

# DS90UB638-Q1 MIPI CSI-2 出力搭載、FPD-Link III 4.16Gbps シングル入力デシリアライザ

## 1 特長

- 車載アプリケーション用に AEC-Q100 認定済み:
  - デバイス温度グレード 2: -40°C ~ +105°C の動作時周囲温度範囲
- ISO 10605 および IEC 61000-4-2 ESD に準拠
- Power-over-Coax (PoC) 対応トランシーバ
- MIPI DPHY バージョン 1.2 / CSI-2 バージョン 1.3 準拠:
  - CSI-2 のデータ速度は、データ・レーンごとに 400Mbps / 800Mbps / 1.5Gbps / 1.6Gbps にスケール可能
  - 1、2、3、4 データ・レーンをサポート
  - 最大 4 つの仮想チャンネルをサポート
  - データ型をプログラミング可能
  - CRC データ保護、センサ・データ整合性チェック、I2C 書き込み保護、電圧および温度の遠隔測定、プログラム可能なアラーム、BIST、パターン生成、ライン・フォルト検出を含む高度なデータ保護および診断機能
- 非常に低いデータおよび制御パスのレイテンシ
- シングルエンド同軸ケーブルをサポート
- シングル入力デシリアライザ
- 適応型受信イコライゼーション
- I2C で最大 1Mbps のファスト・モード・プラスに対応
- 柔軟な GPIO によるカメラ同期および診断
- 偽造防止向けの一意的ダイ ID
- シングルエンドの同軸またはシールド付きツイストペア (STP) ケーブルに対応
- DS90UB635-Q1 および DS90UB633A-Q1 シリアルライザと互換

## 2 アプリケーション

- 先進運転支援システム (ADAS)
  - 自動運転車 (AV)
  - 車載用衛星レーダーおよび LIDAR モジュール
  - カメラ監視システム (CMS)
  - 前方視野カメラ (FC)
  - タイム・オブ・フライト (ToF) および LIDAR センサ・モジュール
  - サラウンド・ビュー・システム (SVS)
  - リアビュー・カメラ (RVC)
  - ドライバー監視システム (DMS)
  - サイド・ミラー・ディスプレイ (SMD)
- セキュリティと監視
- 産業用および医療用イメージング

## 3 概要

DS90UB638-Q1 は、シリアル化されたセンサ・データを FPD-Link III インターフェイスを介してソースから受信できる多用途デシリアライザです。DS90UB635-Q1 シリアルライザと組み合わせる場合、DS90UB638-Q1 はイメージャからデータを受信し、2MP/60fps および 4MP/30fps のカメラをサポートします。さらに衛星レーダーや、ToF や LIDAR などの他のセンサもサポートします。受信したデータは、MIPI CSI-2 準拠で出力され、ダウンストリーム・プロセッサに相互接続できます。2 レーン動作用に CSI-2 インターフェイスを構成すれば、二重の MIPI CSI-2 クロック・レーンを利用して出力を複製できます。複製モードでは、受信したビデオ・ストリームの 2 つのコピーが作成され、データの記録や並列処理に使用できます。

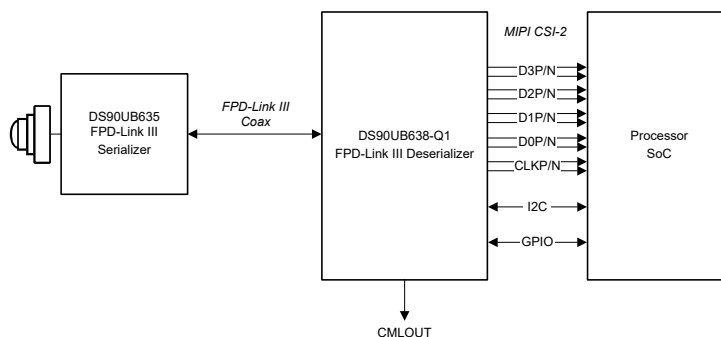
DS90UB638-Q1 は AEC-Q100 認定済みで、50Ω のシングルエンド同軸ケーブルでデータを受信するように設計されています。このデシリアライザは Power-over-Coax アプリケーションに最適であり、受信イコライザは追加プログラミング不要で、ケーブルの損失特性に応じて自動的に補償を行い、経時劣化にも対応します。

各 FPD-Link III インターフェイスには、独立した低レイテンシの双方向制御チャンネル (BCC) が含まれ、I2C、GPIO、その他の制御情報を連続的に伝送します。センサの同期および診断機能に必要な GPIO 信号も、この BCC を使用します。

### 製品情報

部品番号	パッケージ <sup>(1)</sup>	本体サイズ (公称)
DS90UB638-Q1	VQFN (48)	7.00mm × 7.00mm

(1) 利用可能なパッケージについては、このデータシートの末尾にある注文情報を参照してください。



代表的なアプリケーションの回路図



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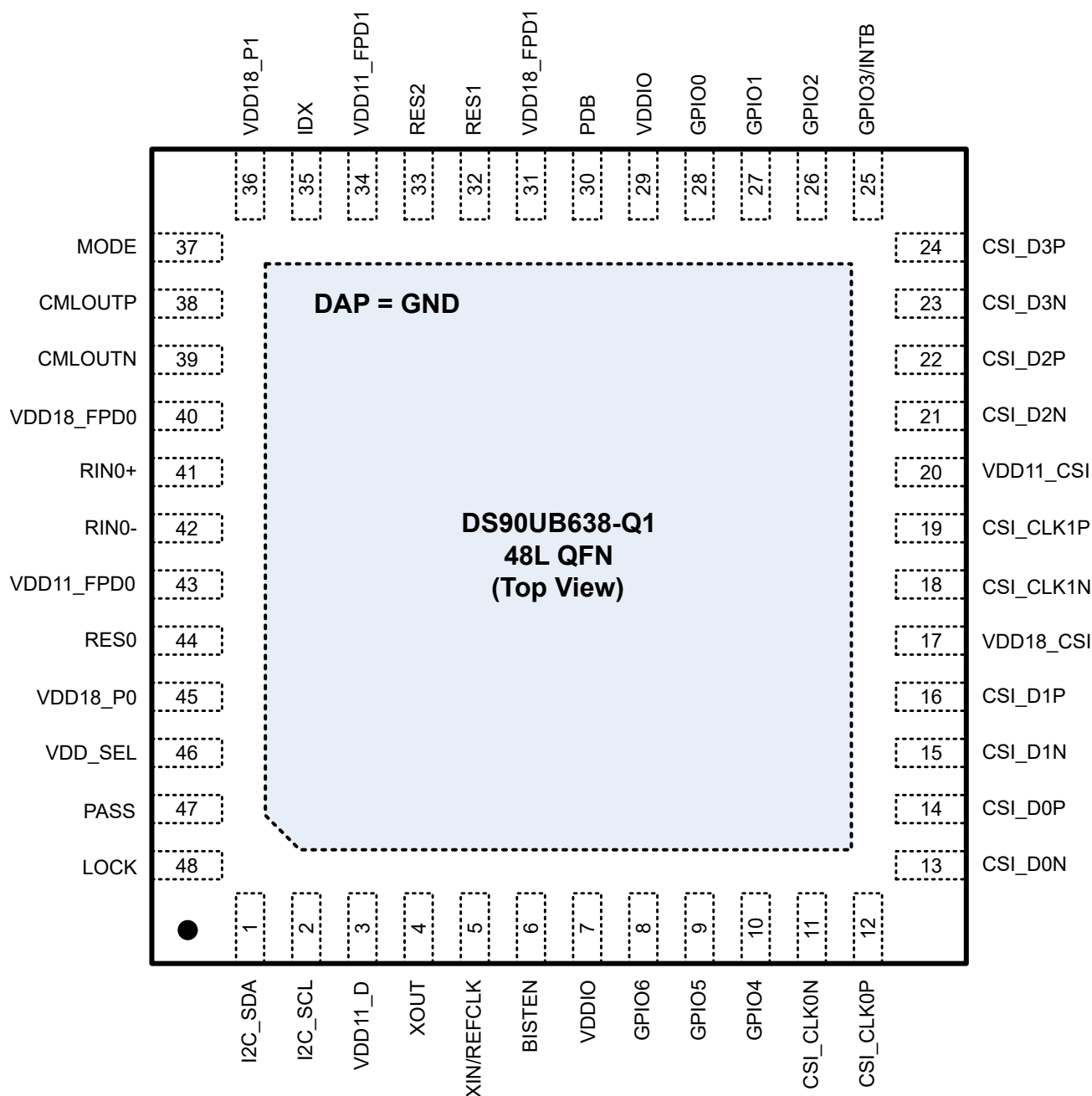
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## 4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

DATE	REVISION	NOTES
February 2023	*	Initial Release

## 5 Pin Configuration and Functions



**5-1. RGZ Package  
48-Pin VQFN  
Top View**

表 5-1. Pin Functions

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
RECEIVE DATA CSI-2 OUTPUT			
CSI_CLK0N	11	O	RECEIVE DATA OUTPUT: This signal carries data from the FPD-LINK III Deserializer to the processor over CSI-2 interface. Receive data is CSI-2 configured with DPHY outputs as one differential clock lane (CSI_CLK0P/N) and up to four differential data lanes (CSI_D0P/N: CSI_D3P/N) or two clock lanes (CSI_CLK0P/N, CSI_CLK1P/N) and two differential data lanes for each clock. When in replicate mode data lanes CSI_D2P/N and CSI_D3P/N are associated with clock lane CSI_CLK1P/N to provide the replicated output. Leave unused outputs as no connect.
CSI_CLK0P	12		
CSI_CLK1N	18		
CSI_CLK1P	19		
CSI_D0N	13		
CSI_D0P	14		
CSI_D1N	15		
CSI_D1P	16		
CSI_D2N	21		
CSI_D2P	22		
CSI_D3N	23		
CSI_D3P	24		
CLOCK INTERFACE			
XIN/REFCLK	5	S, I	Reference clock input or crystal oscillator input. Pin is shared with XIN and REFCLK. Typically REFCLK connected to 23-MHz to 26-MHz reference oscillator output (100 ppm) or XIN configured with external 23-MHz to 26-MHz crystal to XOUT. See <a href="#">セクション 7.4.4</a> .
XOUT	4	O	Crystal oscillator output: Output Pin for providing crystal oscillator reference. Leave this pin NC when reference clock input is driving XIN/REFCLK.
SYNCHRONIZATION AND GPIO			
GPIO0	28	I/O, PD	General-Purpose Input/Output: Pins can be used to control and respond to various commands. They may be configured to be the input signals for the corresponding GPIOs on the serializer, or they may be configured to be outputs to follow local register settings. At power up, the GPIO are disabled and include a 35-kΩ (typical) pulldown resistor by default. See <a href="#">セクション 7.4.12</a> for programmability. Unused GPIOs can be left open or no connect.
GPIO1	27		
GPIO2	26		
GPIO4	10		
GPIO5	9		
GPIO6	8		
GPIO3/INTB	25	I/O, OD	General-Purpose Input/Output: Pin GPIO3 can be configured as input signals for GPOs on the serializer. Pin 25 is shared with INTB. Pul lup with 4.7 kΩ to V <sub>(VDDIO)</sub> . The programmable input and output pin is an active-low open drain and controlled by the status registers. See <a href="#">セクション 7.4.12</a> for programmability. Unused GPIO can be left open or no connect.
FPD-LINK III INTERFACE			
RIN0–	42	I/O	Receive Input Channel 0: Differential FPD-Link receiver and bidirectional control back channel output. The IO must be AC-coupled. For applications that use single-ended, coaxial channel, connect RIN0+ with 33nF, AC-coupling capacitor and terminate RIN0– to GND with a 15nF capacitor and 50-Ω resistor. For STP applications, connect both RIN0+ and RIN0– with 33nF, AC-coupling capacitor. If connecting to a DS90UB633A-Q1, please follow the capacitor values suggested in <i>Design Requirements</i> .
RIN0+	41		
I2C PINS			
I2C_SCL	2	I/O, OD	I2C Serial Clock: Clock line for the bidirectional control bus communication. An external 2-kΩ to 4.7-kΩ pullup resistor to 1.8-V or 3.3-V supply rail is recommended per I2C interface standards. I2C_SCL and I2C_SDA inputs are 3.3-V tolerant. See <a href="#">セクション 7.5.1</a> for more information.
I2C_SDA	1	I/O, OD	I2C Serial Data: Data line for bidirectional control bus communication. An external 2-kΩ to 4.7-kΩ pullup resistor to 1.8-V or 3.3-V supply rail is recommended per I2C interface standards. I2C_SCL and I2C_SDA inputs are 3.3-V tolerant. See <a href="#">セクション 7.5.1</a> for more information.
CONFIGURATION AND CONTROL PINS1.2			
IDX	35	S, PD	Input. I2C Serial Control Bus Primary Device ID Address Select. Once enabled, the voltage at this pin will be sampled to configure the default I2C device address. This pin is typically connected with external pullup resistor to VDD18 and pulldown resistor to GND to create a voltage divider. See <i>Serial Control Bus Addresses for IDX</i> .
MODE	37	S, PD	Mode select configuration input to set operating mode based on input voltage level. This pin is typically connected to voltage divider through an external pullup to VDD18 and pulldown to GND. <a href="#">表 7-2</a> .

**表 5-1. Pin Functions (continued)**

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
PDB	30	I, PD	Power-down inverted Input Pin. This pin is typically connected to processor GPIO with a pulldown resistor. When PDB input is brought HIGH, the device is enabled and internal register and state machines are reset to default values. Asserting PDB signal low will power down the device and consume minimum power. The default function of this pin is PDB = LOW; POWER DOWN with internal 50-k $\Omega$ internal pulldown enabled. PDB should remain low until after power supplies are applied and reach minimum required levels. <b>PDB INPUT IS 3.3-V TOLERANT.</b> See <a href="#">セクション 9.2</a> for more information. PDB = 1.8V, device is enabled (normal operation). PDB = 0 V, device is powered down.
VDD_SEL	46	S, PD	VDD Select: Configuration pin to select internal LDO regulator supply. When VDD_SEL = LOW, internal 1.1V supply mode is selected. Feed 1.8V to VDD18 inputs = 1.8V $\pm$ 5%. An internal 1.1V regulator will supply the VDD11. VDD11 inputs should be terminated with bypass capacitors. When VDD_SEL = HIGH, external 1.1V supply mode is selected. After 1.8V supply is applied to VDD18 inputs, then apply 1.1V to VDD11 inputs = 1.1 V $\pm$ 5%. <b>Voltage at VDD11 supply pins must always be less than main voltage applied to VDD18 when using external 1.1V supply. VDD_SEL IS 3.3V TOLERANT.</b>
<b>DIAGNOSTIC PINS</b>			
BISTEN	6	S, PD	BIST Enable: BISTEN = H, BIST Mode is enabled BISTEN = L, BIST Mode is disabled. If unused, connect BISTEN directly to GND. See <a href="#">セクション 7.5.10</a> for more information.
CMLOUTP	38	O	Monitor Loop-Through Driver differential output which supports functional checks. This pin is typically routed to test points and not connected. For monitoring, CMLOUT should be terminated with 100- $\Omega$ differential load. See <a href="#">セクション 7.4.9</a> for more information.
CMLOUTN	39		
LOCK	48	O	LOCK Status: Output Pin for monitoring lock status of FPD-Link III channel that may be used for Link Status. When LOCK = H, the FPD-Link III receiver is locked and the Rx Port is active. When LOCK = L, the receiver is unlocked. See <a href="#">セクション 7.4.7</a> for more information. Leave pin as no connect if unused.
PASS	47	O	PASS Output: PASS = H indicates pass conditions are met, and PASS = L indicates that pass conditions are not met. This pin is typically routed to the processor input pin or test point for monitoring. See <a href="#">セクション 7.4.7</a> for more information. For BIST operation, PASS = H indicates that ERROR FREE Transmission is in forward channel operation. PASS = L in BIST operation indicates that one or more errors were detected in the received payload. See <a href="#">セクション 7.5.10</a> for more information. Leave pin as no connect if unused.

表 5-1. Pin Functions (continued)

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
POWER AND GROUND			
GND	DAP	G	DAP is the large metal contact at the bottom side, located at the center of the QFN package. Connect to the ground plane (GND).
VDD11_CSI	20	D, P	When VDD_SEL = LOW: <b>Do not connect to 1.1-V power rail</b> Requires 0.1 to 0.01-μF and minimum 4.7-μF capacitors to GND When VDD_SEL = HIGH: <b>Connect to external 1.1-V power rail</b> Requires a 0.01-μF to GND
VDD11_D	3	D, P	When VDD_SEL = LOW: <b>Do not connect to 1.1-V power rail</b> Requires 0.1 to 0.01-μF and minimum 4.7-μF capacitors to GND When VDD_SEL = HIGH: <b>Connect to external 1.1-V power rail</b> Requires a 0.01-μF to GND
VDD11_FPD0	43	D, P	When VDD_SEL = LOW: <b>Do not connect to 1.1-V power rail</b> Requires 0.1 to 0.01-μF and minimum 4.7-μF capacitors to GND When VDD_SEL = HIGH: <b>Connect to external 1.1-V power rail</b> Requires a 0.01-μF to GND Requires a 10-μF to GND shared with VDD11_FPD1 See sections セクション 9and セクション 8.2 for more information
VDD11_FPD1	34	D, P	When VDD_SEL = LOW: <b>Do not connect to 1.1-V power rail</b> Requires 0.1 to 0.01-μF and minimum 4.7-μF capacitors to GND When VDD_SEL = HIGH: <b>Connect to external 1.1-V power rail</b> Requires a 0.01-μF to GND Requires a 10-μF to GND shared with VDD11_FPD0
VDD18_CSI	17	P	1.8-V (±5%) Power Supply. See セクション 8.2 for decoupling capacitor requirements.
VDD18_P0 VDD18_P1	45 36	P	1.8-V (±5%) Power Supplies. See セクション 8.2 for decoupling capacitor requirements.
VDD18_FPD0 VDD18_FPD1	40 31	P	1.8-V (±5%) Analog Power Supplies. See セクション 8.2 for decoupling capacitor requirements.
VDDIO	7, 29	P	VDDIO voltage supply input. The single-ended outputs and control input are powered from VDDIO. VDDIO can be connected to either a 1.8-V or 3.3-V supply rail. When VDDIO is connected to 1.8-V supply, VDDIO must be within ±100 mV of VDD18 to ensure the output timing requirements are met. See セクション 8.2 for decoupling capacitor requirements.
OTHER			
RES0	44	PD	RES0 must be tied to GND for normal operation.
RES1	32,	—	These pins should be left floating.
RES2	33	—	These pins should be left floating.

The definitions below define the functionality of the I/O cells for each pin. TYPE:

- I = Input
- O = Output
- I/O = Input/Output
- S = Configuration pin (All strap pins have internal pulldowns. If the default strap value is needed to be changed then use an external resistor.)
- PD = Internal pulldown
- OD = Open Drain
- P, G = Power supply, ground
- D = Decoupling pin for internal voltage rail

## 6 Specifications

### 6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)<sup>(1) (2)</sup>

			MIN	MAX	UNIT
Supply voltage	VDD18 (VDD18_CSI, VDD18_P1 , VDD18_P0 , VDD18_FPD0, VDD18_FPD1)		−0.3	2.16	V
	VDD11 (VDD11_CSI, VDD11_D , VDD11_FPD0, VDD11_FPD1)		−0.3	1.32 and < V <sub>(VDD18)</sub>	V
	VDDIO		−0.3	3.96	V
FPD-Link III input voltage	RIN0+, RIN0−,	Device powered up (VDD18, VDD11 and VDDIO within recommended operating conditions)	−0.3	2.75	V
		Device powered down (VDD18, VDD11 and VDDIO below recommended operating conditions) Transient Voltage	−0.3	1.45	V
		Device powered down (VDD18, VDD11 and VDDIO below recommended operating conditions) DC Voltage	−0.3	1.35	V
LVCMOS IO voltage	GPIO0, GPIO1, GPIO2, GPIO4, GPIO5, GPIO6, XIN/REFCLK, VDD_SEL, XOUT, BISTEN, LOCK, PASS, CSI_D3P/N, CSI_D2P/N, CSI_D1P/N, CSI_D0P/N, CSI_CLK1P/N, CSI_CLK0P/N		−0.3	V <sub>(VDDIO)</sub> + 0.3	V
	PDB		−0.3	3.96	V
Configuration input voltage	MODE, IDX		−0.3	V <sub>(VDD18)</sub> + 0.3	V
Open-drain voltage	GPIO3/INTB, I2C_SDA, I2C_SCL		−0.3	3.96	V
Junction temperature				150	°C
Storage temperature, T <sub>stg</sub>			−65	150	°C

- (1) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office or Distributors for availability and specifications.
- (2) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per AEC Q100-002 <sup>(1)</sup>	All pins except 32, 33, 41 and 42	±4500
			Pins 32, 33, 41 and 42	±8000
		Charged device model (CDM), per AEC Q100-011		±1250
		IEC 61000-4-2, powered-up only R <sub>D</sub> = 330 Ω, C <sub>S</sub> = 150 pF	Contact Discharge (RIN0+, RIN0−)	±8000
			Air Discharge (RIN0+, RIN0−)	±18000
		ISO 10605 R <sub>D</sub> = 330 Ω, C <sub>S</sub> = 150 pF and 330 pF R <sub>D</sub> = 2 kΩ, C <sub>S</sub> = 150 pF and 330 pF	Contact Discharge (RIN0+, RIN0−)	±8000
			Air Discharge (RIN0+, RIN0−)	±18000
				V

- (1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

## 6.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage	$V_{(VDD18)}$	1.71	1.8	1.89	V
	$V_{(VDD11)}$ (VDD_SEL = HIGH ONLY)	1.045	1.1	1.155	V
Supply voltage offset	$V_{(VDD11)} - V_{(VDDIO)}$ , $V_{(VDDIO)} = 1.8V$	-50		50	mV
LVCMOS supply voltage	$V_{(VDDIO)} = 1.8V$	1.71	1.8	1.89	V
	OR $V_{(VDDIO)} = 3.3V$	3	3.3	3.6	V
Open-drain voltage	$GPIO3/INTB = V_{(INTB)}$ , $I2C\_SDA$ , $I2C\_SCL = V_{(I2C)}$	1.71		3.6	V
Operating free-air temperature, $T_A$		-40	25	105	°C
MIPI data rate (per CSI-2 lane)		368		1664	Mbps
MIPI CSI-2 HS clock frequency		184		832	MHz
Reference clock oscillator frequency	REFCLK or XIN/XOUT	23		26	MHz
Spread-spectrum reference clock modulation percentage	Center Spread	-0.5		0.5	%
	Down Spread	-1		0	%
Local I <sup>2</sup> C frequency, $f_{I2C}$				1	MHz
Supply noise <sup>(1)</sup>	$V_{(VDD11)}$			25	mV <sub>P-P</sub>
	$V_{(VDD18)}$			50	mV <sub>P-P</sub>
	$V_{(VDDIO)} = 1.8V$			50	mV <sub>P-P</sub>
	$V_{(VDDIO)} = 3.3V$			100	mV <sub>P-P</sub>
	RIN0+		10		mV <sub>P-P</sub>

(1) DC-50 MHz

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		DS90UB638-Q1	UNIT
		RGC (VQFN)	
		48 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	30.2	°C/W
$R_{\theta JC(TOP)}$	Junction-to-case (top) thermal resistance	15.7	°C/W
$R_{\theta JC(BOT)}$	Junction-to-case (bottom) thermal resistance	1.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	6.7	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	0.2	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	6.7	°C/W

(1) Thermal data in accordance with JESD51. For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report, SPRA953.



## 6.5 DC Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
TOTAL POWER CONSUMPTION							
P <sub>T</sub>	Total power consumption for MIPI CSI-2 output mode, normal operation	FPD-Link III Input, FPD-Link III line-rate = 4.0 Gbps CSI-2 line-rate = 1.6 Gbps, CSI-2 = 4 DATA lanes + 1 CLK lane VDD_SEL = LOW, default registers	V <sub>(VDD18)</sub> = 1.89 V, V <sub>(VDDIO)</sub> = 3.6 V		473	564	mW
		FPD-Link III Input, FPD-Link III line-rate = 4.0 Gbps CSI-2 line-rate = 1.6 Gbps, CS-I2 = 4 DATA lanes + 1 CLK lane VDD_SEL = HIGH, default registers	V <sub>(VDD18)</sub> = 1.89 V, V <sub>(VDD11)</sub> = 1.155 V V <sub>(VDDIO)</sub> = 3.6 V			450	mW
DESERIALIZER SUPPLY CURRENT - FPD-Link III Rx Port0 WITH DS90UB635							
I <sub>DD-R1T4</sub>	Deserializer supply current 1 Rx 4 Tx	1 x FPD-Link III Input, FPD-Link III line-rate = 4.0 Gbps CSI-2 line-rate = 800 Mbps per lane, CSI-2 = 4 DATA lanes + 1 CLK lane VDD_SEL=LOW, default registers, includes CSI-2 load current	VDD18		170	188	mA
			VDDIO		5	10	
		1 x FPD-Link III Input, FPD-Link III line-rate = 4.0 Gbps CSI-2 line-rate = 800 Mbps per lane, CSI2 = 4 DATA lanes + 1 CLK lane VDD_SEL=HIGH, default registers, includes CSI-2 load current	VDD18		65	80	mA
			VDD11		80	100	
			VDDIO		5	10	
DESERIALIZER SUPPLY CURRENT - FPD-Link III Rx Port0 WITH DS90UB633A							
I <sub>DD2-R1T4</sub>	Deserializer supply current 2G 1 Rx 4 Tx	1 x FPD-Link III Input, FPD-Link III line-rate = 1.867 Gbps CSI-2 line-rate = 800 Mbps, CSI-2 = 4 DATA lanes + 1 CLK lane VDD_SEL=LOW, includes CSI-2 load current	VDD18		150	205	mA
			VDDIO		5	10	
		1 x FPD-Link III Input, FPD-Link III line-rate = 1.867 Gbps CSI-2 line-rate = 800 Mbps, CSI-2 = 4 DATA lanes + 1 CLK lane VDD_SEL=HIGH, includes CSI-2 load current	VDD18		65	86	mA
			VDD11		75	110	
			VDDIO		5	10	
DESERIALIZER SUPPLY CURRENT - Power Down							
I <sub>DDZ</sub>	Deserializer shutdown current	PDB = HIGH to LOW, VDD_SEL = LOW	VDD18		82	115	mA
			VDIO		2.5	5	
		PDB = HIGH to LOW, VDD_SEL = HIGH	VDD18		10	15	
			VDD11		30	110	
			VDDIO		2.5	5	
1.8-V LVCMOS I/O							
V <sub>OH</sub>	High level output voltage	I <sub>OH</sub> = −2 mA, V <sub>(VDDIO)</sub> = 1.71 to 1.89 V; V <sub>(VDDIO)</sub> = VDD18 ±50 mV	GPIO[6:4], GPIO[2:0], LOCK, PASS	V <sub>(VDDIO)</sub> − 0.45	V <sub>(VDDIO)</sub>		V
V <sub>OL</sub>	Low level output voltage	I <sub>OL</sub> = 2 mA, V <sub>(VDDIO)</sub> = 1.71 to 1.89 V; V <sub>(VDDIO)</sub> = VDD18 ±50 mV	GPIO[6:0], LOCK, PASS	GND	0.45		V
V <sub>IH</sub>	High level input voltage	V <sub>(VDDIO)</sub> = 1.71 to 1.89 V; V <sub>(VDDIO)</sub> = VDD18 ±50 mV	GPIO[6:0], BISTEN	0.65 × V <sub>(VDDIO)</sub>	V <sub>(VDDIO)</sub>		V
		V <sub>(VDDIO)</sub> = 1.71 to 1.89 V; V <sub>(VDDIO)</sub> = VDD18 ±50 mV	PDB, VDD_SEL	1.17	V <sub>(VDDIO)</sub>		V
		V <sub>(VDDIO)</sub> = 1.71 to 1.89 V; V <sub>(VDDIO)</sub> = VDD18 ±50 mV	XIN/REFCLK	1.15	V <sub>(VDDIO)</sub>		V

## 6.5 DC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS		PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
V <sub>IL</sub>	Low level input voltage	V <sub>(VDDIO)</sub> = 1.71 to 1.89V; V(VDDIO) = VDD18 ±50 mV		GPIO[6:0], BISTEN	GND		0.35 × V <sub>(VDDIO)</sub>	V
		V <sub>(VDDIO)</sub> = 1.71 to 1.89V; V(VDDIO) = VDD18 ±50 mV		PDB, VDD_SEL	GND		0.63	V
		V <sub>(VDDIO)</sub> = 1.71 to 1.89V; V(VDDIO) = VDD18 ±50 mV		XIN/REFCLK	GND		0.7	V
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = V <sub>(VDDIO)</sub> = 1.71 to 1.89 V,	Internal pulldown enabled	GPIO[6:0], PDB, BISTEN	–100		100	μA
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = V <sub>(VDDIO)</sub> = 1.71 to 1.89 V,	Internal pulldown disabled	GPIO[6:0], XIN/REFCLK, VDD_SEL	–20		30	μA
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0V		GPIO[6:0], PDB, XIN/REFCLK, VDD_SEL, BISTEN	–20		30	μA
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0 V	V <sub>OUT</sub> = 0 V			–25		mA
I <sub>OZ</sub>	TRI-STATE Output Current	V <sub>OUT</sub> = 0 V or V <sub>DDIO</sub> , PDB = L	V <sub>OUT</sub> = 0 V or V <sub>DDIO</sub> , PDB = L		–25		25	μA
3.3-V LVCMOS I/O								
V <sub>OH</sub>	High level output voltage	I <sub>OH</sub> = –4 mA, V <sub>(VDDIO)</sub> = 3.0 to 3.6 V		GPIO[6:4], GPIO[2:0], LOCK, PASS	2.4		V <sub>(VDDIO)</sub>	V
V <sub>OL</sub>	Low level output voltage	I <sub>OL</sub> = 4 mA, V <sub>(VDDIO)</sub> = 3.0 to 3.6 V		GPIO[6:0], LOCK, PASS	GND		0.4	V
V <sub>IH</sub>	High level input voltage	V <sub>(VDDIO)</sub> = 3 to 3.6 V		GPIO[6:0], XIN/REFCLK, VDD_SEL, BISTEN	2		V <sub>(VDDIO)</sub>	V
		V <sub>(VDDIO)</sub> = 3 to 3.6 V		PDB, VDD_SEL	1.17		V <sub>(VDDIO)</sub>	V
		V <sub>(VDDIO)</sub> = 3 to 3.6 V		XIN/REFCLK	1.15		V <sub>(VDDIO)</sub>	V
V <sub>IL</sub>	Low level input voltage	V <sub>(VDDIO)</sub> = 3 to 3.6 V		GPIO[6:0], XIN/REFCLK, VDD_SEL, BISTEN	GND		0.8	V
		V <sub>(VDDIO)</sub> = 3 to 3.6 V		PDB, VDD_SEL	GND		0.63	V
		V <sub>(VDDIO)</sub> = 3 to 3.6 V		XIN/REFCLK	GND		0.7	V
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = 3 to 3.6 V, internal pulldown enabled		GPIO[6:0], PDB, BISTEN	–190		190	μA
		V <sub>IN</sub> = 3 to 3.6 V, internal pulldown disabled		GPIO[6:0], XIN/REFCLK, VDD_SEL	–20		30	μA
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0 V		GPIO[6:0], PDB, XIN/REFCLK, VDD_SEL, BISTEN	–20		30	μA
I <sub>OS</sub>	Output short circuit current	V <sub>OUT</sub> = 0 V		GPIO[7:0], LOCK, PASS		–40		mA
I <sub>OZ</sub>	TRI-STATE output current	V <sub>OUT</sub> = 0 V or V <sub>(VDDIO)</sub> , PDB = L		GPIO[7:0], LOCK, PASS	–25		35	μA
SERIAL CONTROL BUS <sup>(1)</sup>								

## 6.5 DC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS		PIN OR FREQUENCY	MIN	TYP	MAX	UNIT	
V <sub>IH</sub>	Input high level			I2C_SDA, I2C_SCL	0.7 × V <sub>(I2C)</sub>		V <sub>(I2C)</sub>	V	
V <sub>IL</sub>	Input low level				GND		0.3 × V <sub>(I2C)</sub>	V	
V <sub>HY</sub>	Input hysteresis					50		mV	
V <sub>OL</sub>	Output low level	Standard-mode/Fast-mode I <sub>OL</sub> = 3 mA				0		0.4	V
		Fast-mode Plus I <sub>OL</sub> = 20 mA				0		0.4	V
I <sub>IH</sub>	Input high current	V <sub>IN</sub> = V <sub>(I2C)</sub>				−10		10	μA
I <sub>IL</sub>	Input low current	V <sub>IN</sub> = 0V				−10		10	μA
C <sub>IN</sub>	Input capacitance						5		pF
FPD-LINK III INPUT									
V <sub>CM</sub>	Common mode voltage			RIN0+, RIN0-		1.2		V	
R <sub>T</sub>	Internal termination resistor	Single-ended		RIN0+	40	50	60	Ω	
		Differential		RIN0+, RIN0-	80	100	120	Ω	
FPD-LINK III BIDIRECTIONAL CONTROL CHANNEL									
V <sub>OUT-BC</sub>	Back Channel Output Single-ended voltage	RL = 50 Ω, coaxial configuration, forward channel disabled		RIN0+, RIN0-	190	225	260	mV	
V <sub>OD-BC</sub>	Back channel output differential	RL = 100 Ω, STP configuration, forward channel disabled			380	450	520	mV	
HSTX DRIVER									
V <sub>CMTX</sub>	HS transmit static common-mode voltage			CSI_D3P/N, CSI_D2P/N, CSI_D1P/N, CSI_D0P/N, CSI_CLK1P/N, CSI_CLK0P/N	150	200	250	mV	
ΔV <sub>CMTX(1,0)</sub>	VCMTX mismatch when output is 1 or 0						5	mV <sub>P-P</sub>	
V <sub>OD</sub>	HS transmit differential voltage				140	200	270	mV	
ΔV <sub>OD</sub>	VOD mismatch when output is 1 or 0						14	mV	
V <sub>OHHS</sub>	HS output high voltage						360	mV	
Z <sub>Os</sub>	Single-ended output impedance				40	50	62.5	Ω	
ΔZ <sub>Os</sub>	Mismatch in single-ended output impedance						10	%	
LPTX DRIVER									
V <sub>OH</sub>	High level output voltage	Applicable when the supported data rate is ≤ 1.5 Gbps		CSI_D3P/N, CSI_D2P/N, CSI_D1P/N, CSI_D0P/N, CSI_CLK1P/N, CSI_CLK0P/N	1.1	1.2	1.3	V	
		Applicable when the supported data rate is > 1.5 Gbps			0.95		1.3	V	
V <sub>OL</sub>	Low level output voltage				−50		50	mV	
Z <sub>OLP</sub>	Output impedance				110			Ω	

(1) V<sub>(VDDIO)</sub> = 1.8 V ± 5% **OR** 3.3 V ± 10%

## 6.6 AC Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>LVCNOS I/O</b>							
t <sub>CLH</sub>	LVCNOS low-to-high transition time	V <sub>(VDDIO)</sub> = 1.71 to 1.89 V = VDD18 ±50 mV OR V <sub>(VDDIO)</sub> = 3V to 3.6 V, C <sub>L</sub> = 8pF	GPIO[6:0]		2.5		ns
t <sub>CHL</sub>	LVCNOS high-to-low transition time				2.5		ns
t <sub>PDB</sub>	PDB reset pulse width	Voltage supplies applied and stable	PDB	2			ms

## 6.6 AC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>FPD-LINK III RECEIVER INPUT</b>							
$V_{IN}$	Single ended input voltage	Coaxial configuration, attenuation = 20dB @ 2.1 GHz	RIN0+	40			mV
$V_{ID}$	Differential input voltage	STP configuration, attenuation = 25dB @ 2.1 GHz	RIN0+, RIN0-	80			mV
$t_{DDL1}$	Deserializer data lock time	CSI mode paired with DS90UB635-Q1, coaxial cable, attenuation = 20 dB @ 2.1GHz	AEQ full range 0x00 to 0x3F, SFILTER_CFG = 0xA9		20	300	ms
$t_{DDL2}$	Deserializer data lock time	CSI mode paired with DS90UB635-Q1, coaxial cable, attenuation = 20 dB @ 2.1GHz	AEQ range +/- 3, SFILTER_CFG = 0xA9		15	30	ms
$t_{DDL3}$	Deserializer data lock time	RAW mode paired with DS90UB633A-Q1, coaxial cable, attenuation = 14 dB @ 1.2 GHz	AEQ full range 0x00 to 0x3F, SFILTER_CFG = 0xA9		15	200	ms
$t_{DDL4}$	Deserializer data lock time	RAW mode paired with DS90UB633A-Q1, coaxial cable, attenuation = 14 dB @ 1.2 GHz	AEQ range +/- 3, SFILTER_CFG = 0xA9		15	30	ms
$t_{JIT}$	Input Jitter	CSI-2 mode paired with DS90UB635-Q1, coaxial configuration (attenuation = 20 dB) or STP configuration (attenuation = 25 dB) @ 2.1 GHz	Jitter Frequency > FPD3_PLCK/15			0.4	UI
<b>FPD-LINK III BI-DIRECTIONAL CONTROL CHANNEL</b>							
$E_{H-BC}$	Back channel output eye height	Coaxial configuration, $f_{BC} = 52$ MHz	RIN0+	130	160		mV
		STP configuration, $f_{BC} = 52$ MHz	RIN0+, RIN0-	260	320		mV
$E_{W-BC}$	Back channel output eye width	Coaxial or STP configuration, $f_{BC} = 52$ MHz	RIN0+, RIN0-	0.7	0.8		UI
$f_{BC}$	Back channel datarate <sup>(1)</sup>	Synchronous CSI-2 input mode, default register settings	Signal applied to REFCLK input		2× REFCLK		Mbps
			No signal present at REFCLK input	46		56	Mbps

- (1) The backchannel data rate (Mbps) listed is for the encoded back channel data stream. The internal reference frequency used to generate the encoded back channel data stream is two times the back channel datarate.

## 6.7 AC Electrical Characteristics CSI-2

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>HSTX DRIVER AC SPECIFICATIONS</b>							
$HSTX_{DBR}$	Data bit rate	REFCLK = 23 MHz	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	368	736	1472	Mbps
		REFCLK = 25 MHz		400	800	1600	Mbps
		REFCLK = 26 MHz		416	832	1664	Mbps
$f_{CLK}$	DDR clock frequency	REFCLK = 23 MHz	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	184	368	736	MHz
		REFCLK = 25 MHz		200	400	800	MHz
		REFCLK = 26 MHz		208	416	832	MHz
$\Delta V_{CMTX(HF)}$	Common mode voltage variations HF	Common-level variations above 450MHz	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N			15	mV <sub>RMS</sub>
$\Delta V_{CMTX(LF)}$	Common mode voltage variations LF	Common-level variations between 50 and 450MHz				25	mV <sub>RMS</sub>

## 6.7 AC Electrical Characteristics CSI-2 (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS		PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
$t_{RHS}$ $t_{FHS}$	20% to 80% rise and fall HS	HS bit rates $\leq 1$ Gbps (UI $\geq 1$ ns)		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N			0.3	UI
		HS bit rates $> 1$ Gbps (UI)					0.35	UI
		Applicable for all HS bit rates. However, to avoid excessive radiation, bit rates $\leq 1$ Gbps (UI $\geq 1$ ns), should not use values below 150 ps			100			ps
		Applicable for all HS bit rates when supporting $> 1.5$ Gbps					0.4	UI
		Applicable for all HS bit rates when supporting $> 1.5$ Gbps. However, to avoid excessive radiation, bit rates $\leq 1.5$ Gbps should not use values below 100 ps and bit rates $\leq 1$ Gbps should not use values below 150 ps.			50			ps
SDD <sub>TX</sub>	TX differential return loss	$f_{LPMAX}$		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N		−18		dB
		$f_H$	HSData rates $< 1.5$ Gbps			−9		dB
			HSData rates $> 1.5$ Gbps			−4.5		dB
		$f_{MAX}$	HSData rates $< 1.5$ Gbps			−3		dB
			HSData rates $> 1.5$ Gbps			−2.5		dB
SCC <sub>TX</sub>	TX common mode return loss	$f_{LPMAX}$		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N		−20		dB
		$f_H$				−15		dB
		$f_{MAX}$				−9		dB
LPTX DRIVER AC SPECIFICATIONS								
$t_{RLP}$	Rise time LP	15% to 85% rise time		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N			25	ns
$t_{FLP}$	Fall time LP	15% to 85% fall time					25	ns
$t_{REOT}$	Rise time post-EoT	30%-85% rise time					35	ns
$t_{LP}^{PULSE-TX}$	Pulse width of the LP exclusive-OR clock	First LP exclusive-OR clock pulse after Stop state or last pulse before Stop state		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	40			ns
		All other pulses			20			ns
$t_{LP}^{PER-TX}$	Pulse width of the LP exclusive-OR clock			CSI_CLK0P/N, CSI_CLK1P/N	90			ns

## 6.7 AC Electrical Characteristics CSI-2 (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
DV/DtSR	Slew rate	CLoad = 0pF	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N			500	mV/ns
		CLoad = 5pF				300	mV/ns
		CLoad = 20pF				250	mV/ns
		CLoad = 70pF				150	mV/ns
		CLoad = 0 to 70pF (Falling Edge Only) Data rate < 1.5 Gbps		30		mV/ns	
		CLoad = 0 to 70pF (Rising Edge Only) Data rate < 1.5 Gbps		30		mV/ns	
		CLoad = 0 to 70pF (Falling Edge Only) Data rate > 1.5 Gbps		25		mV/ns	
		CLoad = 0 to 70pF (Rising Edge Only) Data rate > 1.5 Gbps		25		mV/ns	
		CLoad = 0 to 70pF (Rising Edge Only) Applicable when the supported Data rate is < 1.5 Gbps		0 - 0.075 × (V <sub>O,INST</sub> - 700)		mV/ns	
		CLoad = 0 to 70pF (Rising Edge Only) Applicable when the supported Data rate is > 1.5 Gbps		25 - 0.0625 × (V <sub>O,INST</sub> - 550)		mV/ns	
CLOAD	Load capacitance		CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	0		50	pF
DATA-CLOCK TIMING SPECIFICATIONS							
UI <sub>INST</sub>	UI instantaneous	In 1, 2, 3, or 4 Lane Configuration	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	0.6		2.7	ns
ΔUI	UI variation	UI ≥ 1ns	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	-10%		10%	UI
		0.667ns ≤ UI		-5%		5%	UI
t <sub>SKEW(TX)</sub>	Data to Clock Skew (measured at transmitter) Skew between clock and data from ideal center	Data rate ≤ 1 Gbps	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	-0.15		0.15	UI <sub>INST</sub>
		Data rate: 1 Gbps to 1.5 Gbps	CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	-0.2		0.2	UI <sub>INST</sub>
t <sub>SKEW(TX)STATIC</sub>	Static Data to Clock Skew (TX)	Data rate > 1.5 Gbps		-0.2		0.2	UI <sub>INST</sub>
t <sub>SKEW(TX)DYNAMIC</sub>	Dynamic Data to Clock Skew (TX)			-0.15		0.15	UI <sub>INST</sub>
ISI	Channel ISI					0.2	UI <sub>INST</sub>
CSI-2 TIMING SPECIFICATIONS							

## 6.7 AC Electrical Characteristics CSI-2 (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS		PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
$t_{\text{CLK-MISS}}$	Timeout for receiver to detect absence of clock transitions and disable the clock lane HS-RX			CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N	60			ns
$t_{\text{CLK-POST}}$	HS exit				$60 + 52 \times \text{UI}$			ns
$t_{\text{CLK-PRE}}$	Time HS clock shall be driver prior to any associated data lane beginning the transition from LP to HS mode				8			UI
$t_{\text{CLK-PREPARE}}$	Clock lane HS entry				38		95	ns
$t_{\text{CLK-SETTLE}}$	Time interval during which the HS receiver shall ignore any clock lane HS transitions				95		300	ns
$t_{\text{CLK-TERM-EN}}$	Time-out at clock lane display module to enable HS termination						38	ns
$t_{\text{CLK-TRAIL}}$	Time that the transmitter drives the HS-0 state after the last payload clock bit of a HS transmission burst				60			ns
$t_{\text{CLK-PREPARE}} + t_{\text{CLK-ZERO}}$	TCLK-PREPARE + time that the transmitter drives the HS-0 state prior to starting the clock				300			ns
$t_{\text{D-TERM-EN}}$	Time for the data lane receiver to enable the HS line termination			CSI_D0P/N, CSI_D1P/N, CSI_D2P/N, CSI_D3P/N, CSI_CLK0P/N, CSI_CLK1P/N			$35 + 4 \times \text{UI}$	ns
$t_{\text{EOT}}$	Transmitted time interval from the start of tHS-TRAIL to the start of the LP-11 state following a HS burst						$105 + 12 \times \text{UI}$	ns
$t_{\text{HS-EXIT}}$	Time that the transmitter drives LP-11 following a HS burst				100			ns
$t_{\text{HS-PREPARE}}$	Data lane HS entry				$40 + 4 \times \text{UI}$		$85 + 6 \times \text{UI}$	ns
$t_{\text{HS-PREPARE}} + t_{\text{HS-ZERO}}$	$t_{\text{HS-PREPARE}}$ + time that the transmitter drives the HS-0 state prior to transmitting the Sync sequence				$145 + 10 \times \text{UI}$			ns
$t_{\text{HS-SETTLE}}$	Time interval during which the HS receiver shall ignore any data lane HS transitions, starting from the beginning of tHS-SETTLE				$85 + 6 \times \text{UI}$		$145 + 10 \times \text{UI}$	ns
$t_{\text{HS-SKIP}}$	Time interval during which the HS-RX should ignore any transitions on the data lane, following a HS burst. The end point of the interval is defined as the beginning of the LP-11 state following the HS burst.				40		$55 + 4 \times \text{UI}$	ns
$t_{\text{HS-TRAIL}}$	Data lane HS exit				$60 + 4 \times \text{UI}$			ns
$t_{\text{LPX}}$	Transmitted length of LP state				50			ns
$t_{\text{WAKEUP}}$	Recovery Time from Ultra Low Power State (ULPS)				1			ms
$t_{\text{INIT}}$	Initialization period				100			μs

## 6.8 Recommended Timing for the Serial Control Bus

Over I<sup>2</sup>C supply and temperature ranges unless otherwise specified.

			MIN	TYP	MAX	UNIT
$f_{\text{SCL}}$	SCL Clock Frequency	Standard-mode	>0		100	kHz
		Fast-mode	>0		400	kHz
		Fast-mode Plus	>0		1	MHz

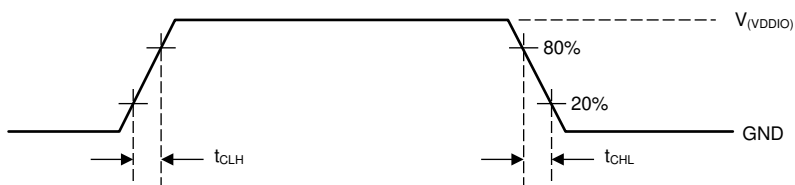
## 6.8 Recommended Timing for the Serial Control Bus (continued)

Over I<sup>2</sup>C supply and temperature ranges unless otherwise specified.

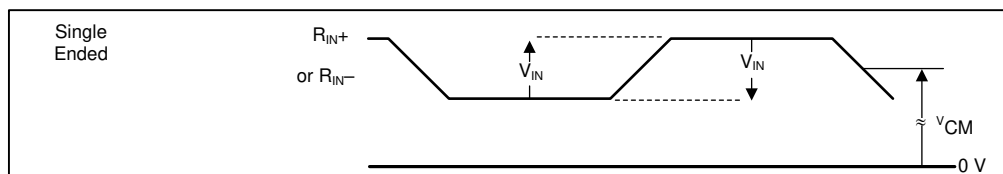
			MIN	TYP	MAX	UNIT
t <sub>LOW</sub>	SCL Low Period	Standard-mode	4.7			μs
		Fast-mode	1.3			μs
		Fast-mode Plus	0.5			μs
t <sub>HIGH</sub>	SCL High Period	Standard-mode	4.0			μs
		Fast-mode	0.6			μs
		Fast-mode Plus	0.26			μs
t <sub>HD,STA</sub>	Hold time for a start or a repeated start condition	Standard-mode	4.0			μs
		Fast-mode	0.6			μs
		Fast-mode Plus	0.26			μs
t <sub>SU,STA</sub>	Set up time for a start or a repeated start condition	Standard-mode	4.7			μs
		Fast-mode	0.6			μs
		Fast-mode Plus	0.26			μs
t <sub>HD,DAT</sub>	Data hold time	Standard-mode	0			μs
		Fast-mode	0			μs
		Fast-mode Plus	0			μs
t <sub>SU,DAT</sub>	Data set up time	Standard-mode	250			ns
		Fast -mode	100			ns
		Fast-mode Plus	50			ns
t <sub>SU,STO</sub>	Set up time for STOP condition	Standard-mode	4.0			μs
		Fast-mode	0.6			μs
		Fast-mode Plus	0.26			μs
t <sub>BUF</sub>	Bus free time between STOP and START	Standard-mode	4.7			μs
		Fast-mode	1.3			μs
		Fast-mode Plus	0.5			μs
t <sub>r</sub>	SCL & SDA rise time	Standard-mode			1000	ns
		Fast-mode			300	ns
		Fast-mode Plus			120	ns
t <sub>f</sub>	SCL & SDA fall time	Standard-mode			300	ns
		Fast-mode			300	ns
		Fast-mode Plus			120	ns
C <sub>b</sub>	Capacitive load for each bus line	Standard-mode			400	pF
		Fast-mode			400	pF
		Fast-mode Plus			550	pF
t <sub>VD,DAT</sub>	Data valid time	Standard-mode			3.45	μs
		Fast-mode			0.9	μs
		Fast-mode Plus			0.45	μs
t <sub>VD,ACK</sub>	Data valid acknowledge time	Standard-mode			3.45	μs
		Fast-mode			0.9	μs
		Fast-mode Plus			0.45	μs
t <sub>SP</sub>	Input filter	Fast-mode			50	ns
		Fast-mode Plus			50	ns



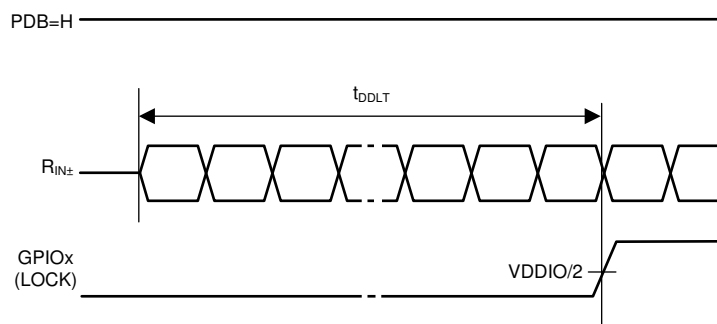
## 6.9 Timing Diagrams



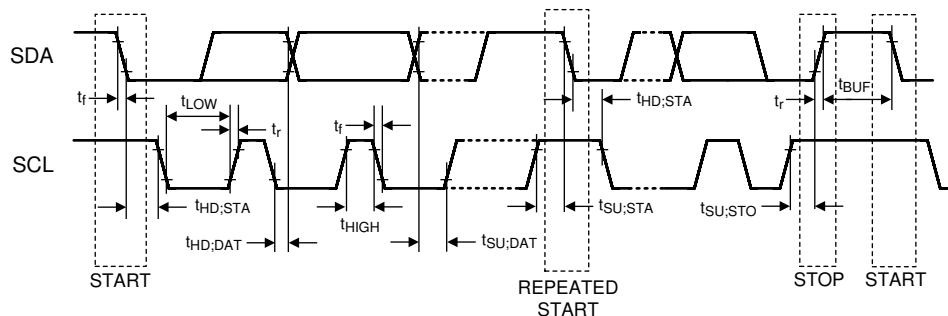
**Figure 6-1. LVCMOS Transition Times**



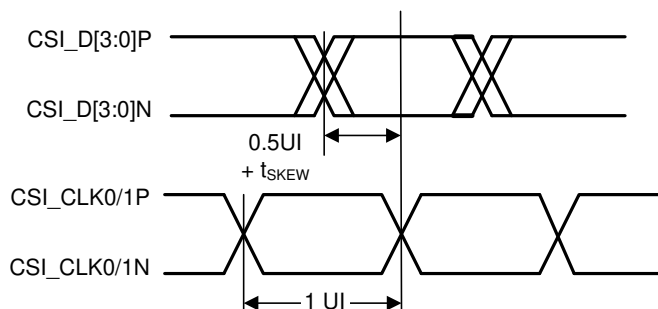
**Figure 6-2. FPD-Link III Receiver  $V_{ID}$ ,  $V_{IN}$ ,  $V_{CM}$**



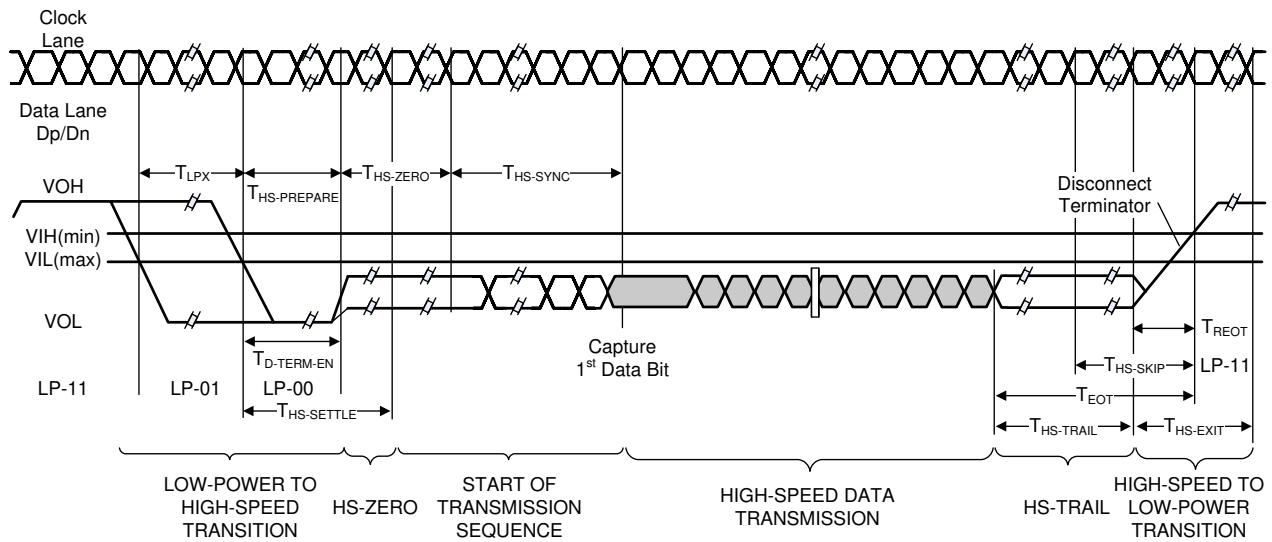
**Figure 6-3. Deserializer Data Lock Time**



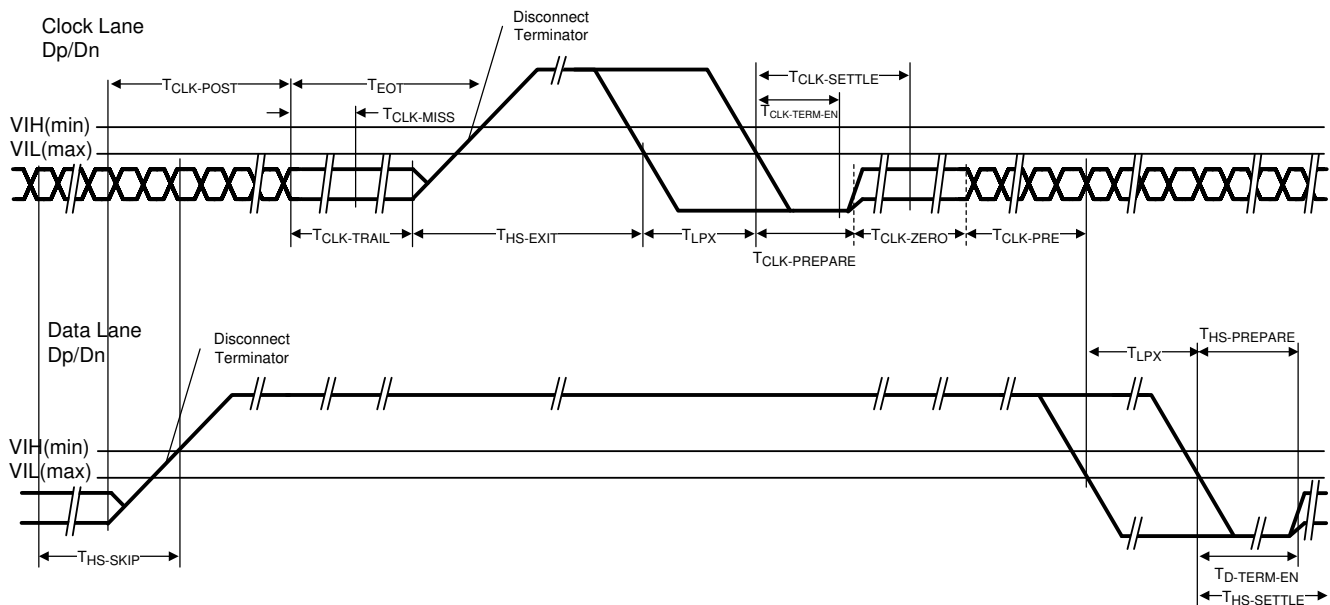
**Figure 6-4. I2C Serial Control Bus Timing**



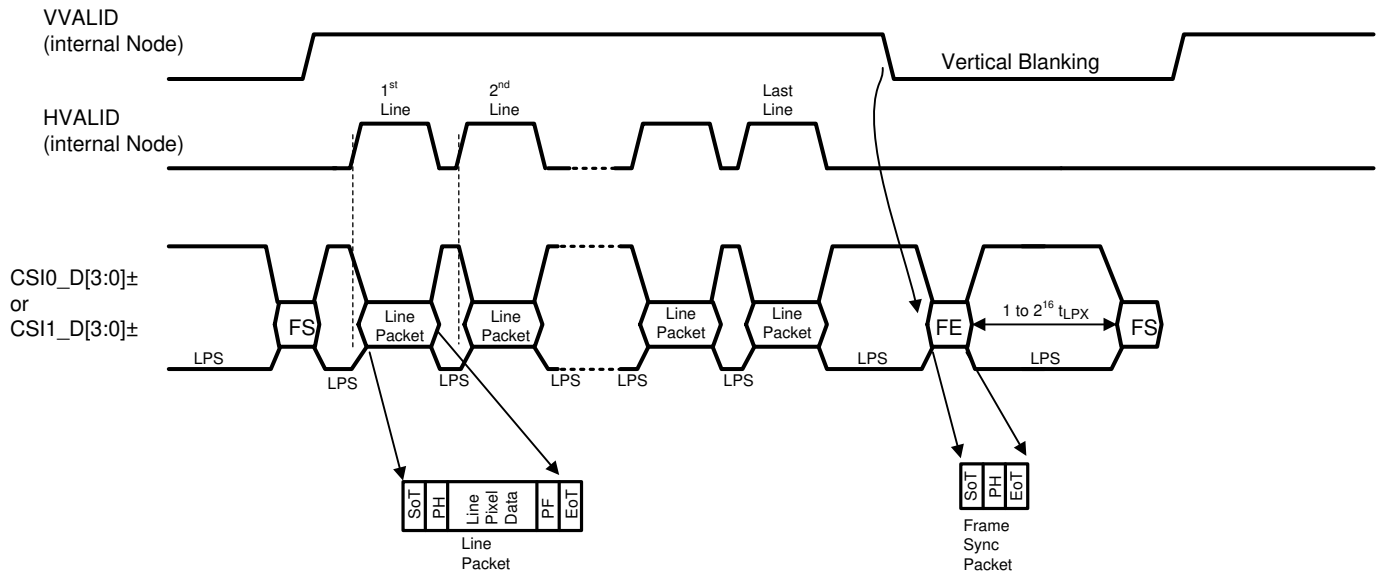
**Figure 6-5. Clock and Data Timing in HS Transmission**



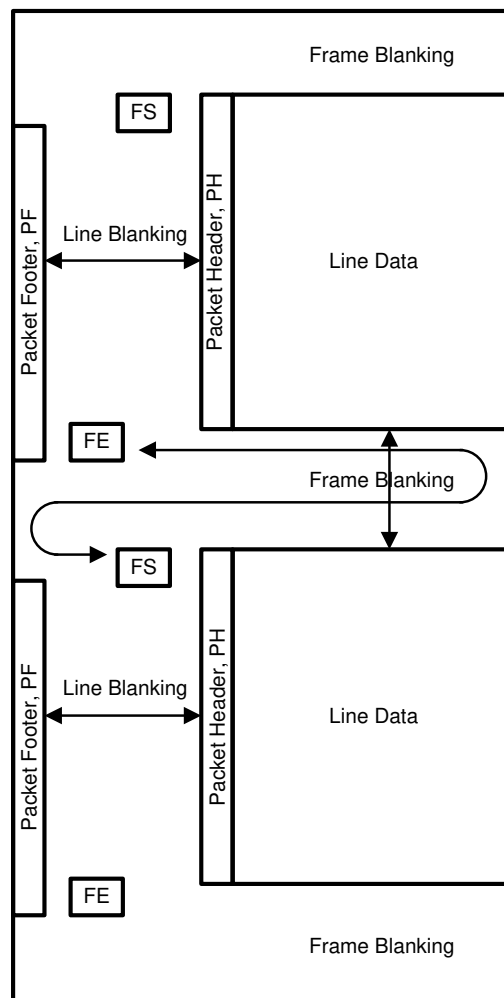
### Figure 6-6. High-Speed Data Transmission Burst



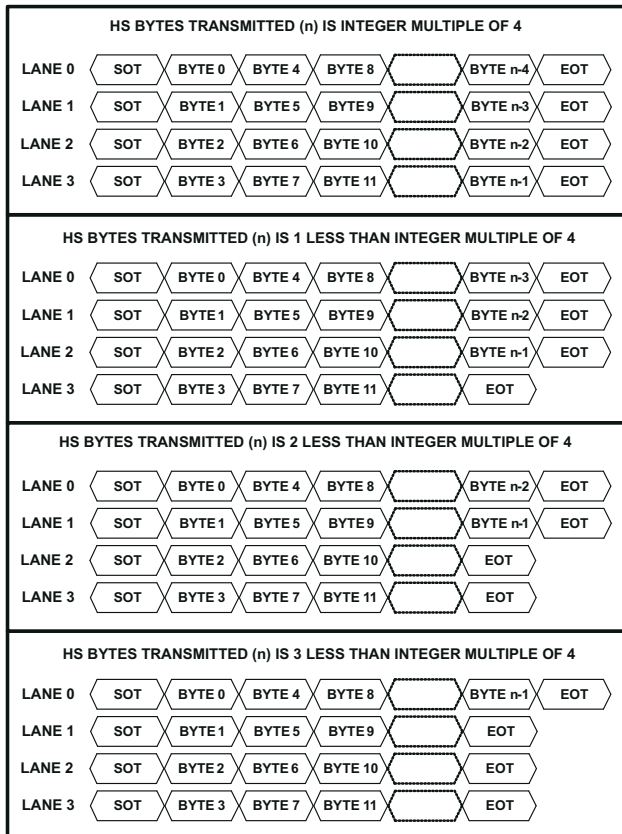
### ☒ 6-7. Switching the Clock Lane Between Clock Transmission and Low-Power Mode



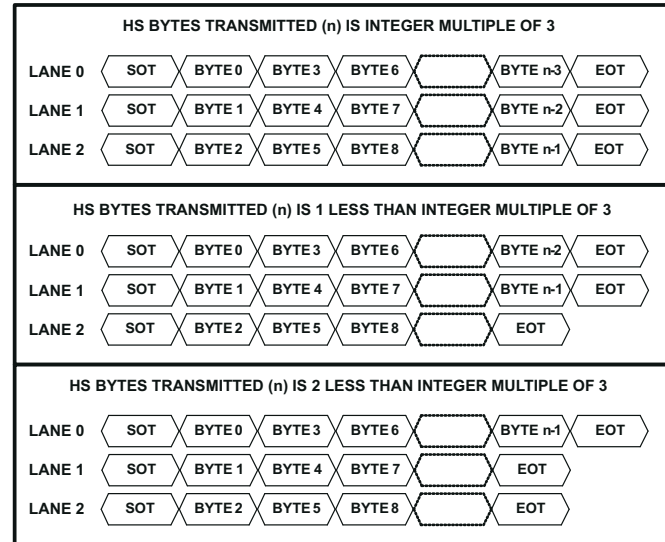
**6-8. Long Line Packets and Short Frame Sync Packets**



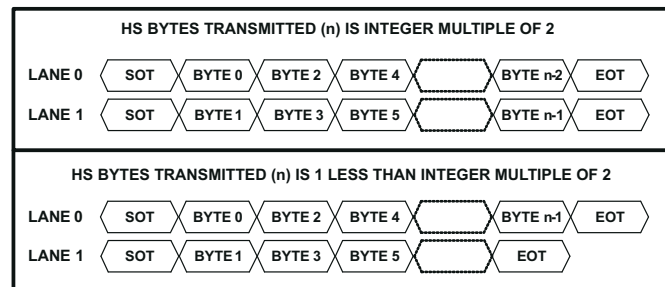
**6-9. CSI-2 General Frame Format**



4 CSI-2 Data Lane Configuration (default)



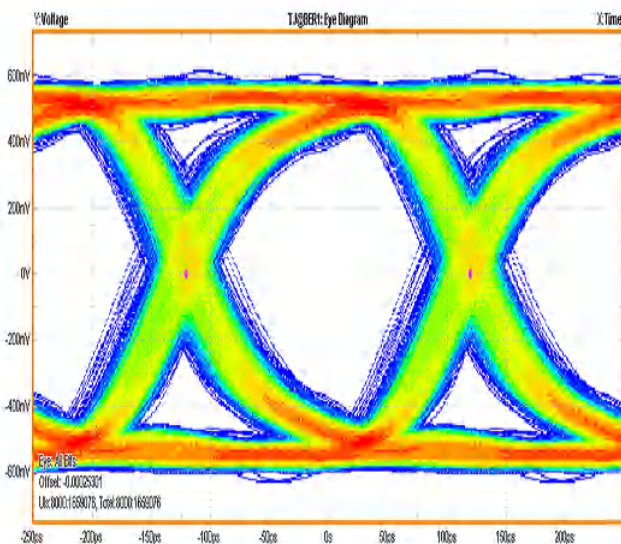
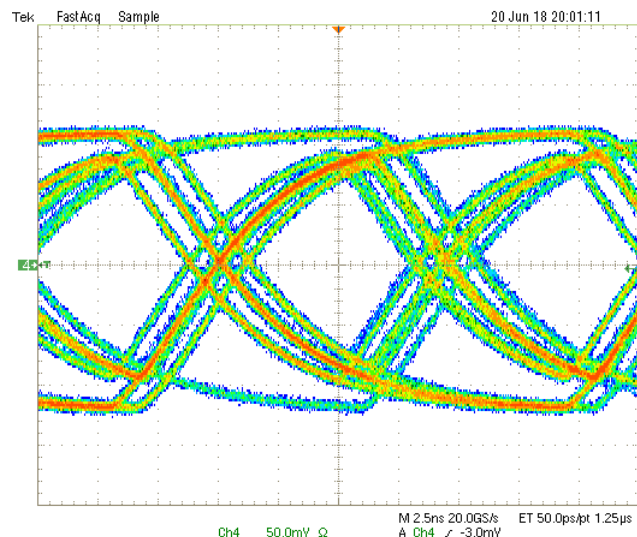
3 CSI-2 Data Lane Configuration



2 CSI-2 Data Lane Configuration


**FIG 6-10. MIPI CSI-2 Data Lane Configuration**

## 6.10 Typical Characteristics


**FIG 6-11. Forward Channel Monitor Loop Through Typical Rx Waveform (CMLOUT)**

**FIG 6-12. Back Channel Output Typical Waveform**

## 7 Detailed Description

### 7.1 Overview

The DS90UB638-Q1 is a versatile deserializer that receives an FPD-Link III stream and transmits the received data over a MIPI camera serial interface (CSI-2). When coupled with an ADAS FPD-Link III serializer (DS90UB63x-Q1), the DS90UB638-Q1 receives a data stream from a remote sensor, then transmits that data as one or two CSI-2 outputs.

**表 7-1. Serializer Compatibility**

SERIALIZER	DS90UB635-Q1	DS90UB633A-Q1
Compatibility	Yes	Yes

#### 7.1.1 Functional Description

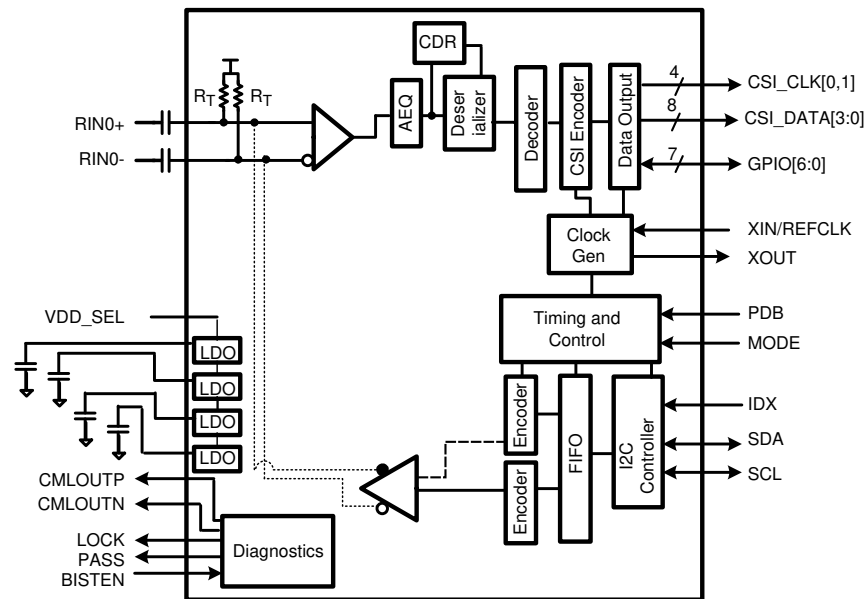
The DS90UB638-Q1 FPD-Link III deserializer, in conjunction with an ADAS FPD-Link III serializer, supports the video transport needs with an ultra-high speed forward channel and an embedded bidirectional control channel. The DS90UB638-Q1 received data is output from a configurable MIPI CSI-2 port. The CSI-2 port may be configured as either a single CSI-2 output with four lanes up to 1.6 Gbps per lane or as two 2 lane CSI-2 outputs for sending replicated data on both ports. A second differential clock is available for the second replicated output when configured for dual CSI-2 outputs supporting one clock lane and one or two data lanes each. The DS90UB638-Q1 can support multiple data formats and different resolutions as provided by the sensor. Conversion between different data formats is not supported. The CSI-2 Tx module accommodates both image data and non-image data (including synchronization or embedded data packets).

The DS90UB638-Q1 CSI-2 interface converts the sensor data stream into packets designated for each virtual channel. Each virtual channel is identified by a unique channel identification number in the packet header.

When the DS90UB638-Q1 is paired with a DS90UB635-Q1 serializer, the received FPD-Link III forward channel is constructed in 40-bit long frames and when it is paired with a DS90UB633A-Q1 serializer, the received FPD-Link III forward channel is constructed in 28-bit long frames. Each encoded frame contains video payload data, I2C forward channel data, and additional information on framing, data integrity and link diagnostics. The high-speed, serial bit stream from the DS90UB635-Q1 or DS90UB633A-Q1 contains an embedded clock and DC-balancing ensuring sufficient data line transitions for enhanced signal quality. When paired with ADAS serializers in RAW input mode, the received FPD-Link III forward channel is similarly constructed at a lower line rate in 28-bit long frames. The DS90UB638-Q1 device recovers a high-speed, FPD-Link III forward channel signal and generates a bidirectional control channel control signal in the reverse channel direction. The DS90UB638-Q1 converts the FPD-Link III stream into a MIPI CSI-2 output interface designed to support automotive sensors, including 2MP/60fps and 4MP/30fps image sensors.

The DS90UB638-Q1 device has one receive input port to accept sensor streams. The control channel function of the DS90UB63x-Q1 chipset provides bidirectional communication between the image sensors and ECU. The integrated control channel transfers data bidirectionally over the same cable used for video data interface. This interface offers advantages over other chipsets by eliminating the need for additional wires for programming and control. The bidirectional control channel bus is controlled through an I2C port. The bidirectional control channel offers continuous low latency communication and is not dependent on video blanking intervals.

## 7.2 Functional Block Diagram



## 7.3 Feature Description

The DS90UB638-Q1 provides a flexible deserializer for automotive sensor applications.

## 7.4 Device Functional Modes

The DS90UB638-Q1 supports two main the FPD-Link III operating modes:

- CSI-2 Mode (DS90UB635-Q1 compatible)
- RAW Mode (DS90UB633A-Q1 compatible)

The two modes mainly control the FPD-Link III receiver operation of the device. In both cases, the output format for the device is CSI-2 through the CSI-2 transmit port.

The input mode of operation is controlled by the FPD3\_MODE (Register 0x6D[1:0]) setting in the Port Configuration register. The input mode may also be controlled by the MODE strap pin.

### 7.4.1 CSI-2 Mode

The DS90UB638-Q1 receives CSI-2 formatted data on an FPD-Link III input port and forwards the data to the CSI-2 transmit port. The deserializer can operate in CSI-2 mode with synchronous back channel reference or non-synchronous mode. The forward channel line rate is independent of the CSI-2 rate in synchronous or non-synchronous with external clock mode. The CSI-2 mode supports remapping of Virtual Channel IDs at the input of the receive port.

The deserializer Rx Port can support an FPD-Link line rate up to 4.16 Gbps, where the forward channel and back channel rates are based on the reference frequency used for the serializer:

- In Synchronous mode based on REFCLK input frequency reference, the FPD-Link line rate is a fixed value of  $160 \times \text{REFCLK}$ .  $\text{FPD3\_PCLK} = 4 \times \text{REFCLK}$  and Back channel rate =  $2 \times \text{REFCLK}$ . For example with  $\text{REFCLK} = 25 \text{ MHz}$ , line rate = 4.0 Gbps,  $\text{FPD3\_PCLK} = 100 \text{ MHz}$ , back channel data rate = 50 Mbps. The sensor CSI-2 rate is independent of the line rate and Tx CSI-2 rate in synchronous clocking mode and can be up to 3.328 Gbps.
- In Non-synchronous clocking mode when the DS90UB635-Q1 uses external reference clock ( $f_{\text{CLKIN}}$ ) the FPD-Link line rate is typically  $f_{\text{CLKIN}} \times 80$ ,  $\text{FPD3\_PCLK} = 2 \times f_{\text{CLKIN}}$  or  $1 \times f_{\text{CLKIN}}$  and back channel data rate is set to 10 Mbps. For example, with  $f_{\text{CLKIN}} = 50 \text{ MHz}$ , line rate = 4Gbps,  $\text{FPD3\_PCLK} = 100 \text{ MHz}$ , and the back channel rate is 10 Mbps. The sensor CSI-2 rate is independent of the  $f_{\text{CLKIN}}$ .

## 7.4.2 RAW Mode

When operating in Raw FPD-Link IV input mode, the DS90UB638-Q1 receives RAW10 or RAW12 data from a DS90UB633A-Q1 serializer. The data is translated into a RAW10 or RAW12 CSI-2 video stream for forwarding to the CSI-2 transmit port. For each input port, the CSI-2 packet header VC-ID and Data Type are programmable.

In RAW mode the DS90UB638-Q1 deserializer each Rx Port can support up to:

- 12-bits of DATA + 2 SYNC bits for an input PCLK range of 56.25 MHz to 100 MHz in the 12-bit high frequency mode. Line rate =  $\text{FPD3\_PCLK} \times (2/3) \times 28$ ; for example,  $\text{FPD3\_PCLK} = 100 \text{ MHz}$ , line rate =  $(100 \text{ MHz}) \times (2/3) \times 28 = 1.87 \text{ Gbps}$ . Note: No HS/VS restrictions (raw).
- 10-bits of DATA + 2 SYNC bits for an input PCLK range of 75 MHz to 100 MHz in the 10-bit mode. Line rate =  $\text{FPD3\_PCLK}/2 \times 28$ ; for example,  $\text{FPD3\_PCLK} = 100 \text{ MHz}$ , line rate =  $(100 \text{ MHz}/2) \times 28 = 1.40 \text{ Gbps}$  Note: HS/VS restricted to no more than one transition per 10 PCLK cycles.

When operating with DVP serializer, the DS90UB638-Q1 deserializer also supports DVP formats such as YUV-422 which have the same pixel packing as RAW8, RAW10 or RAW12. For example; there are 3 YUV CSI-2 data types that have the same pixel packing as RAW10: YUV420 10 bit, YUV420 10 bit Chroma shifted or YUV422 10bit. These formats can be used as well as 8 bit and 12 bit YUV formats which adhere to the same structure as RAW8 and RAW12 respectively.

## 7.4.3 RX MODE Pin

Configuration of the FPD-Link III operating input mode may be done through the MODE input strap pin, or through the configuration register bits. A pullup resistor and a pull-down resistor of suggested values may be used to set the voltage ratio of the MODE input ( $V_{\text{TARGET}}$ ) and  $V_{(\text{VDD18})}$  to select one of the mode. The DS90UB638-Q1 waits 1 ms after PDB goes high to allow time for power supply transients before sampling the MODE pin strap value and configuring the device to set the I2C address. The DS90UB638-Q1 operating mode is:

- CSI-2 input Rx REFCLK mode
- 12-bit HF / 10-bit DVP Rx modes

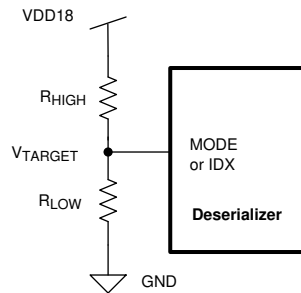


图 7-1. Strap Pin Connection Diagram

表 7-2. Strap Configuration Mode Select

MODE NO.	V <sub>TARGET</sub> VOLTAGE RANGE			V <sub>TARGET</sub> STRAP VOLTAGE	SUGGESTED STRAP RESISTORS (1% TOL)		RX MODE
	V <sub>MIN</sub>	V <sub>TYP</sub>	V <sub>MAX</sub>	VDD18 = 1.8 V	R <sub>HIGH</sub> (kΩ)	R <sub>LOW</sub> (kΩ)	
0	0	0	$0.131 \times V_{(\text{VDD18})}$	0	OPEN	10.0	CSI-2 non-synchronous Back Channel
1	$0.296 \times V_{(\text{VDD18})}$	$0.330 \times V_{(\text{VDD18})}$	$0.362 \times V_{(\text{VDD18})}$	0.582	75.0	35.7	RAW12 HF
	$0.761 \times V_{(\text{VDD18})}$	$0.792 \times V_{(\text{VDD18})}$	$0.823 \times V_{(\text{VDD18})}$	1.420	25.5	95.3	
2	$0.412 \times V_{(\text{VDD18})}$	$0.443 \times V_{(\text{VDD18})}$	$0.474 \times V_{(\text{VDD18})}$	0.792	71.5	56.2	RAW10
	$0.876 \times V_{(\text{VDD18})}$	$V_{(\text{VDD18})}$	$V_{(\text{VDD18})}$	1.8	10.0	OPEN	
3	$0.525 \times V_{(\text{VDD18})}$	$0.559 \times V_{(\text{VDD18})}$	$0.592 \times V_{(\text{VDD18})}$	0.995	78.7	97.6	CSI-2 Synchronous Back Channel



The strapped values can be viewed and modified in the following locations:

- RX Mode – Port Configuration FPD3\_MODE (Register 0x6D[1:0])
- Clock Mode – Device Status and CSI\_PLL\_CTL (Register bits 0x04[4] and 0x1F[0])

#### 7.4.4 REFCLK

A valid 23-MHz to 26-MHz reference clock is required on the REFCLK pin 5 for precise frequency operation. The REFCLK frequency defines all internal clock timers, including the back channel rate, I2C timers, CSI-2 datarate, FrameSync signal parameters, and other timing critical internal circuitry. REFCLK input must be continuous. If the REFCLK input does not detect a transition more than 20  $\mu$ S, this may cause a disruption in the CSI-2 output. REFCLK should be applied to the DS90UB638-Q1 only when the supply rails are above minimum levels (see [セクション 9.2](#)). At start-up, the DS90UB638-Q1 defaults to an internal oscillator to generate an backup internal reference clock at nominal frequency of 25 MHz  $\pm$ 10%.

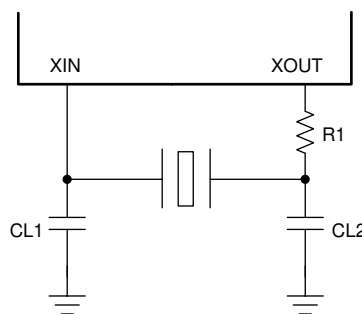
The REFCLK LVCMOS input oscillator specifications are listed in [表 7-3](#).

**表 7-3. REFCLK Oscillator Specifications**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>REFERENCE CLOCK</b>					
Frequency tolerance	$-40^{\circ}\text{C} \leq T_A \leq 105^{\circ}\text{C}$			$\pm 50$	ppm
Frequency stability	Aging			$\pm 50$	ppm
Amplitude		800	1200	$V_{(VDDIO)}$	mVp-p
Symmetry	Duty Cycle	40%	50%	60%	
Rise and fall time	10% – 90%			5	ns
Jitter	200 kHz – 10 MHz		50	1000	ps p-p
Frequency		23	25	26	MHz

#### 7.4.5 Crystal Recommendations

A 25-MHz, parallel, 18-pF load crystal resonator should be used if a crystal source is desired. [図 7-2](#) shows a typical connection for a crystal resonator circuit. The load capacitor values will vary with the crystal vendors; check with the vendor for the recommended loads.



**図 7-2. Crystal Oscillator Circuit**

As a starting point for evaluating an oscillator circuit, if the requirements for the crystal are not known, CL1 and CL2 should be set at 27 pF and R1 should be set at 0  $\Omega$ . Specification for 25-MHz crystal are listed in [表 7-4](#).

**表 7-4. 25-MHz Crystal Specifications**

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>REFERENCE CLOCK</b>					
Frequency			25		MHz
Frequency Tolerance and Stability	Across operational temperature and aging			$\pm 100$	ppm



### 7.4.6 Receiver Port Control

After the device is powered up, the following register settings are required:

- Program 0x0c bit 1 to 0
- Program 0x21 bit 6 to 1

The DS90UB638-Q1 can support a single input to Rx port 0. The Receiver port control register RX\_PORT\_CTL 0x0C (表 7-29) allows for disabling the Rx input when not in use. These bits can only be written by a local I2C controller at the deserializer side of the FPD-Link.

As an alternative to paging to access FPD-Link III Receive unique port registers, separate I2C addresses may be enabled to allow direct access to the port-specific registers. The Port I2C address registers allow programming a separate 7-bit I2C address to allow access to unique, port-specific registers without paging. I2C commands to these assigned I2C addresses are also allowed access to all shared registers (see 表 7-161).

### 7.4.7 LOCK and PASS Status

The DS90UB638-Q1 provides dedicated PASS and LOCK outputs for monitoring status as well as through the DEVICE\_STS register (address 0x04). The source of the deserializer LOCK and PASS signals for pin monitoring and interrupt operation is also controlled by the LOCK\_SEL and PASS\_SEL fields in the RX\_PORT\_CTL register. The source of the LOCK and PASS can be allocated to Port 0 Receiver. At start-up, the deserializer will synchronize with the input signal provided by the serializer and assert the LOCK indication once stable. The lock detect circuit includes an option to check for link bit errors as part of the lock detection and determine if LOCK is lost. The Receive Port Lock status is available through the RX\_PORT\_STS1 register 0x4D. The LOCK status may also be used to enable video forwarding and other options. I2C communication across the FPD-Link should be attempted only during LOCK condition.

If the deserializer loses LOCK, the receiver will reset and perform the LOCK algorithm again to reacquire the serial data stream sent by the serializer. The receive port will truncate video frames containing errors and resume forwarding the video when LOCK is re-established.

The Receive port will indicate Pass status once specific conditions are met, including a number of valid frames received. Valid frames may include requiring no link bit errors and consistent frame size including video line length or number of video lines. The receive port may be programmed to truncate video frames containing errors and prevent the forwarding of video until the Pass conditions are met.

### 7.4.8 Adaptive Equalizer

The FPD-Link III receiver inputs incorporates an adaptive equalizer (AEQ), to compensate for signal degradation from the communications channel and interconnect components. The RX port signal path continuously monitors cable characteristics for long-term cable aging and temperature changes. The AEQ is primarily intended to adapt and compensate for channel losses over the lifetime of a cable installed in an automobile. The AEQ attempts to optimize the equalization setting of the RX receiver. This adaption includes compensating insertion loss from temperature effects and aging degradation due to bending and flexion. To determine the maximum cable reach, factors that affect signal integrity such as jitter, skew, inter-symbol interference (ISI), crosstalk, and so forth, must also be considered. The equalization configuration and status are programmed in registers 0xD2–0xD3 (see 表 7-141).

#### 7.4.8.1 Adaptive Equalizer Algorithm

The AEQ process steps through allowed values of the equalizer controls find a value that allows the Clock Data Recovery (CDR) circuit to maintain valid lock condition. For each EQ setting, the circuit waits for a programmed re-lock time period, then checks results for valid lock. If valid lock is detected, the circuit will stop at the current EQ setting and maintain constant value as long as lock state persists. If the deserializer loses LOCK, the adaptive equalizer will resume the LOCK algorithm and the EQ setting is incremented to the next valid state. Once lock is lost, the circuit will continue searching EQ settings to find a valid setting to reacquire the serial data stream sent by the serializer that remains locked.

### 7.4.8.2 AEQ Settings

#### 7.4.8.2.1 AEQ Start-Up and Initialization

The AEQ circuit can be restarted at any time by setting the AEQ\_RESTART bit in the AEQ\_CTL2 register 0xD2 (see [表 7-141](#)). Once the deserializer is powered on, the AEQ is continually searching through EQ settings and could be at any setting when signal is supplied from the serializer. If the Rx Port CDR locks to the signal, it may be good enough for low bit errors, but could be not optimized or overequalized. The DS90UB638-Q1 when connected to an ADAS serializer (DS90UB63x-Q1) will by default restart the AEQ adaption upon achieving first positive lock indication in order to provide more consistent start-up from known conditions. With this feature disabled, the AEQ may lock at a relatively random EQ setting based on when the FPD-Link III input signal is initially present. Alternatively, AEQ\_RESTART or DIGITAL\_RESET0 could be applied once the ADAS serializer input signal frequency is stable to restart adaption from the minimum EQ gain value. These techniques allow for a more consistent initial EQ setting following adaption.

#### 7.4.8.2.2 AEQ Range

AEQ Min/Max settings: The AEQ circuit can be programmed with minimum and maximum settings used during the EQ adaption. Using the full AEQ range will provide the most flexible solution, however if the channel conditions are known an improved deserializer lock time can be achieved by narrowing the search window for allowable EQ gain settings. For example in a system use case with a longer cable and multiple interconnects creating higher channel attenuation, the AEQ would not adapt to the minimum EQ gain settings. Likewise in a system use case with short cable and low channel attenuation AEQ would not generally adapt to the highest EQ gain settings. The AEQ range is determined by the AEQ\_MIN\_MAX register 0xD5 (see [セクション 7.7.128](#)) where AEQ\_MAX sets the maximum value of EQ gain. The ADAPTIVE\_EQ\_FLOOR\_VALUE determines the starting value for EQ gain adaption. To enable the minimum AEQ limit, SET\_AEQ\_FLOOR bit in the AEQ\_CTL2 register 0xD2[2] must also be set. An AEQ range (AEQ\_MAX - AEQ\_FLOOR) to allow a variation around the nominal setting of  $-2/+4$  or  $\pm 3$  around the nominal AEQ value specific to Rx port channel characteristics provides a good trade off in lock time and adaptability. The setting for the AEQ after adaption can be readback from the AEQ\_STATUS register 0xD3 (see [セクション 7.7.126](#)).

#### 7.4.8.2.3 AEQ Timing

The dwell time for AEQ to wait for lock or error free status is also programmable. When checking each EQ setting the AEQ will wait for a time interval, controlled by the ADAPTIVE\_EQ\_RELOCK\_TIME field in the AEQ\_CTL2 register (see [表 7-141](#)) before incrementing to the next allowable EQ gain setting. The default wait time is set to 2.62 ms based on REFCLK = 25 MHz. Once the maximum setting is reached, if there is no lock acquired during the programmed relock time, the AEQ will restart adaption at the minimum setting or AEQ\_FLOOR value.

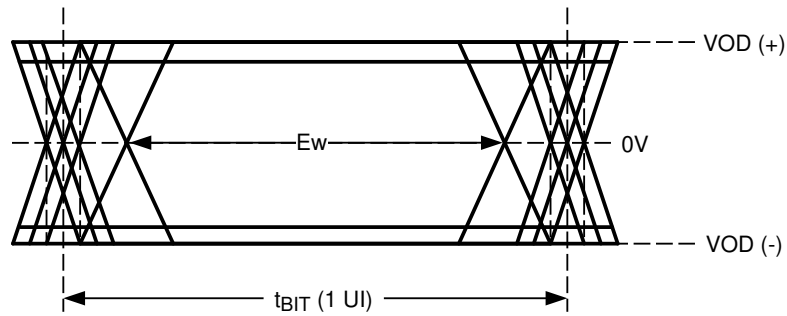
#### 7.4.8.2.4 AEQ Threshold

The DS90UB638-Q1 receiver will by default adapt based on FPD-Link error checking during the Adaptive Equalization process. The specific errors linked to equalizer adaption, FPD-Link III clock recovery error, packet encoding error, and parity error can be individually selected in AEQ\_CTL1 register 0x42 (see [セクション 7.7.54](#)). Errors are accumulated over 1/2 of the period of the timer set by the ADAPTIVE\_EQ\_RELOCK\_TIME. If the number of errors is greater than the programmed threshold (AEQ\_ERR\_THOLD), the AEQ will attempt to increase the EQ setting.

### 7.4.9 Channel Monitor Loop-Through Output Driver (CMLOUT)

The DS90UB638-Q1 includes an internal Channel Monitor Loop-through output on the CMLOUTP and CMLOUTN pins which provides a functional check of the receive channel. CMLOUT is useful in identifying gross signal conditioning issues.

[表 7-5](#) includes details on selecting the RX receiver of CMLOUTP and CMLOUTN configuration. To disable the CMLOUT, either follow the instructions in table to reload register default values, or reset the DS90UB638-Q1.




**7-3. CMLOUT Output Driver**

**表 7-5. Channel Monitor Loop-Through Output Configuration**

	FPD-Link III RX Port 0
<b>ENABLE MAIN LOOP-THROUGH DRIVER</b>	0xB0 = 0x14; 0xB1 = 0x00; 0xB2 = 0x80 0xB1 = 0x03; 0xB2 = 0x28 0xB1 = 0x04; 0xB2 = 0x28
<b>SELECT CHANNEL MUX</b>	0xB1 = 0x02; 0xB2 = 0x20
<b>SELECT RX PORT</b>	0xB0 = 0x04; 0xB1 = 0x0F; 0xB2 = 0x01 0xB1 = 0x10; 0xB2 = 0x02
<b>DISABLE MAIN LOOP-THROUGH DRIVER</b>	0xB0 = 0x14; 0xB1 = 0x00; 0xB2 = 0x00 0xB1 = 0x03; 0xB2 = 0x08 0xB1 = 0x04; 0xB2 = 0x08
<b>DESELECT CHANNEL MUX</b>	0xB1 = 0x02; 0xB2 = 0x20
<b>DESELECT RX PORT</b>	0xB0 = 0x04; 0xB1 = 0x0F; 0xB2 = 0x00 0xB1 = 0x10; 0xB2 = 0x00

**7.4.9.1 Code Example for CML OUT FPD-Link III RX Port 0:**

```

WriteI2C(0xB0,0x14)    # FPD-Link III RX Shared, page 0
WriteI2C(0xB1,0x00)    # Offset 0
WriteI2C(0xB2,0x80)    # Enable loop through driver
WriteI2C(0xB1,0x03)    #
WriteI2C(0xB2,0x28)    #
WriteI2C(0xB1,0x04)    #
WriteI2C(0xB2,0x28)    #
#
WriteI2C(0xB1,0x02)    #
WriteI2C(0xB2,0x20)    #
#
WriteI2C(0xB0,0x04)    # Offset 4
WriteI2C(0xB1,0x0F)    #
WriteI2C(0xB2,0x01)    #
WriteI2C(0xB1,0x10)    #
WriteI2C(0xB2,0x02)    # Enable CML data output

```

**7.4.10 RX Port Status**

In addition to the Lock and PASS indications, the deserializer is able to monitor and detect several other RX conditions and interrupt states. This information is latched into the RX port status registers RX\_PORT\_STS1 (0x4D) and RX\_PORT\_STS2 (0x4E). There are bits to flag any change in LOCK status (LOCK\_STS\_CHG) or detect any errors in the control channel over the forward link (BCC\_CRC\_ERROR, BCC\_SEQ\_ERROR) which are cleared upon read. The Rx Port status registers also allow the user to monitor the presence of the stable input signal, along with parity and CRC errors, line length, and lines per video frame.

**7.4.10.1 RX Parity Status**

The FPD-Link III receiver checks the decoded data parity to detect any errors in the received FPD-Link III frame. Parity errors are counted up and accessible through the RX\_PAR\_ERR\_HI and RX\_PAR\_ERR\_LO registers 0x55 and 0x56 to provide combined 16-bit error counter. In addition, a parity error flag can be set once a programmed number of parity errors have been detected. This condition is indicated by the PARITY\_ERROR flag in the RX\_PORT\_STS1 register. Reading the counter value will clear the counter value and PARITY\_ERROR flag. An interrupt may also be generated based on assertion of the parity error flag. By default, the parity error counter will be cleared and the flag will be cleared on loss of Receiver lock. To ensure an exact read of the parity error counter, parity checking should be disabled in the GENERAL\_CFG register 0x02 before reading the counter.

**7.4.10.2 FPD-Link Decoder Status**

The FPD-Link III receiver also checks the decoded data for encoding or sequence errors in the received FPD-Link III frame. If either of these error conditions are detected the FPD3\_ENC\_ERROR bit will be latched in the RX\_PORT\_STS2 register 0x4E[5]. An interrupt may also be generated based on assertion of the encoded error flag. To detect FPD-Link III Encoder errors, the LINK\_ERROR\_COUNT must be enabled with a

LINK\_ERR\_THRESH value greater than 1. Otherwise, the loss of Receiver Lock will prevent detection of the Encoder error. The FPD3\_ENC\_ERROR flag is cleared on read.

When partnered with a DS90UB63x-Q1, the FPD3 Encoder may be configured to include a CRC check of the FPD3 encoder sequence. The CRC check provides an extra layer of error checking on the encoder sequence. This CRC checking adds protection to the encoder sequence used to send link information comprised of Datapath Control (registers 0x59 and 0x5A), Sensor Status (registers 0x51-0x54), and Serializer ID (register 0x5B). TI recommends enabling the CRC error checking on the FPD3 Encoder sequence to prevent any updates of link information values from encoded packets that do not pass CRC check. The FPD3 Encoder CRC is enabled by setting the FPD3\_ENC\_CRC\_DIS (register 0xBA[7] 表 7-137) to 0. In addition, the FPD3\_ENC\_CRC\_CAP flag should be set in register 0x4A[4] (see セクション 7.7.57).

#### 7.4.10.3 RX Port Input Signal Detection

The DS90UB638-Q1 can detect and measure the approximate input frequency and frequency stability of the RX input port and indicate status in bits [2:1] of RX\_PORT\_STS2. Frequency measurement stable `FREQ_STABLE` indicates the FPD-Link III input clock frequency is stable. When no FPD-Link III input clock is detected at the RX input port the `NO_FPD3_CLK` bit indicates that condition has occurred. Setting of these error flags is dependent on the stability control settings in the `FREQ_DET_CTL` register 0x77. The `NO_FPD3_CLK` bit will be set if the input frequency is below the setting programmed in the `FREQ_LO_THR` setting in the `FREQ_DET_CTL` register. A change in frequency `FREQ_STABLE = 0`, is defined as any change in MHz greater than the value programmed in the `FREQ_HYST` value. The frequency is continually monitored and provided for readback through the I2C interface less than every 1 ms. A 16-bit value is used to provide the frequency in units of 2 to 8 MHz. An interrupt can also be generated for the RX ports to indicate if a change in frequency is detected on the port.

#### 7.4.10.4 Line Counter

For each video frame received, the deserializer will count the number of video lines in the frame. In CSI-2 input mode, any long packet will be counted as a video line. The `LINE_COUNT_1` and `LINE_COUNT_0` registers in 0x73 and 0x74 can be used to read the line count for the most recent video frame. An interrupt may be enabled based on a change in the `LINE_COUNT` value. If interrupts are enabled, the `LINE_COUNT` registers will be latched at the interrupt and held until read back by the processor through I2C.

#### 7.4.10.5 Line Length

For each video line, the length (in bytes) will be determined. The `LINE_LEN_1` and `LINE_LEN_0` registers 0x75 and 0x76 can be used to read the line count for the most recent video frame. If the line length is not stable throughout the frame, the length of the last line of the frame will be reported. Line Count may not be consistent when receiving multiple CSI-2 video streams differentiated by VC-ID. An interrupt may be enabled based on a change in the `LINE_LEN` value. If interrupts are enabled, the `LINE_LEN` registers will be latched at the interrupt and held until read by the processor through I2C.

#### 7.4.11 Sensor Status

When paired with the DS90UB63x-Q1 CSI-2 serializer, the DS90UB638-Q1 is capable of receiving diagnostic indicators from the serializer. The sensor alarm and status diagnostic information are reported in the `SENSOR_STS_X` registers (0x51 to 0x54 in 表 7-80). The interrupt capability from detected status changes in sensor are described in セクション 7.5.7.2.2. Sensor Status This interrupt condition will be cleared by reading the `SEN_INT_RISE_STS` and `SEN_INT_FALL_STS` registers.

#### 7.4.12 GPIO Support

In addition to the dedicated LOCK and PASS output pins, the DS90UB638-Q1 supports seven pins, GPIO0 through GPIO6, which can be monitored, configured, and controlled through I2C in registers 0x0E - 0x16. GPIO3 programmable I/O pin is an active-low open drain and is shared with INTB. The current status of all GPIO can be readback from register 0x0E. Each GPIO is programmable for multiple uses options through the `GPIOx_PIN_CTL` registers 0x10 - 0x16.

### 7.4.12.1 GPIO Input Control and Status

Upon initialization GPIO0 through GPIO6 are enabled as inputs by default. Each GPIO pin has an input disable and a pulldown disable control bit, with the exception of GPIO3 which is open drain. By default, the GPIO pin input paths are enabled and the internal pulldown circuit for the GPIO is enabled. The GPIO\_INPUT\_CTL (0x0F) and GPIO\_PD\_CTL (0xBE) registers allow control of the input enable and the pulldown, respectively. For example, to disable GPIO1 and GPIO2 as inputs the user would program in register 0x0F[2:1] = 11. For most applications, there is no need to modify the default register settings for the pulldown resistors. The status HIGH or LOW of each GPIO pin 0 through 6 may be read through the GPIO\_PIN\_STS register 0x0E. This register read operation provides the status of the GPIO pin independent of whether the GPIO pin is configured as an input or output.

### 7.4.12.2 GPIO Output Pin Control

Individual GPIO output pin control is programmable through the GPIOx\_PIN\_CTL registers 0x10 to 0x16 (表 7-33). To enable any of the GPIO as output, set bit 0 = 1 in the respective register 0x10 to 0x16 after clearing the corresponding input enable bit in register 0x0F (表 7-32). The configuration register for each GPIO is listed in 表 7-6.

图 7-4. GPIOx Register Content (0x10 - 0x16)

7	6	5	4	3	2	1	0
GPIOX_OUTPUT_SEL[2:0]			GPIOX_OUT_SRC[2:0]			GPIOX_OUT_VAL	GPIOX_OUT_EN

表 7-6. GPIOx Output Function Programming

GPIO OUTPUT FUNCTION	GPIOX OUTPUT SOURCE SELECT GPIOX_OUT_SRC[2:0]		GPIOX OUTPUT FUNCTION SELECT GPIOX_OUTPUT_SEL[2:0]	GPIOX OUTPUT VALUE (GPIOX_OUT_VAL)	GPIO OUTPUT ENABLE (GPIOX_OUT_EN)
	VALUE	OUTPUT SIGNAL SOURCE			
GPIOX output disabled	X	No output. GPIO is Disabled or set to input mode	X	X	0
GPIOX linked to Forward channel received GPIO0 from RX Port 0 Serializer	000	RX Port 0	000	X	1
GPIOX linked to Forward channel received GPIO1 from RX Port 0 Serializer			001	X	1
GPIOX linked to Forward channel received GPIO2 from RX Port 0 Serializer			010	X	1
GPIOX linked to Forward channel received GPIO3 from RX Port 0 Serializer			011	X	1
RX Port 0 Lock indication			100	X	1
RX Port 0 Pass indication			101	X	1
RX Port 0 Frame Valid signal			110	X	1
RX Port 0 Line Valid signal			111	X	1
Reserved	010	Reserved	X	X	X
Set GPIOX = LOW value programmed by register	100	Device Status	000	0	1
Set GPIOX = HIGH value programmed by register			000	1	1
Lock indication from enabled RX port			001	X	1
RESERVED			010	X	1
RESERVED			011	X	1
FrameSync signal (internal or external)			100	X	1
Device interrupt active high			101	X	1
Device interrupt active low			110	X	1
Reserved	100	Reserved	111	X	X



**表 7-6. GPIOx Output Function Programming (continued)**

GPIO OUTPUT FUNCTION	GPIOX OUTPUT SOURCE SELECT GPIOX_OUT_SRC[2:0]		GPIOX OUTPUT FUNCTION SELECT GPIOX_OUTPUT_SEL[2:0]	GPIOX OUTPUT VALUE (GPIOX_OUT_VAL)	GPIO OUTPUT ENABLE (GPIOX_OUT_EN)
	VALUE	OUTPUT SIGNAL SOURCE			
Reserved	101	CSI-2 Tx Port	000	X	1
Pass			001	X	1
Frame Valid signal corresponding to video frame recovered at deserializer			010	X	1
Line Valid signal corresponding to video frame recovered at deserializer			011	X	1
Reserved			100	X	1
CSI-2 TX Port Interrupt active high			101	X	1
Reserved	101	Reserved	110	X	X
Reserved	101	Reserved	111	X	X
Reserved	110	Reserved	X	X	X
Reserved	111	Reserved	X	X	X

#### 7.4.12.3 Forward Channel GPIO

The DS90UB638-Q1 seven GPIO pins can output data received from the forward channel when paired with the DS90UB63x-Q1 CSI-2 serializer. The remote Serializer GPIO are mapped to a Deserializer GPIO. Each GPIO pin can be programmed for output mode and mapped. Up to four GPIOs are supported in the forward direction on the FPD-Link III Receive port (see 表 7-87). Each forward channel GPIO can be mapped to any GPIO output pin.

The timing for the forward channel GPIO is dependant on the number of GPIOs assigned at the serializer. When a single GPIO input from the DS90UB63x-Q1 CSI-2 serializer is linked to a DS90UB638-Q1 deserializer, the GPIO output value is sampled every forward channel transmit frame. Two linked GPIO are sampled every two forward channel frames and three or four linked GPIO are sampled every five frames. The typical minimum latency for the GPIO remains consistent (approximately 200 ns), but as the information gets spread over multiple frames, the jitter is typically increased on the order of the sampling period (number of forward channel frames). TI recommends maintaining a 4x over-sampling ratio for linked GPIO throughput. For example, when operating at -Gbps with REFCLK = 25 MHz, the maximum recommended GPIO input frequency based on the number of GPIO linked over the forward channel is shown in 表 7-7.

**表 7-7. Forward Channel GPIO Typical Timing**

NUMBER OF LINKED FORWARD CHANNEL GPIOs (FC_GPIO_EN 表 7-87)	SAMPLING FREQUENCY (MHz) AT FPD-Link III LINE RATE = Gbps	MAXIMUM RECOMMENDED FORWARD CHANNEL GPIO FREQUENCY (MHz)	TYPICAL JITTER (ns)
1	121.25	30.31	12
2	60.63	15.15	24
4	30.31	7.57	60

In addition to mapping remote serializer GPI, an internally generated FrameSync (see セクション 7.4.25) or other control signals may be output from any of the deserializer GPIOs for synchronization with a local processor or another deserializer.

#### 7.4.12.4 Back Channel GPIO

Each DS90UB638-Q1 GPIO pin defaults to input mode at start-up. The deserializer can link GPIO pin input data on up to four available slots to send on the back channel per each remote serializer connection. Any of the seven GPIO pin data can be mapped to send over the available back channel slots for each FPD-Link III Rx port. The same GPIO on the deserializer pin can be mapped to multiple back channel GPIO signals. For 10-Mbps back channel operation, the frame period is 3  $\mu$ s (30 bits  $\times$  100 ns/bit). As the back channel GPIOs are sampled

and sent each back channel frame by the DS90UB638-Q1 deserializer, the latency and jitter timing are each on the order of one back channel frame. The back channel GPIO is effectively sampled at a rate of 1/30 of the back channel rate or 333 kHz at  $f_{BC} = 10$  Mbps. TI recommends that the input to back channel GPIO switching frequency is  $< 1/4$  of the sampling rate or 83 kHz at  $f_{BC} = 10$  Mbps. For example, when operating at Gbps with REFCLK = 25 MHz, the maximum recommended GPIO input frequency based on the data rate when linked over the back channel is shown in 表 7-8.

**表 7-8. Back Channel GPIO Typical Timing**

BACK CHANNEL RATE (Mbps)	SAMPLING FREQUENCY (kHz)	MAXIMUM RECOMMENDED BACK CHANNEL GPIO FREQUENCY (kHz)	TYPICAL LATENCY (us)	TYPICAL JITTER (us)
10	334	83.5	3.2	3

In addition to sending GPIO from pins, an internally generated FrameSync or external FrameSync input signal may be mapped to any of the back channel GPIOs for synchronization of multiple sensors with extremely low skew. (see セクション 7.4.25).

For each port, GPIO control is available through the BC\_GPIO\_CTL0 register 0x6E (see 表 7-108) and BC\_GPIO\_CTL1 register 0x6F (see 表 7-109).

#### 7.4.12.5 Other GPIO Pin Controls

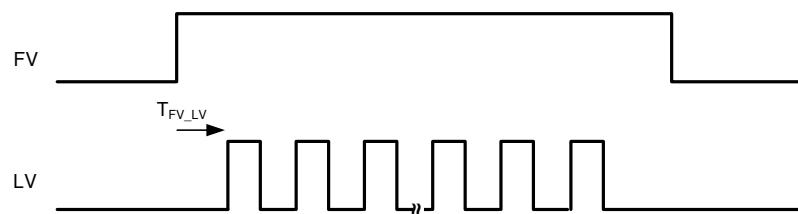
Each GPIO pin can has a input disable and a pulldown disable. By default, the GPIO pin input paths are enabled and the internal pulldown circuit in the GPIO is enabled. The GPIO\_INPUT\_CTL register 0x0F and GPIO\_PD\_CTL register 0xBE allow control of the input enable and the pulldown respectively. For most applications, there is no need to modify the default register settings.

#### 7.4.13 Line Valid and Frame Valid Indicators

The FrameValid (FV) and LineValid (LV) indications from the Receive Port indicate approximate frame and line boundaries at the FPD-Link III Receiver input. These signals may not be accurate if the receiver is in CSI-2 input mode and multiple video streams are present at the Receive Port input. A common example of this scenario would be multiple Virtual Channel IDs received on a single port.

The DS90UB638-Q1 allows setting the polarity of these signals by register programming. The FV and LV polarity are controlled on a per-port basis and can be independently set in the PORT\_CONFIG2 register 0x7C.

To prevent false detection of FrameValid, FV must be asserted for a minimum number of clocks prior to first video line to be considered valid. The minimum FrameValid time is programmable in the FV\_MIN\_TIME register 0xBC.



**図 7-5. Minimum FV to LV**

For other settings of FV\_MIN\_TIME, the required FV to LV setup in Serializer PCLKs can be determined by:

Absolute Min + (FV\_MIN\_TIME × Conversion factor)



**表 7-9. Minimum FV to LV Setup Requirement (in RAW Mode Serializer FPD-Link III PCLKs)**

MODE	FV_MIN_TIME CONVERSION FACTOR	ABSOLUTE MIN (FV_MIN_TIME = 0)	DEFAULT (FV_MIN_TIME = 128)
RAW12 HF	1.5	3	195
RAW10	2	5	261

For other settings of FV\_MIN\_TIME, the required FV to LV setup in Serializer PCLKs can be determined by:

Absolute Min + (FV\_MIN\_TIME × Conversion factor)

#### 7.4.14 CSI-2 Protocol Layer

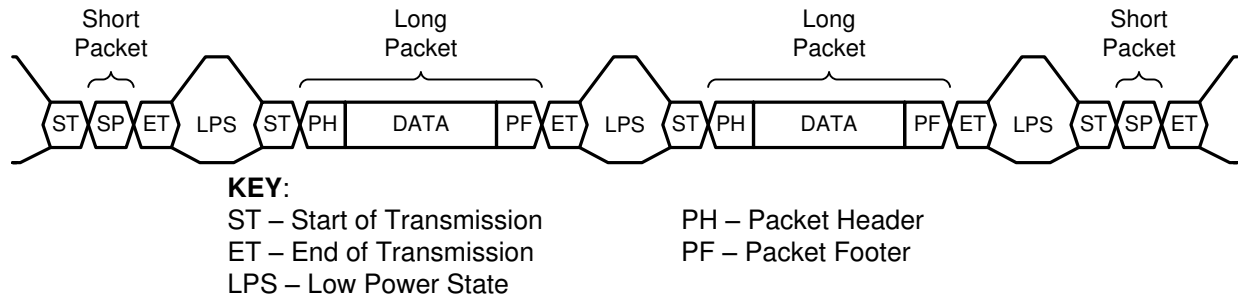
The DS90UB638-Q1 implements High-Speed mode to forward CSI-2 Low Level Protocol data. This includes features as described in the Low Level Protocol section of the MIPI CSI-2 Specification. It supports short and long packet formats.

The feature set of the protocol layer implemented by the CSI-2 TX is:

- Transport of arbitrary data (payload-independent)
- 8-bit word size
- Support for up to four interleaved virtual channels on the same link
- Special packets for frame start, frame end, line start and line end information
- Descriptor for the type, pixel depth and format of the Application Specific Payload data
- 16-bit Checksum Code for error detection

☒ 7-6 shows the CSI-2 protocol layer with short and long packets.

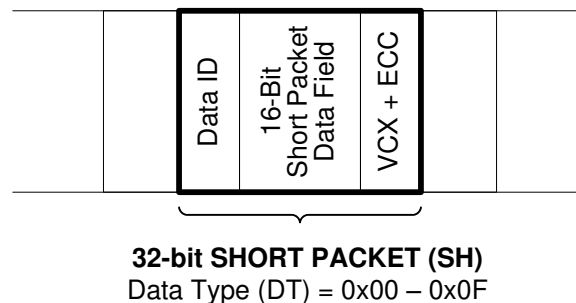
**DATA:**



**☒ 7-6. CSI-2 Protocol Layer With Short and Long Packets**

#### 7.4.15 CSI-2 Short Packet

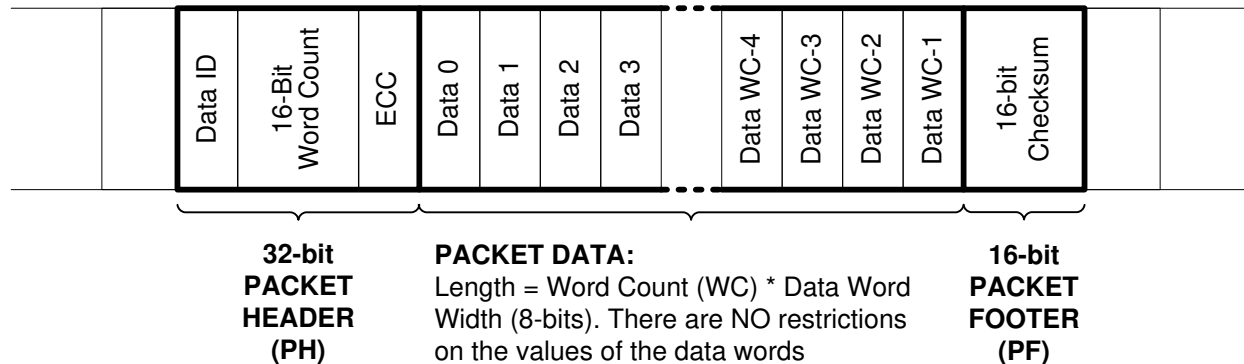
The short packet provides frame or line synchronization. ☒ 7-7 shows the structure of a short packet. A short packet is identified by data types 0x00 to 0x0F.



**☒ 7-7. CSI-2 Short Packet Structure**

### 7.4.16 CSI-2 Long Packet

A long packet consists of three elements: a 32-bit packet header (PH), an application-specific data payload with a variable number of 8-bit data words, and a 16-bit packet footer (PF). The packet header is further composed of three elements: an 8-bit data identifier, a 16-bit word count field, and an 8-bit ECC. The packet footer has one element, a 16-bit checksum. [Figure 7-8](#) shows the structure of a long packet.



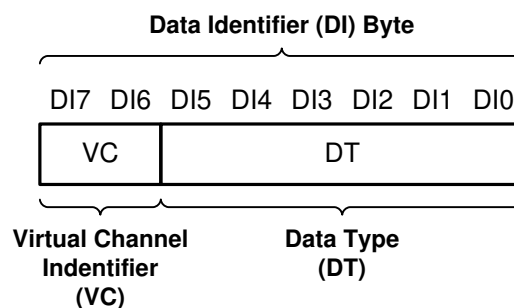
**Figure 7-8. CSI-2 Long Packet Structure**

**表 7-10. CSI-2 Long Packet Structure Description**

PACKET PART	FIELD NAME	SIZE (BIT)	DESCRIPTION
Header	VC / Data ID	8	Contains the virtual channel identifier and the data-type information.
	Word Count	16	Number of data words in the packet data. A word is 8 bits.
	ECC	8	ECC for data ID and WC field. Allows 1-bit error recovery and 2-bit error detection.
Data	Data	WC × 8	Application-specific payload (WC words of 8 bits).
Footer	Checksum	16	16-bit cyclic redundancy check (CRC) for packet data.

### 7.4.17 CSI-2 Data Type Identifier

The DS90UB638-Q1 MIPI CSI-2 protocol interface transmits the data identifier byte containing the values for the virtual channel ID (VC) and data type (DT) for the application specific payload data, as shown in [Figure 7-9](#). The virtual channel ID is contained in the 2 MSBs of the data identifier byte and identify the data as directed to one of four virtual channels. The value of the data type is contained in the six LSBs of the data identifier byte. When partnered with DS90UB63x-Q1 CSI-2 serializer, the Data Type is passed through from the received CSI-2 packets.



**Figure 7-9. CSI-2 Data Identifier Structure**

### 7.4.18 Virtual Channel and Context

The CSI-2 protocol layer transports virtual channels. Each virtual channel is identified by a unique channel identification number in the packet header. Therefore, a CSI-2 TX context can be associated with a virtual

channel and a data type. Virtual channels are defined by a 2-bit field. This channel identification number is encoded in the 2-bit code.

The CSI-2 TX transmits the channel identifier number. The CSI-2 TX supports up to four concurrent virtual channels.

#### 7.4.19 CSI-2 Transmitter Frequency

The CSI-2 Transmitters may operate nominally at 400 or 800 Mbps, or 1.6 Gbps. This operation is controlled through the CSI\_PLL\_CTL 0x1F register (see 表 7-48). The actual CSI-2 rate is proportional to the REFCLK frequency.

**表 7-11. Net CSI-2 Bandwidth Options**

CSI_PLL_CTL[1:0]	CSI-2 TX DATA RATE PER LANE (Mbps)	REFCLK FREQUENCY (MHz)	NET CSI-2 VIDEO BANDWIDTH (Gbps)
00	1664	26	3.328
	1600	25	3.2
	1472	23	2.944
01	Reserved	Reserved	Reserved
10	800	25	1.6
11	400	25	0.8

When configuring to 800 Mbps or 1.6 Gbps, the CSI-2 timing parameters are automatically set based on the CSI\_PLL\_CTL 0x1F register. In the case of alternate settings, the respective CSI-2 timing parameters registers must be programmed, and the appropriate override bit must be set. For the 1.664-Gbps and 1.472-Gbps options, these settings will also affect internal device timing for back channel operation, I2C, Bidirectional Control Channel, and FrameSync operation which scale with the REFCLK frequency.

To operate CSI-2 at speed of 400-Mbps mode, set CSI\_PLL\_CTL to 11b (0x1F[1:0] =11) to enable 400-Mbps operation for the CSI-2 Transmitters. Internal PLL and Timers are then automatically adjusted for the reduced reference clock frequency. Software control of CSI-2 Transmitter timing registers is required to provide proper interface timing on the CSI-2 Output. The following are the recommended timer settings for 400-Mbps operation.

```
# Set CSI-2 Timing parameters
WriteI2C(0xB0,0x2) # set auto-increment, page 0
WriteI2C(0xB1,0x40) # CSI-2 Port 0
WriteI2C(0xB2,0x83) # TCK Prep
WriteI2C(0xB2,0x8D) # TCK Zero
WriteI2C(0xB2,0x87) # TCK Trail
WriteI2C(0xB2,0x87) # TCK Post
WriteI2C(0xB2,0x83) # THS Prep
WriteI2C(0xB2,0x86) # THS Zero
WriteI2C(0xB2,0x84) # THS Trail
WriteI2C(0xB2,0x86) # THS Exit
WriteI2C(0xB2,0x84) # TLPX
```

#### 7.4.20 CSI-2 Replicate Mode

In CSI-2 Replicate mode, both ports can be programmed to output the same data. The output from CSI-2 port 0 is also presented on CSI-2 port 1.

To configure this mode of operation, set the CSI\_REPLICATE bit in the FWD\_CTL2 register (Address 0x21 in 表 7-50). Enabling replicate mode will automatically enable the second CSI-2 Clock output signal. The CSI-2 transmitter must be programmed for one or two lanes only through the CSI\_LANE\_COUNT field in the CSI\_CTL register as only one or two lanes are supported.

#### 7.4.21 CSI-2 Transmitter Output Control

Two register bits allow controlling the CSI-2 Transmitter output state. If the OUTPUT\_SLEEP\_STATE\_SELECT (OSS\_SEL) control is set to 0 in the GENERAL\_CFG 0x02 register (see 表 7-19), the CSI-2 Transmitter outputs

are forced to the HS-0 state. If the OUTPUT\_ENABLE (OEN) register bit is set to 0 in the GENERAL\_CFG register, the CSI-2 pins are set to the high-impedance state.

For normal operation (OSS\_SEL and OEN both set to 1), activity on either of the Rx Port determines the state of the CSI-2 outputs. The CSI-2 Pin State during FPD-Link III inactive includes two options, controlled by the OUTPUT\_EN\_MODE bit in the GENERAL\_CFG register and FWD\_PORTx\_DIS in the FWD\_CTL1 register 0x20. If OUTPUT\_EN\_MODE is set to 0, a lack of activity will force the outputs to Hi-Z condition. If OUTPUT\_EN\_MODE is set to 1, or if the forwarding for the Rx Port is disabled (FWD\_PORTx\_DIS = 1), the output enters LP-11 state as there is no data available to the CSI-2 Transmitter input. The FPD-Link III input is considered active if the Receiver indicates valid lock to the incoming signal. For a CSI-2 TX port, lock is considered valid if the Received port mapped to the TX port is indicating Lock. See section [セクション 7.4.6](#) for description of Rx port forwarding.

**表 7-12. CSI-2 Output Control Options**

PDB PIN	OSS_SEL	OEN	OUTPUT_O EN_MODE	FWD_PORTx_DIS	FPD-Link III INPUT	CSI-2 PIN STATE
0	X	X	X	X	X	Hi-Z
1	0	X	X	X	X	HS-0
1	1	0	X	X	X	Hi-Z
1	1	1	0	X	All inactive	Hi-Z
1	1	1	1	X	All inactive	LP-11
1	1	1	X	1	Any active	LP-11
1	1	1	X	0	Any active	Valid

#### 7.4.22 CSI-2 Transmitter Status

The status of the CSI-2 Transmitter may be monitored by readback of the CSI\_STS register 0x35, or brought to one of the configurable GPIO pins as an output. The TX\_PORT\_PASS 0x35[0] indicates valid CSI-2 data being presented on CSI-2 port. If no data is being forwarded or if error conditions have been detected on the video data, the CSI-2 Pass signal will be cleared. Interrupts may also be generated based on changes in the CSI-2 port status.

#### 7.4.23 Video Buffers

The DS90UB638-Q1 implements a video line buffer and FIFO. The video buffer provides storage of data payload and forward requirements for sending multiple video streams on the CSI-2 transmit ports. The total line buffer memory size is a 16-kB block.

The CSI-2 transmitter waits for an entire packet to be available before pulling data from the video buffer.

#### 7.4.24 CSI-2 Line Count and Line Length

The DS90UB638-Q1 counts the number of received lines (long packets) to determine line count on LINE\_COUNT\_1 and LINE\_COUNT\_0 registers 0x73–74. For received line length, DS90UB638-Q1 reads the number of bytes per line in LINE\_LEN\_1 and LINE\_LEN\_0 registers 0x75–0x76. Line Count and Line Length values are valid when receiving a single video stream. If multiple virtual channels are received on a FPD-Link III Receive port in CSI-2 input mode, the values in registers 0x73-74 may not be accurate

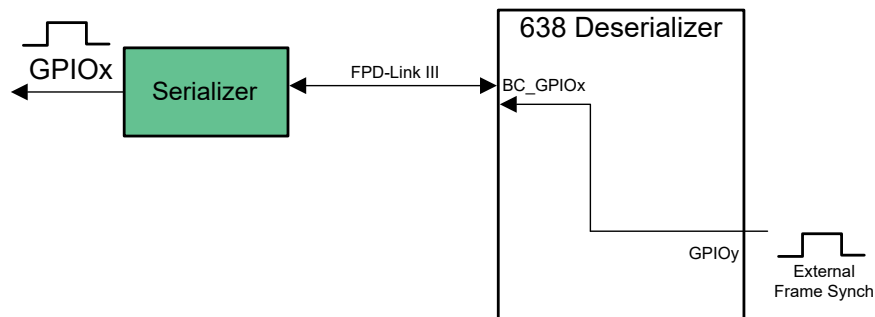
#### 7.4.25 FrameSync Operation

A frame synchronization signal (FrameSync) can be sent through the back channel using any of the back channel GPIOs. The signal can be generated in two different methods. The first option offers sending the external FrameSync using one of the available GPIO pins on the DS90UB638-Q1 and mapping that GPIO to a back channel GPIO on the FPD-Link III port.

The second option is to have the DS90UB638-Q1 internally generate a FrameSync signal to send through the back channel GPIO to the attached Serializer.

### 7.4.25.1 External FrameSync Control

In External FrameSync mode, an external signal is input to the DS90UB638-Q1 through one of the GPIO pins on the device. The external FrameSync signal may be propagated to one or more of the attached FPD-Link III Serializers through a GPIO signal in the back channel. The expected skew timing for external FrameSync mode is on the order of one back channel frame period or 3  $\mu$ s when operating at 10 Mbps.



 **7-10. External FrameSync**

Enabling the external FrameSync mode is done by setting the FS\_MODE control in the FS\_CTL register to a value between 0x8 (GPIO0 pin) to 0xF (GPIO7 pin). Set FS\_GEN\_ENABLE to 0 for this mode.

To send the FrameSync signal on a port's BC\_GPIOx signal, the BC\_GPIO\_CTL0 or BC\_GPIO\_CTL1 register should be programmed for that port to select the FrameSync signal.

### 7.4.25.2 Internally Generated FrameSync

In Internal FrameSync mode, an internally generated FrameSync signal is sent to the attached FPD-Link III Serializer through a GPIO signal in the back channel.

FrameSync operation is controlled by the FS\_CTL 0x18, FS\_HIGH\_TIME\_x, and FS\_LOW\_TIME\_x 0x19–0x1A registers. The resolution of the FrameSync generator clock (FS\_CLK\_PD) is derived from the back channel frame period (see BC\_FREQ\_SELECT[2:0] in 表 7-86). For example, each 50-Mbps back channel operation, the frame period is 600 ns (30 bits × 20 ns/bit), and for 2.5-Mbps back channel operation, the frame period is 12 μs (30 bits × 400 ns/bit).

Once enabled, the FrameSync signal is sent continuously based on the programmed conditions.

Enabling the internal FrameSync mode is done by setting the FS\_GEN\_ENABLE control in the FS\_CTL register to a value of 1. The FS\_MODE field controls the clock source used for the FrameSync generation. The FS\_GEN\_MODE field configures whether the duty cycle of the FrameSync is 50/50 or whether the high and low periods are controlled separately. The FrameSync high and low periods are controlled by the FS\_HIGH\_TIME and FS\_LOW\_TIME registers.

The accuracy of the internally generated FrameSync is directly dependent on the accuracy of the 25-MHz oscillator used as the reference clock and timing values should be scaled if reference other than 25 MHz is used.

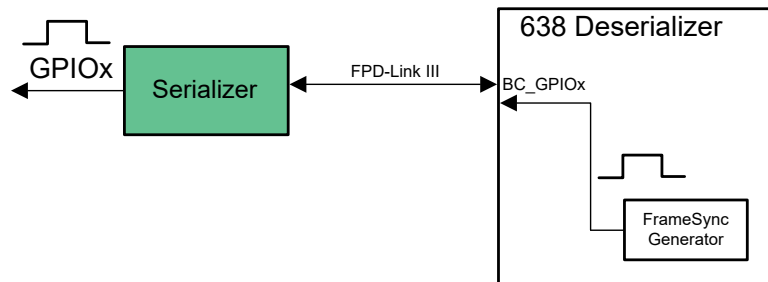


图 7-11. Internal FrameSync With DS90UB638 Deserializer

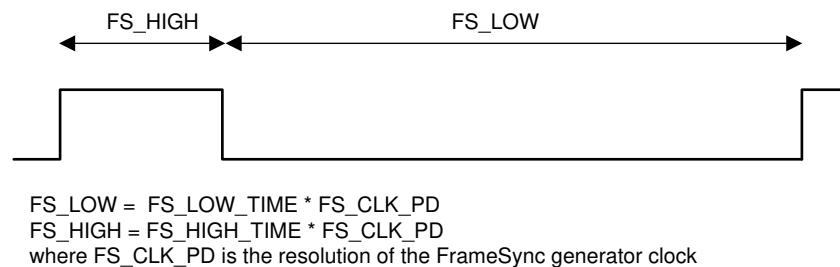


图 7-12. Internal FrameSync Signal

The following example shows generation of a FrameSync signal at 60 pulses per second. Mode settings:

- Programmable High/Low periods: FS\_GEN\_MODE 0x18[1]=0
- Use port 0 back channel frame period: FS\_MODE 0x18[7:4]=0x0
- Back channel rate of 50 Mbps: BC\_FREQ\_SELECT for port 0 0x58[2:0]=110b
- Initial FS state of 0: FS\_INIT\_STATE 0x18[2]=0

Based on mode settings, the FrameSync is generated based upon FS\_CLK\_PD of 600 ns.

The total period of the FrameSync is (1 / 60 hz) / 600 ns or approximately 27778 counts. The high time is programmed to 2778 and the low time is programmed to 25000.

For a 10% duty cycle, set the high time to 2777 (0x0AD9) cycles, and the low time to 24999 (0x61A7) cycles:

- FS\_HIGH\_TIME\_1: 0x19=0x0A
- FS\_HIGH\_TIME\_0: 0x1A=0xD9
- FS\_LOW\_TIME\_1: 0x1B=0x61
- FS\_LOW\_TIME\_0: 0x1C=0xA7

#### 7.4.25.2.1 Code Example for Internally Generated FrameSync

```
WriteI2C(0x4C,0x01) # RX0
WriteI2C(0x6E,0xAA) # BC_GPIO_CTL0: FrameSync signal to GPIO0/1
WriteI2C(0x10,0x91A) # FrameSync signal; Device Status; Enabled
WriteI2C(0x19,0x0A) # FS_HIGH_TIME_1
WriteI2C(0x1A,0xD9) # FS_HIGH_TIME_0
WriteI2C(0x1B,0x61) # FS_LOW_TIME_1
WriteI2C(0x1C,0xA7) # FS_LOW_TIME_0
WriteI2C(0x18,0x01) # Enable FrameSync
```

## 7.5 Programming

### 7.5.1 Serial Control Bus and Bidirectional Control Channel

The DS90UB638-Q1 implements an I2C-compatible serial control bus. The I2C is for local device configuration and incorporates a bidirectional control channel (BCC) that allows communication across the FPD-Link cable with remote serializers, as well as with remote I2C target devices. The DS90UB638-Q1 implements an I2C-compatible target that can be compliant with the standard, fast, and fast-plus modes of operation. This allows I2C operation at up to 1-MHz clock frequencies. When paired with a DS90UB63x-Q1 serializer, the DS90UB638-Q1 supports a combined format I2C read and write access. The timing for the I2C interface is shown in [Figure 6-4](#).

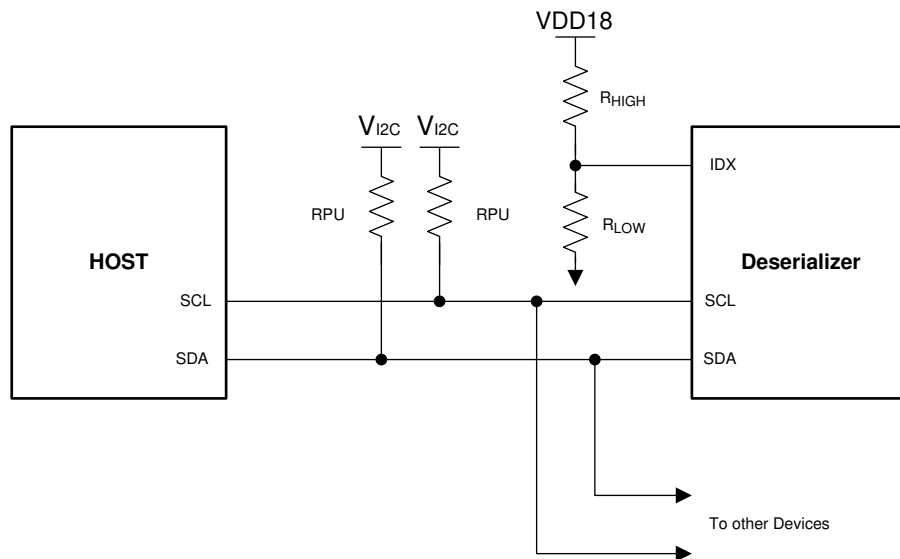
For accesses to local registers, the I2C target operates without stretching the clock. Accesses to remote devices over the bidirectional control channel results in clock stretching to allow for response time across the link. The DS90UB638-Q1 can also act as I2C controller for regenerating bidirectional control channel accesses originating from the remote devices across FPD-Link. Set I2C\_CONTROLLER\_EN in register 0x02[5] = 1 to enable the proxy controller functionality of the deserializer.

#### 7.5.1.1 Bidirectional Control

The bidirectional control channel (BCC) supports 10-Mbps operation when attached to the DS90UB63x-Q1. The bidirectional control channel is compatible with I2C devices, which allows local I2C target access to device registers as well as bidirectional I2C operation across the link to the serializer and attached devices. I2C access should not be attempted across the link when Rx Port Lock status is low. In addition to providing BCC operation, the back channel signaling also supports GPIO operations and advertising device capabilities to the attached serializer device. The default back channel frequency is selected by the strap setting of the MODE pin.

#### 7.5.1.2 Device Address

The primary device address is set through a resistor divider ( $R_{HIGH}$  and  $R_{LOW}$  — see [Serial Control Bus Connection](#) below) connected to the IDX pin. **Note that the voltage of  $V_{I2C}$  must match the voltage of  $V_{DDIO}$ .** The DS90UB638-Q1 waits 1 ms after PDB goes high to allow time for power supply transients before the IDX value is sampled and the device is configured to set the I2C address. The primary I2C target address is stored in the I2C Device ID register at address 0x0. In addition to the primary I2C target address, the DS90UB638-Q1 may be programmed to respond to up to 2 other I2C addresses. The two RX Port ID addresses provide direct access to the Receive Port 0 registers, which means the user does not need to set the paging controls normally required to access the port registers. The I2C\_RX0\_ID register is located in register address 0xF8.



**图 7-13. Serial Control Bus Connection**

The IDX pin configures the control interface to one of eight possible device addresses. A pullup resistor and a pulldown resistor may be used to set the appropriate voltage ratio between the IDX input pin ( $V_{IDX}$ ) and  $V_{(VDD18)}$ , each ratio corresponding to a specific device address. See [Serial Control Bus Addresses for IDX](#) for more information.

**表 7-13. Serial Control Bus Addresses for IDX**

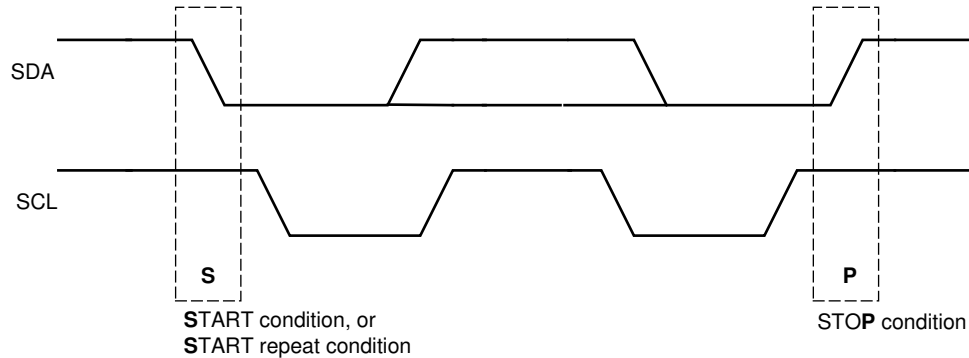
NO	$V_{IDX}$ VOLTAGE RANGE			$V_{IDX}$ TARGET VOLTAGE (V); $VDD1P8 = 1.8\text{ V}$	SUGGESTED STRAP RESISTORS (1% TOL)		PRIMARY ASSIGNED I2C ADDRESS	
	$V_{MIN}$	$V_{TYP}$	$V_{MAX}$		$R_{HIGH}$ (k $\Omega$ )	$R_{LOW}$ (k $\Omega$ )	7-BIT	8-BIT
0	0	0	$0.131 \times V_{(VDD18)}$	0	OPEN	10.0	0x30	0x60
1	$0.179 \times V_{(VDD18)}$	$0.213 \times V_{(VDD18)}$	$0.247 \times V_{(VDD18)}$	0.374	88.7	23.2	0x32	0x64
2	$0.296 \times V_{(VDD18)}$	$0.330 \times V_{(VDD18)}$	$0.362 \times V_{(VDD18)}$	0.582	75.0	35.7	0x34	0x68
3	$0.412 \times V_{(VDD18)}$	$0.443 \times V_{(VDD18)}$	$0.474 \times V_{(VDD18)}$	0.792	71.5	56.2	0x36	0x6C
4	$0.525 \times V_{(VDD18)}$	$0.559 \times V_{(VDD18)}$	$0.592 \times V_{(VDD18)}$	0.995	78.7	97.6	0x38	0x70
5	$0.642 \times V_{(VDD18)}$	$0.673 \times V_{(VDD18)}$	$0.704 \times V_{(VDD18)}$	1.202	39.2	78.7	0x3A	0x74
6	$0.761 \times V_{(VDD18)}$	$0.792 \times V_{(VDD18)}$	$0.823 \times V_{(VDD18)}$	1.420	25.5	95.3	0x3C	0x78
7	$0.876 \times V_{(VDD18)}$	$V_{(VDD18)}$	$V_{(VDD18)}$	1.8	10.0	OPEN	0x3D	0x7A

### 7.5.1.3 Basic I2C Serial Bus Operation

The serial control bus consists of two signals, SCL and SDA. SCL is a serial bus clock input, and SDA is the serial bus data input / output signal. Both SCL and SDA signals require an external pullup resistor to 1.8-V or 3.3-V nominal  $V_{(VDDIO)}$ . For most applications, TI recommends a 4.7-k $\Omega$  pullup resistor to  $V_{(VDDIO)}$ . However, the pullup resistor value may be adjusted for capacitive loading and data rate requirements. The signals are either pulled high or driven low.

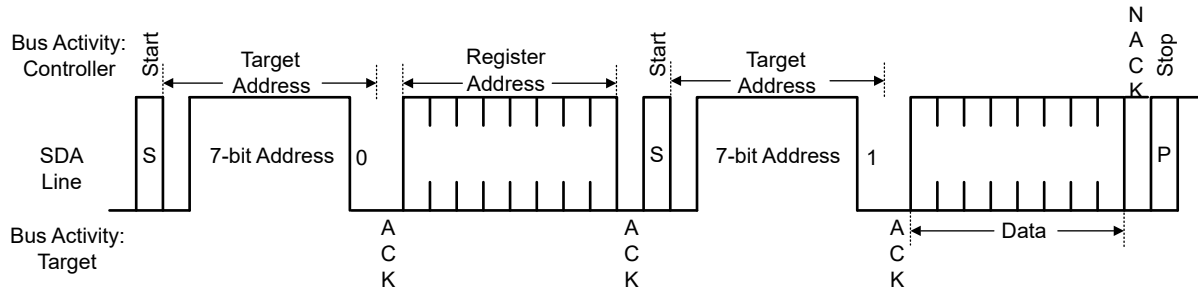


The serial bus protocol is controlled by START, START-Repeated, and STOP phases. A START occurs when SCL transitions low while SDA is high. A STOP occurs when SDA transitions high while SCL is also high. See [Figure 7-14](#).

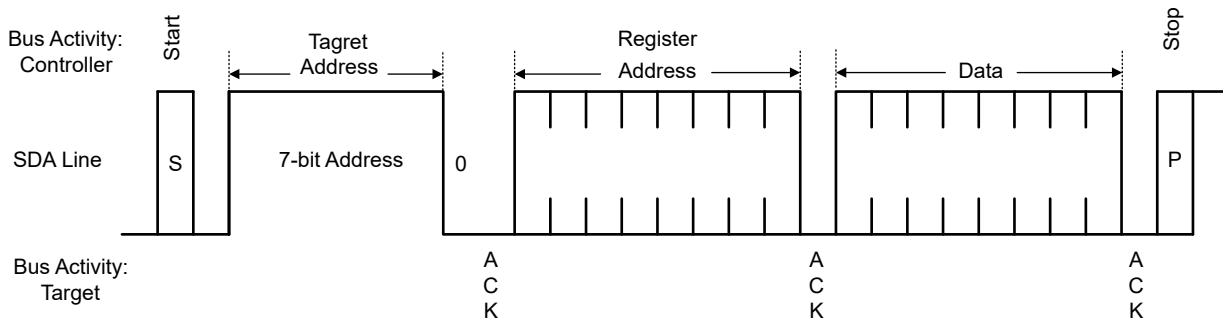


**Figure 7-14. START and STOP Conditions**

To communicate with a target device, the host controller (controller) sends the target address and listens for a response from the target. This response is referred to as an acknowledge bit (ACK). If a target on the bus is addressed correctly, it acknowledges (ACKs) the controller by driving the SDA bus low. If the address does not match the target address of the device, it not-acknowledges (NACKs) the controller by letting SDA be pulled high. ACKs also occur on the bus when data is being transmitted. When the controller is writing data, the target ACKs after every data byte is successfully received. When the controller is reading data, the controller ACKs after every data byte is received to let the target know that the controller wants to receive another data byte. When the controller wants to stop reading, it NACKs after the last data byte and creates a stop condition on the bus. All communication on the bus begins with either a START or START-Repeated condition. All communication on the bus ends with a STOP condition. A READ is shown in [Figure 7-15](#) and a WRITE is shown in [Figure 7-16](#).



**Figure 7-15. Serial Control Bus — READ**



**Figure 7-16. Serial Control Bus — WRITE**

For more information on I2C interface requirements and throughput considerations, refer to [I2C Communication Over FPD-Link III With Bidirectional Control Channel](#) (SNLA131A).

### 7.5.2 I2C Target Operation

The DS90UB638-Q1 implements an I2C-compatible target that can be compliant with the Standard, Fast, and Fast-plus modes of operation, which allows I2C operation at up to 1-MHz clock frequencies. Local I2C transactions to access DS90UB638-Q1 registers can be conducted 2 ms after power supplies are stable and PDB is brought high. For accesses to local registers, the I2C target operates without stretching the clock. The primary I2C target address is set through the IDx pin. The primary I2C target address is stored in the I2C Device ID register at address 0x0. In addition to the primary I2C target address, the DS90UB638-Q1 may be programmed to respond to up to two other I2C addresses. The RX Port ID address provides direct access to the Receive Port registers without the need to set the paging controls normally required to access the port registers.

### 7.5.3 Remote Target Operation

The bidirectional control channel provides a mechanism to read or write I2C registers in remote devices over the FPD-Link III interface. The I2C controller located at the deserializer must support I2C clock stretching. Accesses to serializer or remote target devices over the bidirectional control channel will result in clock stretching to allow for response time across the link. The DS90UB638-Q1 acts as an I2C target on the local bus, forwards read and write requests to the remote device, and returns the response from the remote device to the local I2C bus. To allow for the propagation and regeneration of the I2C transaction at the remote device, the DS90UB638-Q1 will stretch the I2C clock while waiting for the remote response. The I2C address of the serializer is populated in register 0x5B of the DS90UB638-Q1. The BCC\_CONFIG register 0x58 must also have bit 6, I2C\_PASS\_THROUGH, set to 1. If enabled, local I2C transactions with a valid address decode are then forwarded through the bidirectional control channel to the remote I2C bus. When I2C\_PASS\_THROUGH is set, the deserializer will only propagate messages that the deserializer recognizes, such as the registered serializer alias address (SER\_ALIAS), or the registered remote target alias attached to the serializer I2C bus (TARGET\_ALIAS). Setting PASS\_THROUGH\_ALL and AUTO\_ACK are less common use cases and are primarily used for debugging I2C messaging because they will respectively pass all addresses regardless of valid I2C address (PASS\_THROUGH\_ALL) and acknowledge all I2C commands without waiting for a response from serializer (AUTO\_ACK).

### 7.5.4 Remote Target Addressing

Eight pairs of TargetAlias and TargetID registers are allocated for the FPD-Link III Receive port in registers 0x5C through 0x6C. The TargetAlias register allows programming a virtual address which the host controller uses to access the remote device. The TargetID register provides the actual target address for the device on the remote I2C bus.

### 7.5.5 I2C Controller Proxy

The DS90UB638-Q1 implements an I2C controller that acts as a proxy controller to regenerate I2C accesses originating from a remote serializer. By default, the I2C Controller Enable bit (I2C\_CONTROLLER\_EN) in register 0x05[2] = 0 to block controller access to the local deserializer I2C from remote serializers. Set I2C\_CONTROLLER\_EN = 1 if system requires the deserializer to act as proxy controller for remote serializers on the local deserializer I2C bus. The proxy controller is an I2C-compatible controller, capable of operating with Standard-mode, Fast-mode, or Fast-mode Plus I2C timing. The deserializer is also capable of arbitration with other controllers, which allows multiple controllers and targets to exist on the I2C bus. A separate I2C proxy controller is implemented for each Receive port. This allows independent operation for all sources to the I2C interface. Arbitration between multiple sources is handled automatically using I2C multi-controller arbitration.

### 7.5.6 I2C Controller Proxy Timing

The proxy controller timing parameters are based on the REFCLK timing. Timing accuracy for the I2C proxy controller is based on the REFCLK or XTL clock source attached to the DS90UB638-Q1 deserializer. Before REFCLK is applied, the deserializer will default to an internal reference clock with 25-MHz  $\pm 10\%$  accuracy. The I2C controller regenerates the I2C read or write access using timing controls in the registers 0xA and 0xB to regenerate the clock and data signals to meet the desired I2C timing in Standard, Fast, or Fast-plus modes of operation.

I2C controller SCL high time is set in register 0x0A[7:0]. This field configures the high pulse width of the SCL output when the serializer is the controller on the local deserializer I2C bus. The default value is set to provide a minimum 5- $\mu$ s SCL high time with the reference clock at 25 MHz + 100 ppm, including four additional oscillator clock periods or synchronization and response time. Units are 40 ns for the nominal oscillator clock frequency, giving  $\text{Min\_delay} = 40 \text{ ns} \times (\text{SCL\_HIGH\_TIME} + 4)$ .

I2C controller SCL low time is set in register 0x0B[7:0]. This field configures the low pulse width of the SCL output when the serializer is the controller on the local deserializer I2C bus. This value is also used as the SDA setup time by the I2C target to provide data before SCL is released during accesses over the bidirectional control channel. The default value is set to provide a minimum 5- $\mu$ s SCL high time with the reference clock at 25 MHz + 100 ppm, including four additional oscillator clock periods or synchronization and response time. Units are 40 ns for the nominal oscillator clock frequency, giving  $\text{Min\_delay} = 40 \text{ ns} \times (\text{SCL\_HIGH\_TIME} + 4)$ . See [表 7-14](#) example settings for Standard mode, Fast mode, and Fast Mode Plus timing.

**表 7-14. Typical I2C Timing Register Settings**

I2C MODE	SCL HIGH TIME		SCL LOW TIME	
	0x0A[7:0]	NOMINAL DELAY AT REFCLK = 25 MHz	0x0B[7:0]	NOMINAL DELAY AT REFCLK = 25 MHz
Standard	0x7A	5.04 $\mu$ s	0x7A	5.04 $\mu$ s
Fast	0x13	0.920 $\mu$ s	0x25	1.64 $\mu$ s
Fast - Plus	0x06	0.400 s	0x0C	0.640 $\mu$ s

### 7.5.6.1 Code Example for Configuring Fast Mode Plus I2C Operation

```
# "RX0 I2C Controller Fast Plus Configuration"
WriteI2C(0x02,0x3E) # Enable Proxy
WriteI2C(0x4c,0x01) # Select RX_PORT0
# Set SCL High and Low Time delays
WriteI2C(0x0a,0x06) # SCL High
WriteI2C(0x0b,0x0C) # SCL Low
```

### 7.5.7 Interrupt Support

Interrupts can be brought out on the INTB pin as controlled by the INTERRUPT\_CTL 0x23 and INTERRUPT\_STS 0x24 registers. The main interrupt control registers provide control and status for interrupts from the individual sources. Sources include the FPD-Link III receive port as well as the CSI-2 transmit port. Clearing interrupt conditions requires reading the associated status register for the source. The setting of the individual interrupt status bits is not dependent on the related interrupt enable controls. The interrupt enable controls whether an interrupt is generated based on the condition, but does not prevent the interrupt status assertion.

The DS90UB638-Q1 devices have built-in flexibility such that the main interrupt may be brought to any GPIO pin through the GPIOx\_PIN\_CTL register for that pin (see 表 7-33). Note that the GPIO3 pin is the only GPIO that is implemented as open-drain, so this is the preferred pin for signaling the interrupt.

For an interrupt to be generated based on one of the interrupt status assertions, both the individual interrupt enable and the INT\_EN control must be set in the INTERRUPT\_CTL 0x23 register. For example, to generate an interrupt if IS\_RX0 is set, both the IE\_RX0 and INT\_EN bits must be set. If IE\_RX0 is set but INT\_EN is not, the INT status is indicated in the INTERRUPT\_STS register, and the INTB pin does not indicate the interrupt condition.

See the INTERRUPT\_CTL 0x23 and INTERRUPT\_STS 0x24 registers for details.

#### 7.5.7.1 Code Example to Enable Interrupts

```
# "RX0/1 INTERRUPT_CTL enable"
WriteI2C(0x23,0xBF) # RX all & INTB PIN EN
# Individual RX0/1 INTERRUPT_CTL enable
# "RX0 INTERRUPT_CTL enable"
WriteI2C(0x4C,0x01) # RX0
WriteI2C(0x23,0x81) # RX0 & INTB PIN EN
```

#### 7.5.7.2 FPD-Link III Receive Port Interrupts

For the FPD-Link III Receive port, multiple options are available for generating interrupts. Interrupt generation is controlled through the PORT\_ICR\_HI 0xD8 and PORT\_ICR\_LO 0xD9 registers. In addition, the PORT\_ISR\_HI 0xDA and PORT\_ISR\_LO 0xDB registers provide read-only status for the interrupts. Clearing of interrupt conditions is handled by reading the RX\_PORT\_STS1, RX\_PORT\_STS2, and CSI\_RX\_STS registers. The status bits in the PORT\_ISR\_HI/LO registers are copies of the associated bits in the main status registers.

To enable interrupts from one of the receive port interrupt sources:

1. Enable the interrupt source by setting the appropriate interrupt enable bit in the PORT\_ICR\_HI or PORT\_ICR\_LO register.
2. Set the RX Port X Interrupt control bit (IE\_RXx) in the INTERRUPT\_CTL register.
3. Set the INT\_EN bit in the INTERRUPT\_CTL register to allow the interrupt to assert the INTB pin low.

To clear interrupts from one of the receive port interrupt sources:

1. (optional) Read the INTERRUPT\_STS register to determine which RX Port caused the interrupt.
2. (optional) Read the PORT\_ISR\_HI and PORT\_ISR\_LO registers to determine source of interrupt.
3. Read the appropriate RX\_PORT\_STS1, RX\_PORT\_STS2, or CSI\_RX\_STS register to clear the interrupt.

The first two steps are optional. The interrupt can be determined and cleared by just reading the status registers.

#### **7.5.7.2.1 Interrupts on Forward Channel GPIO**

When connected to the DS90UB63x-Q1 CSI-2 serializer, interrupts can be generated on changes in any of the four forward channel GPIOs per port. Interrupts are enabled by setting bits in the FC\_GPIO\_ICR register. Interrupts may be generated on rising and/or falling transitions on the GPIO signal. The GPIO interrupt status is cleared by reading the FC\_GPIO\_STS register.

Interrupts should only be used for GPIO signals operating at less than 10 MHz. High or low pulses that are less than 100 ns might not be detected at the DS90UB638-Q1. To avoid false interrupt indications, the interrupts should not be enabled until after the forward channel GPIOs are enabled at the serializer.

#### **7.5.7.2.2 Interrupts on Change in Sensor Status**

The FPD-Link III receiver can recover 32 bits of sensor status from the attached DS90UB63x-Q1 CSI-2 serializer. Interrupts may be generated based on changes in the sensor status values received from the forward channel. The sensor status has 4 bytes of data, which may be read from the SENSOR\_STS\_x registers for each receive port. Interrupts may be generated based on a change in any of the bits in the first byte (SENSOR\_STS\_0). Each bit can be individually masked for rising and/or falling interrupts.

Two registers control the interrupt masks for the SENSOR\_STS bits: SEN\_INT\_RISE\_CTL and SEN\_INT\_FALL\_CTL.

Two registers provide interrupt status: SEN\_INT\_RISE\_STS, SEN\_INT\_FALL\_STS.

If a mask bit is set, a change in the associated SENSOR\_STS\_0 bit will be detected and latched in the SEN\_INT\_RISE\_STS or SEN\_INT\_FALL\_STS registers. If the mask bit is not set, the associated interrupt status bit will always be 0. If any of the SEN\_INT\_RISE\_STS or SEN\_INT\_FALL\_STS bits is set, the IS\_FC\_SEN\_STS bit will be set in the PORT\_ISR\_HI register.

### 7.5.7.3 Code Example to Readback Interrupts

```

INTERRUPT_STS = ReadI2C(0x24) # 0x24 INTERRUPT_STS
if ((INTERRUPT_STS & 0x80) >> 7):
    print "# GLOBAL INTERRUPT DETECTED "
if ((INTERRUPT_STS & 0x40) >> 6):
    print "# RESERVED "
if ((INTERRUPT_STS & 0x10) >> 4):
    print "# IS_CSI_TX DETECTED "
if ((INTERRUPT_STS & 0x01) ):
    print "# IS_RX0 DETECTED "
# "#####"
# "RX0 status"
# "#####"
WriteReg(0x4C,0x01) # RX0
PORT_ISR_LO = ReadI2C(0xDB)
print "0xDB PORT_ISR_LO : ", hex(PORT_ISR_LO) # readout; cleared by RX_PORT_STS2
if ((PORT_ISR_LO & 0x40) >> 6):
    print "# IS_LINE_LEN_CHG INTERRUPT DETECTED "
if ((PORT_ISR_LO & 0x20) >> 5):
    print "# IS_LINE_CNT_CHG DETECTED "
if ((PORT_ISR_LO & 0x10) >> 4):
    print "# IS_BUFFER_ERR DETECTED "
if ((PORT_ISR_LO & 0x08) >> 3):
    print "# IS_CSI_RX_ERR DETECTED "
if ((PORT_ISR_LO & 0x04) >> 2):
    print "# IS_FPD3_PAR_ERR DETECTED "
if ((PORT_ISR_LO & 0x02) >> 1):
    print "# IS_PORT_PASS DETECTED "
if ((PORT_ISR_LO & 0x01) ):
    print "# IS_LOCK_STS DETECTED "
#####
PORT_ISR_HI = ReadI2C(0xDA)
print "0xDA PORT_ISR_HI : ", hex(PORT_ISR_HI) # readout; cleared by RX_PORT_STS2
if ((PORT_ISR_HI & 0x04) >> 2):
    print "# IS_FPD3_ENC_ERR DETECTED "
if ((PORT_ISR_HI & 0x02) >> 1):
    print "# IS_BCC_SEQ_ERR DETECTED "
if ((PORT_ISR_HI & 0x01) ):
    print "# IS_BCC_CRC_ERR DETECTED "
#####
RX_PORT_STS1 = ReadI2C(0x4D) # R/COR
if ((RX_PORT_STS1 & 0xc0) >> 6) == 0:
    print "# RX_PORT_NUM = RX0"

if ((RX_PORT_STS1 & 0x20) >> 5):
    print "# BCC_CRC_ERR DETECTED "
if ((RX_PORT_STS1 & 0x10) >> 4):
    print "# LOCK_STS_CHG DETECTED "
if ((RX_PORT_STS1 & 0x08) >> 3):
    print "# BCC_SEQ_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x04) >> 2):
    print "# PARITY_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x02) >> 1):
    print "# PORT_PASS=1 "
if ((RX_PORT_STS1 & 0x01) ):
    print "# LOCK_STS=1 "
#####
RX_PORT_STS2 = ReadI2C(0x4E)
if ((RX_PORT_STS2 & 0x80) >> 7):
    print "# LINE_LEN UNSTABLE DETECTED "
if ((RX_PORT_STS2 & 0x40) >> 6):
    print "# LINE_LEN_CHG "
if ((RX_PORT_STS2 & 0x20) >> 5):
    print "# FPD3_ENCODE_ERROR DETECTED "
if ((RX_PORT_STS2 & 0x10) >> 4):
    print "# BUFFER_ERROR DETECTED "
if ((RX_PORT_STS2 & 0x08) >> 3):
    print "# CSI_ERR DETECTED "
if ((RX_PORT_STS2 & 0x04) >> 2):
    print "# FREQ_STABLE DETECTED "
if ((RX_PORT_STS2 & 0x02) >> 1):
    print "# NO_FPD3_CLK DETECTED "
if ((RX_PORT_STS2 & 0x01) ):

```

```
print "# LINE_CNT_CHG DETECTED "
#####
```

#### 7.5.7.4 CSI-2 Transmit Port Interrupts

The following interrupts are available for each CSI-2 Transmit Port:

- Pass indication
- Deassertion of Pass indication for an input port assigned to the CSI-2 TX port
- RX Port Interrupt – interrupts from the RX port mapped to this CSI-2 transmit port

See the CSI\_TX\_ICR address 0x36 and CSI\_TX\_ISR address 0x37 registers for details.

The setting of the individual interrupt status bits is not dependent on the related interrupt enable controls. The interrupt enable controls whether an interrupt is generated based on the condition, but does not prevent the interrupt status assertion.

#### 7.5.8 Error Handling

In the DS90UB638-Q1, the FPD-Link III receiver transfers incoming video frames to internal video buffers for forwarding to the CSI-2 Transmit ports. When the DS90UB638-Q1 detects an error condition, the standard operation would be to flag this error condition and truncate sending the CSI-2 frame to avoid sending corrupted data downstream. When the DS90UB638-Q1 recovers from an error condition, it will provide Start of Frame and resume sending valid data. Consequently, when the downstream CSI-2 input receives a repeated Start of Frame condition, this will indicate that the data received in between the prior start of frame is suspect, and the signal processor can then discard the suspected data. The settings in registers PORT\_CONFIG2 (0x7C) and PORT\_PASS\_CTL (0x7D) can be used to change how the DS90UB638-Q1 handles errors when passing video frames. The receive port may be configured to qualify the incoming video, providing a status indication and preventing the forward of video frames until certain error-free conditions are met. The Pass indication may be used to prevent forwarding packets to the internal video buffers by setting the PASS\_DISCARD\_EN bit in the PORT\_PASS\_CTL register. When this bit is set, video input will be discarded until the Pass signal indicates valid receive data. The receive port will indicate Pass status once specific conditions are met, including a number of valid frames received. Valid frames may include requiring no FPD-Link III parity errors and consistent frame size including video line length and/or number of video lines.

In addition, the receive port may be programmed to truncate video frames containing errors or prevent the forwarding of video until the Pass conditions are met. Register settings in PORT\_CONFIG2 register 0x7C can be used to truncate frames on different line/frame sizes or a CSI-2 parity error is detected. When the deserializer truncates frames in cases of different line/frame sizes different line/frame sizes, the video frame will stop immediately with no frame end packet. Often the condition will not be cleared until the next valid frame is received.

The Rx port PASS indication may be used to prevent forwarding packets to the internal video buffers by setting the PASS\_DISCARD\_EN bit in the PORT\_PASS\_CTL register 0x7D. When this bit is set, video input will be discarded until the Pass signal indicates valid receive data. The incoming video frames may be truncated based on error conditions or change in video line size or number of lines. These functions are controlled by bits in the PORT\_CONFIG2 register. When truncating video frames, the video frame may be truncated after sending any number of video lines. A truncated frame will not send a Frame End packet to the CSI-2 transmit port.

##### 7.5.8.1 Receive Frame Threshold

The FPD-Link III receiver may be programmed to require a specified number of valid video frames prior to indicating a Pass condition and forwarding video frames. The number of required valid video frames is programmable through the PASS\_THRESH field in the PORT\_PASS\_CTL register 0x7D (表 7-121). The threshold can be programmed from 0 to 3 video frames. If set to 0, Pass will typically be indicated as soon as the FPD-Link III receiver reports lock to the incoming signal. If set greater than 0, the receiver will require that number of valid frames before indicating Pass. Determination of valid frames will be dependent on the control bits in the PORT\_PASS\_CTL register. In the case of a parity error, when PASS\_PARITY\_ERR is set to 1, forwarding will be enabled one frame early. To ensure at least one good frame occurs following a parity error, the counter should be set to 2 or higher when PASS\_PARITY\_ERR = 1.



### 7.5.8.2 Port PASS Control

When the PASS\_LINE\_SIZE control is set in the PORT\_PASS\_CTL register, the receiver will qualify received frames based on having a consistent video line size. For PASS\_LINE\_SIZE to be clear, the deserializer checks that the received line length remains consistent during the frame and between frames. For each video line, the length (in bytes) will be determined. If it varies then we will flag this condition. Each video line in the packet must be the same size, and the line size must be consistent across video frames. A change in video line size will restart the valid frame counter.

When the PASS\_LINE\_CNT control is set in the PORT\_PASS\_CTL register, the receiver will qualify received frames based on having a consistent frame size in number of lines. A change in number of video lines will restart the valid frame counter.

When the PASS\_PARITY\_ERR control is set in the PORT\_PASS\_CTL register, the receiver will clear the Pass indication on receipt of a parity error on the FPD-Link III interface. The valid frame counter will also be cleared on the parity error event. When PASS\_PARITY\_ERR is set to 1, TI also recommends setting PASS\_THRESHOLD to 2 or higher to ensure at least one good frame occurs following a parity error.

### 7.5.9 Pattern Generation

The DS90UB638-Q1 supports an internal pattern generation feature to provide a simple way to generate video test patterns for the CSI-2 transmitter outputs. Two types of patterns are supported—the reference color bar pattern and fixed color pattern—accessed by the Pattern Generator page 0 in the indirect register set.

Prior to enabling the Pattern Generator, the following should be done:

1. Disable video forwarding by setting bits [5:4] of the FWD\_CTL1 register (that is, set register 0x20 to 0x30).
2. Configure CSI-2 Transmitter operating speed using the CSI\_PLL\_CTL register.
3. Enable the CSI-2 Transmitter for port 0 using the CSI\_CTL register.

#### 7.5.9.1 Reference Color Bar Pattern

The reference color bar patterns are based on the pattern defined in Appendix D of the mipi\_CTS\_for\_D-PHY\_v1-1\_r03 specification. The pattern is an eight color bar pattern designed to provide high, low, and medium frequency outputs on the CSI-2 transmit data lanes.

The CSI-2 reference pattern provides eight color bars by default with the following byte data for the color bars: X bytes of 0xAA (high-frequency pattern, inverted) X bytes of 0x33 (mid-frequency pattern) X bytes of 0xF0 (low-frequency pattern, inverted) X bytes of 0x7F (lone 0 pattern) X bytes of 0x55 (high-frequency pattern) X bytes of 0xCC (mid-frequency pattern, inverted) X bytes of 0x0F (low-frequency pattern) Y bytes of 0x80 (lone 1 pattern) In most cases, Y will be the same as X. For certain data types, the last color bar may need to be larger than the others to properly fill the video line dimensions.

The Pattern Generator is programmable with the following options:

- Number of color bars (1, 2, 4, or 8)
- Number of bytes per line
- Number of bytes per color bar
- CSI-2 DataType field and VC-ID
- Number of active video lines per frame
- Number of total lines per frame (active plus blanking)
- Line period (possibly program in units of 10 ns)
- Vertical front porch – number of blank lines prior to FrameEnd packet
- Vertical back porch – number of blank lines following FrameStart packet

The pattern generator relies on proper programming by software to ensure the color bar widths are set to multiples of the block (or word) size required for the specified DataType. For example, the block size is 3 bytes for RGB888, which also matches the pixel size. In this case, the number of bytes per color bar must be a multiple of 3. The Pattern Generator is implemented in the CSI-2 transmit clock domain, providing the pattern directly to the CSI-2 transmitter. The circuit generates the CSI-2 formatted data.



### 7.5.9.2 Fixed Color Patterns

When programmed for fixed color pattern mode, Pattern Generator can generate a video image with a programmable fixed data pattern. The basic programming fields used for image dimensions are the same ones used for the color bar patterns. When sending fixed color patterns, the color bar controls allow alternating between the fixed pattern data and the bit-wise inverse of the fixed pattern data.

The fixed color patterns assume a fixed block size for the byte pattern to be sent. The block size is programmable through the register and is designed to support most 8-bit, 10-bit, and 12-bit pixel formats. The block size should be set based on the pixel size converted to blocks that are an integer multiple of bytes. For example, an RGB888 pattern would consist of 3-byte pixels and therefore require a 3-byte block size. A 2x12-bit pixel image would also require 3-byte block size, while a 3x12-bit pixel image would require nine bytes (two pixels) to send an integer number of bytes.

The fixed color patterns support block sizes up to 16 bytes in length, allowing additional options for patterns in some conditions. For example, an RGB888 image could alternate between four different pixels by using a twelve-byte block size. An alternating black and white RGB888 image could be sent with a block size of 6-bytes and setting first three bytes to 0xFF and next three bytes to 0x00.

To support up to 16-byte block sizes, a set of sixteen registers are implemented to allow programming the value for each data byte. The line period is calculated in units of 10 ns, unless the CSI-2 mode is set to 400-Mb operation in which case the unit time dependency is 20 ns.

### 7.5.9.3 Packet Generator Programming

The information in this section provides details on how to program the Pattern Generator to provide a specific color bar pattern, based on datatype, frame size, and line size.

Most basic configuration information is determined directly from the expected video frame parameters. The requirements should include the datatype, frame rate (frames per second), number of active lines per frame, number of total lines per frame (active plus blanking), and number of pixels per line.

- PGEN\_ACT\_LPF – Number of active lines per frame
- PGEN\_TOT\_LPF – Number of total lines per frame
- PGEN\_LSIZE – Video line length size in bytes. Compute based on pixels per line multiplied by pixel size in bytes
- CSI-2 DataType field and VC-ID
- Optional: PGEN\_VBP – Vertical back porch. This is the number of lines of vertical blanking following Frame Valid
- Optional: PGEN\_VFP – Vertical front porch. This is the number of lines of vertical blanking preceding Frame Valid
- PGEN\_LINE\_PD – Line period in 10-ns units. Compute based on Frame Rate and total lines per frame
- PGEN\_BAR\_SIZE – Color bar size in bytes. Compute based on datatype and line length in bytes (see details below)

#### 7.5.9.3.1 Determining Color Bar Size

The color bar pattern should be programmed in units of a block or word size dependent on the datatype of the video being sent. The sizes are defined in the Mipi CSI-2 specification. For example, RGB888 requires a 3-byte block size which is the same as the pixel size.

When programming the Pattern Generator, software should compute the required bar size in bytes based on the line size and the number of bars. For the standard eight color bar pattern, that would require the following algorithm:

- Select the desired datatype, and a valid length for that datatype (in pixels).
- Convert pixels/line to blocks/line (by dividing by the number of pixels/block, as defined in the datatype specification).
- Divide the blocks/line result by the number of color bars (8), giving blocks/bar
- Round result down to the nearest integer

- Convert blocks/bar to bytes/bar and program that value into the PGEN\_BAR\_SIZE register

As an alternative, the blocks/line can be computed by converting pixels/line to bytes/line and divide by bytes/block.

#### 7.5.9.4 Code Example for Pattern Generator

```
#Patgen Fixed Colorbar 1280x720p30
WriteI2C(0x33,0x01) # CSI0 enable
WriteI2C(0xB0,0x00) # Indirect Pattern Gen Registers
WriteI2C(0xB1,0x01) # PGEN_CTL
WriteI2C(0xB2,0x01)
WriteI2C(0xB1,0x02) # PGEN_CFG
WriteI2C(0xB2,0x33)
WriteI2C(0xB1,0x03) # PGEN_CSI_DI
WriteI2C(0xB2,0x24)
WriteI2C(0xB1,0x04) # PGEN_LINE_SIZE1
WriteI2C(0xB2,0x0F)
WriteI2C(0xB1,0x05) # PGEN_LINE_SIZE0
WriteI2C(0xB2,0x00)
WriteI2C(0xB1,0x06) # PGEN_BAR_SIZE1
WriteI2C(0xB2,0x01)
WriteI2C(0xB1,0x07) # PGEN_BAR_SIZE0
WriteI2C(0xB2,0xE0)
WriteI2C(0xB1,0x08) # PGEN_ACT_LPF1
WriteI2C(0xB2,0x02)
WriteI2C(0xB1,0x09) # PGEN_ACT_LPF0
WriteI2C(0xB2,0xD0)
WriteI2C(0xB1,0x0A) # PGEN_TOT_LPF1
WriteI2C(0xB2,0x04)
WriteI2C(0xB1,0x0B) # PGEN_TOT_LPF0
WriteI2C(0xB2,0x1A)
WriteI2C(0xB1,0x0C) # PGEN_LINE_PD1
WriteI2C(0xB2,0x0C)
WriteI2C(0xB1,0x0D) # PGEN_LINE_PD0
WriteI2C(0xB2,0x67)
WriteI2C(0xB1,0x0E) # PGEN_VBP
WriteI2C(0xB2,0x21)
WriteI2C(0xB1,0x0F) # PGEN_VFP
WriteI2C(0xB2,0x0A)
```

#### 7.5.10 FPD-Link BIST Mode

An optional at-speed built-in self test (BIST) feature supports high-speed serial link and the back channel testing without external data connections. The BIST mode is enabled by either applying a logic high level to the BISTEN pin or by programming the BIST configuration register 0xB3. This is useful in the prototype stage, equipment production, in-system test, and system diagnostics.

When BIST is activated, the DS90UB638-Q1 sends register writes to the serializer through the back channel. The control channel register writes configure the serializer for BIST mode operation. The serializer outputs a continuous stream of a pseudo-random sequence and drives the link at speed. The deserializer detects the test pattern and monitors it for errors. The serializer also tracks errors indicated by the CRC fields in each back channel frame.

The LOCK, PASS and CMLOUT output functions are all available during BIST mode. While the lock indications are required to identify the beginning of proper data reception, for any link failures or data corruption, the best indication is the contents of the error counter in the BIST\_ERR\_COUNT register 0x57 for the RX port. The test may select whether the serializer uses an external or internal clock as reference for the BIST pattern frequency.

##### 7.5.10.1 BIST Operation Through BISTEN Pin

One method to enable BIST is by driving a logic high level on the BISTEN pin. During pin control BIST, the values on GPIO1 and GPIO0 pins will control whether the serializer uses an external or internal clock for the BIST pattern. The values on GPIO1 and GPIO0 will be written to the serializer register 0x14[2:1]. A value of 00 will select an external clock. A non-zero value will enable an internal clock of the frequency defined in the serializer register 0x14. The GPIO1 and GPIO0 values are sampled at the start of BIST (when BISTEN pin transitions to high). Changing this value after BIST is enabled will not change operation. Link BIST can also be enabled by register control through the BIST control register (address 0xB3)

### 7.5.10.2 BIST Operation Through Register Control

The FPD-Link III BIST is configured and enabled by programming the BIST control register (address 0xB3). BIST pass or fail status may be brought to GPIO pins by selecting the Pass indication for each receive port using the GPIOx\_PIN\_CTL registers. The Pass/Fail status will be deasserted low for each data error detected on the selected port input data. In addition, it is advisable to bring the receiver lock status for the port to the GPIO pins as well. After completion of BIST, the BIST error counter may be read to determine if errors occurred during the test. If the DS90UB638-Q1 failed to lock to the input signal or lost lock to the input signal, the BIST error counter will indicate 0xFF. The maximum normal count value will be 0xFE. The SER\_BIST\_ACT register bit 0xD0[5] can be monitored during testing to ensure BIST is activated in the serializer.

During BIST, DS90UB638-Q1 output activity are gated by BIST\_Control[7:6] (BIST\_OUT\_MODE[1:0]), as follows:

00 : Outputs disabled during BIST

10 : Outputs enabled during BIST

When enabling the outputs by setting BIST\_OUT\_MODE = 10, the CSI-2 will be inactive by default (LP11 state). To exercise the CSI-2 interface during BIST mode, it is possible to enable the Pattern Generator to send a video data pattern on the CSI-2 outputs.

The BIST clock frequency is controlled by the BIST\_CLOCK\_SOURCE field in the BIST Control register. This 2-bit value will be written to the serializer register 0x14[2:1]. A value of 00 will select an external clock. A non-zero value will enable an internal clock of the frequency defined in the serializer register 0x14. The BIST\_CLOCK\_SOURCE field is sampled at the start of BIST. Changing this value after BIST is enabled will not change operation.

## 7.6 Unique ID

Each device is programmed with a Unique DIE-ID that is burnt into devices at wafer level; Unique DIE-ID with a 16 bytes customer readable value indicating wafer lot and position of each IC inside a wafer. Combination of Unique DIE-IDs can be read and maintained by customer in a database or in a Hash table. Each system can be identified by the Unique DIE-ID programmed into the devices. Authenticity of the overall system can be established at the powerup/initialization or periodically by checking the Unique DIE-ID.

A Unique DIE-ID is programmed into each device and can be read using I<sup>2</sup>C reads. To read the Unique DIE-ID, set the IA\_SEL (0xB0[5:2]) register to Unique DIE ID Registers (1001), then set register IND\_ACC\_ADDR (0xB1) address to the Unique ID register being read, and then read the IND\_ACC\_DATA (0xB2) register to get the Unique DIE-ID. There are 16 Unique ID registers, each of the registers contain 8 bits of the total unique DIE-ID. The table below lists the Unique ID registers addresses.

**表 7-15. Unique ID Registers**

Unique ID register	IND_ACC_ADDR address
UNIQUE_ID_0	0x00
UNIQUE_ID_1	0x01
UNIQUE_ID_2	0x02
UNIQUE_ID_3	0x03
UNIQUE_ID_4	0x04
UNIQUE_ID_5	0x05
UNIQUE_ID_6	0x06
UNIQUE_ID_7	0x07
UNIQUE_ID_8	0x08
UNIQUE_ID_9	0x09
UNIQUE_ID_10	0x0A
UNIQUE_ID_11	0x0B
UNIQUE_ID_12	0x0C
UNIQUE_ID_13	0x0D
UNIQUE_ID_14	0x0E
UNIQUE_ID_15	0x0F

## 7.7 Register Maps

The DS90UB638-Q1 implements the following register blocks, accessible via I2C as well as the bi-directional control channel:

- Main Registers
- FPD-Link III RX Port Registers
- CSI-2 Port Registers (separate register block for each of the CSI-2 ports)

**表 7-16. Main Register Map Descriptions**

ADDRESS RANGE	DESCRIPTION	ADDRESS MAP
0x00-0x31	Digital Shared Registers	Shared
0x32-0x3A	Digital CSI-2 Tx Port Registers	Shared
0x3B - 0x4B	Reserved	Reserved
0x4C-0x7F	Digital RX Port Registers (paged, broadcast write allowed)	FPD3 RX Port 0 R: 0x4C[5:4]=00 W: 0x4C[0]=1
0x80-0x9F	Reserved	Reserved
0xA0-0xAF	Reserved	Reserved
0xB0-0xB2	Indirect Access Registers	Shared
0xB0-0xBF	Digital Share Registers	Shared
0xC0-0xCF	Reserved	Reserved
0xD0-0xDF	Digital RX Port Debug Registers	FPD3 RX Port 0
0xE0-0xEF	Reserved	Reserved
0xF0-0xF5	FPD3 RX ID	Shared
0xF8-0xFB	Port I2C Addressing	Shared
0xF6-0xF7 0xFC-0xFF	Reserved	Reserved

**LEGEND:**

- RW = Read Write
- RW/SC = RW/SC = Read Write access/Self Clearing bit
- R = Read Only, Permanent value
- R/COR = Read Only, Clear On Read

### 7.7.1 I2C Device ID Register

The I2C Device ID Register field always indicates the current value of the I2C ID. When bit 0 of this register is 0, this field is read-only and shows the strapped ID from device initialization after power on. When bit 0 of this register is 1, this field is read/write and can be used to assign any valid I2C ID address to the deserializer.

**表 7-17. I2C Device ID (Address 0x00)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	DEVICE_ID	RW	0x3D	7-bit I2C ID of Deserializer.
0	DES_ID	RW	0	0: Device ID is from strap 1: Register I2C Device ID overrides strapped value

### 7.7.2 Reset Register

The Reset register allows for soft digital reset of the DS90UB638-Q1 device internal circuitry without using PDB hardware analog reset. Digital Reset 0 is recommended if desired to reset without overwriting configuration registers to default values.

**表 7-18. Reset (Address 0x01)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:3	RESERVED	RW	0x00	Reserved
2	RESTART_AUTOLOAD	RW/SC	0	Restart Auto-load Setting this bit to 1 causes a re-load of the default settings including MODE and IDX. This bit is self-clearing. Software may check for Auto-load complete by checking the CFG_INIT_DONE bit in the DEVICE_STS register.
1	DIGITAL_RESET1	RW/SC	0	Digital Reset 1 Resets the entire digital block including registers. This bit is self-clearing. 1: Reset 0: Normal operation
0	DIGITAL_RESET0	RW/SC	0	Digital Reset 0 Resets the entire digital block except registers. This bit is self-clearing. 1: Reset 0: Normal operation

### 7.7.3 General Configuration Register

The general configuration register enables and disables high level block functionality.

**表 7-19. General Configuration (Address 0x02)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	RW	0	Reserved
5	I2C_CONTROLLER_ENABLE	RW	0	I2C Controller Enable. This bit must be set if system requires the deserializer to act as proxy controller for remote I2C access to the local I2C bus from remote serializers. 0: Block proxy Controller access to local I2C from remote serializers 1: Enable proxy Controller access to local I2C from remote serializers
4	OUTPUT_EN_MODE	RW	1	Output Enable Mode. If set to 0, the CSI TX output port will be forced to the high-impedance state the assigned RX port has an active Receiver lock. If set to 1 and no assigned RX ports have an active Receiver lock the CSI TX output port will continue in normal operation and enter the LP-11 state. CSI TX operation will remain under register control via the CSI_CTL register for each port.
3	OUTPUT_ENABLE	RW	1	Output Enable Control (usage dependant on Output Sleep State Select). If OUTPUT_SLEEP_STATE_SEL is set to 1 and OUTPUT_ENABLE is set to 0, the CSI TX outputs will be forced into a high impedance state.
2	OUTPUT_SLEEP_STATE_SELECT	RW	1	OSS Select to control output state when LOCK is low (usage dependant on Output Enable) When OUTPUT_SLEEP_STATE_SELECT is set to 0, the CSI TX outputs will be forced into a HS-0 state.
1	RX_PARITY_CHECKER_ENABLE	RW	1	FPD-Link III Parity Checker Enable 0: Disable 1: Enable
0	FORCE_REFCLK_DET	RW	0	Force indication of external reference clock 0: Normal operation, reference clock detect circuit indicates the presence of an external reference clock 1: Force reference clock to be indicated present

### 7.7.4 Revision/Mask ID Register

Revision ID field for production silicon version can be read back from this register.

**表 7-20. Revision/Mask ID (Address 0x03)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	REVISION_ID	R	0x2	Revision ID field (Note that 0x2 corresponds to the production device)
3:0	MASK_ID	R	0	Mask ID

### 7.7.5 DEVICE\_STS Register

Device status register provides read back access to high level link diagnostics.

**表 7-21. DEVICE\_STS (Address 0x04)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	CFG_CKSUM_STS	R	1	Configuration Checksum Passed. CFG_CKSUM_STS bit is set to one following initialization if the Configuration data had a valid checksum
6	CFG_INIT_DONE	R	1	Power-up initialization complete. CFG_INIT_DONE bit is set to one after Initialization is complete.
5	RESERVED	R	0	Reserved
4	REFCLK_VALID	R	0	REFCLK valid frequency bit indicates when a valid frequency has been detected on the REFCLK pin. 0 : Invalid frequency detected 1 : REFCLK frequency between 12MHz and 64MHz.
3	PASS	R	0	Device PASS status This bit indicates the PASS status for the device. The value in this register matches the indication on the PASS pin.
2	LOCK	R	0	Device LOCK status This bit indicates the LOCK status for the device. The value in this register matches the indication on the LOCK pin.
1:0	RESERVED	R	11	Reserved

### 7.7.6 PAR\_ERR\_THOLD\_HI Register

If the FPD-Link III receiver detects a number of parity errors greater than or equal to total value in PAR\_ERR\_THOLD[15:0], the PARITY\_ERROR flag is set in the RX\_PORT\_STS1 register. PAR\_ERR\_THOLD\_HI contains bits [15:8] of the 16 bit parity error threshold PAR\_ERR\_THOLD[15:0].

**表 7-22. PAR\_ERR\_THOLD\_HI (Address 0x05)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PAR_ERR_THOLD_HI	RW	0x01	FPD3 Parity Error Threshold High byte This register provides the 8 most significant bits [15:8] of the Parity Error Threshold value PAR_ERR_THOLD[15:0].

### 7.7.7 PAR\_ERR\_THOLD\_LO Register

If the FPD-Link III receiver detects a number of parity errors greater than or equal to total value in PAR\_ERR\_THOLD[15:0], the PARITY\_ERROR flag is set in the RX\_PORT\_STS1 register. PAR\_ERR\_THOLD\_LO contains bits [7:0] of the 16 bit parity error threshold PAR\_ERR\_THOLD[15:0].

**表 7-23. PAR\_ERR\_THOLD\_LO (Address 0x06)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PAR_ERR_THOLD_LO	RW	0	FPD3 Parity Error Threshold Low byte This register provides the 8 least significant bits [7:0] of the Parity Error Threshold value PAR_ERR_THOLD[15:0].

### 7.7.8 BCC Watchdog Control Register

The BCC watchdog timer allows termination of a control channel transaction if it fails to complete within a programmed amount of time.

**表 7-24. BCC Watchdog Control (Address 0x07)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	BCC_WATCHDOG_TIMER	RW	0x7F	Sets the Bidirectional Control Channel Watchdog Timeout value in units of 2 milliseconds. This field should not be set to 0.
0	BCC_WATCHDOG_TIMER_DISABLE	RW	0	Disable Bidirectional Control Channel Watchdog Timer 1: Disables BCC Watchdog Timer operation 0: Enables BCC Watchdog Timer operation

### 7.7.9 I2C Control 1 Register

**表 7-25. I2C Control 1 (Address 0x08)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	LOCAL_WRITE_DISABLE	RW	0	Disable Remote Writes to Local Registers Setting this bit to a 1 will prevent remote writes to local device registers from across the control channel. This prevents writes to the Deserializer registers from an I2C controller attached to the Serializer. Setting this bit does not affect remote access to I2C targets at the Deserializer.
6:4	I2C_SDA_HOLD	RW	0x1	Internal SDA Hold Time This field configures the amount of internal hold time provided for the SDA input relative to the SCL input. Units are 40 nanoseconds.
3:0	I2C_FILTER_DEPTH	RW	0xC	I2C Glitch Filter Depth This field configures the maximum width of glitch pulses on the SCL and SDA inputs that will be rejected. Units are 5 nanoseconds.

### 7.7.10 I2C Control 2 Register

**表 7-26. I2C Control 2 (Address 0x09)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	SDA_OUTPUT_SETUP	RW	1	Remote Ack SDA Output Setup When a Control Channel (remote) access is active, this field configures setup time from the SDA output relative to the rising edge of SCL during ACK cycles. Setting this value will increase setup time in units of 640ns. The nominal output setup time value for SDA to SCL are: 00 : 80ns 01: 720ns 10: 1400ns 11: 2080ns
3:2	SDA_OUTPUT_DELAY	RW	0	SDA Output Delay This field configures additional delay on the SDA output relative to the falling edge of SCL. Setting this value increases output delay in units of 40ns. Nominal output delay values for SCL to SDA are: 00 : 240ns 01: 280ns 10: 320ns 11: 360ns
1	I2C_BUS_TIMER_SPEEDUP	RW	0	Speed up I2C Bus Watchdog Timer 1: Watchdog Timer expires after approximately 50 microseconds 0: Watchdog Timer expires after approximately 1 second.



**表 7-26. I2C Control 2 (Address 0x09) (continued)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
0	I2C_BUS_TIMER_DISABLE	RW	0	Disable I2C Bus Watchdog Timer When enabled, the I2C Watchdog Timer may be used to detect when the I2C bus is free or hung up following an invalid termination of a transaction. If SDA is high and no signalling occurs for approximately 1 second, the I2C bus is assumed to be free. If SDA is low and no signaling occurs, the device will attempt to clear the bus by driving 9 clocks on SCL

### 7.7.11 SCL High Time Register

The SCL High Time register field configures the high pulse width of the I2C SCL output when the Serializer is the Controller on the local I2C bus. Units are 40 ns for the nominal oscillator clock frequency. The default value is set to approximately 100 kHz with the internal oscillator clock running at nominal 25 MHz. Delay includes 4 additional oscillator clock periods. The internal oscillator has  $\pm 10\%$  variation when REFCLK is not applied, which must be taken into account when setting the SCL High and Low Time registers.

**表 7-27. SCL High Time (Address 0x0A)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SCL_HIGH_TIME	RW	0x7A	I2C Controller SCL high time Default set to approximately 100 kHz when REFCLK = 25 MHz. Nominal High Time = 40 ns $\times$ (SCL HIGH TIME + 4)

### 7.7.12 SCL Low Time Register

The SCL Low Time register field configures the low pulse width of the SCL output when the serializer is the controller on the local I2C bus. This value is also used as the SDA setup time by the I2C Target for providing data prior to releasing SCL during accesses over the Bidirectional control channel. Units are 40 ns for the nominal oscillator clock frequency. The default value is set to approximately 100 kHz with the internal oscillator clock running at nominal 25 MHz. Delay includes 4 additional oscillator clock periods. The internal oscillator has  $\pm 10\%$  variation when REFCLK is not applied, which must be taken into account when setting the SCL High and Low Time registers.

**表 7-28. SCL Low Time (Address 0x0B)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SCL_LOW_TIME	RW	0x7A	I2C SCL low time Default set to approximately 100 kHz when REFCLK = 25 MHz. Nominal low time = 40 ns $\times$ (SCL LOW TIME + 4)

### 7.7.13 RX\_PORT\_CTL Register

Receiver port control register assigns rules for lock and pass in the general status register and allows for enabling and disabling the Rx port.

**表 7-29. RX\_PORT\_CTL (Address 0x0C)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	R	10	Reserved
5:4	PASS_SEL	RW	00	Pass Output Select This field controls the source of the PASS output. 00: Port 0 Receiver Pass This field can only be written via a local I2C controller.
3:2	LOCK_SEL	RW	0	Lock Output Select This field controls the source of the LOCK output. 00: Port 0 Receiver Lock This field can only be written via a local I2C controller.
1	RESERVED	RW	1	<b>Set to 0 at start-up. This is required for proper operation.</b>
0	PORT0_EN	RW	1	Port 0 Receiver Enable 0: Disable Port 0 Receiver 1: Enable Port 0 Receiver

## 7.7.14 IO\_CTL Register

**表 7-30. IO\_CTL (Address 0x0D)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	SEL3P3V	RW	0	3.3V I/O Select on I2C_SCL, I2C_SDA, and INTB pins. 0: 1.8V I/O Supply 1: 3.3V I/O Supply If IO_SUPPLY_MODE_OV is 0, a read of this register will return the detected I/O voltage level.
6	IO_SUPPLY_MODE_OV	RW	0	Override I/O Supply Mode bit 0: Detected I/O voltage level will be used for both SEL3P3V and IO_SUPPLY_MODE controls. 1: Register values written to the SEL3P3V and IO_SUPPLY_MODE fields will be used.
5:4	IO_SUPPLY_MODE	RW	0x0	I/O Supply Mode 00: 1.8V 01: Reserved 10: Reserved 11: 3.3V If IO_SUPPLY_MODE_OV is 0, a read of this register will return the detected I/O voltage level.
3:0	RESERVED	RW	0x9	Reserved

## 7.7.15 GPIO\_PIN\_STS Register

This register reads the current values on each of the 7 GPIO pins.

**表 7-31. GPIO\_PIN\_STS (Address 0x0E)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6:0	GPIO_STS	R	0x0	GPIO Pin High/ Low Status. Bit 6 reads GPIO6 and bit 0 reads GPIO0.

### 7.7.16 GPIO\_INPUT\_CTL Register

表 7-32. GPIO\_INPUT\_CTL (Address 0x0F)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	GPIO6_INPUT_EN	RW	1	GPIO6 Input Enable. Must be set to zero if GPIO6 is configured as an output by setting 0x16[0] = 1 0: Disabled 1: Enabled
5	GPIO5_INPUT_EN	RW	1	GPIO5 Input Enable. Must be set to zero if GPIO5 is configured as an output by setting 0x15[0] = 1 0: Disabled 1: Enabled
4	GPIO4_INPUT_EN	RW	1	GPIO4 Input Enable. Must be set to zero if GPIO4 is configured as an output by setting 0x14[0] = 1 0: Disabled 1: Enabled
3	GPIO3_INPUT_EN	RW	1	GPIO3 Input Enable. Must be set to zero if GPIO3 is configured as an output by setting 0x13[0] = 1 0: Disabled 1: Enabled
2	GPIO2_INPUT_EN	RW	1	GPIO2 Input Enable. Must be set to zero if GPIO2 is configured as an output by setting 0x12[0] = 1 0: Disabled 1: Enabled
1	GPIO1_INPUT_EN	RW	1	GPIO1 Input Enable. Must be set to zero if GPIO1 is configured as an output by setting 0x11[0] = 1 0: Disabled 1: Enabled
0	GPIO0_INPUT_EN	RW	1	GPIO0 Input Enable. Must be set to zero if GPIO0 is configured as an output by setting 0x10[0] = 1 0: Disabled 1: Enabled

### 7.7.17 GPIO0\_PIN\_CTL Register

表 7-33. GPIO0\_PIN\_CTL (Address 0x10)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO0_OUT_SEL	RW	0x0	GPIO0 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO0_OUT_SRC	RW	0x0	GPIO0 Output Source Select Selects output source for GPIO0 data: See <a href="#">表 7-6</a> .
1	GPIO0_OUT_VAL	RW	0	GPIO0 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO0_OUT_SRC[2:0] = 100 and GPIO0_OUT_SEL[2:0] = 000.
0	GPIO0_OUT_EN	RW	0	GPIO0 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[0] = 1 0: Disabled 1: Enabled

## 7.7.18 GPIO1\_PIN\_CTL Register

**表 7-34. GPIO1\_PIN\_CTL (Address 0x11)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO1_OUT_SEL	RW	0x0	GPIO1 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO1_OUT_SRC	RW	0x0	GPIO1 Output Source Select Selects output source for GPIO1 data: See <a href="#">表 7-6</a> .
1	GPIO1_OUT_VAL	RW	0	GPIO1 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO1_OUT_SRC[2:0] = 100 and GPIO1_OUT_SEL[2:0] = 000
0	GPIO1_OUT_EN	RW	0	GPIO1 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[1] = 1. 0: Disabled 1: Enabled

## 7.7.19 GPIO2\_PIN\_CTL Register

**表 7-35. GPIO2\_PIN\_CTL (Address 0x12)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO2_OUT_SEL	RW	0x0	GPIO2 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO2_OUT_SRC	RW	0x0	GPIO2 Output Source Select Selects output source for GPIO2 data: See <a href="#">表 7-6</a> .
1	GPIO2_OUT_VAL	RW	0	GPIO2 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO2_OUT_SRC[2:0] = 100 and GPIO2_OUT_SEL[2:0] = 00
0	GPIO2_OUT_EN	RW	0	GPIO2 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[2] = 1. 0: Disabled 1: Enabled

## 7.7.20 GPIO3\_PIN\_CTL Register

**表 7-36. GPIO3\_PIN\_CTL (Address 0x13)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO3_OUT_SEL	RW	0x0	GPIO3 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO3_OUT_SRC	RW	0x0	GPIO3 Output Source Select Selects output source for GPIO3 data: See <a href="#">表 7-6</a> .
1	GPIO3_OUT_VAL	RW	0	GPIO3 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO3_OUT_SRC[2:0] = 100 and GPIO3_OUT_SEL[2:0] = 000
0	GPIO3_OUT_EN	RW	0	GPIO3 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[3] = 1. 0: Disabled 1: Enabled

## 7.7.21 GPIO4\_PIN\_CTL Register

表 7-37. GPIO4\_PIN\_CTL (Address 0x14)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO4_OUT_SEL	RW	0x0	GPIO4 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO4_OUT_SRC	RW	0x0	GPIO4 Output Source Select Selects output source for GPIO4 data. See <a href="#">セクション 7.4.12.2</a> .
1	GPIO4_OUT_VAL	RW	0	GPIO4 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO4_OUT_SRC[2:0] = 100 and GPIO4_OUT_SEL[2:0] = 000
0	GPIO4_OUT_EN	RW	0	GPIO4 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[4] = 1. 0: Disabled 1: Enabled

## 7.7.22 GPIO5\_PIN\_CTL Register

表 7-38. GPIO5\_PIN\_CTL (Address 0x15)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO5_OUT_SEL	RW	0x0	GPIO5 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO5_OUT_SRC	RW	0x0	GPIO5 Output Source Select Selects output source for GPIO5 data: See <a href="#">表 7-6</a> .
1	GPIO5_OUT_VAL	RW	0	GPIO5 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO5_OUT_SRC[2:0] = 100 and GPIO5_OUT_SEL[2:0] = 00
0	GPIO5_OUT_EN	RW	0	GPIO5 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[5] = 1. 0: Disabled 1: Enabled

## 7.7.23 GPIO6\_PIN\_CTL Register

表 7-39. GPIO6\_PIN\_CTL (Address 0x16)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	GPIO6_OUT_SEL	RW	0x0	GPIO6 Output Select Determines the output data for the selected source. See <a href="#">セクション 7.4.12.2</a> .
4:2	GPIO6_OUT_SRC	RW	0x0	GPIO6 Output Source Select Selects output source for GPIO6 data: See <a href="#">表 7-6</a>
1	GPIO6_OUT_VAL	RW	0	GPIO6 Output Value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value by setting GPIO6_OUT_SRC[2:0] = 100 and GPIO6_OUT_SEL[2:0] = 00
0	GPIO6_OUT_EN	RW	0	GPIO6 Output Enable. Must be set to zero when configured as an input in GPIO Input Control register, 0x0F[6] = 1. 0: Disabled 1: Enabled

## 7.7.24 RESERVED Register

**表 7-40. RESERVED (Address 0x17)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x0	Reserved.

## 7.7.25 FS\_CTL Register

**表 7-41. FS\_CTL (Address 0x18)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	FS_MODE	RW	0	FrameSync Mode 0000: Internal Generated FrameSync, use back channel frame clock from port 0 0001: Reserved 0010: Reserved. 0011: Reserved 01xx: Internal Generated FrameSync, use 25MHz clock 1000: External FrameSync from GPIO0 1001: External FrameSync from GPIO1 1010: External FrameSync from GPIO2 1011: External FrameSync from GPIO3 1100: External FrameSync from GPIO4 1101: External FrameSync from GPIO5 1110: External FrameSync from GPIO6 1111: Reserved
3	FS_SINGLE	RW, SC	0	Generate Single FrameSync pulse When this bit is set, a single FrameSync pulse will be generated. The system should wait for the full duration of the desired pulse before generating another pulse. When using this feature, the FS_GEN_ENABLE bit should remain set to 0. This bit is self-clearing and will always return 0.
2	FS_INIT_STATE	RW	0	Initial State. This register controls the initial state of the FrameSync signal. 0: FrameSync initial state is 0 1: FrameSync initial state is 1
1	FS_GEN_MODE	RW	0	FrameSync Generation Mode This control selects between Hi/Lo and 50/50 modes. In Hi/Lo mode, the FrameSync generator uses the FS_HIGH_TIME [15:0] and FS_LOW_TIME [15:0] register values to separately control the High and Low periods for the generated FrameSync signal. In 50/50 mode, the FrameSync generator uses the values in the FS_HIGH_TIME_0, FS_LOW_TIME_1 and FS_LOW_TIME_0 registers as a 24-bit value for both the High and Low periods of the generated FrameSync signal. 0: Hi/Lo 1: 50/50
0	FS_GEN_ENABLE	RW	0	FrameSync Generation Enable 0: Disabled 1: Enabled

## 7.7.26 FS\_HIGH\_TIME\_1 Register

**表 7-42. FS\_HIGH\_TIME\_1 (Address 0x19)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FRAMESYNC_HIGH_TIME_1	RW	0	FrameSync High Time bits 15:8 The value programmed to the FS_HIGH_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_HIGH_TIME field will result in a 1 cycle high pulse on the FrameSync signal.

**7.7.27 FS\_HIGH\_TIME\_0 Register****表 7-43. FS\_HIGH\_TIME\_0 (Address 0x1A)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FRAMESYNC_HIGH_TIME_0	RW	0	FrameSync High Time bits 7:0 The value programmed to the FS_HIGH_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_HIGH_TIME field will result in a 1 cycle high pulse on the FrameSync signal.

**7.7.28 FS\_LOW\_TIME\_1 Register****表 7-44. FS\_LOW\_TIME\_1 (Address 0x1B)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FRAMESYNC_LOW_TIME_1	RW	0	FrameSync Low Time bits 15:8 The value programmed to the FS_LO_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_LO_TIME field will result in a 1 cycle high pulse on the FrameSync signal.

**7.7.29 FS\_LOW\_TIME\_0 Register****表 7-45. FS\_LOW\_TIME\_0 (Address 0x1C)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FRAMESYNC_LOW_TIME_0	RW	0	FrameSync Low Time bits 7:0 The value programmed to the FS_LO_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_LO_TIME field will result in a 1 cycle high pulse on the FrameSync signal.

**7.7.30 MAX\_FRM\_HI Register****表 7-46. MAX\_FRM\_HI (Address 0x1D)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	MAX_FRAME_HI	RW	0x00	CSI-2 Maximum Frame Count bits 15:8

**7.7.31 MAX\_FRM\_LO Register****表 7-47. MAX\_FRM\_LO (Address 0x1E)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	MAX_FRAME_LO	RW	0x04	CSI-2 Maximum Frame Count bits 7:0

**7.7.32 CSI\_PLL\_CTL Register****表 7-48. CSI\_PLL\_CTL (Address 0x1F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	RESERVED	R	0	Reserved
3:2	RESERVED	RW	0	Reserved
1:0	CSI_TX_SPEED	RW	10	CSI Transmitter Speed select: Controls the CSI Transmitter frequency. 00 : 1.6 Gbps serial rate 01 : Reserved 10 : 800 Mbps serial rate 11 : 400 Mbps serial rate

**7.7.33 FWD\_CTL1 Register**

Forwarding control enables or disables video stream from the Rx Port.



**表 7-49. FWD\_CTL1 (Address 0x20)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	RW	0x0	Reserved.
5	RESERVED	RW	1	Reserved.
4	FWD_PORT0_DIS	RW	1	Disable forwarding of RX Port 0 0: Forwarding enabled for RX Port 0 1: Forwarding disabled for RX Port 0
3:0	RESERVED	R	0x0	Reserved.

### 7.7.34 FWD\_CTL2 Register

**表 7-50. FWD\_CTL2 (Address 0x21)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	CSI_REPLICATE	RW	0	CSI Replicate Mode. When set to a 1, the CSI output from port 0 will also be generated on CSI port 1. In this mode, each CSI port may be one or two lanes only. The same output data will be presented on both ports.
6	FWD_SYNC_AS_AVAIL	RW	0	Set to 1 after start-up
5:4	RESERVED	R	0	Reserved.
3:2	RESERVED	RW	00	Reserved
1	RESERVED	RW	0	Reserved.
0	CSI0_RR_FWD	RW	1	Enable round robin forwarding for CSI TX output port. When this mode is enabled, no attempt is made to synchronize the video traffic. When multiple sources have data available to forward, the data will tend to be forwarded in a round-robin fashion. 0: Round robin forwarding disabled 1: Round robin forwarding enabled Only one of CSI0_RR_FWD and CSI0_SYNC_FWD must be set at a time.

### 7.7.35 FWD\_STS Register

表 7-51. FWD\_STS (Address 0x22)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0	Reserved

### 7.7.36 INTERRUPT\_CTL Register

表 7-52. INTERRUPT\_CTL (Address 0x23)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	INT_EN	RW	0	Global Interrupt Enable: Enables interrupt on the interrupt signal to the controller.
6:5	RESERVED	R	0	Reserved
4	IE_CSI_TX0	RW	0	CSI Transmit Port Interrupt: Enable interrupt from CSI Transmitter Port.
3:2	RESERVED	R	0	Reserved
1	RESERVED	RW	0	Reserved
0	IE_RX0	RW	0	RX Port 0 Interrupt: Enable interrupt from Receiver Port 0.

## 7.7.37 INTERRUPT\_STS Register

**表 7-53. INTERRUPT\_STS (Address 0x24)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	INTERRUPT_STS	R	0	Global Interrupt: Set if any enabled interrupt is indicated in the individual status bits in this register. The setting of this bit is not dependent on the INT_EN bit in the INTERRUPT_CTL register but does depend on the IE_xxx bits. For example, if IE_RX0 and IS_RX0 are both asserted, the INTERRUPT_STS bit is set to 1.
6:5	RESERVED	R	0	Reserved
4	IS_CSI_TX0	R	0	CSI Transmit Port Interrupt: An interrupt has occurred for CSI Transmitter Port 0. This interrupt is cleared upon reading the CSI_TX_ISR register for CSI Transmit Port.
3:2	RESERVED	R	0	Reserved
1	RESERVED	R	0	Reserved
0	IS_RX0	R	0	RX Port 0 Interrupt: An interrupt has occurred for Receive Port 0. This interrupt is cleared by reading the associated status register(s) for the event(s) that caused the interrupt. The status registers are RX_PORT_STS1, RX_PORT_STS2, and CSI_RX_STS.

## 7.7.38 RESERVED Register

**表 7-54. RESERVED (Address 0x25 – 0x32)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x00	Reserved

### 7.7.39 CSI\_CTL Register

**表 7-55. CSI\_CTL (Address 0x33)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	CSI_CAL_EN	RW	0	Enable initial CSI Skew-Calibration sequence When the initial skew-calibration sequence is enabled, the CSI Transmitter will send the sequence at initialization, prior to sending any HS data. This bit should be set when operating at 1.6 Gbps CSI speed (as configured in the CSI_PLL_CTL register). 0: Disabled 1: Enabled
5:4	CSI_LANE_COUNT	RW	0x0	CSI lane count 00: 4 lanes 01: 3 lanes 10: 2 lanes 11: 1 lane If CSI_REPLICATE is set in the FWD_CTL2 register, the device must be programmed for 1 or 2 lanes only.
3:2	CSI_ULP	RW	0	Force LP00 state on data/clock lanes 00: Normal operation 01: LP00 state forced only on data lanes 10: Reserved 11: LP00 state forced on data and clock lanes
1	CSI_CONTS_CLOCK	RW	0	Enable CSI continuous clock mode. CSI-2 Tx outputs will provide a continuous clock output signal once first packet is received. 0: Disabled 1: Enabled
0	CSI_ENABLE	RW	0	Enable CSI output 0: Disabled 1: Enabled

### 7.7.40 CSI\_CTL2 Register

**表 7-56. CSI\_CTL2 (Address 0x34)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	RESERVED	R	0x4	RESERVED
3	RESERVED	RW	0	RESERVED
2	CSI_CAL_INV	RW	0	CSI Calibration Inverted Data pattern During the CSI skew-calibration pattern, the CSI Transmitter will send a sequence of 01010101 data (first bit 0). Setting this bit to a 1 will invert the sequence to 10101010 data.
1	CSI_CAL_SINGLE	RW/SC	0	Enable single periodic CSI Skew-Calibration sequence Setting this bit will send a single skew-calibration sequence from the CSI Transmitter. The skew-calibration sequence is the 1010 bit sequence required for periodic calibration. The calibration sequence is sent at the next idle period on the CSI interface. This bit is self-clearing and will reset to 0 after the calibration sequence is sent.
0	CSI_CAL_PERIODIC	RW	0	Enable periodic CSI Skew-Calibration sequence When the periodic skew-calibration sequence is enabled, the CSI Transmitter will send the periodic skew-calibration sequence following the sending of Frame End packets. 0: Disabled 1: Enabled

## 7.7.41 CSI\_STS Register

**表 7-57. CSI\_STS (Address 0x35)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:2	RESERVED	R	0x0	Reserved
1	RESERVED	R	0	Reserved
0	TX_PORT_PASS	R	0	TX Port Pass The TX_PORT_PASS indicates the CSI port is actively delivering valid video data. The status is cleared based on detection of an error condition that interrupts transmission.

## 7.7.42 CSI\_TX\_ICR Register

**表 7-58. CSI\_TX\_ICR (Address 0x36)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	RESERVED	R	0x0	Reserved
4	IE_RX_PORT_INT	RW	0	RX Port Interrupt Enable Enable interrupt based on receiver port interrupt for the RX Port being forwarded to the CSI Transmit Port.
3	IE_CSI_SYNC_ERROR	RW	0	CSI Sync Error interrupt Enable Enable interrupt on CSI Synchronization enable.
2	IE_CSI_SYNC	RW	0	CSI Synchronized interrupt Enable Enable interrupts on CSI Transmit Port assertion of CSI Synchronized Status.
1	IE_CSI_PASS_ERROR	RW	0	CSI RX Pass Error interrupt Enable Enable interrupt on CSI Pass Error
0	IE_CSI_PASS	RW	0	CSI Pass interrupt Enable Enable interrupt on CSI Transmit Port assertion of CSI Pass.

### 7.7.43 CSI\_TX\_ISR Register

**表 7-59. CSI\_TX\_ISR (Address 0x37)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	RESERVED	R	0x0	Reserved
4	IS_RX_PORT_INT	R	0	RX Port Interrupt A Receiver port interrupt has been generated for the RX Ports being forwarded to the CSI Transmit Port. A read of the associated port receive status registers will clear this interrupt. See the PORT_ISR_HI and PORT_ISR_LO registers for details.
3	IS_CSI_SYNC_ERROR	R/COR	0	CSI Sync Error interrupt A synchronization error has been detected for multiple video stream inputs to the CSI Transmitter.
2	IS_CSI_SYNC	R/COR	0	CSI Synchronized interrupt CSI Transmit Port assertion of CSI Synchronized Status. Current status for CSI Sync can be read from the TX_PORT_SYNC flag in the CSI_STS register.
1	IS_CSI_PASS_ERROR	R/COR	0	CSI RX Pass Error interrupt A deassertion of CSI Pass has been detected on the RX Ports being forwarded to the CSI Transmit Port
0	IS_CSI_PASS	R/COR	0	CSI Pass interrupt CSI Transmit Port assertion of CSI Pass detected. Current status for the CSI Pass indication can be read from the TX_PORT_PASS flag in the CSI_STS register

### 7.7.44 CSI\_TEST\_CTL Register

**表 7-60. CSI\_TEST\_CTL (Address 0x38)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.45 CSI\_TEST\_PATT\_HI Register

**表 7-61. CSI\_TEST\_PATT\_HI (Address 0x39)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	CSI_TEST_PATT	RW	0x00	Bits 15:8 of fixed pattern for characterization test

### 7.7.46 CSI\_TEST\_PATT\_LO Register

**表 7-62. CSI\_TEST\_PATT\_LO (Address 0x3A)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	CSI_TEST_PATT	RW	0x00	Bits 7:0 of fixed pattern for characterization test

### 7.7.47 RESERVED Register

**表 7-63. RESERVED (Address 0x3B)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x01	Reserved

### 7.7.48 RESERVED Register

**表 7-64. RESERVED (Address 0x3C)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x14	Reserved

## 7.7.49 RESERVED Register

**表 7-65. RESERVED (Address 0x3D)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x6F	Reserved

## 7.7.50 RESERVED Register

**表 7-66. RESERVED (Address 0x3E)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

## 7.7.51 RESERVED Register

**表 7-67. RESERVED (Address 0x3F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x40	Reserved

## 7.7.52 RESERVED Register

**表 7-68. RESERVED (Address 0x40)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.53 SFILTER\_CFG Register

The SFilter configuration register controls the minimum and maximum values allow for the clock to data sample timing. It is recommended to program this register to 0xA9 during initialization for optimal startup time and ensure consistent AEQ performance across different channel characteristics.

**表 7-69. SFILTER\_CFG (Address 0x41)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	SFILTER_MAX	RW	0xA	SFILTER maximum setting This field controls the maximum SFILTER setting. Allowed values are 0-14 with 7 being the mid point. These values are used for both AEQ adaption and dynamic SFILTER control. The maximum setting must be greater than or equal to the minimum setting.
3:0	SFILTER_MIN	RW	0x7	SFILTER minimum setting. This field controls the maximum SFILTER setting. Allowed values are 0-14, where 7 is the mid point. These values are used for both AEQ adaption and dynamic SFILTER control. The minimum setting must be less than or equal to the SFILTER_MAX. Recommend to set SFILTER_MIN = 0x9 for normal operation in typical system use cases.

### 7.7.54 AEQ\_CTL1 Register

**表 7-70. AEQ\_CTL1 (Address 0x42)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6:4	AEQ_ERR_CTL	RW	0x7	AEQ Error Control Setting any bits in AEQ_ERR_CTL will enable FPD3 error checking during the Adaptive Equalization process. Errors are accumulated over 1/2 of the period of the timer set by the ADAPTIVE_EQ_RELOCK_TIME field in the AEQ_CTL2 register. If the number of errors is greater than the programmed threshold (AEQ_ERR_THOLD), the AEQ will attempt to increase the EQ setting. The errors may also be checked as part of EQ setting validation if AEQ_2STEP_EN is set. The following errors are checked based on this three bit field: [6] FPD-Link III clock errors [5] Packet encoding errors [4] Parity errors
3	RESERVED	RW	0	Reserved
2	AEQ_2STEP_EN	RW	0	AEQ 2-step enable This bit enables a two-step operation as part of the Adaptive EQ algorithm. If disabled, the state machine will wait for a programmed period of time, then check status to determine if setting is valid. If enabled, the state machine will wait for 1/2 the programmed period, then check for errors over an additional 1/2 the programmed period. If errors occur during the 2nd step, the state machine will immediately move to the next setting. 0 : Wait for full programmed delay, then check instantaneous lock value 1 : Wait for 1/2 programmed time, then check for errors over 1/2 programmed time. The programmed time is controlled by the ADAPTIVE_EQ_RELOCK_TIME field in the AEQ_CTL2 register
1	AEQ_OUTER_LOOP	RW	0	AEQ outer loop control This bit controls whether the Equalizer or SFILTER adaption is the outer loop when the AEQ adaption includes SFILTER adaption. 0 : AEQ is inner loop, SFILTER is outer loop 1 : AEQ is outer loop, SFILTER is inner loop
0	AEQ_SFILTER_EN	RW	1	Enable SFILTER Adaption with AEQ Setting this bit allows SFILTER adaption as part of the Adaptive Equalizer algorithm.



## 7.7.55 AEQ\_ERR\_THOLD Register

**表 7-71. AEQ\_ERR\_THOLD (Address 0x43)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	AEQ_ERR_THRESHOLD	RW	0x1	AEQ Error Threshold This register controls the error threshold to determine when to re-adapt the EQ settings. This register should not be programmed to a value of 0.

### 7.7.56 RESERVED Register

**表 7-72. RESERVED (Address 0x44 – 0x49)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x00	Reserved

### 7.7.57 FPD3\_CAP Register

Recommended to set bit four in the FPD-Link III capabilities register to one in order to flag errors detected from enhanced CRC on encoded link control information. The FPD-Link III Encoder CRC must also be enabled by setting the FPD3\_ENC\_CRC\_DIS (register 0xBA[7]) to 0.

**表 7-73. FPD3\_CAP (Address 0x4A)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	RESERVED	RW	0x0	Reserved
4	FPD3_ENC_CRC_C AP	RW	0	0: Disable CRC error flag from FPD-Link III encoder 1: Disable CRC error flag from FPD-Link III encoder (recommended)
3:0	RESERVED	RW	0x0	Reserved

## 7.7.58 FPD3\_PORT\_SEL Register

The FPD-Link III Port Select register configures which port is accessed in I2C commands to unique Rx Port registers 0x4D - 0x7F and 0xD0 - 0xDF. A 2-bit RX\_READ\_PORT field provides for reading values from a single port. The 4-bit RX\_WRITE\_PORT field provides individual enables for each port, allowing simultaneous writes broadcast to both of the FPD-Link III Receive port register blocks in unison. The DS90UB638-Q1 maintains separate page control, preventing conflict between sources.

**表 7-74. FPD3\_PORT\_SEL (Address 0x4C)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	RESERVED	R	0	Reserved
5	RESERVED	R	0	Reserved
4	RESERVED	RW	0	Reserved
3:2	RESERVED	R	0	Reserved
1	RESERVED	RW	0	Reserved
0	RX_WRITE_PORT_0	RW	0 1 for RX Port 0	Write Enable for RX port 0 registers This bit enables writes to RX port 0 registers. This applies to all paged FPD-Link III Receiver port registers. 0: Writes disabled 1: Writes enabled When accessed via Bi-directional Control Channel, the default value is 1 if accessed over RX port 0.

### 7.7.59 RX\_PORT\_STS1 Register

RX port specific register dependent on the settings in register 0x4C.

**表 7-75. RX\_PORT\_STS1 (Address 0x4D)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	RX_PORT_NUM	R	0	RX Port Number. This read-only field indicates the number of the currently selected RX read port.
5	BCC_CRC_ERROR	R/COR	0	Bi-directional Control Channel CRC Error Detected This bit indicates a CRC error has been detected in the forward control channel. If this bit is set, an error may have occurred in the control channel operation. This bit is cleared on read.
4	LOCK_STS_CHG	R/COR	0	Lock Status Changed This bit is set if a change in receiver lock status has been detected since the last read of this register. Current lock status is available in the LOCK_STS bit of this register. This bit is cleared on read.
3	BCC_SEQ_ERROR	R/COR	0	Bi-directional Control Channel Sequence Error Detected This bit indicates a sequence error has been detected in the forward control channel. If this bit is set, an error may have occurred in the control channel operation. This bit is cleared on read.
2	PARITY_ERROR	R	0	FPD-Link III parity errors detected This flag is set when the number of parity errors detected is greater than the threshold programmed in the PAR_ERR_THOLD registers. 1: Number of FPD-Link III parity errors detected is greater than the threshold 0: Number of FPD-Link III parity errors is below the threshold This bit is cleared when the RX_PAR_ERR_HI/LO registers are cleared.
1	PORT_PASS	R	0	Receiver PASS indication. This bit indicates the current status of the Receiver PASS indication. The requirements for setting the Receiver PASS indication are controlled by the PORT_PASS_CTL register. 1: Receive input has met PASS criteria 0: Receive input does not meet PASS criteria
0	LOCK_STS	R	0	FPD-Link III receiver is locked to incoming data 1: Receiver is locked to incoming data 0: Receiver is not locked

## 7.7.60 RX\_PORT\_STS2 Register

**表 7-76. RX\_PORT\_STS2 (Address 0x4E)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	LINE_LEN_UNSTABLE	R/COR	0	Line Length Unstable If set, this bit indicates the line length was detected as unstable during a previous video frame. The line length is considered to be stable if all the lines in the video frame have the same length. This flag will remain set until read.
6	LINE_LEN_CHG	R/COR	0	Line Length Changed 1: Change of line length detected 0: Change of line length not detected This bit is cleared on read.
5	FPD3_ENCODE_ERROR	R/COR	0	FPD-Link III Encoder error detected If set, this flag indicates an error in the FPD-Link III encoding has been detected by the FPD-Link III receiver. This bit is cleared on read. Note, to detect FP3 Encoder errors, the LINK_ERROR_COUNT must be enabled with a LINK_ERR_THRESH value greater than 1. Otherwise, the loss of Receiver Lock will prevent detection of the Encoder error.
4	BUFFER_ERROR	R/COR	0	Packet buffer error detected. If this bit is set, an overflow condition has occurred on the packet buffer FIFO. 1: Packet Buffer error detected 0: No Packet Buffer errors detected This bit is cleared on read.
3	CSI_ERROR	R	0	CSI Receive error detected. See the CSI_RX_STS register for details.
2	FREQ_STABLE	R	0	Frequency measurement stable
1	NO_FPD3_CLK	R	0	No FPD-Link III input clock detected
0	LINE_CNT_CHG	R/COR	0	Line Count Changed 1: Change of line count detected 0: Change of line count not detected This bit is cleared on read.

## 7.7.61 RX\_FREQ\_HIGH Register

**表 7-77. RX\_FREQ\_HIGH (Address 0x4F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FREQ_CNT_HIGH	R	0x00	Frequency Counter High Byte (MHz) The Frequency counter reports the measured frequency for the FPD-Link III Receiver. This portion of the field is the integer value in MHz.

## 7.7.62 RX\_FREQ\_LOW Register

**表 7-78. RX\_FREQ\_LOW (Address 0x50)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FREQ_CNT_LOW	R	0x00	Frequency Counter Low Byte (1/256 MHz) The Frequency counter reports the measured frequency for the FPD-Link III Receiver. This portion of the field is the fractional value in 1/256 MHz.

### 7.7.63 SENSOR\_STS\_0 Register

Sensor Status Register 0 field provides additional status information when paired with a DS90UB63x-Q1 CSI-2 Serializer. This field is automatically loaded from the forward channel.

**表 7-79. SENSOR\_STS\_0 (Address 0x51)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	R	00	Reserved
5	CSI_ALARM	R	0	Alarm flag for CSI error from serializer
4	BCC_ALARM	R	0	Alarm flag for back channel error from serializer
3	LINK_DETECT_ALARM	R	0	Alarm flag for link detect from serializer
2	TEMP_SENSE_ALARM	R	0	Alarm flag for temp sensor from serializer
1	VOLT1_SENSE_ALARM	R	0	Alarm flag for voltage sensor 1 from serializer
0	VOLT0_SENSE_ALARM	R	0	Alarm flag for voltage sensor 0 from serializer

### 7.7.64 SENSOR\_STS\_1 Register

Sensor Status Register 1 field provides additional status information when paired with a DS90UB63x-Q1 CSI-2 Serializer. This field is automatically loaded from the forward channel.

**表 7-80. SENSOR\_STS\_1 (Address 0x52)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6:4	VOLT1_SENSE_LEVEL	R	0x0	Voltage sensor sampled value from serializer
3	RESERVED	R	0	Reserved
2:0	VOLT0_SENSE_LEVEL	R	0x0	Voltage sensor sampled value from serializer

### 7.7.65 SENSOR\_STS\_2 Register

Sensor Status Register 2 field provides additional status information when paired with a DS90UB63x-Q1 CSI-2 Serializer. This field is automatically loaded from the forward channel.

**表 7-81. SENSOR\_STS\_2 (Address 0x53)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:3	RESERVED	R	0	
2:0	TEMP_SENSE_LEVEL	R	0x0	Temperature sensor sampled value from serializer

### 7.7.66 SENSOR\_STS\_3 Register

Sensor Status Register 3 field provides additional status information on the CSI-2 input when paired with a DS90UB63x-Q1 CSI-2 Serializer. This field is automatically loaded from the forward channel.

**表 7-82. SENSOR\_STS\_3 (Address 0x54)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	RESERVED	R	0	Reserved
4	CSI_ECC_2BIT_ERR	R	0	CSI -2 ECC error flag from serializer
3	CSI_CHKSUM_ERR	R	0	CSI-2 checksum error from serializer
2	CSI_SOT_ERR	R	0	CSI-2 start of transmission error from serializer
1	CSI_SYNC_ERR	R	0	CSI-2 synchronization error from serializer
0	CSI_CNTRL_ERR	R	0	CSI-2 control error from serializer

### 7.7.67 RX\_PAR\_ERR\_HI Register

**表 7-83. RX\_PAR\_ERR\_HI (Address 0x55)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PAR_ERROR line _BYTE_1	R/COR	0	Number of FPD-Link III parity errors 8 most significant bits. The parity error counter registers return the number of data parity errors that have been detected on the FPD-Link III Receiver data since the last detection of valid lock or last read of the RX_PAR_ERR_LO register. For accurate reading of the parity error count, disable the RX_PARITY_CHECKER_ENABLE bit in register 0x02 prior to reading the parity error count registers. This register is cleared upon reading the RX_PAR_ERR_LO register.

### 7.7.68 RX\_PAR\_ERR\_LO Register

**表 7-84. RX\_PAR\_ERR\_LO (Address 0x56)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PAR_ERROR _BYTE_0	R/COR	0	Number of FPD-Link III parity errors 8 least significant bits. The parity error counter registers return the number of data parity errors that have been detected on the FPD-Link III Receiver data since the last detection of valid lock or last read of the RX_PAR_ERR_LO register. For accurate reading of the parity error count, disable the RX_PARITY_CHECKER_ENABLE bit in register 0x02 prior to reading the parity error count registers. This register is cleared on read.

### 7.7.69 BIST\_ERR\_COUNT Register

**表 7-85. BIST\_ERR\_COUNT (Address 0x57)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	BIST_ERROR _COUNT	R	0	Bist Error Count Returns BIST error count

### 7.7.70 BCC\_CONFIG Register

**表 7-86. BCC\_CONFIG (Address 0x58)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	I2C_PASS_THROUGH_ALL	RW	0	I2C Pass-Through All Transactions 0: Disabled 1: Enabled
6	I2C_PASS_THROUGH	RW	0	I2C Pass-Through to Serializer if decode matches 0: Pass-Through Disabled 1: Pass-Through Enabled
5	AUTO_ACK_ALL	RW	0	Automatically Acknowledge all I2C writes independent of the forward channel lock state or status of the remote Acknowledge 1: Enable 0: Disable
4	BC_ALWAYS_ON	RW	1	Back channel enable 1: Back channel is always enabled independent of I2C_PASS_THROUGH and I2C_PASS_THROUGH_ALL 0: Back channel enable requires setting of either I2C_PASS_THROUGH and I2C_PASS_THROUGH_ALL This bit may only be written through a local I2C controller.
3	BC_CRC_GENERATOR_ENABLE	RW	1	Back Channel CRC Generator Enable 0: Disable 1: Enable
2:0	BC_FREQ_SELECT	RW	S	Back Channel Frequency Select. Default value set by strap condition upon asserting PDB = HIGH. 000- 001: Reserved 010: 10 Mbps (select for non-synchronous back channel compatibility) 011-111: Reserved 000: 2.5 Mbps (select for DS90UB633-Q1 compatibility) 001- 011: Reserved 010: 10 Mbps (select for non-synchronous back channel compatibility) 101: 25 Mbps 110: 50 Mbps (default for DS90UB635-Q1 CSI Synchronous back channel compatibility) 111: Reserved Note that changing this setting will result in some errors on the back channel for a short period of time. If set over the control channel, the Deserializer should first be programmed to Auto-Ack operation to avoid a control channel timeout due to lack of response from the Serializer.

### 7.7.71 DATAPATH\_CTL1 Register

**表 7-87. DATAPATH\_CTL1 (Address 0x59)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	OVERRIDE_FC_CONFIG	RW	0	1: Disable loading of the DATAPATH_CTL registers from the forward channel, keeping locally written values intact 0: Allow forward channel loading of DATAPATH_CTL registers
6:2	RESERVED	RW	0	Reserved
1:0	FC_GPIO_EN	RW	0	Forward Channel GPIO Enable Configures the number of enabled forward channel GPIOs 00: GPIOs disabled 01: One GPIO 10: Two GPIOs 11: Four GPIOs This field is normally loaded from the remote serializer. It can be overwritten if the OVERRIDE_FC_CONFIG bit in this register is 1.



## 7.7.72 DATAPATH\_CTL2 Register

**表 7-88. DATAPATH\_CTL2 (Address 0x5A)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0	Reserved

## 7.7.73 SER\_ID Register

**表 7-89. SER\_ID (Address 0x5B)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	SER_ID	RW	0x00	Remote Serializer ID This field is normally loaded automatically from the remote Serializer.
0	FREEZE_DEVICE_ID	RW	0	Freeze Serializer Device ID Prevent auto-loading of the Serializer Device ID from the Forward Channel. The ID is frozen at the value written.

## 7.7.74 SER\_ALIAS\_ID Register

**表 7-90. SER\_ALIAS\_ID (Address 0x5C)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	SER_ALIAS_ID	RW	0	7-bit Remote Serializer Alias ID Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote serializer. The transaction is remapped to the address specified in the Target ID register. A value of 0 in this field disables access to the remote I2C Target.
0	SER_AUTO_ACK	RW	0	Automatically Acknowledge all I2C writes to the remote Serializer independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

## 7.7.75 TargetID[0] Register

**表 7-91. TargetID[0] (Address 0x5D)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID0	RW	0	7-bit Remote Target Device ID 0 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID0, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

## 7.7.76 TargetID[1] Register

**表 7-92. TargetID[1] (Address 0x5E)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID1	RW	0	7-bit Remote Target Device ID 1 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID1, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.77 TargetID[2] Register

**表 7-93. TargetID[2] (Address 0x5F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID2	RW	0	7-bit Remote Target Device ID 2 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID2, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.78 TargetID[3] Register

**表 7-94. TargetID[3] (Address 0x60)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID3	RW	0	7-bit Remote Target Device ID 3 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID3, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.79 TargetID[4] Register

**表 7-95. TargetID[4] (Address 0x61)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID4	RW	0	7-bit Remote Target Device ID 4 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID4, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.80 TargetID[5] Register

**表 7-96. TargetID[5] (Address 0x62)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID5	RW	0	7-bit Remote Target Device ID 5 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID5, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.81 TargetID[6] Register

**表 7-97. TargetID[6] (Address 0x63)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID6	RW	0	7-bit Remote Target Device ID 6 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID6, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.82 TargetID[7] Register

**表 7-98. TargetID[7] (Address 0x64)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ID7	RW	0	7-bit Remote Target Device ID 7 Configures the physical I2C address of the remote I2C Target device attached to the remote Serializer. If an I2C transaction is addressed to the Target Alias ID7, the transaction is remapped to this address before passing the transaction across the Bidirectional Control Channel to the Serializer.
0	RESERVED	R	0	Reserved.

### 7.7.83 TargetAlias[0] Register

**表 7-99. TargetAlias[0] (Address 0x65)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID0	RW	0	7-bit Remote Target Device Alias ID 0 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID0 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK_0	RW	0	Automatically Acknowledge all I2C writes to the remote Target 0 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

### 7.7.84 TargetAlias[1] Register

**表 7-100. TargetAlias[1] (Address 0x66)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID1	RW	0	7-bit Remote Target Device Alias ID 1 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID1 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK_1	RW	0	Automatically Acknowledge all I2C writes to the remote Target 1 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

### 7.7.85 TargetAlias[2] Register

**表 7-101. TargetAlias[2] (Address 0x67)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID2	RW	0	7-bit Remote Target Device Alias ID 2 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID2 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK 2	RW	0	Automatically Acknowledge all I2C writes to the remote Target 2 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

### 7.7.86 TargetAlias[3] Register

**表 7-102. TargetAlias[3] (Address 0x68)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID3	RW	0	7-bit Remote Target Device Alias ID 3 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID3 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK _3	RW	0	Automatically Acknowledge all I2C writes to the remote Target 3 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

### 7.7.87 TargetAlias[4] Register

**表 7-103. TargetAlias[4] (Address 0x69)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID4	RW	0	7-bit Remote Target Device Alias ID 4 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID4 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_AK K_4	RW	0	Automatically Acknowledge all I2C writes to the remote Target 4 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

## 7.7.88 TargetAlias[5] Register

**表 7-104. TargetAlias[5] (Address 0x6A)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID5	RW	0	7-bit Remote Target Device Alias ID 5 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID5 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK_5	RW	0	Automatically Acknowledge all I2C writes to the remote Target 5 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

## 7.7.89 TargetAlias[6] Register

**表 7-105. TargetAlias[6] (Address 0x6B)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID6	RW	0	7-bit Remote Target Device Alias ID 6 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID6 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK_6	RW	0	Automatically Acknowledge all I2C writes to the remote Target 6 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

## 7.7.90 TargetAlias[7] Register

**表 7-106. TargetAlias[7] (Address 0x6C)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	TARGET_ALIAS_ID7	RW	0	7-bit Remote Target Device Alias ID 7 Configures the decoder for detecting transactions designated for an I2C Target device attached to the remote Serializer. The transaction is remapped to the address specified in the Target ID7 register. A value of 0 in this field disables access to the remote I2C Target.
0	TARGET_AUTO_ACK_7	RW	0	Automatically Acknowledge all I2C writes to the remote Target 7 independent of the forward channel lock state or status of the remote Serializer Acknowledge 1: Enable 0: Disable

## 7.7.91 PORT\_CONFIG Register

表 7-107. PORT\_CONFIG (Address 0x6D)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	CSI_WAIT_FS1	RW	0	CSI Wait for FrameStart packet with count 1 The CSI Receiver will wait for a Frame Start packet with count of 1 before accepting other packets
6	CSI_WAIT_FS	RW	1	CSI Wait for FrameStart packet CSI-2 Receiver will wait for a Frame Start packet before accepting other packets
5	CSI_FWD_CKSUM	RW	1	Forward CSI packets with checksum errors 0: Do not forward packets with errors 1: Forward packets with errors
4	CSI_FWD_ECC	RW	1	Forward CSI packets with ECC errors 0: Do not forward packets with errors 1: Forward packets with errors
3	CSI_FWD_LEN/ DISCARD_1ST _LINE_ON_ERR	RW	1	In CSI FPD-Link III Input Mode, Forward CSI packets with length errors. 0: CSI: Do not forward packets with errors. 1: CSI: Forward packets with errors.
2	RESERVED	RW	S	Reserved
1:0	FPD3_MODE	RW	S	FPD-Link III Input Mode Default value set by strap condition of MODE pin upon asserting PDB = HIGH at start-up. 00: CSI Mode (DS90UB635-Q1 compatible) 10: RAW12 HF Mode (56.25-100MHz) (DS90UB633A compatible) 11: RAW10 Mode (75-100MHz) (DS90UB633A compatible)

## 7.7.92 BC\_GPIO\_CTL0 Register

表 7-108. BC\_GPIO\_CTL0 (Address 0x6E)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	BC_GPIO1_SEL	RW	0x8	Back channel GPIO1 Select: Determines the data sent on GPIO1 for the port back channel. 0xxx : Pin GPIOx where x is BC_GPIO1_SEL[2:0] 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved
3:0	BC_GPIO0_SEL	RW	0x8	Back channel GPIO0 Select: Determines the data sent on GPIO0 for the port back channel. 0xxx : Pin GPIOx where x is BC_GPIO0_SEL[2:0] 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved

## 7.7.93 BC\_GPIO\_CTL1 Register

**表 7-109. BC\_GPIO\_CTL1 (Address 0x6F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	BC_GPIO3_SEL	RW	0x8	Back channel GPIO3 Select: Determines the data sent on GPIO3 for the port back channel. 0xxx : Pin GPIOx where x is BC_GPIO3_SEL[2:0] 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved
3:0	BC_GPIO2_SEL	RW	0x8	Back channel GPIO2 Select: Determines the data sent on GPIO2 for the port back channel. 0xxx : Pin GPIOx where x is BC_GPIO2_SEL[2:0] 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved

### 7.7.94 CSI\_VC\_MAP Register

CSI virtual channel mapping only applies when FPD-Link III operating in CSI-2 input mode..

**表 7-110. CSI\_VC\_MAP (Address 0x72)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	CSI_VC_MAP	RW	0xE4	CSI-2 Virtual Channel Mapping Register This register provides a method for replacing the Virtual Channel Identifier (VC-ID) of incoming CSI packets. [7:6] : Map value for VC-ID of 3 [5:4] : Map value for VC-ID of 2 [3:2] : Map value for VC-ID of 1 [1:0] : Map value for VC-ID of 0

### 7.7.95 LINE\_COUNT\_HI Register

**表 7-111. LINE\_COUNT\_HI (Address 0x73)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	LINE_COUNT_HI	R	0x0	High byte of Line Count The Line Count reports the line count for the most recent video frame. When interrupts are enabled for the Line Count (via the IE_LINE_CNT_CHG register bit), the Line Count value is frozen until read.

### 7.7.96 LINE\_COUNT\_LO Register

**表 7-112. LINE\_COUNT\_LO (Address 0x74)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	LINE_COUNT_LO	R	0x0	Low byte of Line Count The Line Count reports the line count for the most recent video frame. When interrupts are enabled for the Line Count (via the IE_LINE_CNT_CHG register bit), the Line Count value is frozen until read. In addition, when reading the LINE_COUNT registers, the LINE_COUNT_LO is latched upon reading LINE_COUNT_HI to ensure consistency between the two portions of the Line Count.

### 7.7.97 LINE\_LEN\_1 Register

**表 7-113. LINE\_LEN\_1 (Address 0x75)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	LINE_LEN_HI	R	0	High byte of Line Length The Line Length reports the line length recorded during the most recent video frame. If line length is not stable during the frame, this register will report the length of the last line in the video frame. When interrupts are enabled for the Line Length (via the IE_LINE_LEN_CHG register bit), the Line Length value is frozen until read.



## 7.7.98 LINE\_LEN\_0 Register

**表 7-114. LINE\_LEN\_0 (Address 0x76)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	LINE_LEN_LO	R	0	Low byte of Line Length The Line Length reports the length of the most recent video line. When interrupts are enabled for the Line Length (via the IE_LINE_LEN_CHG register bit), the Line Length value is frozen until read. In addition, when reading the LINE_LEN registers, the LINE_LEN_LO is latched upon reading LINE_LEN_HI to ensure consistency between the two portions of the Line Length.

## 7.7.99 FREQ\_DET\_CTL Register

**表 7-115. FREQ\_DET\_CTL (Address 0x77)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	FREQ_HYST	RW	0x3	Frequency Detect Hysteresis The Frequency detect hysteresis setting allows ignoring minor fluctuations in frequency. A new frequency measurement will be captured only if the measured frequency differs from the current measured frequency by more than the FREQ_HYST setting. The FREQ_HYST setting is in MHz.
5:4	FREQ_STABLE_THR	RW	0x0	Frequency Stable Threshold The Frequency detect circuit can be used to detect a stable clock frequency. The Stability Threshold determines the amount of time required for the clock frequency to stay within the FREQ_HYST range to be considered stable: 00 : 40 $\mu$ s 01 : 80 $\mu$ s 10 : 320 $\mu$ s 11 : 1.28 ms
3:0	FREQ_LO_THR	RW	0x5	Frequency Low Threshold Sets the low threshold for the Clock frequency detect circuit in MHz. This value is used to determine if the clock frequency is too low for proper operation.

**7.7.100 MAILBOX\_1 Register****表 7-116. MAILBOX\_1 (Address 0x78)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	MAILBOX_0	RW	0x00	Mailbox Register This register is an unused read/write register that can be used for any purpose such as passing messages between I2C controllers on opposite ends of the link.

**7.7.101 MAILBOX\_2 Register****表 7-117. MAILBOX\_2 (Address 0x79)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	MAILBOX_1	RW	0x01	Mailbox Register This register is an unused read/write register that can be used for any purpose such as passing messages between I2C controllers on opposite ends of the link.

**7.7.102 CSI\_RX\_STS Register****表 7-118. CSI\_RX\_STS (Address 0x7A)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	RESERVED	R	0	Reserved
3	LENGTH_ERR	R/COR	0	Packet Length Error detected for received CSI packet If set, this bit indicates a packet length error was detected on at least one CSI packet received from the sensor. Packet length errors occur if the data length field in the packet header does not match the actual data length for the packet. 1: One or more Packet Length errors have been detected 0: No Packet Length errors have been detected This bit is cleared on read.
2	CKSUM_ERR	R/COR	0	Data Checksum Error detected for received CSI packet If set, this bit indicates a data checksum error was detected on at least one CSI packet received from the sensor. Data checksum errors indicate an error was detected in the packet data portion of the CSI packet. 1: One or more Data Checksum errors have been detected 0: No Data Checksum errors have been detected This bit is cleared on read.
1	ECC2_ERR	R/COR	0	2-bit ECC Error detected for received CSI packet If set, this bit indicates a multi-bit ECC error was detected on at least one CSI packet received from the sensor. Multi-bit errors are not corrected by the device. 1: One or more multi-bit ECC errors have been detected 0: No multi-bit ECC errors have been detected This bit is cleared on read.
0	ECC1_ERR	R/COR	0	1-bit ECC Error detected for received CSI packet If set, this bit indicates a single-bit ECC error was detected on at least one CSI packet received from the sensor. Single-bit errors are corrected by the device. 1: One or more 1-bit ECC errors have been detected 0: No 1-bit ECC errors have been detected This bit is cleared on read.

### 7.7.103 CSI\_ERR\_COUNTER Register

**表 7-119. CSI\_ERR\_COUNTER (Address 0x7B)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	CSI_ERR_CNT	R/COR	0x00	CSI Error Counter Register This register counts the number of CSI packets received with errors since the last read of the counter.

### 7.7.104 PORT\_CONFIG2 Register

**表 7-120. PORT\_CONFIG2 (Address 0x7C)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	RW	0x0	RESERVED
5	DISCARD_ON_PAR_ERR	RW	1	Discard frames on Parity Error 0 : Forward packets with parity errors 1 : Truncate Frames if a parity error is detected
4	DISCARD_ON_LINE_SIZE	RW	0	Discard frames on Line Size 0 : Allow changes in Line Size within packets 1 : Truncate Frames if a change in line size is detected
3	DISCARD_ON_FRAME_SIZE	RW	0	Discard frames on change in Frame Size When enabled, a change in the number of lines in a frame will result in truncation of the packet. The device will resume forwarding video frames based on the PASS_THRESHOLD setting in the PORT_PASS_CTL register. 0 : Allow changes in Frame Size 1 : Truncate Frames if a change in frame size is detected
2	RESERVED	RW	0	Reserved
1	RESERVED	RW	0	Reserved
0	RESERVED	RW	0	Reserved

## 7.7.105 PORT\_PASS\_CTL Register

表 7-121. PORT\_PASS\_CTL (Address 0x7D)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	PASS_DISCARD_EN	RW	0	Pass Discard Enable Discard packets if PASS is not indicated. 0 : Ignore PASS for forwarding packets 1 : Discard packets when PASS is not true
6	RESERVED	RW	0	Reserved
5	PASS_LINE_CNT	RW	0	Pass Line Count Control This register controls whether the device will include line count in qualification of the Pass indication: 0 : Don't check line count 1 : Check line count When checking line count, Pass is deasserted upon detection of a change in the number of video lines per frame. Pass will not be reasserted until the PASS_THRESHOLD setting is met.
4	PASS_LINE_SIZE	RW	0	Pass Line Size Control This register controls whether the device will include line size in qualification of the Pass indication: 0 : Don't check line size 1 : Check line size When checking line size, Pass is deasserted upon detection of a change in video line size. Pass will not be reasserted until the PASS_THRESHOLD setting is met.
3	PASS_PARITY_ERR	RW	0	Parity Error Mode If this bit is set to 0, the port Pass indication is deasserted for every parity error detected on the FPD-Link III Receive interface. If this bit is set to a 1, the port Pass indication is cleared on a parity error and remain clear until the PASS_THRESHOLD is met. When PASS_PARITY_ERR is set to 1, TI also recommends setting PASS_THRESHOLD to 2 or higher to ensure at least one good frame occurs following a parity error
2	PASS_WDOG_DIS	RW	0	RX Port Pass Watchdog disable When enabled, if the FPD Receiver does not detect a valid frame end condition within two video frame periods, the Pass indication is deasserted. The watchdog timer will not have any effect if the PASS_THRESHOLD is set to 0. 0 : Enable watchdog timer for RX Pass 1 : Disable watchdog timer for RX Pass
1:0	PASS_THRESHOLD	RW	0x0	Pass Threshold Register This register controls the number of valid frames before asserting the port Pass indication. If set to 0, PASS is asserted after Receiver Lock detect. If non-zero, PASS is asserted following reception of the programmed number of valid frames.

## 7.7.106 SEN\_INT\_RISE\_CTL Register

表 7-122. SEN\_INT\_RISE\_CTL (Address 0x7E)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SEN_INT_RISE_MASK	RW	0x0	Sensor Interrupt Rise Mask This register provides the interrupt mask for detecting rising edge transitions on the bits in SENSOR_STS_0. If a mask bit is set in this register, a rising edge transition on the corresponding SENSOR_STS_0 bit will generate an interrupt that will be latched in the SEN_INT_RISE_STS register.

### 7.7.107 SEN\_INT\_FALL\_CTL Register

**表 7-123. SEN\_INT\_FALL\_CTL (Address 0x7F)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SEN_INT_FALL_MASK	RW	0x0	Sensor Interrupt Fall Mask This register provides the interrupt mask for detecting falling edge transitions on the bits in SENSOR_STS_0. If a mask bit is set in this register, a falling edge transition on the corresponding SENSOR_STS_0 bit will generate an interrupt that will be latched in the SEN_INT_FALL_STS register.

### 7.7.108 RESERVED Register

**表 7-124. RESERVED (Address 0xA0 – 0xA4)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.109 REFCLK\_FREQ Register

**表 7-125. REFCLK\_FREQ (Address 0xA5)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	REFCLK_FREQ	R	0x00	REFCLK frequency measurement in MHz. REFCLK_FREQ measurement is not synchronized. Value in this register should read twice and only considered valid if REFCLK_FREQ is unchanged between reads.

### 7.7.110 RESERVED Register

**表 7-126. RESERVED (Address 0xA7 – 0xAF)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x00	Reserved

## 7.7.111 IND\_ACC\_CTL Register

表 7-127. IND\_ACC\_CTL (Address 0xB0)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	R	0x0	Reserved
5:2	IA_SEL	RW	0x0	Indirect Access Register Select: Selects target for register access 0000 : CSI-2 Pattern Generator & Timing Registers 0001 : FPD-Link III RX Port 0 Reserved Registers 00011–0100: Reserved 0101 : FPD-Link III RX Shared Reserved Registers 0110 : Simultaneous write to FPD-Link III RX Reserved Registers 0111 : CSI-2 Reserved Registers 1000 : Reserved Registers 1001 : Unique DIE ID Registers 1010–1111 : Reserved
1	IA_AUTO_INC	RW	0	Indirect Access Auto Increment: Enables auto-increment mode. Upon completion of a read or write, the register address will automatically be incremented by 1
0	IA_READ	RW	0	Indirect Access Read: Setting this allows generation of a read strobe to the selected register block upon setting of the IND_ACC_ADDR register. In auto-increment mode, read strobes will also be asserted following a read of the IND_ACC_DATA register. This function is only required for blocks that need to pre-fetch register data.

## 7.7.112 IND\_ACC\_ADDR Register

表 7-128. IND\_ACC\_ADDR (Address 0xB1)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	IA_ADDR	RW	0x0	Indirect Access Register Offset: This register contains the 8-bit register offset for the indirect access.

## 7.7.113 IND\_ACC\_DATA Register

表 7-129. IND\_ACC\_DATA (Address 0xB2)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	IA_DATA	RW	0x0	Indirect Access Data: Writing this register will cause an indirect write of the IND_ACC_DATA value to the selected analog block register. Reading this register will return the value of the selected block register

### 7.7.114 BIST Control Register

**表 7-130. BIST Control (Address 0xB3)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	BIST_OUT_MODE	RW	0x0	BIST Output Mode 00 : Outputs disabled during BIST 01 : Reserved 10 : Outputs enabled during BIST 11 : Reserved
5:4	RESERVED	RW	0x0	Reserved
3	BIST_PIN_CONFIG	RW	1	Bist Configured through Pin. 1: Bist configured through pin. 0: Bist configured through bits 2:0 in this register
2:1	BIST_CLOCK_SOURCE	RW	00	BIST Clock Source This register field selects the BIST Clock Source at the Serializer. These register bits are automatically written to the CLOCK SOURCE bits (register offset 0x14) in the Serializer after BIST is enabled. See the appropriate Serializer register descriptions for details.
0	BIST_EN	RW	0	BIST Control 1: Enabled 0: Disabled

### 7.7.115 RESERVED Register

**表 7-131. RESERVED (Address 0xB4)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x25	Reserved

### 7.7.116 RESERVED Register

**表 7-132. RESERVED (Address 0xB5)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.117 RESERVED Register

**表 7-133. RESERVED (Address 0xB6)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x18	Reserved

### 7.7.118 RESERVED Register

**表 7-134. RESERVED (Address 0xB7)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.119 MODE\_IDX\_STS Register

**表 7-135. MODE\_IDX\_STS (Address 0xB8)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	IDX_DONE	R	1	IDX Done If set, indicates the IDX decode has completed and latched into the IDX status bits.
6:4	IDX	R	S	IDX Decode 3-bit decode from IDX pin
3	MODE_DONE	R	1	MODE Done If set, indicates the MODE decode has completed and latched into the MODE status bits.
2:0	MODE	R	S	MODE Decode 3-bit decode from MODE pin

### 7.7.120 LINK\_ERROR\_COUNT Register

**表 7-136. LINK\_ERROR\_COUNT (Address 0xB9)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	R	0x0	Reserved
5	LINK_SFIL_WAIT	RW	1	During SFILTER adaption, setting this bit will cause the Lock detect circuit to ignore errors during the SFILTER wait period after the SFILTER control is updated. 1: Errors during SFILTER Wait period will be ignored 0: Errors during SFILTER Wait period will not be ignored and may cause loss of Lock
4	LINK_ERR_COUNT_EN	RW	1	Enable serial link data integrity error count 1: Enable error count 0: DISABLE
3:0	LINK_ERR_THRESH	RW	0x3	Link error count threshold. The Link Error Counter monitors the forward channel link and determines when link will be dropped. The link error counter is pixel clock based. FPD Link parity, clock, and control are monitored for link errors. If the error counter is enabled, the deserializer will lose lock once the error counter reaches the LINK_ERR_THRESH value. If the link error counter is disabled, the deserializer will lose lock after one error. The control bits in DIGITAL_DEBUG_2 register can be used to disable error conditions individually.

### 7.7.121 FPD3\_ENC\_CTL Register

Recommended to set bit seven in the FPD-Link III encoder control register to 0 in order to prevent any updates of link information values from encoded packets that do not pass CRC check. The FPD-Link III Encoder CRC flag must also be in place by setting FPD3\_ENC\_CRC\_DIS (register 0x4A[4]) to 1.

**表 7-137. FPD3\_ENC\_CTL (Address 0xBA)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	RW	1	0: Enable FPD-Link III encoder CRC (recommended) 1: Disable FPD-Link III encoder CRC
6:0	RESERVED	RW	0x03	Reserved

### 7.7.122 RESERVED Register

**表 7-138. RESERVED (Address 0xBD)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved



### 7.7.123 GPIO\_PD\_CTL Register

表 7-139. GPIO\_PD\_CTL (Address 0xBE)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	GPIO6_PD_DIS	RW	0	<p>GPIOX Pulldown Resistor Disable: The GPIO pins by default include a 35-kΩ typical pulldown resistor that is automatically enabled when the GPIO is not in an output mode. When this bit is set, the corresponding pulldown resistor will also be disabled when the GPIO pin is in an input only mode.</p> <p>1 : Disable GPIO pulldown resistor 0 : Enable GPIO pulldown resistor</p>
5	GPIO5_PD_DIS	RW	0	
4	GPIO4_PD_DIS	RW	0	
3	GPIO3_PD_DIS	RW	0	
2	GPIO2_PD_DIS	RW	0	
1	GPIO1_PD_DIS	RW	0	
0	GPIO0_PD_DIS	RW	0	

### 7.7.124 PORT\_DEBUG Register

表 7-140. PORT\_DEBUG (Address 0xD0)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	RW	0	Reserved
6	RESERVED	RW	0	Reserved
5	SER_BIST_ACT	R	0	<p>Serializer BIST active This register indicates the Serializer is in BIST mode. When in BIST mode this flag can be checked to ensure BIST is activated in the serializer during the test. If the Deserializer is not in BIST mode, this could indicate an error condition.</p>
4:2	RESERVED	RW	0x0	Reserved
1	FORCE_BC_ERRORS	RW	0	Setting this bit introduces continuous single bit errors into Back Channel Frames
0	FORCE_1_BC_ERROR	RW	0	Setting this bit introduces a single bit error into one Back Channel Frame

### 7.7.125 AEQ\_CTL2 Register

表 7-141. AEQ\_CTL2 (Address 0xD2)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	ADAPTIVE_EQ_RELOCK_TIME	RW	0x4	<p>Time to wait for lock before incrementing the EQ to next setting</p> <p>000 : 164 μs 001 : 328 μs 010 : 655 μs 011 : 1.31 ms 100 : 2.62 ms 101 : 5.24 ms 110 : 10.5 ms 111 : 21.0 ms</p>
4	AEQ_1ST_LOCK_MODE	RW	1	<p>AEQ First Lock Mode. This register bit controls the Adaptive Equalizer algorithm operation at initial Receiver Lock.</p> <p>0 : Initial AEQ lock may occur at any value 1 : Initial Receiver lock will restart AEQ at 0, providing a more deterministic initial AEQ value</p>
3	AEQ_RESTART	RW/SC	0	Set high to restart AEQ adaptation from initial value. This bit is self clearing. Adaption is restarted.
2	SET_AEQ_FLOOR	RW	1	AEQ adaptation starts from a pre-set floor value rather than from zero - good in long cable situations
1:0	RESERVED	R	0x0	Reserved

### 7.7.126 AEQ\_STATUS Register

表 7-142. AEQ\_STATUS (Address 0xD3)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	RESERVED	R	0x0	Reserved
5:0	EQ_STATUS	R	0x00	Adaptive EQ Status

### 7.7.127 ADAPTIVE EQ BYPASS Register

表 7-143. ADAPTIVE EQ BYPASS (Address 0xD4)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	EQ_STAGE_1_SELECT_VALUE	RW	0x3	EQ select value [5:3] - Used if adaptive EQ is bypassed.
4	AEQ_LOCK_MODE	RW	0	Adaptive Equalizer lock mode When set to a 1, Receiver Lock status requires the Adaptive Equalizer to complete adaption. When set to a 0, Receiver Lock is based only on the Lock circuit itself. AEQ may not have stabilized.
3:1	EQ_STAGE_2_SELECT_VALUE	RW	0x0	EQ select value [2:0] - Used if adaptive EQ is bypassed.
0	ADAPTIVE_EQ_BYPASS	RW	0	1: Disable adaptive EQ 0: Enable adaptive EQ

### 7.7.128 AEQ\_MIN\_MAX Register

**表 7-144. AEQ\_MIN\_MAX (Address 0xD5)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:4	AEQ_MAX	RW	0xF	Adaptive Equalizer Maximum value This register sets the maximum value for the Adaptive EQ algorithm. Must be higher than ADAPTIVE_EQ_FLOOR_VALUE when SET_AEQ_FLOOR is enabled.
3:0	ADAPTIVE_EQ_FLOOR_VALUE	RW	0x2	When AEQ floor is enabled by register 0xD2[2] the starting EQ gain setting for AEQ adaption is given by this register.

### 7.7.129 RESERVED Register

**表 7-145. RESERVED (Address 0xD6)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.130 RESERVED Register

**表 7-146. RESERVED (Address 0xD7)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x00	Reserved

### 7.7.131 PORT\_ICR\_HI Register

**表 7-147. PORT\_ICR\_HI (Address 0xD8)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:3	RESERVED	R	0x0	Reserved
2	IE_FPD3_ENC_ERR	RW	0	Interrupt on FPD-Link III Receiver Encoding Error When enabled, an interrupt is generated on detection of an encoding error on the FPD-Link III interface for the receive port as reported in the FPD3_ENC_ERROR bit in the RX_PORT_STS2 register
1	IE_BCC_SEQ_ERR	RW	0	Interrupt on BCC SEQ Sequence Error. When enabled, an interrupt is generated if a Sequence Error is detected for the Bi-directional Control Channel forward channel receiver as reported in the BCC_SEQ_ERROR bit in the RX_PORT_STS1 register.
0	IE_BCC_CRC_ERR	RW	0	Interrupt on BCC CRC error detect When enabled, an interrupt is generated if a CRC error is detected on a Bi-directional Control Channel frame received over the FPD-Link III forward channel as reported in the BCC_CRC_ERROR bit in the RX_PORT_STS1 register.

## 7.7.132 PORT\_ICR\_LO Register

表 7-148. PORT\_ICR\_LO (Address 0xD9)

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	RW	0	Reserved
6	IE_LINE_LEN_CHG	RW	0	Interrupt on Video Line length When enabled, an interrupt is generated if the length of the video line changes. Status is reported in the LINE_LEN_CHG bit in the RX_PORT_STS2 register.
5	IE_LINE_CNT_CHG	RW	0	Interrupt on Video Line count When enabled, an interrupt is generated if the number of video lines per frame changes. Status is reported in the LINE_CNT_CHG bit in the RX_PORT_STS2 register.
4	IE_BUFFER_ERR	RW	0	Interrupt on Receiver Buffer Error When enabled, an interrupt is generated if the Receive Buffer overflow is detected as reported in the BUFFER_ERROR bit in the RX_PORT_STS2 register.
3	IE_CSI_RX_ERR	RW	0	Interrupt on CSI Receiver Error. When enabled, an interrupt will be generated on detection of an error by the CSI Receiver. CSI Receiver errors are reported in the CSI_RX_STS register (address 0x7A).
2	IE_FPD3_PAR_ERR	RW	0	Interrupt on FPD-Link III Receiver Parity Error When enabled, an interrupt is generated on detection of parity errors on the FPD-Link III interface for the receive port. Parity error status is reported in the PARITY_ERROR bit in the RX_PORT_STS1 register.
1	IE_PORT_PASS	RW	0	Interrupt on change in Port PASS status When enabled, an interrupt is generated on a change in receiver port valid status as reported in the PORT_PASS bit in the PORT_STS1 register.
0	IE_LOCK_STS	RW	0	Interrupt on change in Lock Status When enabled, an interrupt is generated on a change in lock status. Status is reported in the LOCK_STS_CHG bit in the RX_PORT_STS1 register.

## 7.7.133 PORT\_ISR\_HI Register

**表 7-149. PORT\_ISR\_HI (Address 0xDA)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:5	Reserved	R	0x0	Reserved
4	IE_FC_GPIO	R	0	FC GPIO Interrupt Status A change in forward channel GPIO signal has been detected. Forward Channel GPIO status is reported in the FC_GPIO_STS register. This interrupt condition will be cleared by reading the FC_GPIO_STS register.
3	IE_FC_SENS_STS	R	0	Interrupt on change in Sensor Status A change in Sensor Status has been detected. Camera Status is reported in the SENSOR_STS_X registers. This interrupt condition will be cleared by reading the SEN_INT_RISE_STS and SEN_INT_FALL_STS registers.
2	IS_FPD3_ENC_ERR	R	0	FPD-Link III Receiver Encode Error Interrupt Status An encoding error on the FPD-Link III interface for the receive port has been detected. Status is reported in the FPD3_ENC_ERROR bit in the RX_PORT_STS2 register. This interrupt condition is cleared by reading the RX_PORT_STS2 register.
1	IS_BCC_SEQ_ERR	R	0	BCC CRC Sequence Error Interrupt Status A Sequence Error has been detected for the Bi-directional Control Channel forward channel receiver. Status is reported in the BCC_SEQ_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
0	IS_BCC_CRC_ERR	R	0	BCC CRC error detect Interrupt Status A CRC error has been detected on a Bi-directional Control Channel frame received over the FPD-Link III forward channel. Status is reported in the BCC_CRC_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.

## 7.7.134 PORT\_ISR\_LO Register

表 7-150. PORT\_ISR\_LO (Address 0xDB)

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	RESERVED	R	0	Reserved
6	IS_LINE_LEN_CHG	R	0	Video Line Length Interrupt Status A change in video line length has been detected. Status is reported in the LINE_LEN_CHG bit in the RX_PORT_STS2 register. This interrupt condition is cleared by reading the RX_PORT_STS2 register.
5	IS_LINE_CNT_CHG	R	0	Video Line Count Interrupt Status A change in number of video lines per frame has been detected. Status is reported in the LINE_CNT_CHG bit in the RX_PORT_STS2 register. This interrupt condition is cleared by reading the RX_PORT_STS2 register.
4	IS_BUFFER_ERR	R	0	Receiver Buffer Error Interrupt Status A Receive Buffer overflow has been detected as reported in the BUFFER_ERROR bit in the RX_PORT_STS2 register. This interrupt condition is cleared by reading the RX_PORT_STS2 register.
3	IS_CSI_RX_ERR	R	0	CSI Receiver Error Interrupt Status The CSI Receiver has detected an error. CSI Receiver errors are reported in the CSI_RX_STS register (address 0x7A). This interrupt condition will be cleared by reading the CSI_RX_STS register.
2	IS_FPD3_PAR_ERR	R	0	FPD-Link III Receiver Parity Error Interrupt Status A parity error on the FPD-Link III interface for the receive port has been detected. Parity error status is reported in the PARITY_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
1	IS_PORT_PASS	R	0	Port Valid Interrupt Status A change in receiver port valid status as reported in the PORT_PASS bit in the PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
0	IS_LOCK_STS	R	0	Lock Interrupt Status A change in lock status has been detected. Status is reported in the LOCK_STS_CHG bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.

### 7.7.135 FC\_GPIO\_STS Register

**表 7-151. FC\_GPIO\_STS (Address 0xDC)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	GPIO3_INT_STS	R/COR	0	GPIO3 Interrupt Status. This bit indicates an interrupt condition has been met for GPIO3. This bit is cleared on read.
6	GPIO2_INT_STS	R/COR	0	GPIO2 Interrupt Status. This bit indicates an interrupt condition has been met for GPIO2. This bit is cleared on read.
5	GPIO1_INT_STS	R/COR	0	GPIO1 Interrupt Status. This bit indicates an interrupt condition has been met for GPIO1. This bit is cleared on read.
4	GPIO0_INT_STS	R/COR	0	GPIO0 Interrupt Status. This bit indicates an interrupt condition has been met for GPIO0. This bit is cleared on read.
3	FC_GPIO3_STS	R	0	Forward Channel GPIO3 Status. This bit indicates the current value for forward channel GPIO3.
2	FC_GPIO2_STS	R	0	Forward Channel GPIO2 Status. This bit indicates the current value for forward channel GPIO2.
1	FC_GPIO1_STS	R	0	Forward Channel GPIO1 Status. This bit indicates the current value for forward channel GPIO1.
0	FC_GPIO0_STS	R	0	Forward Channel GPIO0 Status. This bit indicates the current value for forward channel GPIO0.

### 7.7.136 FC\_GPIO\_ICR Register

**表 7-152. FC\_GPIO\_ICR (Address 0xDD)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	GPIO3_FALL_IE	RW	0	GPIO3 Fall Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a falling edge on GPIO3.
6	GPIO3_RISE_IE	RW	0	GPIO3 Rise Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a rising edge on GPIO3.
5	GPIO2_FALL_IE	RW	0	GPIO2 Fall Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a falling edge on GPIO2.
4	GPIO2_RISE_IE	RW	0	GPIO2 Rise Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a rising edge on GPIO2.
3	GPIO1_FALL_IE	RW	0	GPIO1 Fall Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a falling edge on GPIO1.
2	GPIO1_RISE_IE	RW	0	GPIO1 Rise Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a rising edge on GPIO1.
1	GPIO0_FALL_IE	RW	0	GPIO0 Fall Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a falling edge on GPIO0.
0	GPIO0_RISE_IE	RW	0	GPIO0 Rise Interrupt Enable. If this bit is set, an interrupt will be generated based on detection of a rising edge on GPIO0.

### 7.7.137 SEN\_INT\_RISE\_STS Register

**表 7-153. SEN\_INT\_RISE\_STS (Address 0xDE)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SEN_INT_RISE	R/COR	0x00	Sensor Interrupt Rise Status. This register provides the interrupt status for rising edge transitions on the bits in SENSOR_STS_0. If a mask bit is set in the SEN_INT_RISE_MASK register, a rising edge transition on the corresponding SENSOR_STS_0 bit will generate an interrupt that will be latched in this register.

### 7.7.138 SEN\_INT\_FALL\_STS Register

**表 7-154. SEN\_INT\_FALL\_STS (Address 0xDF)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	SEN_INT_FALL	R/COR	0x00	Sensor Interrupt Fall Status. This register provides the interrupt status for falling edge transitions on the bits in SENSOR_STS_0. If a mask bit is set in the SEN_INT_RISE_MASK register, a falling edge transition on the corresponding SENSOR_STS_0 bit will generate an interrupt that will be latched in this register.

### 7.7.139 FPD3\_RX\_ID0 Register

**表 7-155. FPD3\_RX\_ID0 (Address 0xF0)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID0	R	0x5F	FPD3_RX_ID0: First byte ID code: '_'

### 7.7.140 FPD3\_RX\_ID1 Register

**表 7-156. FPD3\_RX\_ID1 (Address 0xF1)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID1	R	0x55	FPD3_RX_ID1: 2nd byte of ID code: 'U'

### 7.7.141 FPD3\_RX\_ID2 Register

**表 7-157. FPD3\_RX\_ID2 (Address 0xF2)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID2	R	0x42	FPD3_RX_ID2: 3rd byte of ID code: 'B'

### 7.7.142 FPD3\_RX\_ID3 Register

**表 7-158. FPD3\_RX\_ID3 (Address 0xF3)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID3	R	0x39	FPD3_RX_ID3: 4th byte of ID code: '9'

### 7.7.143 FPD3\_RX\_ID4 Register

**表 7-159. FPD3\_RX\_ID4 (Address 0xF4)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID4	R	0x35	FPD3_RX_ID4: 5th byte of ID code: '5'



### 7.7.144 FPD3\_RX\_ID5 Register

**表 7-160. FPD3\_RX\_ID5 (Address 0xF5)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	FPD3_RX_ID5	R	0x34	FPD3_RX_ID5: 6th byte of ID code: '4'

### 7.7.145 I2C\_RX0\_ID Register

As an alternative to paging to access FPD-Link III receive port0 registers, a separate I2C address may be enabled to allow direct access to the port 0 specific registers. The I2C\_RX\_0\_ID register provides a simpler method of accessing device registers specifically for port 0 without having to use the paging function to select the register page. Using this address also allows access to all shared registers.

**表 7-161. I2C\_RX0\_ID (Address 0xF8)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	RX_PORT0_ID	RW	0x0	7-bit Receive Port 0 I2C ID Configures the decoder for detecting transactions designated for Receiver port 0 registers. A value of 0x00 in this field disables the Port0 decoder.
0	RESERVED	R	0	Reserved

### 7.7.146 RESERVED Register

**表 7-162. RESERVED (Address 0xFA)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x00	Reserved

### 7.7.147 RESERVED Register

**表 7-163. RESERVED (Address 0xFB)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x00	Reserved

### 7.7.148 Indirect Access Registers

Several functional blocks include register sets contained in the Indirect Access map (表 7-164); that is, Pattern Generator, CSI-2 timing, and Analog controls. Register access is provided via an indirect access mechanism through the Indirect Access registers (IND\_ACC\_CTL, IND\_ACC\_ADDR, and IND\_ACC\_DATA). These registers are located at offsets 0xB0-0xB2 in the main register space.

The indirect address mechanism involves setting the control register to select the desired block, setting the register offset address, and reading or writing the data register. In addition, an auto-increment function is provided in the control register to automatically increment the offset address following each read or write of the data register.

For writes, the process is as follows:

1. Write to the IND\_ACC\_CTL register to select the desired register block
2. Write to the IND\_ACC\_ADDR register to set the register offset
3. Write the data value to the IND\_ACC\_DATA register

If auto-increment is set in the IND\_ACC\_CTL register, repeating step 3 will write additional data bytes to subsequent register offset locations

For reads, the process is as follows:

1. Write to the IND\_ACC\_CTL register to select the desired register block
2. Write to the IND\_ACC\_ADDR register to set the register offset
3. Read from the IND\_ACC\_DATA register

If auto-increment is set in the IND\_ACC\_CTL register, repeating step 3 will read additional data bytes from subsequent register offset locations.

### 7.7.149

**表 7-164. Indirect Register Map Description**

IA SELECT 0xB0[5:2]	PAGE/BLOCK	INDIRECT REGISTERS	ADDRESS RANGE	DESCRIPTION
0000	0	Digital Page 0 Indirect Registers	0x01-0x1F	Pattern Gen Registers
			0x40-0x48	CSI TX port 0 Timing Registers
0001	1	FPD-Link III Channel 0 Reserved Registers	0x00-0x14	Test and Debug registers
0010	2	Reserved	0x00-0x14	Reserved
0011	3	Reserved	0x00-0x14	Reserved
0100	4	Reserved	0x00-0x14	Reserved
0101	5	FPD-Link III Share Reserved Registers	0x00-0x04	Test and Debug registers
0110	6	Write All FPD-Link III Reserved Registers	0x00-0x14	Test and Debug registers
0111	7	CSI TX Reserved Registers	0x00-0x1D	Test and Debug registers
1001	9	Unique DIE ID Registers	0x00-0x0F	Hold 16 bytes that correspond to Die ID data.

### 7.7.150 Reserved Register

**表 7-165. Reserved (Indirect Address Page 0x00; Register 0x00)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	R	0x0	Reserved

### 7.7.151 PGEN\_CTL Register

**表 7-166. PGEN\_CTL (Indirect Address Page 0x00; Register 0x01)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:1	RESERVED	RW	0x0	Reserved
0	PGEN_ENABLE	RW	0	Pattern Generator Enable 1: Enable Pattern Generator 0: Disable Pattern Generator

### 7.7.152 PGEN\_CFG Register

**表 7-167. PGEN\_CFG (Indirect Address Page 0x00; Register 0x02)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	PGEN_FIXED_EN	RW	0	Fixed Pattern Enable Setting this bit enables Fixed Color Patterns. 0 : Send Color Bar Pattern 1 : Send Fixed Color Pattern
6	RESERVED	RW	0	Reserved
5:4	NUM_CBARS	RW	0x3	Number of Color Bars 00 : 1 Color Bar 01 : 2 Color Bars 10 : 4 Color Bars 11 : 8 Color Bars
3:0	BLOCK_SIZE	RW	0x3	Block Size For Fixed Color Patterns, this field controls the size of the fixed color field in bytes. Allowed values are 1 to 15.

### 7.7.153 PGEN\_CSI\_DI Register

**表 7-168. PGEN\_CSI\_DI (Indirect Address Page 0x00; Register 0x03)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:6	PGEN_CSI_VC	RW	0x0	CSI Virtual Channel Identifier This field controls the value sent in the CSI packet for the Virtual Channel Identifier
5:0	PGEN_CSI_DT	RW	0x24	CSI Data Type This field controls the value sent in the CSI packet for the Data Type. The default value (0x24) indicates RGB888.

### 7.7.154 PGEN\_LINE\_SIZE1 Register

**表 7-169. PGEN\_LINE\_SIZE1 (Indirect Address Page 0x00; Register 0x04)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_LINE_SIZE[15:8]	RW	0x07	Most significant byte of the Pattern Generator line size. This is the active line length in bytes. Default setting is for 1920 bytes for a 640 pixel line width.

### 7.7.155 PGEN\_LINE\_SIZE0 Register

**表 7-170. PGEN\_LINE\_SIZE0 (Indirect Address Page 0x00; Register 0x05)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_LINE_SIZE[7:0]	RW	0x80	Least significant byte of the Pattern Generator line size. This is the active line length in bytes. Default setting is for 1920 bytes for a 640 pixel line width.

### 7.7.156 PGEN\_BAR\_SIZE1 Register

**表 7-171. PGEN\_BAR\_SIZE1 (Indirect Address Page 0x00; Register 0x06)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_BAR_SIZE[15:8]	RW	0x0	Most significant byte of the Pattern Generator color bar size. This is the active length in bytes for the color bars. This value is used for all except the last color bar. The last color bar is determined by the remaining bytes as defined by the PGEN_LINE_SIZE value.

### 7.7.157 PGEN\_BAR\_SIZE0 Register

**表 7-172. PGEN\_BAR\_SIZE0 (Indirect Address Page 0x00; Register 0x07)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_BAR_SIZE[7:0]	RW	0xF0	Least significant byte of the Pattern Generator color bar size. This is the active length in bytes for the color bars. This value is used for all except the last color bar. The last color bar is determined by the remaining bytes as defined by the PGEN_LINE_SIZE value.

### 7.7.158 PGEN\_ACT\_LPF1 Register

**表 7-173. PGEN\_ACT\_LPF1 (Indirect Address Page 0x00; Register 0x08)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_ACT_LPF[15:8]	RW	0x01	Active Lines Per Frame Most significant byte of the number of active lines per frame. Default setting is for 480 active lines per frame.

### 7.7.159 PGEN\_ACT\_LPF0 Register

**表 7-174. PGEN\_ACT\_LPF0 (Indirect Address Page 0x00; Register 0x09)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_ACT_LPF[7:0]	RW	0xE0	Active Lines Per Frame Least significant byte of the number of active lines per frame. Default setting is for 480 active lines per frame.

### 7.7.160 PGEN\_TOT\_LPF1 Register

**表 7-175. PGEN\_TOT\_LPF1 (Indirect Address Page 0x00; Register 0x0A)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_TOT_LPF[15:8]	RW	0x02	Total Lines Per Frame Most significant byte of the number of total lines per frame including vertical blanking

### 7.7.161 PGEN\_TOT\_LPF0 Register

**表 7-176. PGEN\_TOT\_LPF0 (Indirect Address Page 0x00; Register 0x0B)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_TOT_LPF[7:0]	RW	0x0D	Total Lines Per Frame Least significant byte of the number of total lines per frame including vertical blanking

### 7.7.162 PGEN\_LINE\_PD1 Register

**表 7-177. PGEN\_LINE\_PD1 (Indirect Address Page 0x00; Register 0x0C)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_LINE_PD[15:8]	RW	0x0C	Line Period Most significant byte of the line period in 10ns units. The default setting for the line period registers sets a line period of 31.75 microseconds.

### 7.7.163 PGEN\_LINE\_PD0 Register

**表 7-178. PGEN\_LINE\_PD0 (Indirect Address Page 0x00; Register 0x0D)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_LINE_PD[7:0]	RW	0x67	Line Period Least significant byte of the line period in 10ns units. The default setting for the line period registers sets a line period of 31.75 microseconds.

### 7.7.164 PGEN\_VBP Register

**表 7-179. PGEN\_VBP (Indirect Address Page 0x00; Register 0x0E)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_VBP	RW	0x21	Vertical Back Porch This value provides the vertical back porch portion of the vertical blanking interval. This value provides the number of blank lines between the FrameStart packet and the first video data packet.

### 7.7.165 PGEN\_VFP Register

**表 7-180. PGEN\_VFP (Indirect Address Page 0x00; Register 0x0F)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_VFP	RW	0x0A	Vertical Front Porch This value provides the vertical front porch portion of the vertical blanking interval. This value provides the number of blank lines between the last video line and the FrameEnd packet.

### 7.7.166 PGEN\_COLOR0 Register

**表 7-181. PGEN\_COLOR0 (Indirect Address Page 0x00; Register 0x10)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR0	RW	0xAA	Pattern Generator Color 0 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 0. For Fixed Color Patterns, this register controls the first byte of the fixed color pattern.

### 7.7.167 PGEN\_COLOR1 Register

**表 7-182. PGEN\_COLOR1 (Indirect Address Page 0x00; Register 0x11)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR1	RW	0x33	Pattern Generator Color 1 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 1. For Fixed Color Patterns, this register controls the second byte of the fixed color pattern.

### 7.7.168 PGEN\_COLOR2 Register

**表 7-183. PGEN\_COLOR2 (Indirect Address Page 0x00; Register 0x12)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR2	RW	0xF0	Pattern Generator Color 2 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 2. For Fixed Color Patterns, this register controls the third byte of the fixed color pattern.

### 7.7.169 PGEN\_COLOR3 Register

**表 7-184. PGEN\_COLOR3 (Indirect Address Page 0x00; Register 0x13)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR3	RW	0x7F	Pattern Generator Color 3 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 3. For Fixed Color Patterns, this register controls the fourth byte of the fixed color pattern.

### 7.7.170 PGEN\_COLOR4 Register

**表 7-185. PGEN\_COLOR4 (Indirect Address Page 0x00; Register 0x14)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR4	RW	0x55	Pattern Generator Color 4 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 4. For Fixed Color Patterns, this register controls the fifth byte of the fixed color pattern.

### 7.7.171 PGEN\_COLOR5 Register

**表 7-186. PGEN\_COLOR5 (Indirect Address Page 0x00; Register 0x15)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR5	RW	0xCC	Pattern Generator Color 5 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 5. For Fixed Color Patterns, this register controls the sixth byte of the fixed color pattern.

### 7.7.172 PGEN\_COLOR6 Register

**表 7-187. PGEN\_COLOR6 (Indirect Address Page 0x00; Register 0x16)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR6	RW	0x0F	Pattern Generator Color 6 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 6. For Fixed Color Patterns, this register controls the seventh byte of the fixed color pattern.

### 7.7.173 PGEN\_COLOR7 Register

**表 7-188. PGEN\_COLOR7 (Indirect Address Page 0x00; Register 0x17)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR7	RW	0x80	Pattern Generator Color 7 For Reference Color Bar Patterns, this register controls the byte data value sent during color bar 7. For Fixed Color Patterns, this register controls the eighth byte of the fixed color pattern.

**7.7.174 PGEN\_COLOR8 Register****表 7-189. PGEN\_COLOR8 (Indirect Address Page 0x00; Register 0x18)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR8	RW	0x0	Pattern Generator Color 8 For Fixed Color Patterns, this register controls the ninth byte of the fixed color pattern.

**7.7.175 PGEN\_COLOR9 Register****表 7-190. PGEN\_COLOR9 (Indirect Address Page 0x00; Register 0x19)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR9	RW	0x0	Pattern Generator Color 9 For Fixed Color Patterns, this register controls the tenth byte of the fixed color pattern.

**7.7.176 PGEN\_COLOR10 Register****表 7-191. PGEN\_COLOR10 (Indirect Address Page 0x00; Register 0x1A)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR10	RW	0x0	Pattern Generator Color 10 For Fixed Color Patterns, this register controls the eleventh byte of the fixed color pattern.

**7.7.177 PGEN\_COLOR11 Register****表 7-192. PGEN\_COLOR11 (Indirect Address Page 0x00; Register 0x1B)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR11	RW	0x0	Pattern Generator Color 11 For Fixed Color Patterns, this register controls the twelfth byte of the fixed color pattern.

**7.7.178 PGEN\_COLOR12 Register****表 7-193. PGEN\_COLOR12 (Indirect Address Page 0x00; Register 0x1C)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR12	RW	0x0	Pattern Generator Color 12 For Fixed Color Patterns, this register controls the thirteenth byte of the fixed color pattern.

**7.7.179 PGEN\_COLOR13 Register****表 7-194. PGEN\_COLOR13 (Indirect Address Page 0x00; Register 0x1D)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR13	RW	0x0	Pattern Generator Color 13 For Fixed Color Patterns, this register controls the fourteenth byte of the fixed color pattern.

**7.7.180 PGEN\_COLOR14 Register****表 7-195. PGEN\_COLOR14 (Indirect Address Page 0x00; Register 0x1E)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	PGEN_COLOR14	RW	0x0	Pattern Generator Color 14 For Fixed Color Patterns, this register controls the fifteenth byte of the fixed color pattern.



## 7.7.181 RESERVED Register

**表 7-196. RESERVED (Indirect Address Page 0x00; Register 0x1F)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7:0	RESERVED	RW	0x0	Reserved

## 7.7.182 CSI0\_TCK\_PREP Register

**表 7-197. CSI0\_TCK\_PREP (Indirect Address Page 0x00; Register 0x40)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_TCK_PREP_OV	RW	0	Override CSI Tck-prep parameter 0: Tck-prep is automatically determined 1: Override Tck-prep with value in bits 6:0 of this register
6:0	MR_TCK_PREP	R RW	0x0	Tck-prep value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.183 CSI0\_TCK\_ZERO Register

**表 7-198. CSI0\_TCK\_ZERO (Indirect Address Page 0x00; Register 0x41)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_TCK_ZERO_OV	RW	0	Override CSI Tck-zero parameter 0: Tck-zero is automatically determined 1: Override Tck-zero with value in bits 6:0 of this register
6:0	MR_TCK_ZERO	R RW	0x0	Tck-zero value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.184 CSI0\_TCK\_TRAIL Register

**表 7-199. CSI0\_TCK\_TRAIL (Indirect Address Page 0x00; Register 0x42)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_TCK_TRAIL_OV	RW	0	Override CSI Tck-trail parameter 0: Tck-trail is automatically determined 1: Override Tck-trail with value in bits 6:0 of this register
6:0	MR_TCK_TRAIL	R RW	0x0	Tck-trail value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.185 CSI0\_TCK\_POST Register

**表 7-200. CSI0\_TCK\_POST (Indirect Address Page 0x00; Register 0x43)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_TCK_POST_OV	RW	0	Override CSI Tck-post parameter 0: Tck-post is automatically determined 1: Override Tck-post with value in bits 6:0 of this register
6:0	MR_TCK_POST	R RW	0x0	Tck-post value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

**7.7.186 CSI0\_THS\_PREP Register****表 7-201. CSI0\_THS\_PREP (Indirect Address Page 0x00; Register 0x44)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_THS_PREP_OV	RW	0	Override CSI Ths-prep parameter 0: Ths-prep is automatically determined 1: Override Ths-prep with value in bits 6:0 of this register
6:0	MR_THS_PREP	R RW	0x0	Ths-prep value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

**7.7.187 CSI0\_THS\_ZERO Register****表 7-202. CSI0\_THS\_ZERO (Indirect Address Page 0x00; Register 0x45)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_THS_ZERO_OV	RW	0	Override CSI Ths-zero parameter 0: Ths-zero is automatically determined 1: Override Ths-zero with value in bits 6:0 of this register
6:0	MR_THS_ZERO	R RW	0x0	Ths-zero value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

**7.7.188 CSI0\_THS\_TRAIL Register****表 7-203. CSI0\_THS\_TRAIL (Indirect Address Page 0x00; Register 0x46)**

BITS	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_THS_TRAIL_OV	RW	0	Override CSI Ths-trail parameter 0: Ths-trail is automatically determined 1: Override Ths-trail with value in bits 6:0 of this register
6:0	MR_THS_TRAIL	R RW	0x0	Ths-trail value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.189 CSI0\_THS\_EXIT Register

**表 7-204. CSI0\_THS\_EXIT (Indirect Address Page 0x00; Register 0x47)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_THS_EXIT_OV	RW	0	Override CSI Ths-exit parameter 0: Ths-exit is automatically determined 1: Override Ths-exit with value in bits 6:0 of this register
6:0	MR_THS_EXIT	R RW	0x0	Ths-exit value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.190 CSI0\_TPLX Register

**表 7-205. CSI0\_TPLX (Indirect Address Page 0x00; Register 0x48)**

BIT	FIELD	TYPE	DEFAULT	DESCRIPTION
7	MR_TPLX_OV	RW	0	Override CSI Tplx parameter 0: Tplx is automatically determined 1: Override Tplx with value in bits 6:0 of this register
6:0	MR_TPLX	R RW	0x0	Tplx value If bit 7 of this register is 0, this field is read-only, indicating current automatically determined value. If bit 7 of this register is 1, this field is read/write.

## 7.7.191

### LEGEND:

- RW = Read Write
- RW/SC = RW/SC = Read Write access/Self Clearing bit
- R/P = Read Only, Permanent value
- R/COR = Read Only, Clear On Read

## 8 Application and Implementation

### 注

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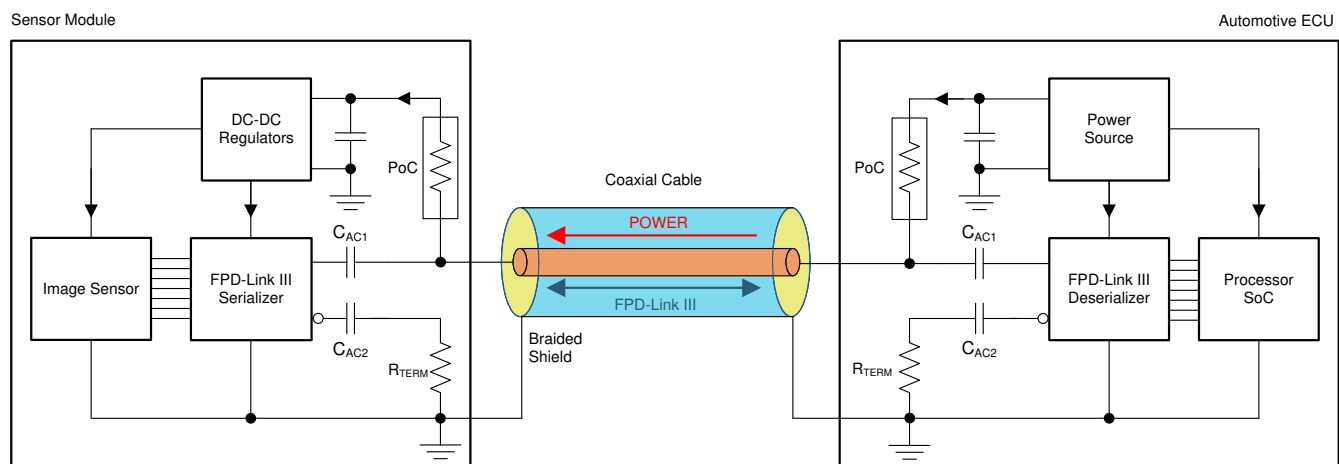
### 8.1 Application Information

#### 8.1.1 System

The DS90UB638-Q1 is a highly integrated sensor chip that includes a single FPD-Link III input targeted at ADAS applications, such as front, rear, and surround view cameras, camera monitoring systems, and sensor fusion.

#### 8.1.2 Power-over-Coax

The DS90UB638-Q1 is designed to support the Power-over-Coax (PoC) method of powering remote sensor systems. With this method, the power is delivered over the same medium (a coaxial cable) used for high-speed digital video data and bidirectional control and diagnostics data transmission. The method uses passive networks or filters that isolate the transmission line from the loading of the DC/DC regulator circuits and their connecting power traces on both sides of the link as shown in [Figure 8-1](#).

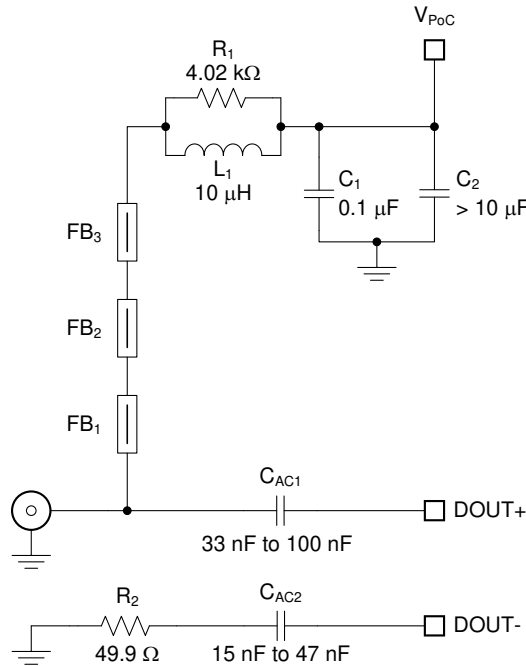


**図 8-1. Power-over-Coax (PoC) System Diagram**

The PoC networks' impedance of  $\geq 1 \text{ k}\Omega$  over a specific frequency band is recommended to isolate the transmission line from the loading of the regulator circuits. Higher PoC network impedance will contribute to favorable insertion loss and return loss characteristics in the high-speed channel. The lower limit of the frequency band is defined as  $\frac{1}{2}$  of the frequency of the back channel,  $f_{BC}$ . The upper limit of the frequency band is the frequency of the forward high-speed channel,  $f_{FC}$ . However, the main criteria that need to be met in the total high-speed channel, which consists of a serializer PCB, a deserializer PCB, and a cable, are the insertion loss and return loss limits defined in the Total Channel Requirements, while the system is under maximum current load and extreme temperature conditions [List item. List item.](#)

1. Contact TI for more information on the required Channel Specifications defined for each individual FPD-Link device.
2. The PoC network and any components along the high-speed trace on the PCB will contribute to the PCB loss budget. TI has recommendations for the loss budget allocation for each individual PCB and cable component in the overall high-speed channel, but the loss limits defined for the total channel in the Channel Specifications must be met.

8-2 shows a PoC network recommended for a 4G FPD-Link III consisting of DS90UB63x-Q1 CSI-2 and DS90UB638-Q1 pair with the bidirectional channel operating at 50 Mbps ( $f_{BCC} = 50$  MHz) and the forward channel operating at 4.16 Gbps ( $f_{FC} \approx 2$  GHz).



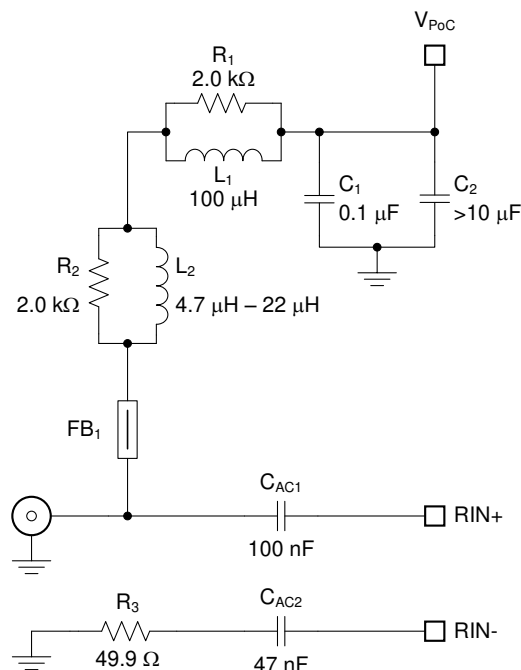
**8-2. Typical PoC Network for a 4 Gbps FPD-Link III**

表 8-1 lists essential components for this particular PoC network.

**表 8-1. Suggested Components for a 4 Gbps FPD-Link PoC Network**

COUNT	REF DES	DESCRIPTION	PART NUMBER	MFR
1	L1	Inductor, 10 $\mu$ H, 0.288 $\Omega$ maximum, 530 mA minimum (Isat, Itemp) 30-MHz SRF min, 3 mm $\times$ 3 mm, General-Purpose	LQH3NPN100MJR	Murata
		Inductor, 10 $\mu$ H, 0.288 $\Omega$ maximum, 530 mA minimum (Isat, Itemp) 30-MHz SRF min, 3 mm $\times$ 3 mm, AEC-Q200	LQH3NPZ100MJR	Murata
		Inductor, 10 $\mu$ H, 0.360 $\Omega$ maximum, 450 mA minimum (Isat, Itemp) 30-MHz SRF min, 3.2 mm $\times$ 2.5 mm, AEC-Q200	NLCV32T-100K-EFD	TDK
		Inductor, 10 $\mu$ H, 0.400 $\Omega$ typical, 550 mA minimum (Isat, Itemp) 39-MHz SRF typ, 3 mm $\times$ 3 mm, AEC-Q200	TYS3010100M-10	Laird
		Inductor, 10 $\mu$ H, 0.325 $\Omega$ maximum, 725 mA minimum (Isat, Itemp) 41-MHz SRF typ, 3 mm $\times$ 3 mm, AEC-Q200	TYS3015100M-10	Laird
3	FB1-FB3	Ferrite Bead, 1.5 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500 mA at 85°C, SM0603, General Purpose	BLM18HE152SN1	Murata
		Ferrite Bead, 1.5 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500 mA at 85°C, SM0603, AEC-Q200	BLM18HE152SZ1	Murata

8-3 shows a PoC network recommended for a 2G FPD-Link III consisting of a DS90UB633A-Q1 serializer and DS90UB638-Q1 with the bidirectional channel operating at the data rate of 2.5 Mbps ( $\frac{1}{2} f_{BCC} = 2.5$  MHz) and the forward channel operating at the data rate as high as 1.87 Gbps ( $f_{FC} \approx 1$  GHz).



**图 8-3. Typical PoC Network for a 2G FPD-Link III**

表 8-2 lists essential components for this particular PoC network.

**表 8-2. Suggested Components for a 2G FPD-Link III PoC Network**

COUNT	REF DES	DESCRIPTION	PART NUMBER	MFR
1	L1	Inductor, 100 $\mu$ H, 0.310 $\Omega$ maximum, 710 mA minimum (Isat, Itemp) 7.2-MHz SRF typical, 6.6 mm $\times$ 6.6 mm, AEC-Q200	MSS7341-104ML	Coilcraft
1	L2	Inductor, 4.7 $\mu$ H, 0.350 $\Omega$ maximum, 700 mA minimum (Isat, Itemp) 160-MHz SRF typical, 3.8 mm $\times$ 3.8 mm, AEC-Q200	1008PS-472KL	Coilcraft
		Inductor, 4.7 $\mu$ H, 0.130 $\Omega$ maximum, 830 mA minimum (Isat, Itemp), 70-MHz SRF typical, 3.2 mm $\times$ 2.5 mm, AEC-Q200	CBC3225T4R7MRV	Taiyo Yuden
1	FB1	Ferrite Bead, 1.5 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500-mA at 85°C, SM0603, General-Purpose	BLM18HE152SN1	Murata
		Ferrite Bead, 1.5 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500-mA at 85°C, SM0603, AEC-Q200	BLM18HE152SZ1	Murata

Application report [Power-over-Coax Design Guidelines for DS90UB953-Q1](#) (SNLA272) discusses PoC networks in more detail.

In addition to the PoC network components selection, their placement and layout play a critical role as well.

- Place the smallest component, typically a ferrite bead or a chip inductor, as close to the connector as possible. Route the high-speed trace through one of its pads to avoid stubs.
- Use the smallest component pads as allowed by manufacturer's design rules. Add anti-pads in the inner planes below the component pads to minimize impedance drop.

- Consult with the connector manufacturer for the optimized connector footprint. If the connector is mounted on the same side as the IC, minimize the impact of the through-hole connector stubs by routing the high-speed signal traces on the opposite side of the connector mounting side.
- Use coupled 100-Ω differential signal traces from the device pins to the AC-coupling caps. Use 50-Ω single-ended traces from the AC-coupling capacitors to the connector.
- Terminate the inverting signal traces close to the connectors with standard 49.9-Ω resistors.

The suggested characteristics for single-ended PCB traces (microstrips or striplines) for serializer or deserializer boards are detailed in 表 8-3. The effects of the PoC networks must be accounted for when testing the traces for compliance to the suggested limits.

**表 8-3. Suggested Characteristics for Single-Ended PCB Traces With Attached PoC Networks**

PARAMETER		MIN	TYP	MAX	UNIT
$L_{\text{trace}}$	Single-ended PCB trace length from the device pin to the connector pin			5	cm
$Z_{\text{trace}}$	Single-ended PCB trace characteristic impedance	45	50	55	Ω
$Z_{\text{con}}$	Connector (mounted) characteristic impedance	40	50	60	Ω
$t_{\Delta Z_{\text{con}}}$	Allowable electrical length of the connector impedance discontinuity as measured with a TDR (100 ps edge)			20	ps

The  $V_{\text{POC}}$  noise must be kept 10 mVp-p or lower on the source / deserializer side of the system. The  $V_{\text{POC}}$  fluctuations on the serializer side, caused by the transient current draw of the sensor and the DC resistance of cables and PoC components, must be kept at minimum as well. Increasing the  $V_{\text{POC}}$  voltage and adding extra decoupling capacitance ( $> 10 \mu\text{F}$ ) help reduce the amplitude and slew rate of the  $V_{\text{POC}}$  fluctuations.

## 8.2 Typical Application

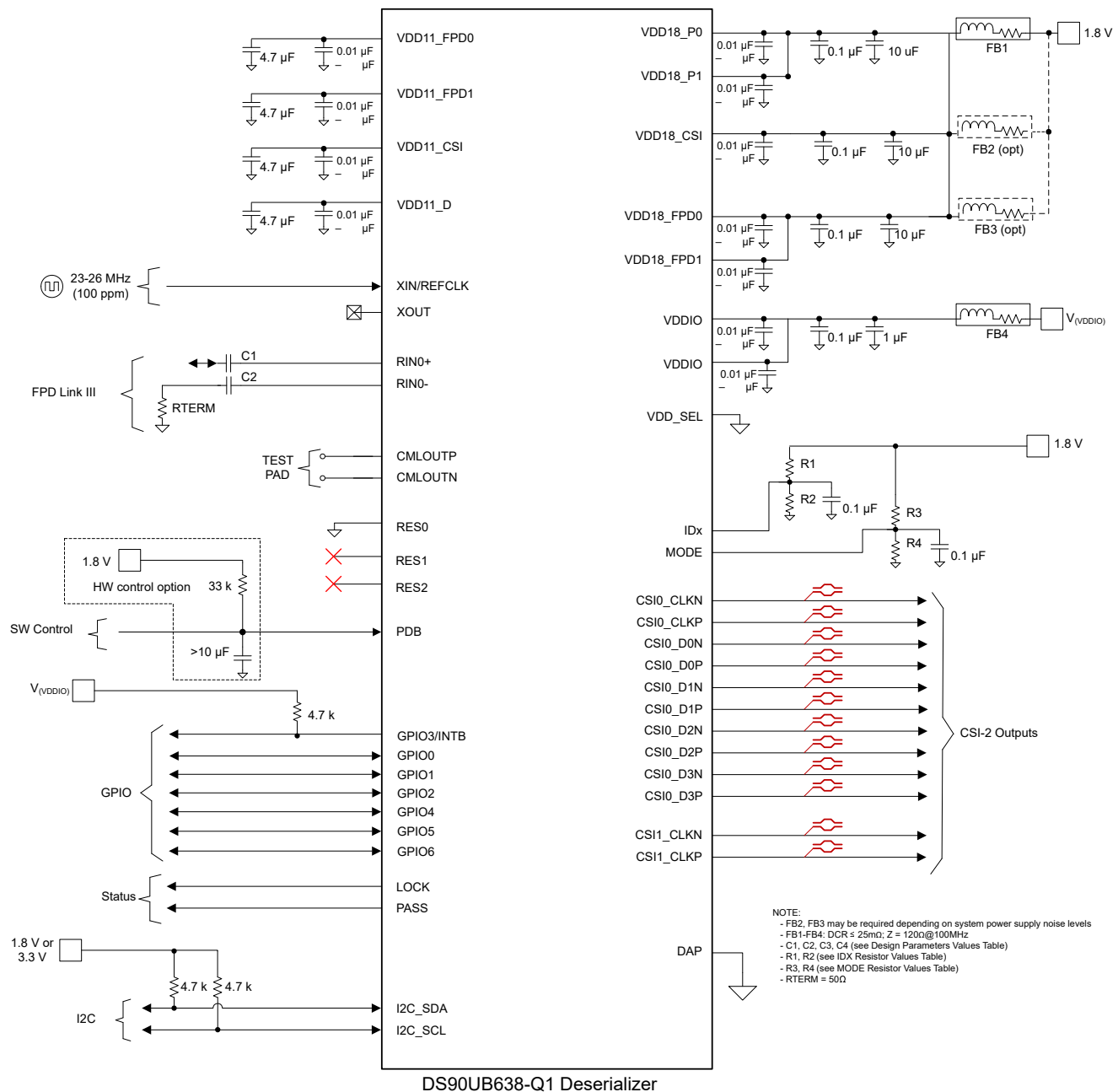
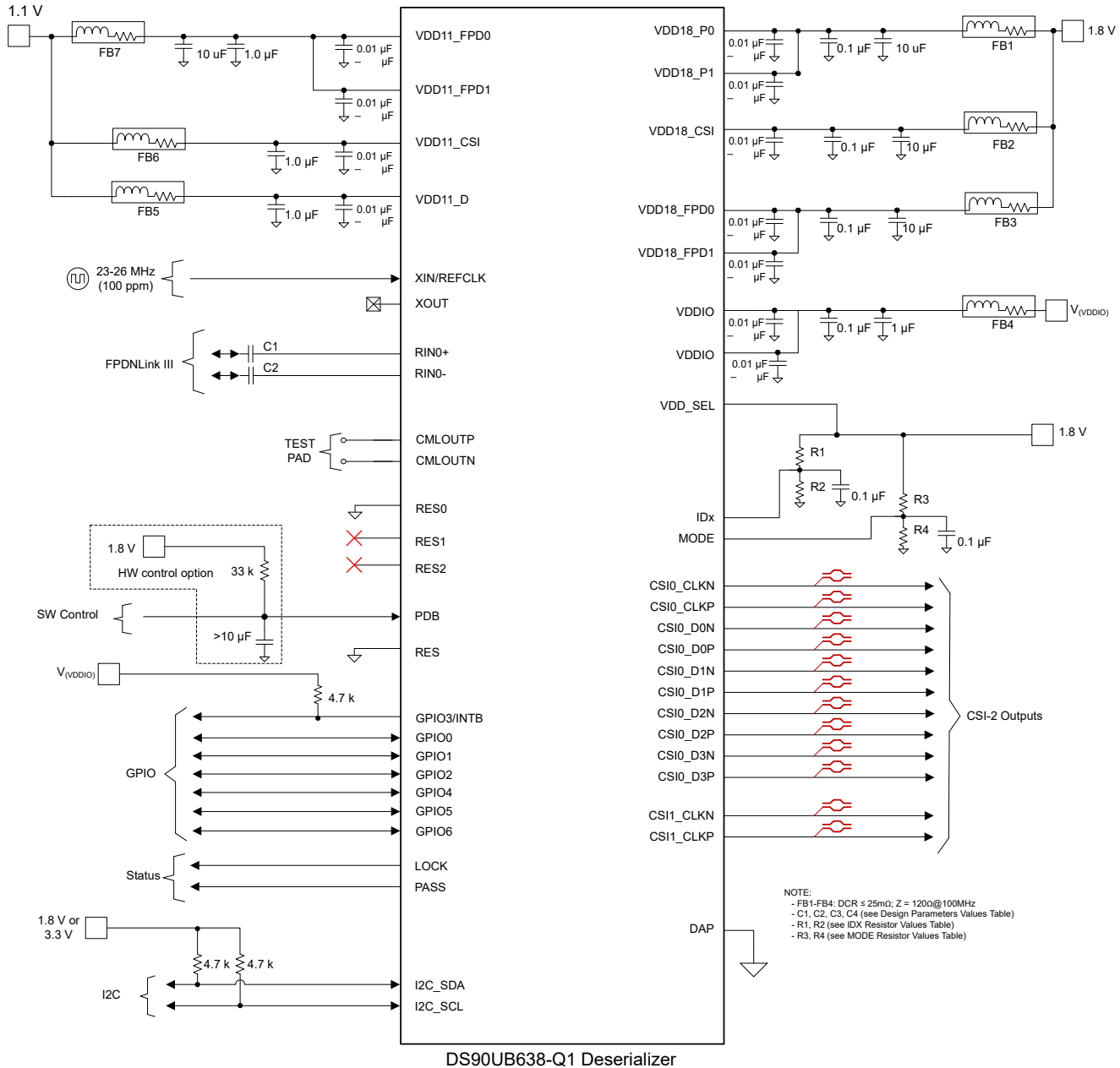


図 8-4. Typical Connection Diagram Coaxial With Internal 1.1-V LDO

注

- The decoupling capacitors for VDD11 are different between the two typical application diagrams because VDD\_SEL is pulled to different levels. See the セクション 5 table for more information.





**図 8-5. Typical Connection Diagram With External 1.1-V Supply**

注

- The decoupling capacitors for VDD11 are different between the two typical application diagrams because VDD\_SEL is pulled to different levels. See the [セクション 5](#) table for more information.

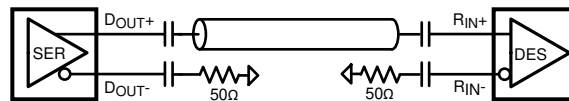
## 8.2.1 Design Requirements

For the typical design application, use the parameters listed in 表 8-4.

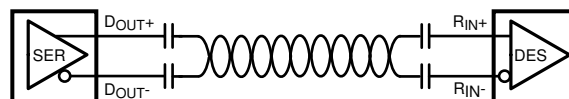
**表 8-4. Design Parameters**

DESIGN PARAMETER	EXAMPLE VALUE
$V_{(VDDIO)}$	1.8 V or 3.3 V
$V_{(VDD18)}$	1.8 V
$V_{(VDD11)}$ ( When VDD_SEL = HIGH)	1.1 V
AC-coupling Capacitor for Synchronous Modes, Coaxial Connection: RIN+	33 nF - 100nF (50 WV 0402)
AC-coupling Capacitor for Synchronous Modes, Coaxial Connection: RIN-	15 nF - 47nF (50 WV 0402)
AC-coupling Capacitor for Synchronous Modes, STP Connection: RIN±	33 nF - 100nF (50 WV 0402)
AC-coupling Capacitor for Non-Synchronous and DVP Backwards Compatible Modes, Coaxial Connection: RIN+	100 nF (50 WV 0402)
AC-coupling Capacitor for Non-Synchronous and DVP Backwards Compatible Modes, Coaxial Connection: RIN-	47 nF (50 WV 0402)
AC-coupling Capacitor for Non-Synchronous and DVP Backwards Compatible Modes, STP Connection: RIN±	100 nF (50 WV 0402)

The SER/DES supports only AC-coupled interconnects through an integrated DC-balanced decoding scheme. External AC-coupling capacitors must be placed in series in the FPD-Link III signal path as shown in 图 8-6. For applications using single-ended, 50-Ω coaxial cable, terminate the unused data pins (RIN-) with an AC-coupling capacitor and a 50-Ω resistor.



**图 8-6. AC-Coupled Connection (Coaxial)**



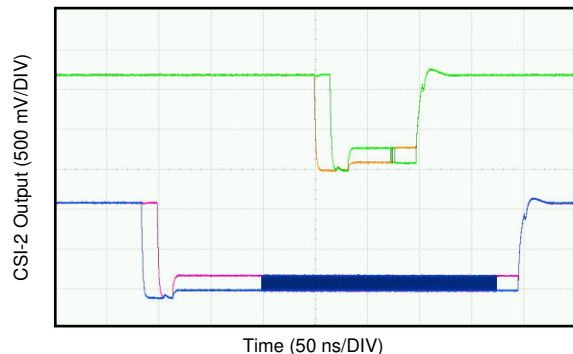
**图 8-7. AC-Coupled Connection (STP)**

For high-speed FPD-Link III transmissions, use the smallest available package for the AC-coupling capacitor to help minimize degradation of signal quality due to package parasitics.

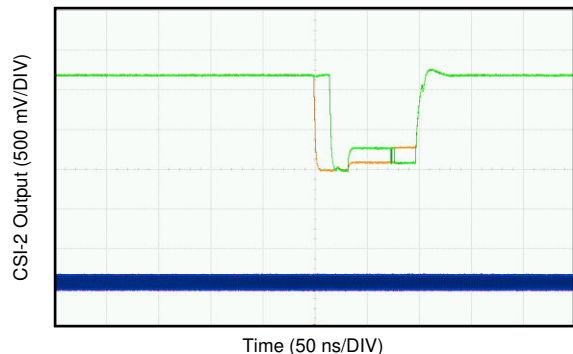
## 8.2.2 Detailed Design Procedure

图 8-4 and 图 8-5 show typical applications of the DS90UB638-Q1 for multi-camera surround view system. From 图 8-4, the FPD-Link III is AC-coupled external 100-nF and 47-nF capacitors for coaxial interconnects. The same AC-coupling capacitor values should be matched on the paired serializer boards. The deserializer has an internal termination. Bypass capacitors are placed near the power supply pins. At a minimum, 0.1-μF or 0.01-μF capacitors should be used for each of the core supply pins for local device bypassing. Additional bulk decoupling capacitors and ferrite beads are placed on the VDD18 supplies for effective noise suppression.

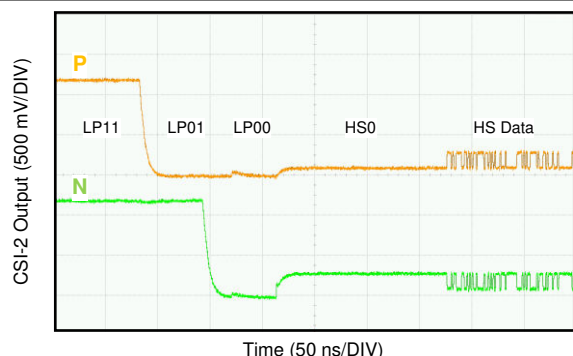
### 8.2.3 Application Curves



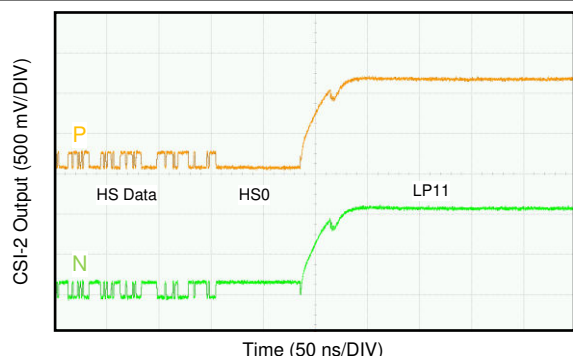
8-8. CSI-2 DATA and CLK Output



8-9. CSI-2 DATA and Continuous CLK Output



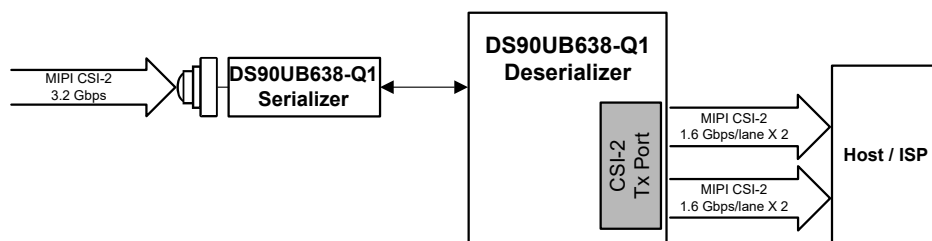
8-10. CSI-2 Start of Transmission (SoT)



8-11. CSI-2 End of Transmission (EoT)

### 8.3 System Examples

The DS90UB638-Q1 has one input port that is capable of operating a single sensor. (8-12). Rx data can be replicated onto two 2-lane CSI-2 outputs for interconnect to two separate CSI-2 Rx inputs for parallel downstream processing.



8-12. DS90UB638-Q1 Sensor Data Replicated onto 2x 2-Lane CSI-2

## 9 Power Supply Recommendations

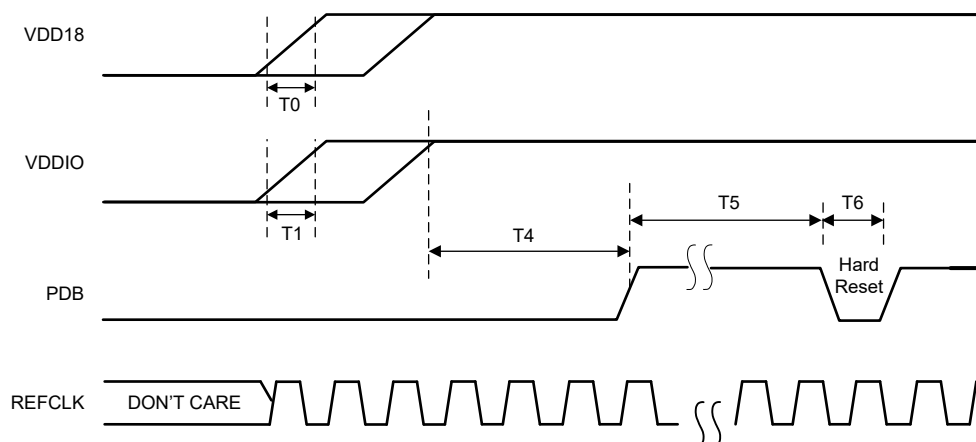
This device provides separate power and ground pins for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. The [Pin Configuration and Functions](#) section provides guidance on which circuit blocks are connected to which power pin pairs. In some cases, an external filter may be used to provide clean power to sensitive circuits such as PLLs.

### 9.1 VDD and VDDIO Power Supply

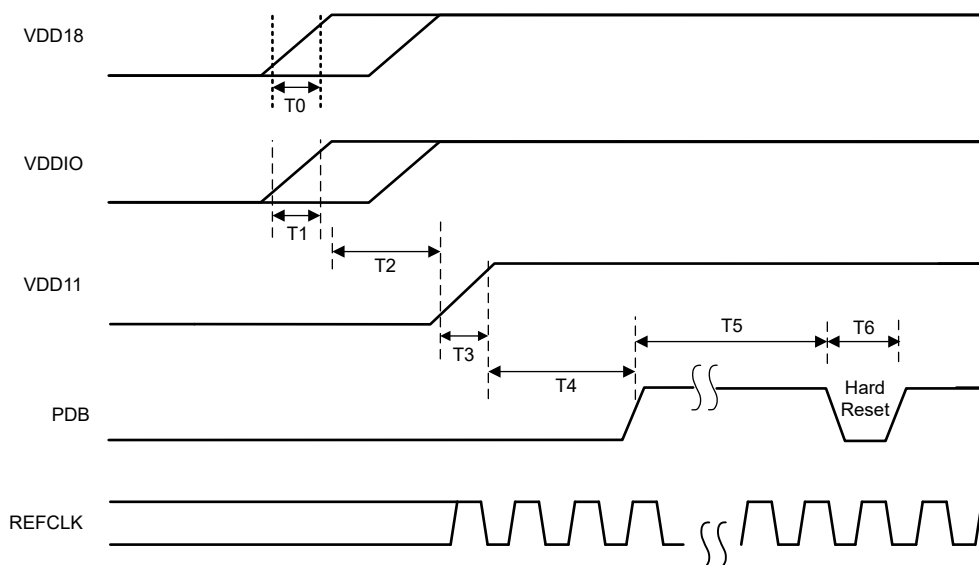
Each VDD power supply pin must have a 10-nF (or 100-nF) capacitor to ground connected as close as possible to DS90UB638-Q1 device. When operating VDDIO at 1.8-V nominal supply, the voltage at VDDIO must be within  $\pm 100$  mV of VDD18 to ensure VIH, VIL specifications. TI recommends having additional decoupling capacitors (1  $\mu$ F or 10  $\mu$ F) connected to a common GND plane. Note that although average current for VDDIO is less than 10 mA maximum, the peak current into VDDIO may exceed 100 mA on device start-up.

### 9.2 Power-Up Sequencing

The power-up sequence for the DS90UB638-Q1 is as follows:



**Figure 9-1. Power Supply Sequencing VDD\_SEL = LOW, Internal VDD 1.1-V Supply**



**Figure 9-2. Power Supply Sequencing VDD\_SEL = HIGH, External VDD 1.1-V Supply**

**表 9-1. Timing Diagram for the Power Supply Start-Up Sequence**

	PARAMETER	MIN	TYP	MAX	UNIT	NOTES
T0	VDD18 rise time	0.05			ms	at 10/90%
T1	VDDIO rise time	0.2	1		ms	at 10/90%
T2	VDD18 High to VDD11 applied	0			ms	N/A when VDD_SEL = LOW
T3	VDD11 rise time	0.2	1		ms	at 10/90%
T4	VDD to PDB	0			ms	After all VDD are stable
T5	PDB high time before PDB hard reset	1			ms	
T6	PDB high to low pulse width	2			ms	Hard reset (optional)
T7	PDB to I2C ready (IDX and MODE valid) delay	2			ms	

**Note:** VDDIO can come up either before or after VDD18.

### 9.2.1 PDB Pin

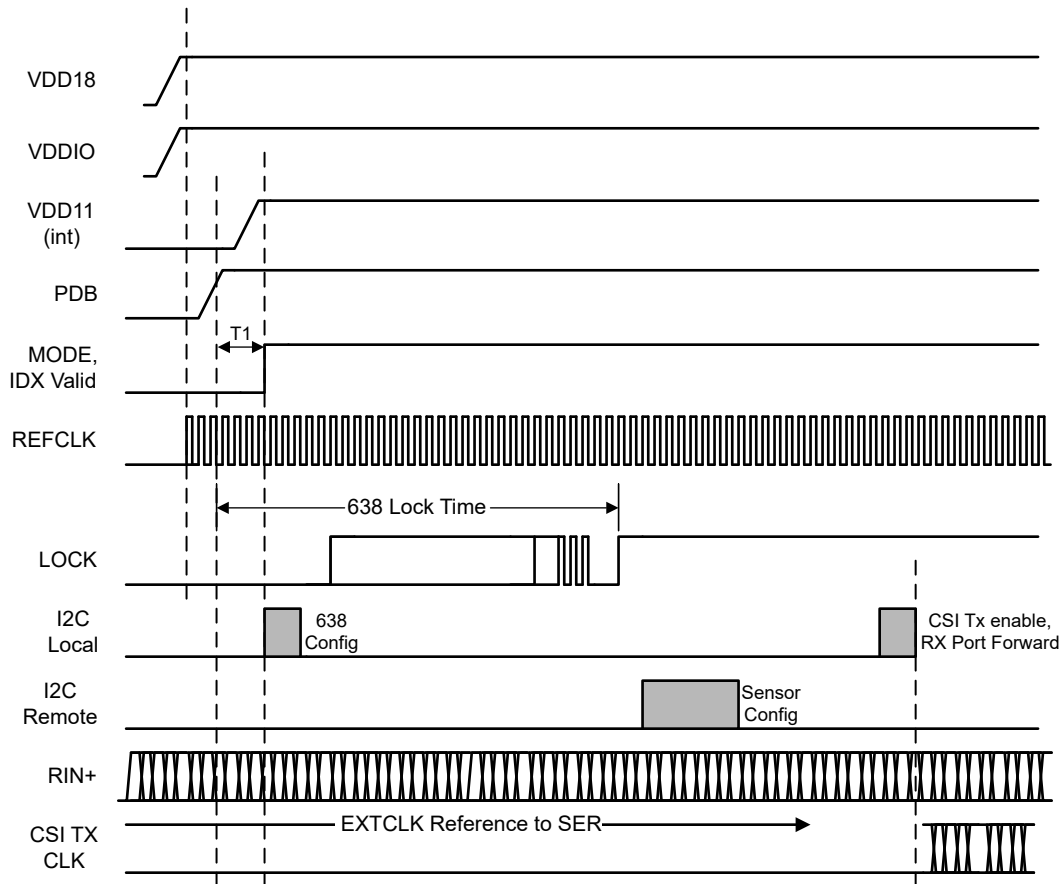
The PDB pin is active high and has internal 50-kΩ pulldown resistor. PDB input must remain low while the VDD pin power supplies are in transition. Typically PDB will be connected to GPIO from processor also with internal pulldown. Alternatively, when VDD\_SEL = LOW, an external RC network on the PDB pin may be connected to ensure PDB arrives after all the supply pins have settled to the recommended operating voltage. When PDB pin is pulled up to VDD18, a 10-kΩ pullup and a > 10-μF capacitor to GND are recommended to delay the PDB input signal rise. All inputs must not be driven until both power supplies have reached steady state. When VDD\_SEL = HIGH, TI does not recommend that the user connect the PDB pin through a RC circuit, as this may conflict with the sequencing of the external 1.1-V supply rail.

**表 9-2. PDB Pin Pulse Width**

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>PDB</b>						
tLRST	PDB Reset Low Pulse		2	3		ms

### 9.2.2 System Initialization

When initializing the communications link between the DS90UB638-Q1 deserializer and a DS90UB635-Q1 serializer, the sensor module requires a local reference clock and timing would follow [Figure 9-3](#).



**FIG 9-3. Power-Up Sequencing Non-Synchronous Back Channel Clocking Mode, VDD\_SEL = LOW**

Note that in order to speed up access to remote target I2C devices attached to the DS90UB635-Q1 from the DS90UB638-Q1 deserializer device, it is recommended to program the watchdog timer speedup by setting 0x0A = 0x12 after LOCK is established with the remote deserializer.

## 10 Layout

### 10.1 PCB Layout Guidelines

Circuit board layout and stack-up for the FPD-Link III devices must be designed to provide low-noise power feed to the device. Good layout practice also separates high-frequency or high-level inputs and outputs to minimize unwanted noise pickup, feedback, and interference. Power system performance may be greatly improved by using thin dielectrics (2 to 4 mils) for power or ground sandwiches. This arrangement provides plane capacitance for the PCB power system with low-inductance parasitics, which has proven especially effective at high frequencies, and makes the value and placement of external bypass capacitors less critical. External bypassing should be low-ESR ceramic capacitors with high-quality dielectric. The voltage rating of the ceramic capacitors must be at least 2× the power supply voltage being used.

TI recommends surface-mount capacitors due to their smaller parasitics. When using multiple capacitors per supply pin, place the smaller value closer to the pin. A large bulk capacitor is recommended at the point of power entry. This is typically in the 47-μF to 100-μF range, which smooths low frequency switching noise. TI recommends connecting power and ground pins directly to the power and ground planes with bypass capacitors connected to the plane with via on both ends of the capacitor. Connecting power or ground pins to an external bypass capacitor increases the inductance of the path.

A small body size X7R chip capacitor, such as 0603 or 0402, is recommended for external bypass. Its small body size reduces the parasitic inductance of the capacitor. The user must pay attention to the resonance frequency of these external bypass capacitors, usually in the range of 20 to 30 MHz. To provide effective bypassing, multiple capacitors are often used to achieve low impedance between the supply rails over the frequency of interest. At high frequency, it is also a common practice to use two vias from power and ground pins to the planes, reducing the impedance at high frequency.


Some devices provide separate power and ground pins for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. Pin Description tables typically provide guidance on which circuit blocks are connected to which power pin pairs (see [Pin Configuration and Functions](#)). In some cases, an external filter may be used to provide clean power to sensitive circuits such as PLLs.

Use at least a four-layer board with a power and ground plane. Place the CSI-2 signals away from the single-ended or differential FPD RX input traces to prevent coupling from the CSI-2 signals to the RX inputs. The following sections provide important details for routing the FPD-Link III traces, PoC filter, and CSI-2 traces.

#### 10.1.1 Ground

TI recommends that a consistent ground plane reference for the high-speed signals in the PCB design to provide the best image plane for signal traces running parallel to the plane. Connect the thermal pad of the DS90UB638-Q1 to the GND plane with vias.

#### 10.1.2 Routing FPD-Link III Signal Traces and PoC Filter

Routing the FPD-Link III signal traces between the  $R_{IN}$  pins and the connector as well as connecting the PoC filter to these traces are the most critical pieces of a successful DS90UB638-Q1 PCB layout.  10-1 shows an example PCB layout of the DS90UB638-Q1 configured for interface to remote sensor modules over coaxial cables. The layout example also uses a footprint of an edge-mount FAKRA connector provided by Rosenberger (P/N: 59S20X-40ML5-Z). For additional PCB layout details of the example, check the DS90UB954-Q1EVM user's guide. Note that the DS90UB638-Q1 shares this user guide with other related products such as the DS90UB954-Q1.

The following list provides essential recommendations for routing the FPD-Link III signal traces between the DS90UB638-Q1 receiver input pins ( $R_{IN}$ ) and the FAKRA connector, and connecting the PoC filter.

- The routing of the FPD-Link III traces may be all on the top layer (as shown in the example) or partially embedded in middle layers if EMI is a concern.
- The AC-coupling capacitors should be on the top layer and very close to the DS90UB638-Q1 receiver input pins to minimize the length of coupled differential trace pair between the pins and the capacitors.

- Route the RIN+ trace between the AC-coupling capacitor and the FAKRA connector as a 50- $\Omega$  single-ended micro-strip with tight impedance control ( $\pm 10\%$ ). Calculate the proper width of the trace for a 50- $\Omega$  impedance based on the PCB stack-up. Ensure that the trace can carry the PoC current for the maximum load presented by the remote sensor module.
- The PoC filter should be connected to the RIN+ trace through the first ferrite bead (FB<sub>1</sub>). The FB<sub>1</sub> should be touching the high-speed trace to minimize the stub length seen by the transmission line. Create an anti-pad or a moat under the FB<sub>1</sub> pad that touches the trace. The anti-pad should be a plane cutout of the ground plane directly underneath the top layer without cutting out the ground reference under the trace. The purpose of the anti-pad is to maintain the impedance as close to 50  $\Omega$  as possible.
- Route the RIN– trace loosely coupled to the RIN+ trace for the length similar to the RIN+ trace length when possible. This will help the differential nature of the receiver to cancel out any common-mode noise that may be present in the environment that may couple on to the RIN+ and RIN– signal traces. When routing on inner layers, length matching for single-ended traces does not provide as significant benefit.



### 10.1.3 Routing CSI-2 Signal Traces

Routing the CSI-2 signal traces between the CSI-2 pins and the CSI-2 connector is also important for a successful DS90UB638-Q1 PCB layout. [Figure 10-3](#) shows essential details for routing the CSI-2 traces. Additional recommendations are given in the following list:

1. Route CSI\_D0N, CSI\_D0P, CSI\_D1N, and CSI\_D1P pairs as differential coupled striplines with controlled 100- $\Omega$  differential impedance ( $\pm 10\%$ )
2. Keep the trace length difference between CSI-2 traces to 5 mils of each other.
3. Length matching should be near the location of mismatch.
4. Each pair should be separated at least by 5 times the signal trace width.
5. Keep away from other high-speed signals.
6. Keep the use of bends in differential traces to a minimum. When bends are used, the number of left and right bends must be as equal as possible, and the angle of the bend should be  $\geq 135$  degrees. This arrangement minimizes any length mismatch caused by the bends and therefore minimizes the impact that bends have on EMI.
7. Route all differential pairs on one or two inner layers.
8. Keep the number of signal vias to a minimum — TI recommends keeping the via count to the maximum of two per CSI-2 trace.
9. Keep traces on layers adjacent to ground plane.
10. Do NOT route differential pairs over any plane split.
11. Adding Test points causes impedance discontinuity and therefore negatively impacts signal performance. If test points are used, place them in series and symmetrically. Test points must not be placed in a manner that causes a stub on the differential pair.

## 10.2 Layout Examples

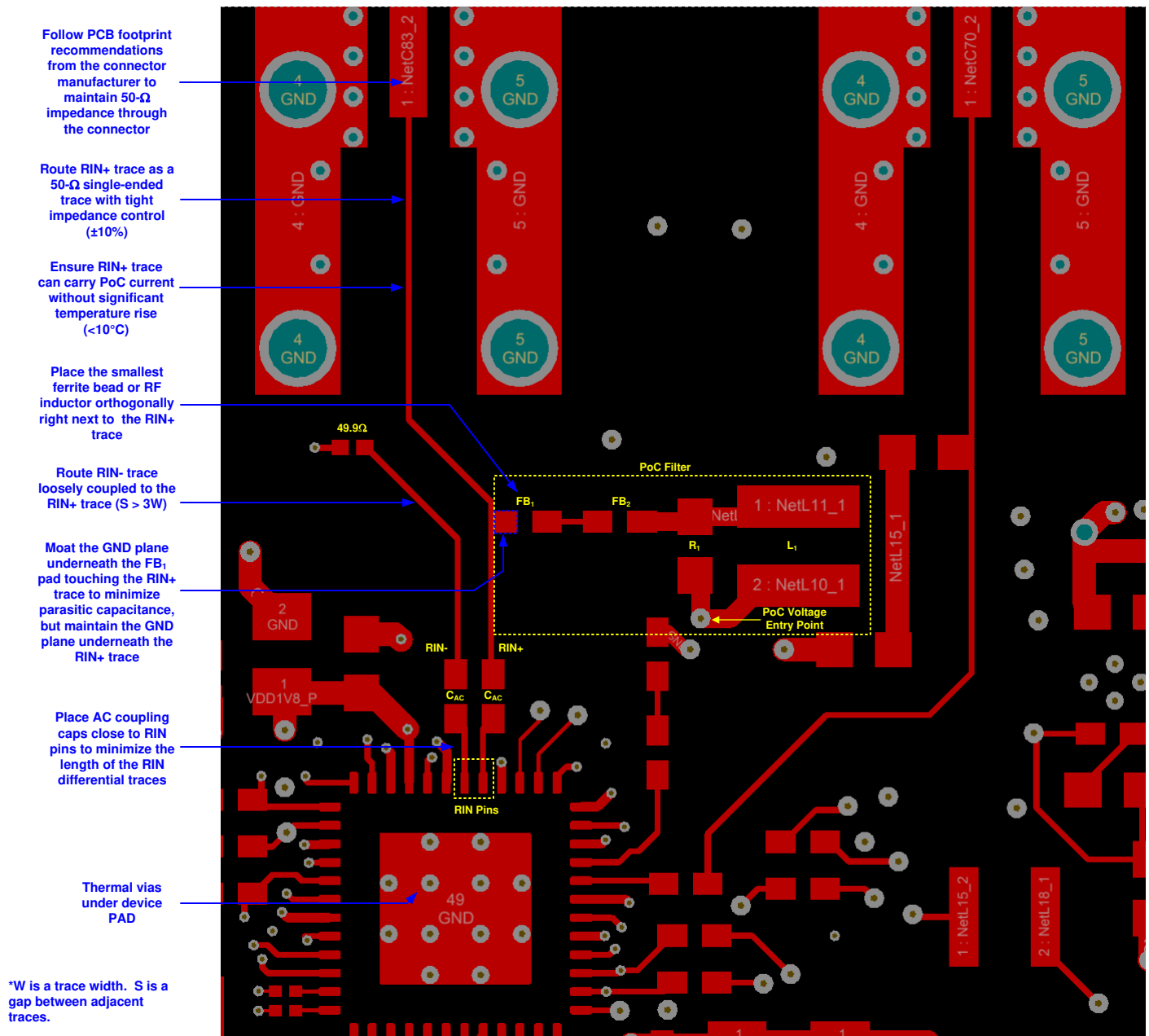


FIG 10-1. DS90UB638-Q1 PCB Layout Example: FPD-Link III Signal Traces and PoC Filter

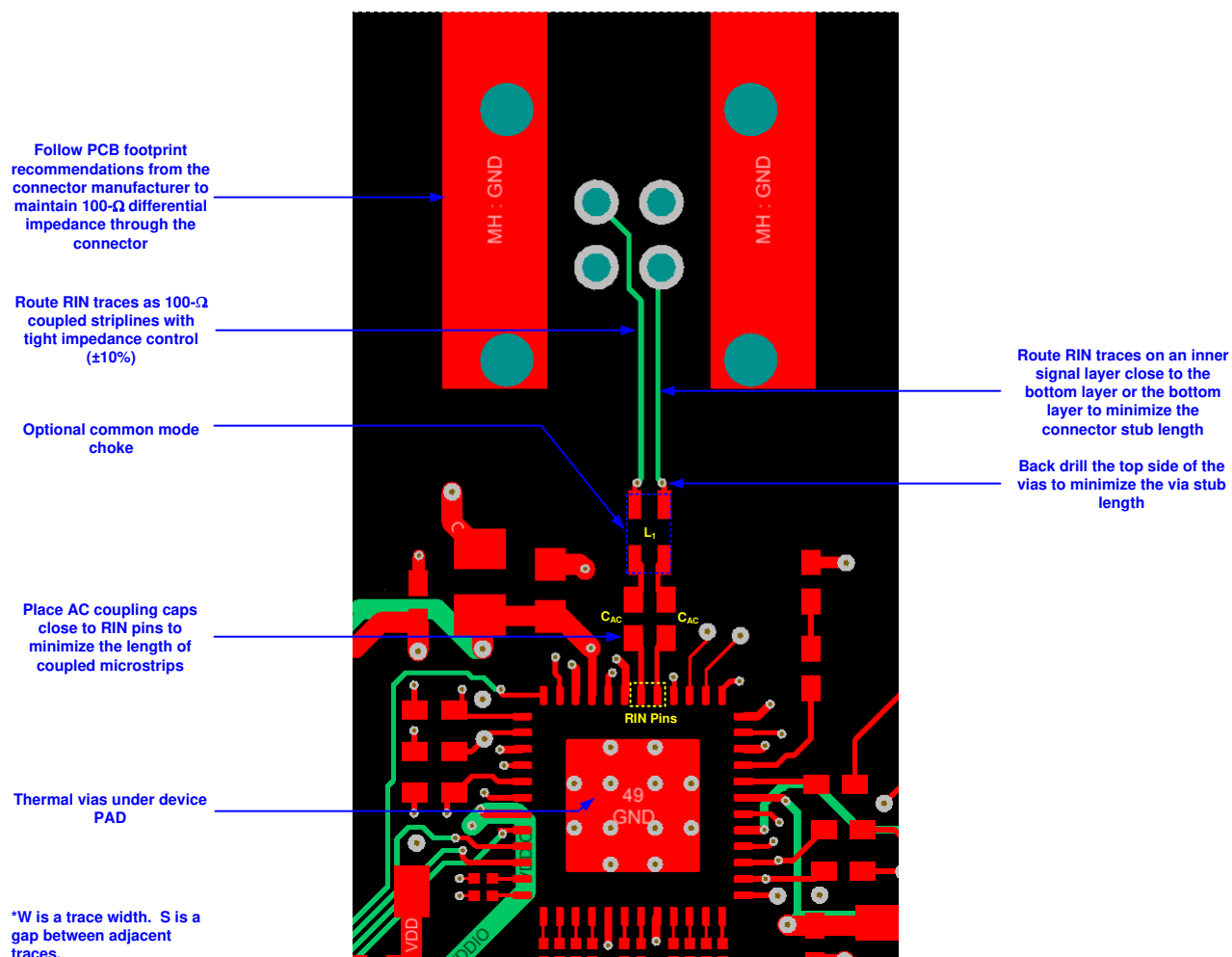


図 10-2. DS90UB638-Q1 PCB Layout Example: FPD-Link III Differential Signal Traces

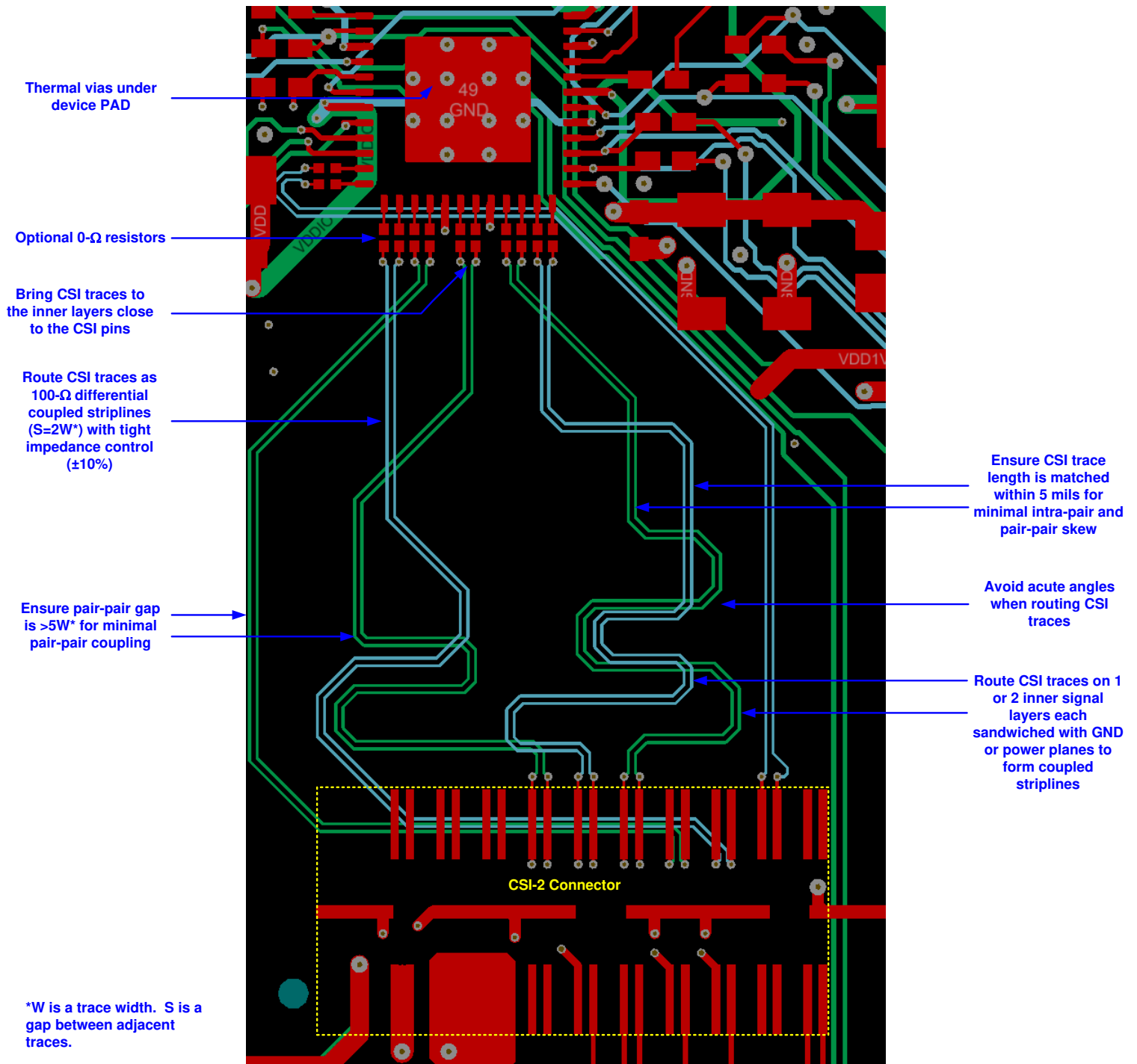


图 10-3. DS90UB638-Q1 PCB Layout Example: CSI-2 Traces

## 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation see the following:

- [I2C Communication Over FPD-Link III With Bidirectional Control Channel](#) (SNLA131)
- [I2C Bus Pullup Resistor Calculation](#) (SLVA689)
- [FPD-Link Learning Center](#)
- [An EMC/EMI System-Design and Testing Methodology for FPD-Link III SerDes](#) (SLYT719)
- [Ten Tips for Successfully Designing With Automotive EMC/EMI Requirements](#) (SLYT636)
- [Power-over-Coax Design Guidelines for DS90UB953-Q1](#)
- [DS90UB953-Q1EVM user's guide](#) (SNLU224)

#### 11.2 サポート・リソース

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#### 11.3 Trademarks

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#### 11.4 静電気放電に関する注意事項



この IC は、ESD によって破損する可能性があります。テキサス・インスツルメンツは、IC を取り扱う際には常に適切な注意を払うことを推奨します。正しい取り扱いおよび設置手順に従わない場合、デバイスを破損するおそれがあります。

ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

#### 11.5 用語集

##### [テキサス・インスツルメンツ用語集](#)

この用語集には、用語や略語の一覧および定義が記載されています。

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

## PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package   Pins	Package qty   Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
<a href="#">DS90UB638TRGZRQ1</a>	Active	Production	VQFN (RGZ)   48	2500   LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q
DS90UB638TRGZRQ1.A	Active	Production	VQFN (RGZ)   48	2500   LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q
DS90UB638TRGZRQ1.B	Active	Production	VQFN (RGZ)   48	2500   LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q
<a href="#">DS90UB638TRGZTQ1</a>	Active	Production	VQFN (RGZ)   48	250   SMALL T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q
DS90UB638TRGZTQ1.A	Active	Production	VQFN (RGZ)   48	250   SMALL T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q
DS90UB638TRGZTQ1.B	Active	Production	VQFN (RGZ)   48	250   SMALL T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 105	UB638Q

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

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## GENERIC PACKAGE VIEW

**RGZ 48**

**VQFN - 1 mm max height**

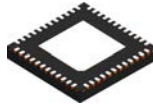
7 x 7, 0.5 mm pitch

PLASTIC QUADFLAT PACK- NO LEAD

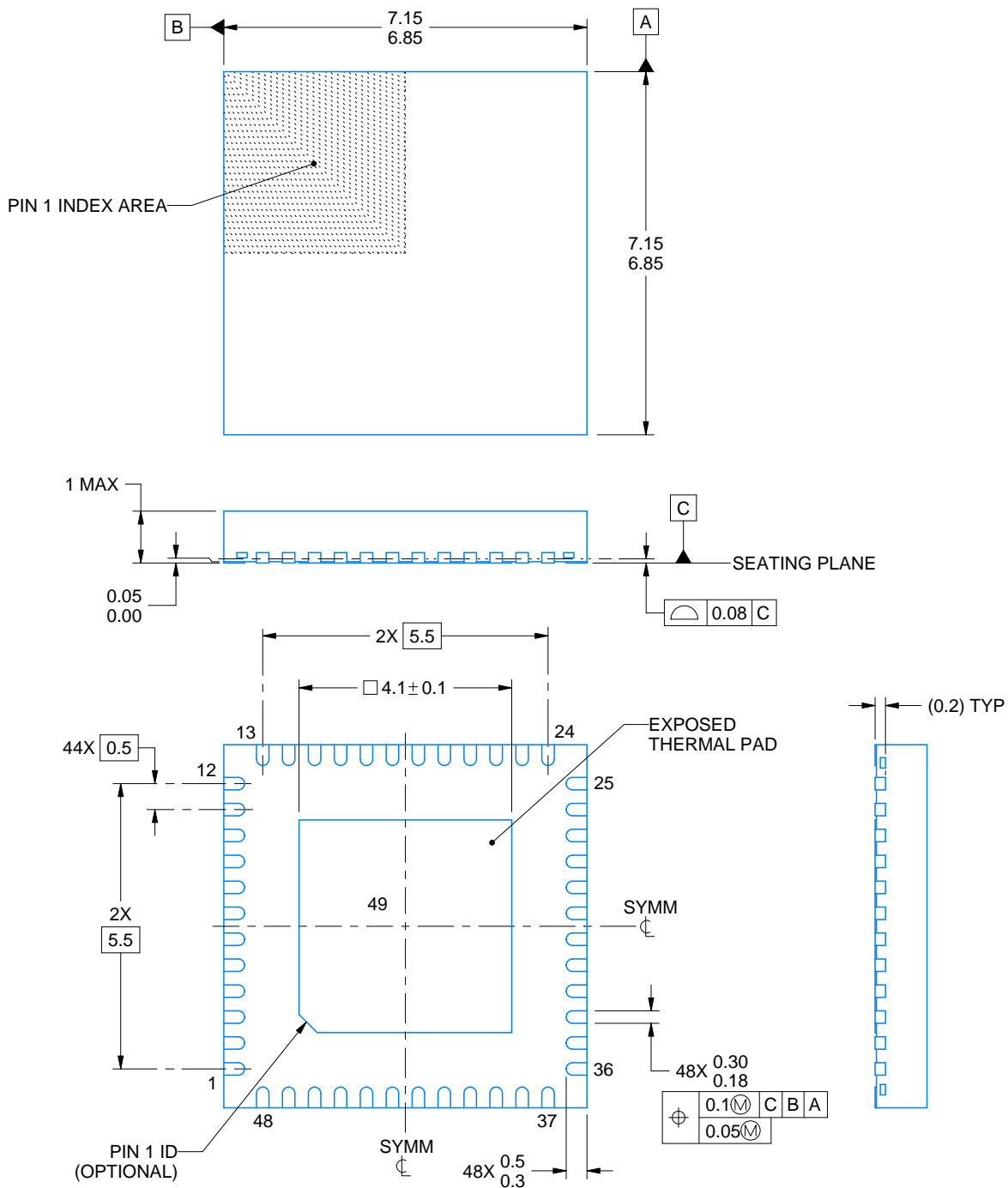


Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

4224671/A

**RGZ0048B****PACKAGE OUTLINE****VQFN - 1 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



4218795/B 02/2017

**NOTES:**

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



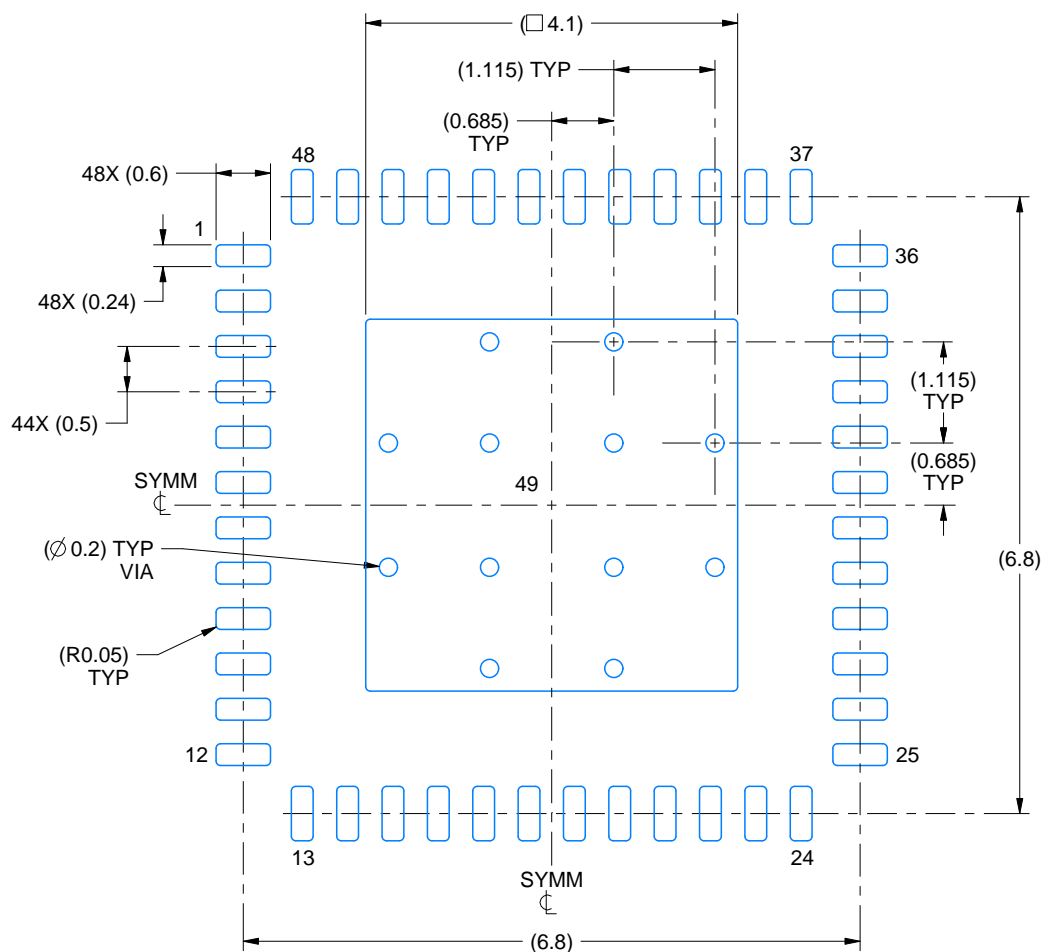
**TEXAS  
INSTRUMENTS**  
www.ti.com

# EXAMPLE BOARD LAYOUT

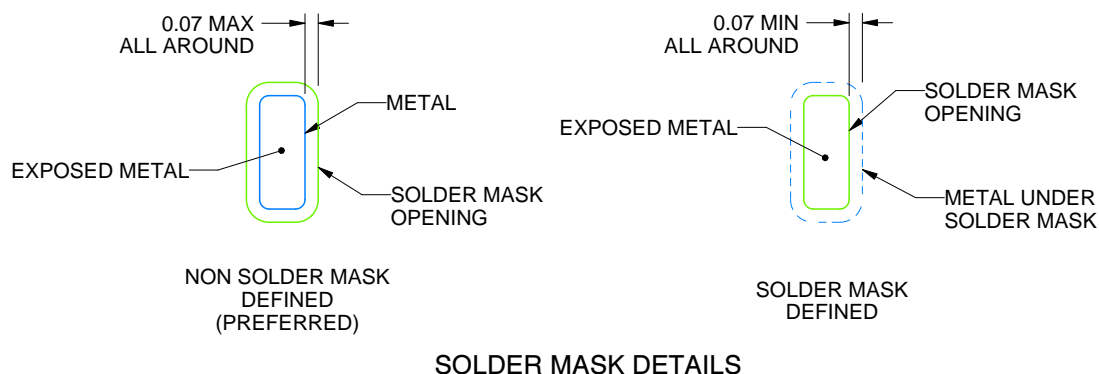
RGZ0048B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:12X



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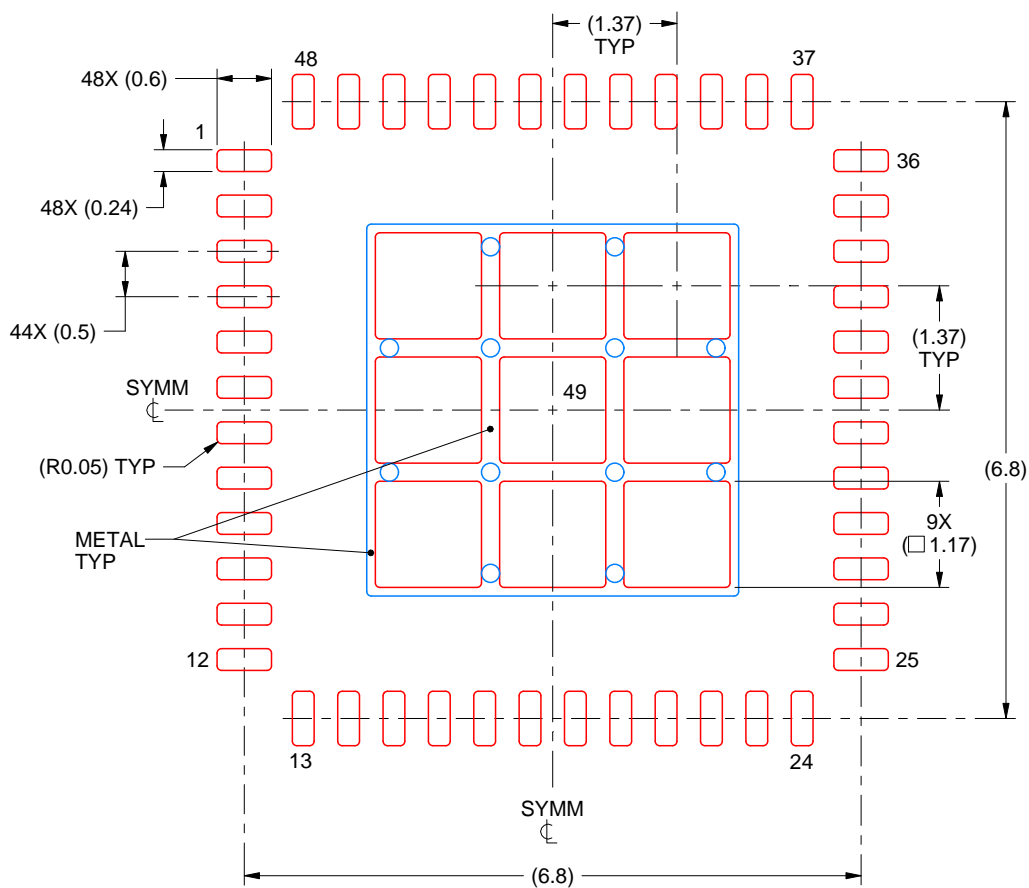
NOTES: (continued)

- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/sluea271](http://www.ti.com/lit/sluea271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

**RGZ0048B**

### VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



## SOLDER PASTE EXAMPLE BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 49  
73% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE  
SCALE:12X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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