Hz (max)		
INPUT CHAR	ACTERISTICS				·		
V <sub>OS</sub>	Offset Voltage		±0.2	±1	mV (max)		
$\Delta V_{OS} / \Delta Temp$	Average Input Offset Voltage Drift vs Temperature	–40°C ≤ T <sub>A</sub> ≤ 85°C	0.5		µV/°C		
I <sub>B</sub>	Input Bias Current	$V_{CM} = 0V$	60	200	nA (max)		
I <sub>OS</sub>	Input Offset Current	$V_{CM} = 0V$	10	65	nA (max)		
ΔI <sub>OS</sub> /ΔTemp	Input Bias Current Drift vs Temperature	–40°C ≤ T <sub>A</sub> ≤ 85°C	0.1		nA/°C		
V <sub>IN-CM</sub>	Common-Mode Input Voltage Range		±14	V <sub>CC</sub> – 1.5 V <sub>EE</sub> + 1.5	V (min) V (min)		
CMRR	Common-Mode Rejection	$-10V < V_{CM} < 10V$	102	95	dB (min)		
7	Differential Input Impedance		16		kΩ		
Z <sub>IN</sub>	Common-Mode Input Impedance	$-10V < V_{CM} < 10V$	500		MΩ		
OUTPUT CHA	RACTERISTICS				÷		
		$R_L = 600\Omega$	52	50	V <sub>P-P</sub> (min)		
V <sub>OUTMAX</sub>	Maximum Output Voltage Swing	$R_L = 2k\Omega$	52		V <sub>P-P</sub>		
		$R_L = 10k\Omega$	53		V <sub>P-P</sub>		
I <sub>OUT-CC</sub>	Instantaneous Short Circuit Current		80		mA		
R <sub>OUT</sub>	Output Impedance	f <sub>IN</sub> = 10kHz Closed-Loop Open-Loop	0.01 23		ΩΩ		
C <sub>LOAD</sub>	Capacitive Load Drive Overshoot	C <sub>L</sub> = 100pF	5		%		

# LME49724

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0.01

0.003

0.001

(%) Z 0.0003

0.00003

0.00001

0.000005

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· 문 0.0001

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(%) N +

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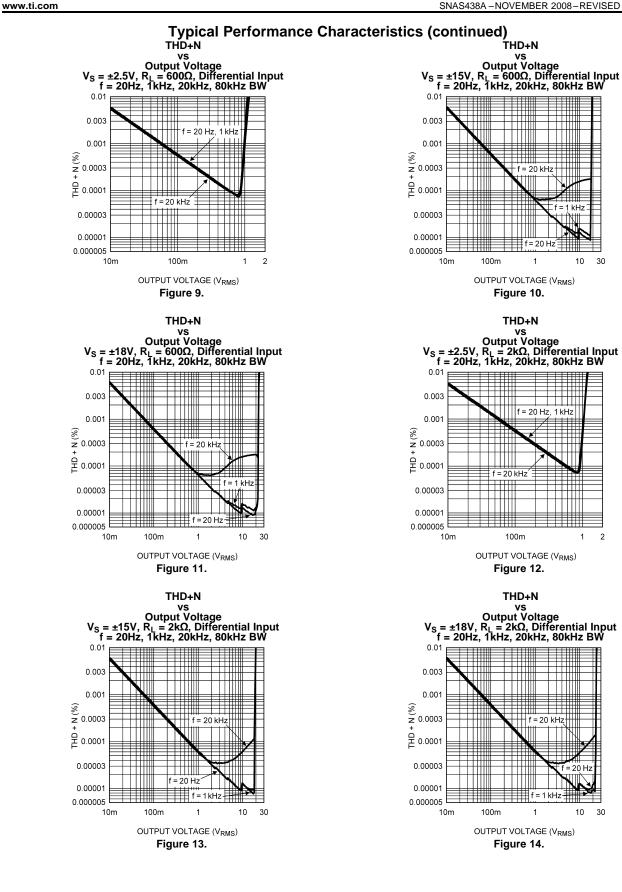
**Typical Performance Characteristics** THD+N THD+N vs vs  $Frequency \\ V_S = \pm 2.5V, V_O = 0.5V_{RMS}, Differential Input \\ R_L = 600\Omega, 2k\Omega, 10k\Omega, 80kHz BW$ Frequency =  $\pm 2.5V$ , V<sub>O</sub> =  $0.8V_{RMS}$ , Differential Input R<sub>L</sub> =  $600\Omega$ , 2k $\Omega$ , 10k $\Omega$ , 80kHz BW ٧s 0.01 0.003 0.001 (%) Z 0.0003 = 600Ω, 2 kΩ, 10 kΩ R = 600Ω · H 0.0001 0.00003 10 kΩ = 2 kΩ 0.00001 0.000005 10 100 1k 10k 20k 10 100 10k 20k 1k FREQUENCY (Hz) FREQUENCY (Hz) Figure 3. Figure 4. THD+N THD+N vs vs Frequency V<sub>S</sub> =  $\pm 15V$ , V<sub>O</sub> =  $3V_{RMS}$ , Differential Input R<sub>L</sub> =  $600\Omega$ ,  $2k\Omega$ ,  $10k\Omega$ , 80kHz BW Frequency  $V_S = \pm 15V, V_O = 10V_{RMS}, Differential Input$   $R_L = 600\Omega, 2k\Omega, 10k\Omega, 80kHz BW$ 0.01 0.003 0.001 ⊗ Z 0.0003 = 600Ω RL = 6000 R · H 0.0001 ''Rı = 2 kΩ Rı = 2 kΩ R 10 0.00003 Rı  $= 10 \, \mathrm{k}\Omega$ 0.00001 0.000005 10 100 1k 10k 20k 10 100 10k 20k 1k FREQUENCY (Hz) FREQUENCY (Hz) Figure 5. Figure 6. THD+N THD+N vs VS Frequency V<sub>S</sub> = ±18V, V<sub>O</sub> = 3V<sub>RMS</sub>, Differential Input R<sub>L</sub> = 600Ω, 2kΩ, 10kΩ, 80kHz BW  $V_{S} = \pm 18V, V_{O} = 10V_{RMS}, Differential Input R_{L} = 600\Omega, 2k\Omega, 10k\Omega, 80kHz BW$ 0.01 0.003 0.001 (%) N + 0.0003 = 600C R ٠Rı = 600Ω 문 0.0001 111) R  $= 2 k\Omega$ R = 10 k 0.00003 = 10 kΩ 0.00001 Rı 11111 0.000005 10k 20k 10 100 1k 10 100 1k 10k 20k

> FREQUENCY (Hz) Figure 7.

FREQUENCY (Hz) Figure 8.

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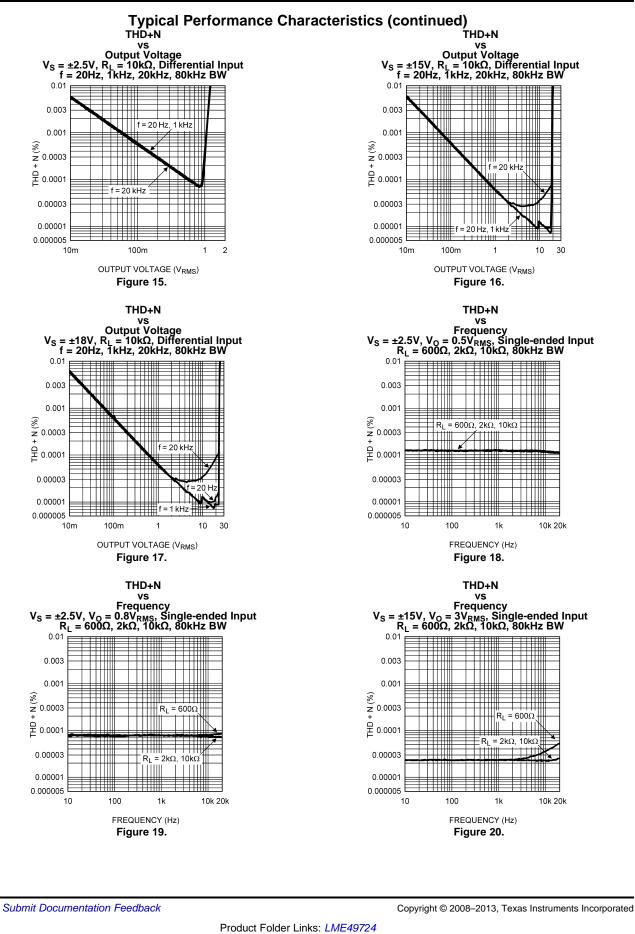




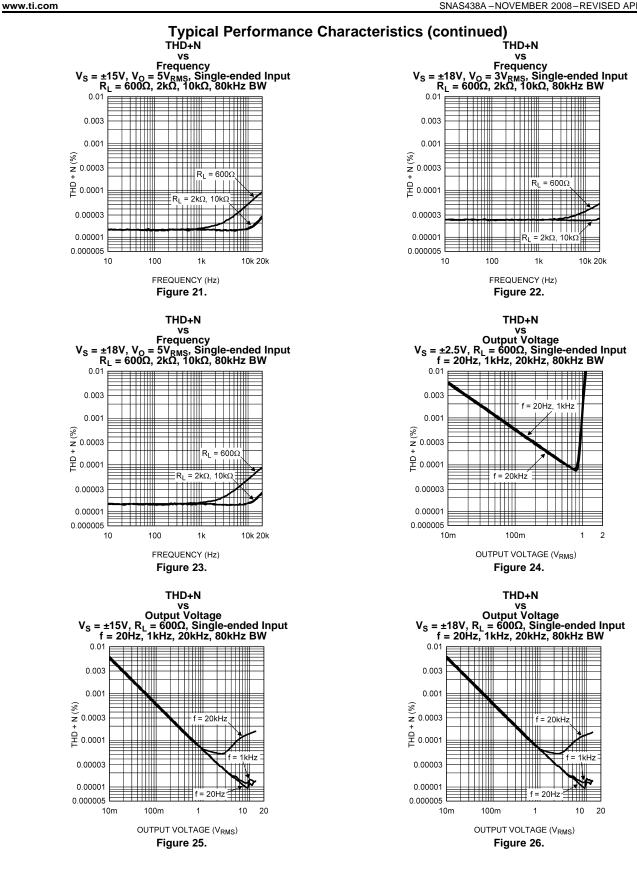
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0.01 0.003

0.001

0.0003

日 0.0001

0.00003

0.00001

0.000005

10m

V<sub>S</sub> = ±18V,

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0.001 (%) N + 0.0003

· E 0.0001

0.00003 0.00001

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0.01

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0.001

(%) N + 0.0003

문 0.0001

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0.00001

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10

10m

10m

100m

1

20

f = 20Hz, 1kHz

Figure 31.

1

100m

Figure 29.

THD+N

f = 20Hz,

+ N (%)

THD+N

f = 20 kHz

100m

Figure 27.

THD+N

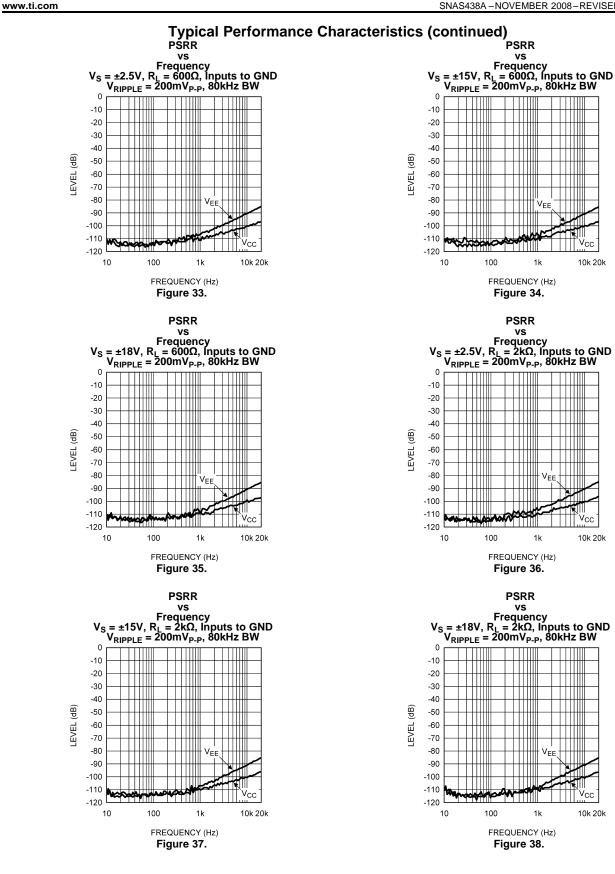
**Typical Performance Characteristics (continued)** THD+N  $v_{S}$ Output Voltage  $V_{S} = \pm 2.5V, R_{L} = 2k\Omega, Single-ended Input$  f = 20Hz, 1kHz, 20kHz, 80kHz BWvs Output Voltage V<sub>S</sub> = ±15V, R<sub>L</sub> = 2k $\Omega$ , Single-ended Input f = 20Hz, 1kHz, 20kHz, 80kHz BW 0.01 0.003 f = 20Hz 1kHz0.00 + N (%) 0.0003 = 20kH 日 0.0001 0.00003 0.00001 20Hz, 1kHz 0.000005 1 2 10m 100m 10 20 1 OUTPUT VOLTAGE (V<sub>RMS</sub>) OUTPUT VOLTAGE (V<sub>RMS</sub>) Figure 28. THD+N vs Output Voltage R<sub>L</sub> = 2kΩ, Single-ended Input Iz, 1kHz, 20kHz, 80kHz BW vs Output Voltage ±2.5V,  $R_L = 10k\Omega$ , Single-ended Input ۷s 1kHz, 20kHz, 80kHz BW f = 20Hz,0.01 1111 0.003 20Hz, 1kHz 0.001 (%) Z 0.0003 20kH · 문 0.0001 f = 20kHz 11111 0.00003 0.00001 = 20Hz, 1kHz 0.000005 10m 10 20 100m 1 2 OUTPUT VOLTAGE (V<sub>RMS</sub>) OUTPUT VOLTAGE (V<sub>RMS</sub>) Figure 30. THD+N vs Output Voltage V<sub>S</sub> = ±15V, R<sub>L</sub> = 10kΩ, Single-ended Input f = 20Hz, 1kHz, 20kHz, 80kHz BW vs Output Voltage V<sub>S</sub> = ±18V, R<sub>L</sub> = 10kΩ, Single-ended Input f = 20Hz, 1kHz, 20kHz, 80kHz BW 0.0 0.003 0.001 (⊗) Z 0.0003 + f = 20 kHz日 0.0001 0.00003 0.00001 f = 20Hz, 1kHz 0.000005 10 20 10m 100m 10 20 1 OUTPUT VOLTAGE (V<sub>RMS</sub>) OUTPUT VOLTAGE (V<sub>RMS</sub>) Figure 32.

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0

-10 -20

-30

-40

-50

-60

-70

-80

-90

-100

-110

-120

0

-10

-20 -30

-40

-50

-60

-70 -80

-90 -100

-110

-120

0

-20

-40

-60

-80

-100

-120

-140 10 = 600Ω.

100

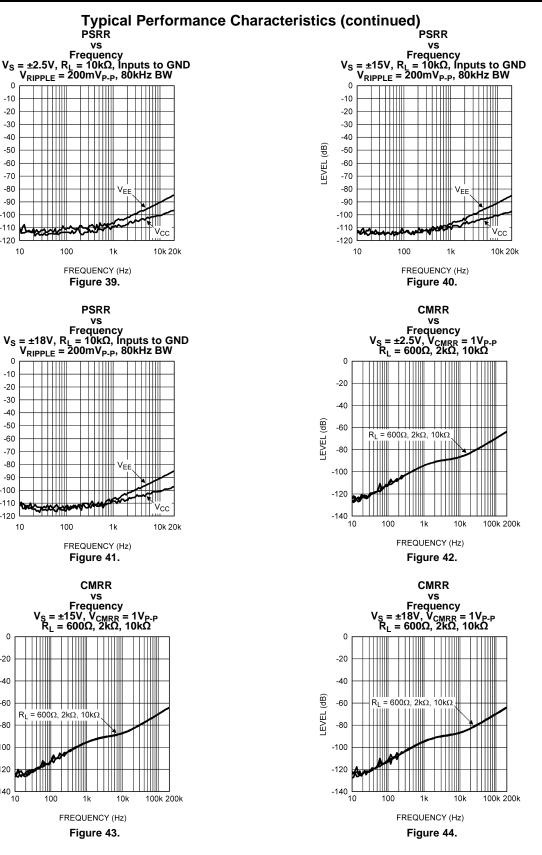
LEVEL (dB)

10

LEVEL (dB)

10

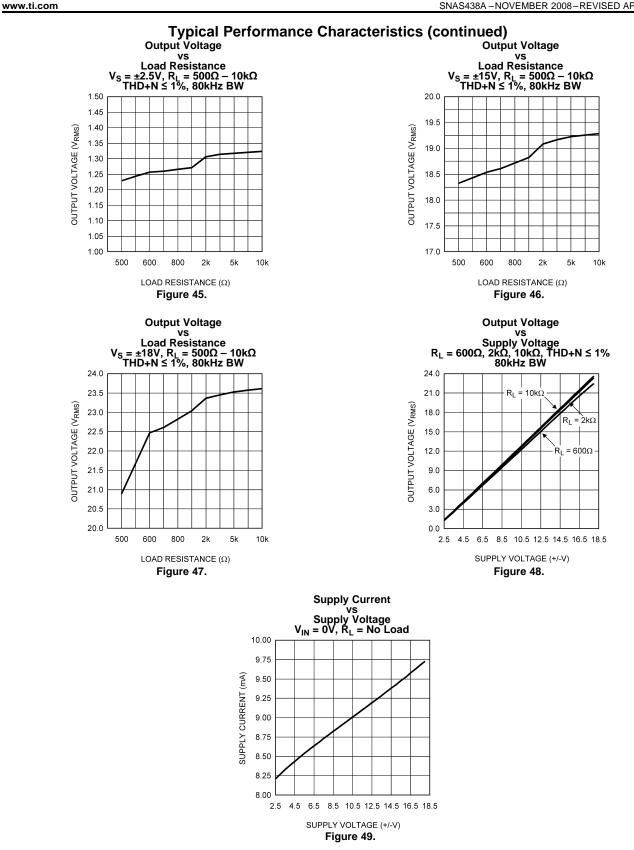
LEVEL (dB)



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## APPLICATION INFORMATION

#### **GENERAL OPERATION**

The LME49724 is a fully differential amplifier with an integrated common-mode reference input (V<sub>OCM</sub>). Fully differential amplification provides increased noise immunity, high dynamic range, and reduced harmonic distortion products.

Differential amplifiers typically have high CMRR providing improved immunity from noise. When input, output, and supply line trace pairs are routed together, noise pick up is common and easily rejected by the LME49724. CMRR performance is directly proportional to the tolerance and matching of the gain configuring resistors. With 0.1% tolerance resistors the worst case CMRR performance will be about 60dB (20LOG(0.001)).

A differential output has a higher dynamic range than a single-ended output because of the doubling of output voltage. The dynamic range is increased by 6dB as a result of the outputs being equal in magnitude but opposite in phase. As an example, a single-ended output with a  $1V_{PP}$  signal will be two  $1V_{PP}$  signals with a differential output. The increase is 20LOG(2) = 6dB. Differential amplifiers are ideal for low voltage applications because of the increase in signal amplitude relative to a single-ended amplifier and the resulting improvement in SNR.

Differential amplifiers can also have reduced even order harmonics, all conditions equal, when compared to a single-ended amplifier. The differential output causes even harmonics to cancel between the two inverted outputs leaving only the odd harmonics. In practice even harmonics do not cancel completely, however there still is a reduction in total harmonic distortion.

### OUTPUT COMMON-MODE VOLTAGE (V<sub>OCM</sub> pin)

The output common-mode voltage is the DC voltage on each output. The output common-mode voltage is set by the V<sub>OCM</sub> pin. The V<sub>OCM</sub> pin can be driven by a low impedance source. If no voltage is applied to the V<sub>OCM</sub> pin, the DC common-mode output voltage will be set by the internal resistor divider to the midpoint of the voltages on the V<sub>CC</sub> and V<sub>EE</sub> pins. The input impedance of the V<sub>OCM</sub> pin is 50k $\Omega$ . The V<sub>OCM</sub> pin can be driven up to V<sub>CC</sub> - 1.5V and V<sub>EE</sub> + 1.5V. The V<sub>OCM</sub> pin should be bypassed to ground with a 0.1µF to 1µF capacitor. The V<sub>OCM</sub> pin should be connected to ground when the desired output common-mode voltage is ground reference. The value of the external capacitor has an effect on the PSRR performance of the LME49724. With the V<sub>OCM</sub> pin only bypassed with a low value capacitor, the PSRR performance of the LME49724 will be reduced, especially at low audio frequencies. For best PSRR performance, the V<sub>OCM</sub> pin should be connected to stable, clean reference. Increasing the value of the bypass capacitor on the V<sub>OCM</sub> pin will also improve PSRR performance.

### ENABLE FUNCTION

The LME49724 can be placed into standby mode to reduce system current consumption by driving the ENABLE pin below  $V_{EE}$  + 1.75V. The LME49724 is active when the voltage on the ENABLE pin is above  $V_{EE}$  + 2.35V. The ENABLE pin should not be left floating. For best performance under all conditions, drive the ENABLE pin to the  $V_{EE}$  pin voltage to enter standby mode and to ground for active operation when operating from split supplies. When operating from a single supply, drive the ENABLE pin to ground for standby mode and to  $V_{CC}$  for active mode.



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#### FULLY DIFFERENTIAL OPERATION

The LME49724 performs best in a fully differential configuration. The circuit shown in Figure 50 is the typical fully differential configuration.

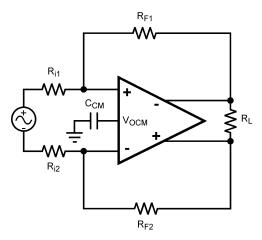


Figure 50. Fully Differential Configuration

The closed-loop gain is shown in Equation 1 below.

$$A_{V} = R_{F} / R_{i} \quad (V/V)$$

where

• R<sub>i1</sub> = R<sub>i2</sub>

• Using low value resistors will give the lowest noise performance

SINGLE-ENDED TO DIFFERENTIAL CONVERSION

For many applications, it is required to convert a single-ended signal to a differential signal. The LME49724 can be used for a high performance, simple single-to-differential converter. Figure 51 shows the typical single-to-differential converter circuit configuration.

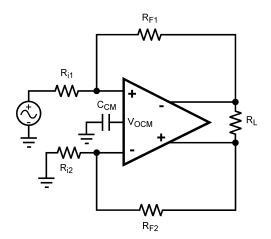


Figure 51. Single-Ended Input to Differential Output

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(1)



# SINGLE SUPPLY OPERATION

The LME49724 can be operated from a single power supply, as shown in Figure 52. The supply voltage range is limited to a minimum of 5V and a maximum of 36V. The common-mode output DC voltage will be set to the midpoint of the supply voltage. The  $V_{OCM}$  pin can be used to adjust the common-mode output DC voltage on the outputs, as described previously, if the supply voltage midpoint is not the desired DC voltage.

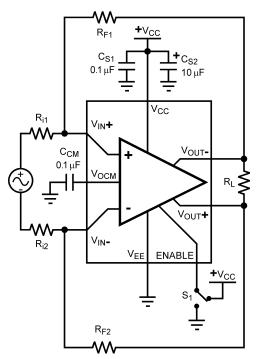


Figure 52. Single Supply Configuration

## DRIVING A CAPACITIVE LOAD

The LME49724 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

### THERMAL PCB DESIGN

The LME49724's high operating supply voltage along with its high output current capability can result in significant power dissipation. For this reason the LME49724 is provided in the exposed DAP SO PowerPad package for improved thermal dissipation performance compared to other surface mount packages. The exposed pad is designed to be soldered to a copper plane on the PCB which then acts as a heat sink. The thermal plane can be on any layer by using multiple thermal vias under and outside the IC package. The vias under the IC should have solder mask openings for the entire pad under the IC on the top layer but cover the vias on the bottom layer. This method prevents solder from being pulled away from the thermal vias during the reflow process resulting in optimum thermal conductivity.

Heat radiation from the PCB plane area is best accomplished when the thermal plane is on the top or bottom copper layers. The LME49724 should always be soldered down to a copper pad on the PCB for both optimum thermal performance as well as mechanical stability.

The exposed pad is for heat transfer and the thermal plane should either be electrically isolated or connected to the same potential as the  $V_{EE}$  pin. For high frequency applications (f > 1MHz) or lower impedance loads, the pad should be connected to a plane that is connected to the  $V_{EE}$  potential.



#### SUPPLY BYPASSING

The LME49724 should have its supply leads bypassed with low-inductance capacitors such as leadless surface mount (SMT) capacitors located as close as possible to the supply pins. It is recommended that a 10µF tantalum or electrolytic capacitor be placed in parallel with a 0.1µF ceramic or film type capacitor on each supply pin. These capacitors should be star routed with a dedicated ground return plane or large trace for best THD performance. Placing capacitors too far from the power supply pins, especially with thin connecting traces, can lead to excessive inductance, resulting in degraded high-frequency bypassing. Poor high-frequency bypassing can result in circuit instabilities. When using high bandwidth power supplies, the value and number of supply bypass capacitors should be reduced for optimal power supply performance.

#### **BALANCE CABLE DRIVER**

With high peak-to-peak differential output voltage and plenty of low distortion drive current, the LME49724 makes an excellent balanced cable driver. Combining the single-to-differential configuration with a balanced cable driver results in a high performance single-ended input to balanced line driver solution.

Although the LME49724 can drive capacitive loads up to 100pF, cable loads exceeding 100pF can cause instability. For such applications, series resistors are needed on the outputs before the capacitive load.

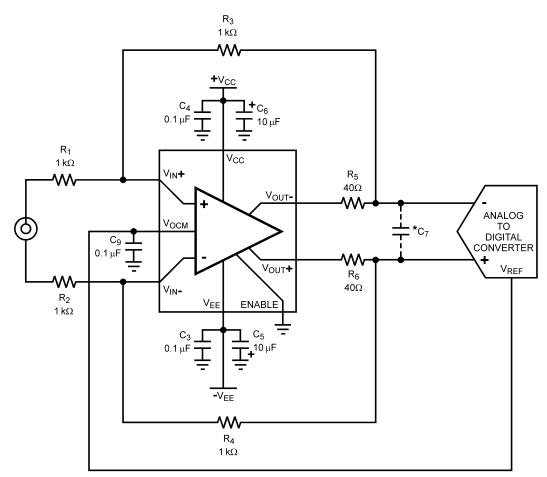
### ANALOG-TO-DIGITAL CONVERTER (ADC) APPLICATION

Figure 53 is a typical fully differential application circuit for driving an analog-to-digital converter (ADC). The additional components of  $R_5$ ,  $R_6$ , and  $C_7$  are optional components and are for stability and proper ADC sampling. ADC's commonly use switched capacitor circuitry at the input. When the ADC samples the signal the current momentarily increases and may disturb the signal integrity at the sample point causing a signal glitch. Component  $C_7$  is significantly larger than the input capacitance of a typical ADC and acts as a charge reservoir greatly reducing the effect of the signal sample by the ADC. Resistors  $R_5$  and  $R_6$  decouple the capacitive load,  $C_7$ , for stability. The values shown are general values. Specific values should be optimized for the particular ADC loading requirements.

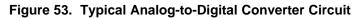
The output reference voltage from the ADC can be used to drive the V<sub>OCM</sub> pin to set the common-mode DC voltage on the outputs of the LME49724. A buffer may be needed to drive the LME49724's V<sub>OCM</sub> pin if the ADC cannot drive the 50k $\Omega$  input impedance of the V<sub>OCM</sub> pin.

In order to minimize circuit distortion when using capacitors in the signal path, the capacitors should be comprised of either NPO ceramic, polystyrene, polypropylene or mica composition. Other types of capacitors may provide a reduced distortion performance but for a cost improvement, so capacitor selection is dependent upon design requirements. The performance/cost tradeoff for a specific application is left up to the user.





\* Value is application and converted dependent.



### DISTORTION MEASUREMENTS

The vanishing low residual distortion produced by the LME49724 is below the capabilities of commercially available equipment. This makes distortion measurements more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49724's low residual distortion is an input referred internal error. As shown in Figure 54, adding a resistor connected between the amplifier's inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is increased. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced, which means that measurement resolution increases. To ensure minimum effects on distortion measurements, keep the value of  $R_5$  low. The distortion reading on the audio analyzer must be divided by a factor of  $(R_3 + R_4)/R_5$ , where  $R_1 = R_2$  and  $R_3 = R_4$ , to get the actual measured distortion of the device under test. The values used for the LME49724 measurements were  $R_1$ ,  $R_2$ ,  $R_3$ ,  $R_4 = 1k\Omega$  and  $R_5 = 20\Omega$ .

This technique is verified by duplicating the measurements with high closed-loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities.



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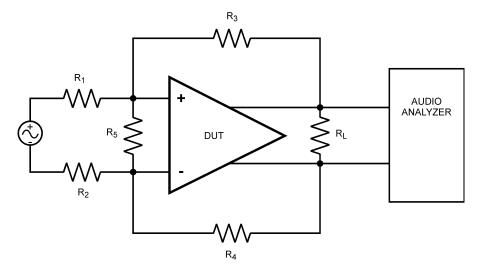
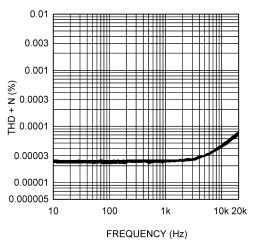
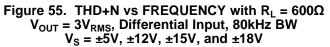


Figure 54. THD+N and IMD Distortion Test Circuit

#### PERFORMANCE VARIATIONS

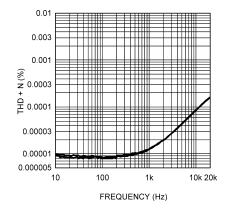
The LME49724 has excellent performance with little variation across different supply voltages, load impedances, and input configuration (single-ended or differential). Inspection of the THD+N vs Frequency and THD+N vs Output Voltage performance graphs (See Typical Performance Characteristics reveals only minimal differences with different load values. Figure 55 and Figure 56 below show the performance across different supply voltages with the same output signal level and load. Figure 55 has plots at ±5V, ±12V, ±15V, and ±18V with a  $3V_{RMS}$  output while Figure 56 has plots at ±12V, ±15V, and ±18V with a  $10V_{RMS}$  output. Both figures use a  $600\Omega$  load. The performance for each different supply voltage under the same conditions is so similar it is nearly impossible to discern the different plots lines.

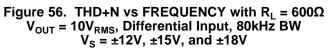




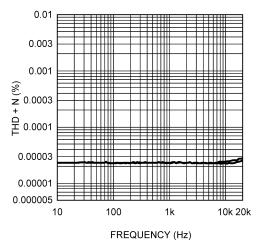


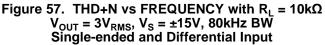
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Whether the input configuration is single-ended or differential has only a minimal affect on THD+N performance at higher audio frequencies or higher signal levels. For easy comparison, Figure 57 and Figure 58 are a combination of the performance graphs found in Typical Performance Characteristics.

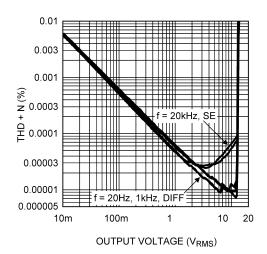






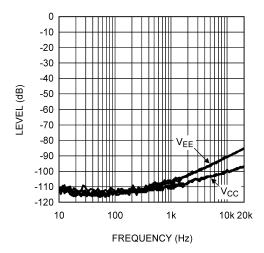


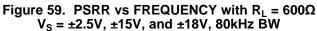
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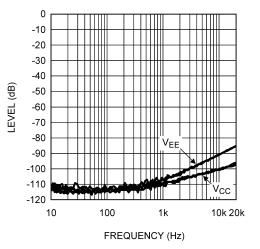


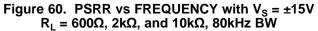
Power Supply Rejection Ratio does not vary with load value nor supply voltage. For easy comparison, Figure 59 and Figure 60 below are created by combining performance graphs found in Typical Performance Characteristics.





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Although supply current may not be a critical specification for many applications, there is also no real variation in supply current with no load or with a  $600\Omega$  load. This is a result of the extremely low offset voltage, typically less than 1mV. Figure 61 shows the supply current under the two conditions with no real difference discernable.

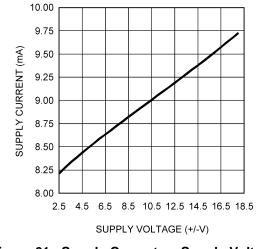
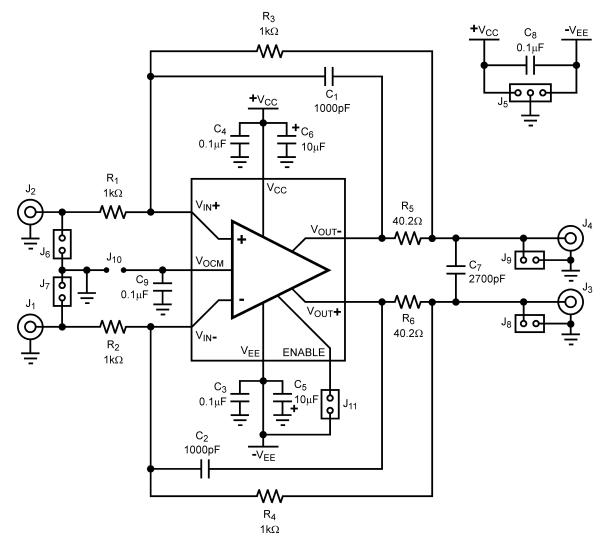


Figure 61. Supply Current vs Supply Voltage  $R_L$  = No Load and 600 $\Omega$ 



# Demo Board Schematic

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#### **Build of Materials**

Table 2. Reference	Demo Board	Bill of Materials
--------------------	------------	-------------------

Designator Value Toler		or Value Tolerance Part Description		
R <sub>1</sub> , R <sub>2</sub> , R <sub>3</sub> , R <sub>4</sub>	1kΩ	1%	1/8W, 0603 Resistor	
R <sub>5</sub> , R <sub>6</sub>	40.2Ω	1%	1/8W, 0603 Resistor	
C <sub>1</sub> , C <sub>2</sub>	1000pF	10%	0603, NPO Ceramic Capacitor, 50V	
C <sub>3</sub> , C <sub>4</sub> , C <sub>8</sub> , C <sub>9</sub>	0.1µF	-20%, +80%	0603, Y5V Ceramic Capacitor, 25V	
C <sub>5</sub> , C <sub>6</sub>	10µF	20%	Size C (6032), Tantalum Capacitor, 25V	
C <sub>7</sub>	2700pF	10%	0805, NPO Ceramic Capacitor, 50V	
U <sub>1</sub>			LME49724MR	
J <sub>1</sub> , J <sub>2</sub> , J <sub>3</sub> , J <sub>4</sub>			SMA coaxial connector	Inputs & Outputs
$J_5$			0.100" 1x3 header, vertical mount	V <sub>DD</sub> , V <sub>EE</sub> , GND
J <sub>6</sub> , J <sub>7</sub> , J <sub>8</sub> , J <sub>9</sub> , J <sub>10</sub> , J <sub>11</sub>			0.100" 1x2 header, vertical mount	Inputs, Outputs, V <sub>OCM</sub> , Enable



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## **REVISION HISTORY**

Rev	Date	Description
1.0	11/12/08	Initial release.
А	04/04/13	Changed layout of National Data Sheet to TI format.



30-Jun-2016

# PACKAGING INFORMATION

Orderable Device	Status	Package Type	•	Pins	-	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
LME49724MR/NOPB	ACTIVE	SO PowerPAD	DDA	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	L49724 MR	Samples
LME49724MRX/NOPB	ACTIVE	SO PowerPAD	DDA	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	L49724 MR	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(<sup>6)</sup> Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE OPTION ADDENDUM

30-Jun-2016

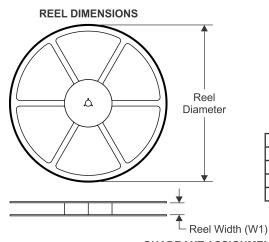
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# PACKAGE MATERIALS INFORMATION

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Texas Instruments

## TAPE AND REEL INFORMATION



LME49724MRX/NOPB



B0

(mm)

5.4

6.5

**P1** 

(mm)

8.0

K0

(mm)

2.0

w

(mm)

12.0

Pin1

Quadrant

Q1

## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



330.0

12.4

*All dimensions are nominal					
Device	Package Drawing		Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)

8

2500

DDA

SO

Power PAD TEXAS INSTRUMENTS

www.ti.com

# PACKAGE MATERIALS INFORMATION

8-Apr-2013



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LME49724MRX/NOPB	SO PowerPAD	DDA	8	2500	367.0	367.0	35.0

# **GENERIC PACKAGE VIEW**

# **DDA 8**

# PowerPAD<sup>™</sup> SOIC - 1.7 mm max height PLASTIC SMALL OUTLINE



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.





#### **PACKAGING INFORMATION**

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking
	(1)	(2)			(3)	(4)	(5)		(6)
LME49724MR/NOPB	Active	Production	SO PowerPAD (DDA)   8	95   TUBE	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR
LME49724MR/NOPB.A	Active	Production	SO PowerPAD (DDA)   8	95   TUBE	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR
LME49724MR/NOPB.B	Active	Production	SO PowerPAD (DDA)   8	95   TUBE	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR
LME49724MRX/NOPB	Active	Production	SO PowerPAD (DDA)   8	2500   LARGE T&R	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR
LME49724MRX/NOPB.A	Active	Production	SO PowerPAD (DDA)   8	2500   LARGE T&R	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR
LME49724MRX/NOPB.B	Active	Production	SO PowerPAD (DDA)   8	2500   LARGE T&R	Yes	SN	Level-3-260C-168 HR	-40 to 85	L49724 MR

<sup>(1)</sup> **Status:** For more details on status, see our product life cycle.

<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.



# PACKAGE OPTION ADDENDUM

23-May-2025

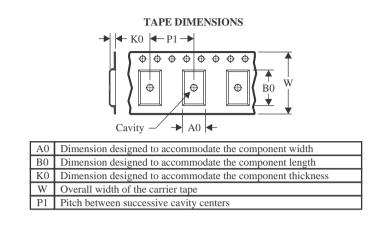
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## TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions a	re nominal
-------------------	------------

Device	-	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LME49724MRX/NOPB	SO PowerPAD	DDA	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1



# PACKAGE MATERIALS INFORMATION

23-May-2025



\*All dimensions are nominal

Device	Device Package Type		Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LME49724MRX/NOPB	SO PowerPAD	DDA	8	2500	356.0	356.0	36.0

## TEXAS INSTRUMENTS

www.ti.com

23-May-2025

## TUBE



# - B - Alignment groove width

#### \*All dimensions are nominal

Device	Device Package Name		Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
LME49724MR/NOPB	DDA	HSOIC	8	95	495	8	4064	3.05
LME49724MR/NOPB.A	DDA	HSOIC	8	95	495	8	4064	3.05
LME49724MR/NOPB.B	DDA	HSOIC	8	95	495	8	4064	3.05

# **GENERIC PACKAGE VIEW**

# **DDA 8**

# PowerPAD<sup>™</sup> SOIC - 1.7 mm max height PLASTIC SMALL OUTLINE



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



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