



### TI Designs

故障指示灯 (FI) 是安装在配电网中高架电源线上的设备，用于检测和指示发生的故障。FI 每隔一定的距离重复出现一次，可以通过确定故障区域，在最大程度上缩短断电时间。对于高架 FI，指示可以是闪烁的 LED。

设计可靠的 FI 具有一定的挑战性，它需要能够抗噪、确保精确测量并具有低功耗的信号链。该设计的突出特点是，提出了一种架构，该架构展示了如何在不牺牲功能和性能的前提下，通过精选的模拟前端 (AFE) 组件最大程度地降低功耗，此外通过占空比控制和根据需要启用特定的 FI 块来实现进一步的电源改进。超低功耗 MCU 与 AFE 配合使用，可实现  $\leq \pm 2.0\%$  的电流和电压测量精度。该设计中还实现了其他功能，如用于执行系统诊断的自检机制。

### 设计资源

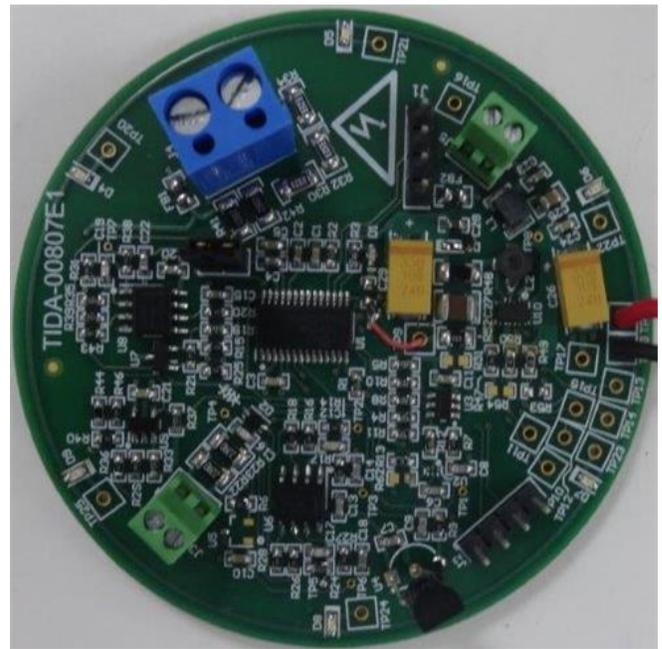
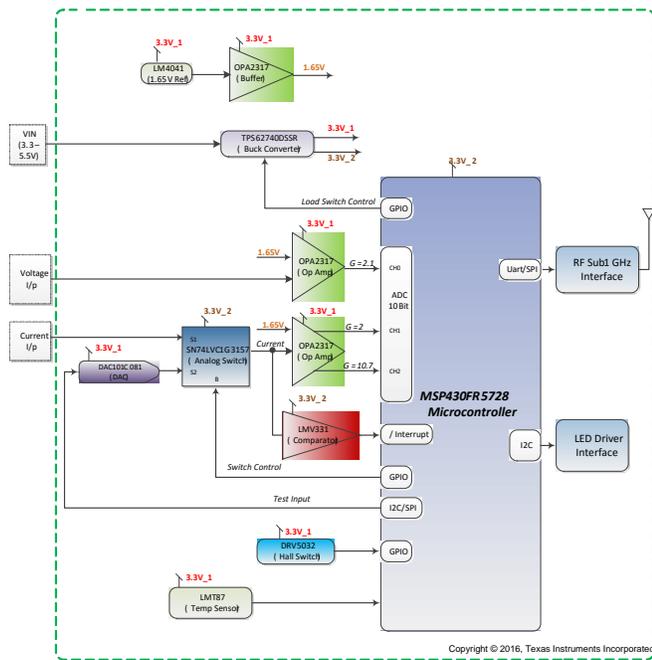
<a href="#">TIDA-00807</a>	设计文件夹
<a href="#">MSP430FR5728</a>	产品文件夹
<a href="#">OPA2317</a>	产品文件夹
<a href="#">SN74LVC1G3157</a>	产品文件夹
<a href="#">TPS62740</a>	产品文件夹
<a href="#">DAC101C081</a>	产品文件夹
<a href="#">LM4041-N</a>	产品文件夹
<a href="#">LMT87-Q1</a>	产品文件夹
<a href="#">DRV5032</a>	产品文件夹



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### 设计特性

- 电流和电压测量
  - 测量范围：0.5-100A、50-300V
  - 测量精度： $\leq \pm 2\%$
  - 使用了多个增益级和 ADC 通道以涵盖宽电流和电压范围
  - 用于故障电流检测的用户可配置阈值
  - 根据变电站的要求支持测量和记录电流和电压级别
- 功耗
  - 选择 AFE 组件和 MCU，使静态电流消耗小于 350 $\mu$ A
  - 双轨、可配置直流/直流转换器（具有集成负载开关）可提高灵活性并节省功耗
  - 基于超低功耗、小引脚数和小尺寸选择的 MCU
  - MCU 支持 LPRF 和高强度 LED 驱动器功能扩展
- 自检和诊断
  - 使用低功耗霍尔开关来检测是否存在磁体
  - 使用消耗低于 2 $\mu$ A 电流的 10 位 DAC 对故障电流进行仿真
  - 使用模拟输出 TMP 传感器完成温度测量



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## 1 System Description

Electric power systems consists of overhead transmission/distribution lines that carry the power from various generation stations to substations and distribute power to regions occupied with commercial and residential buildings. An occurrence of fault on the overhead line can lead to extended durations of power outages, cause damages to portions or entire power networks, resulting in productivity and economic loss thereby affecting consumers. To minimize the duration of power outages smart Fault Indicator (FI) devices are installed on overhead power lines to identify and indicate occurrence of overcurrent faults on the line. Visual indication aids utility crew members in identifying the location of faults from a distance and service the faults with faster turnaround times.

The features and complexities of Fault Indicators vary on operating environment, wide input current ranges, power consumption, self- test requirements and measurement accuracy. This design provides a robust architecture with a low power AFE, an ultra -low power MCU and a programmable DC-DC converter with an integrated load switch to control the switching on-off of discrete analog front end. Measurement accuracy of  $\pm 2\%$  is achieved under low power consumption. An additional feature such as self-test mechanism comprising a DAC and Hall switch is implemented in the design. The self-test mechanism generates a test waveform that simulates faults on the line. The benefit of this mechanism is to validate the working of the FI as expected. To achieve low power consumption, devices in the system have been selected carefully with low power consumption numbers, programmable DC converter and various modes implemented in software. The modes implemented in software are:

- Active Mode: MCU Timer configured to sample analog inputs every sec and place the MCU in Low power mode after sampling. The duty cycle can be varied in firmware to achieve low power consumption.
- Standby Mode: The MCU is placed in Low-power standby mode. A comparator remains active to detect a fault greater than a user-defined threshold level and wake the MCU upon any fault activity.
- Self-Test Mode: If a magnet is detected by the hall switch (when the flux density of the magnet exceeds the operating point threshold of (7.5 mT), the hall switch devices open-drain drives a low voltage. The software then enters Self-test mechanism mode to generate a test waveform to validate the sensing circuitry and software computation.

## 2 Key System Specifications

The key specifications of the system are listed in 表 1.

表 1. Key System Specifications

Number	Parameters	Description	Comments
1	Threshold detection	Range- 3 – 100A	User configurable Threshold detection levels in software
2	Analog Inputs	4 analog inputs	Current Inputs - 2 Voltage Input - 1 Temperature input - 1
3	AC voltage Input range	50-300V AC with resistor divider and fixed-gain amplifier	Measurement Accuracy: $\pm 2\%$
4	Current Input Range	0.50A to 100A AC with fixed gain amplifier for High Gain and Low Gain Paths	Measurement Accuracy: $\pm 2\%$
5	DC Offset Reference	VCC/2	Unipolar analog inputs, VCC input voltage set to 3.3V
6	Power Supply	Wide Vin range 2.2 V-5.5V DC	
7	Minimum Flux density to enter Self-Test Mode	Magnetic Threshold – 4 – 11 mT	Magnet distance > 0.5 mm
8	Temperature Measurement	Temperature Range: -50 – 150 °C	Accuracy: $\pm 0.3^\circ\text{C}$ typical

### 3 Block Diagram

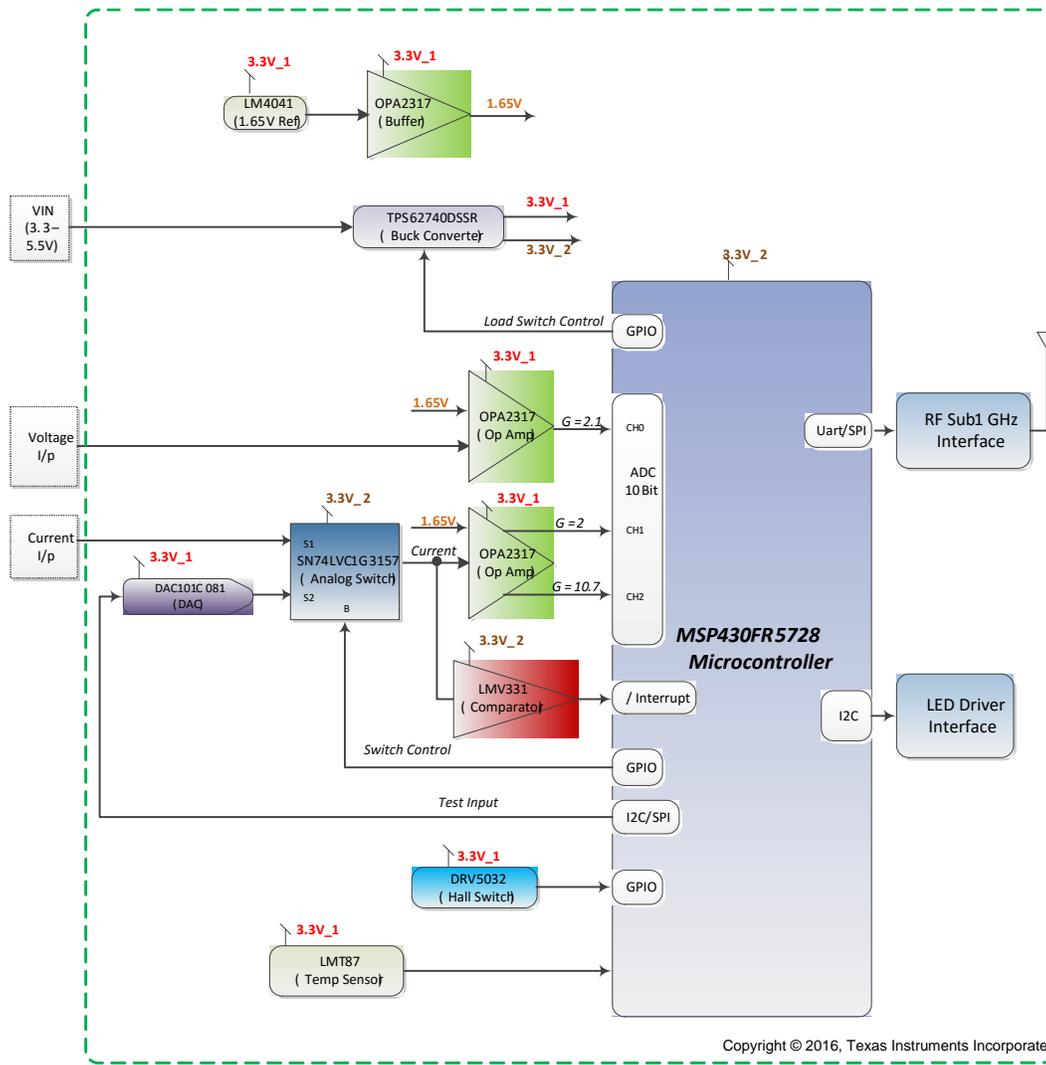


图 1. Block Diagram

The system blocks of this design are based on functionality, details are listed below:

- Current Measurement
  - Current inputs are connected to external current transformers with a turns ratio of 2000 to step down the primary current.
  - Current input circuitry consists of burden resistors, signal conditioning Op-Amps, reference DC level shift and interface to ADC.
  - Current inputs to be sensed by Fault Indicators have wide ranges. To achieve the required accuracy over the wide range of current input, two gain stages are provided for all current inputs.
- Voltage Measurement
  - Voltage input circuit with potential divider circuit divides AC input voltage to a Level that can be measured by ADC. Multiple resistors are used to meet the required voltage rating and also ensure reliability.
  - The output of the potential divider is given as input to a differential amplifier. The output of the

amplifier is connected to the ADCs.

- Single gain stage is provided for the voltage inputs since the expected measurement range is less.

- MCU handles
  - Data sampling by 10-bit Analog-to-Digital converter (ADC).
  - Analysis and processing of data to detect faults.
  - Control of programmable DC power supply to switch ON-OFF devices.
  - Control of analog switch to handle both analog and digital signals.
  - Indication of fault occurrence.
- DC/DC converter
  - Programmable DC/DC converter with integrated load switch is beneficial as it provides flexibility to connect or disconnect other system devices to DC-DC converter and reduce the power consumption during the normal operating conditions without fault.
  - Control of device power supply is achieved by control signal from the microcontroller.
  - Load switch rail is tied to Op-Amp, DAC, Hall Switch, Temperature Sensor, DC offset devices
- Self-Test/Diagnostic Mechanism
  - The self-test mechanism is implemented with Hall Switch and external DAC to identify defective devices on the line.
  - A magnet with operating magnetic flux  $> 7.5$  mT is placed over the hall switch to trigger a flag and place the device in self-test mode. The software then enters self-test mechanism mode to generate a test waveform to validate the sensing circuitry and software computation.
- Reference Offset
  - The reference is configured depending on the input and level shifted by  $VCC/2$ . The DC offset is level shifted for both Voltage and Current measurement to  $VCC/2$  by using a precision shunt voltage reference and Op-Amp for both voltage and current reference.
- LED Driver Interface
  - 4 pin header Interface to connect modular board with high intensity LED's for fault indication.
- Temperature Sensor
  - Temperature sensors with an analog output voltage that is linearly and inversely proportional to temperature to monitor temperature.
  - Interface to ADC
- Comparator
  - User configurable threshold detection levels to wake the MCU from low power mode.

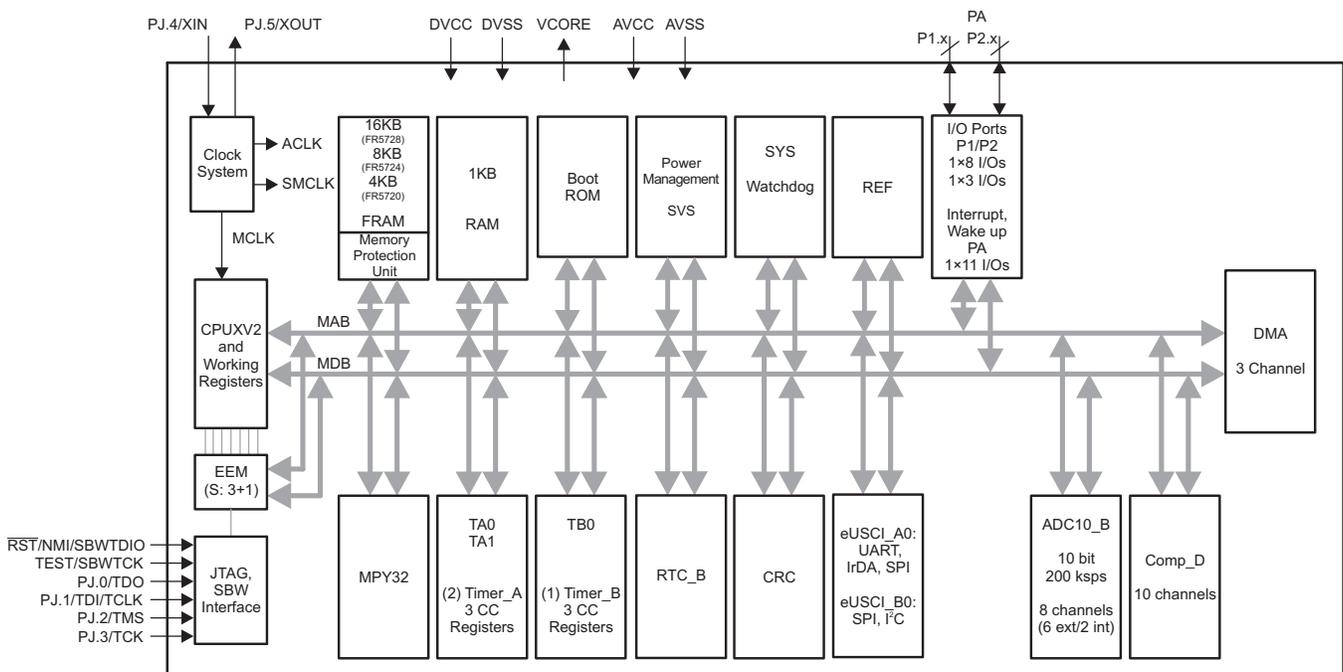
### 3.1 Microcontroller - MSP430FR5728 MCU

The MSP430FR5728 ultra-low-power microcontroller (MCU) consists of multiple features embedded FRAM nonvolatile memory, ultra-low-power 16-bit MSP430™ CPU, and different peripherals targeted for various applications. The architecture, FRAM, and peripherals, combined with seven low-power modes, are optimized to achieve extended battery life in portable and wireless sensing applications. FRAM is a new nonvolatile memory that combines the speed, flexibility, and endurance of SRAM with the stability and reliability of flash, all at lower total power consumption.

Peripherals include a 10-bit ADC, a 16-channel comparator with voltage reference generation and hysteresis capabilities, three enhanced serial channels capable of inter-integrated circuit (I<sup>2</sup>C), serial peripheral interface (SPI), or universal asynchronous receiver/transmitter (UART) protocols, an internal DMA, a hardware multiplier, an real-time clock (RTC), five 16-bit timers, and digital I/Os. 表 2 provides the design features of MSP430FR5728 MCU and 图 2 shows the functional block diagram.

表 2. MSP430FR5728 Design Features

<b>FRAM</b>	16K
<b>SRAM</b>	1K
<b>I/O</b>	Up to 21
<b>ADC</b>	10-bit ADC, 8 Channels externally , 2 channels internal
<b>USCI</b>	1
<b>SPI</b>	1
<b>Comparator</b>	12 Channels
<b>I2C</b>	1
<b>Package</b>	TSSOP (28)



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图 2. MSP430FR5728 Functional Block Diagram

### 3.2 Operational Amplifier – OPA2317

The OPA317 series of CMOS operational amplifiers offer precision performance at a very competitive price. These devices are members of the Zero-Drift family of amplifiers that use a proprietary Auto calibration technique to simultaneously provide low offset voltage (90  $\mu\text{V}$  max) and near-zero drift over time and temperature at only 35  $\mu\text{A}$  (max) of quiescent current.

The OPA317 family features rail-to-rail input and output in addition to near flat 1/f noise, making this amplifier ideal for many applications, and much easier to design into a system. These devices are optimized for low-voltage operation as low as +1.8 V ( $\pm 0.9$  V) and up to +5.5 V ( $\pm 2.75$  V). The OPA317 (single version) is available in the SC70-5, SOT23-5, and SOIC-8 packages. The OPA2317 (dual version) is offered in MSOP-8 and SOIC-8 packages.

The features of rail-to-rail I/O Operational Amplifier are:

- Supply Voltage: 1.8 V to 5.5 V
- Low Offset Voltage: 20  $\mu\text{V}$  (typ)
- CMRR: 108 dB (typ)
- Quiescent Current: 35  $\mu\text{A}$  (max)
- Gain Bandwidth: 300 kHz
- Rail-to-Rail Input/Output
- Internal EMI/RFI Filtering

### 3.3 Power Supply – TPS62740

The TPS62740 is industry's first step down converter featuring typ 360 nA quiescent current and operating with a tiny 2.2  $\mu\text{H}$  inductor and 10  $\mu\text{F}$  output capacitor. This new DCS-control-based device extends the light load efficiency range below 10  $\mu\text{A}$  load currents. This has the following features:

- Input voltage range  $V_{\text{IN}}$  from 2.2 V to 5.5 V
- Typical 360 nA quiescent current
- Up to 90% efficiency at 10- $\mu\text{A}$  output current
- Up to 300 mA output current
- RF friendly DCS-control
- Up to 2 MHz switching frequency
- Low output ripple voltage
- Automatic transition to no ripple 100% mode
- Slew rate controlled load switch
- Discharge function on  $V_{\text{OUT}}/\text{LOAD}$
- Power good output
- Optimized for operation with a tiny 2.2  $\mu\text{H}$  inductor and 10  $\mu\text{F}$   $C_{\text{OUT}}$
- Small 2 x 3 mm<sup>2</sup> SON package

Slew rate controlled switch is used to switch on-off the peripheral supply to reduce the power consumption during the normal operating conditions without fault. This control of peripheral power supply is achieved by control signal from the controller. To obtain the output voltage setting, see 表 3. The output configured in this application is 3.3 V.

**表 3. TPS67240 Output Voltage Settings**

VOUT	VSEL 4	VSEL 3	VSEL 2	VSEL 1
3.0V	1	1	0	0
3.1V	1	1	0	1
3.2V	1	1	1	0
3.3V	1	1	1	1

### 3.4 Current Inputs

The design provides provision to measure one current input and an external current transformer is used to step down the primary current. The current transformer ratio used in the design is 1:2000. The current inputs to be sensed by FCI have wide ranges and may not be able to sense the complete range of current inputs with a single 10-bit ADC range within the required accuracy. To achieve the required accuracy over the wide range of current input, two gain stages are provided for all current inputs.

The gains configured are indicative gains and these gains may change based on the application and the range of currents to be measured.

High gain stage: When the input current is less, the amplification is more. Gain value is currently set to 10.7.

Low gain stage: When the input current is High, the amplification is reduced. Gain value is currently set to 2.

图 3 shows High gain and Low gain stages for one current input.

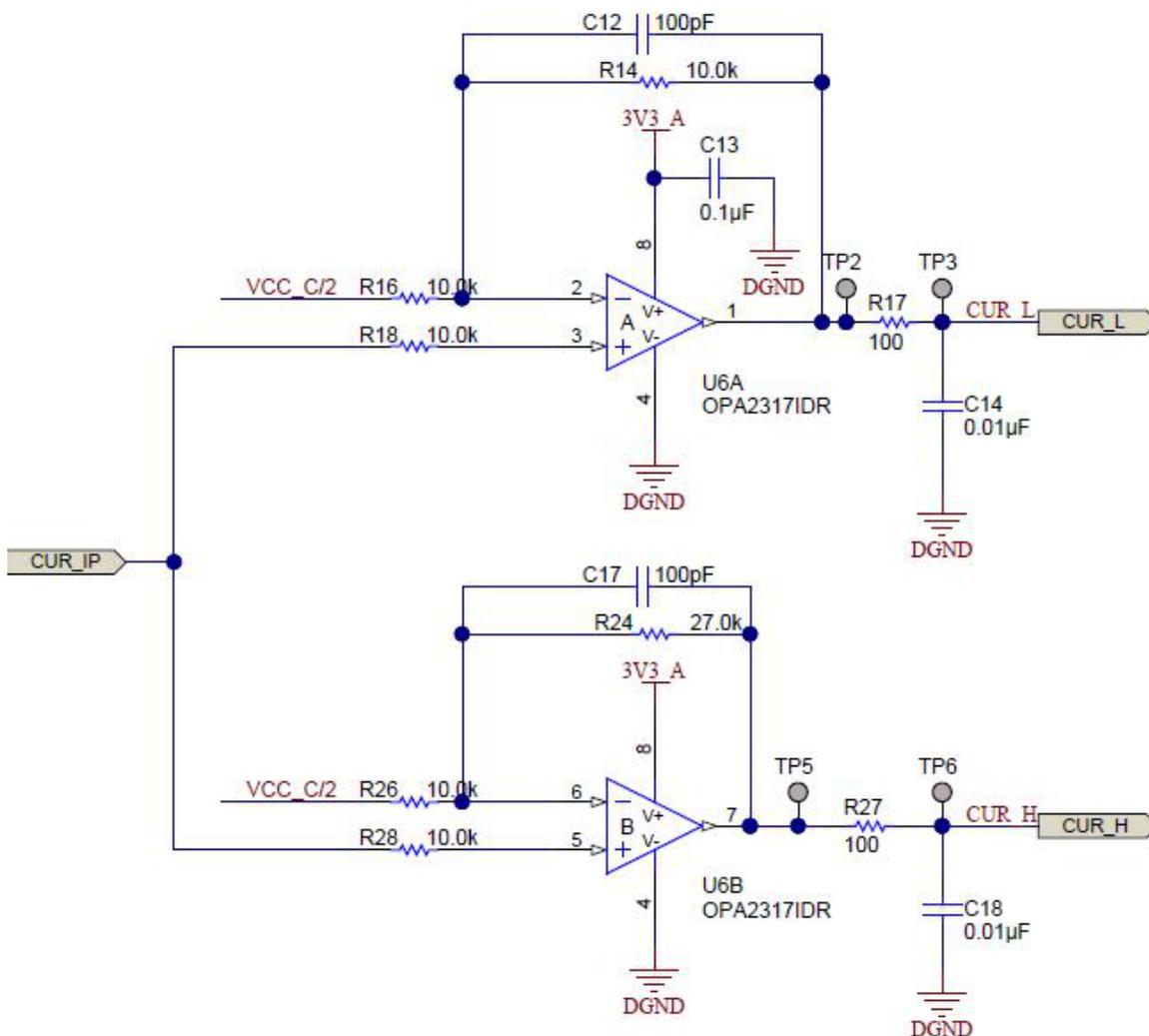


图 3. Current Input Gain Stages

### 3.5 Voltage Inputs

The 120-V or 230-V AC input is divided by the Potential divider to a Level that can be measured by the ADC. Multiple resistors are used to meet the required voltage rating and also ensure reliability. The output of the potential divider is given as input to a differential amplifier. The output of the amplifier is connected to the ADCs. Single gain stage is provided for the voltage inputs since the expected measurement range is less.

### 3.6 Precision Shunt Voltage Reference - LM4041-N

The LM4041 series of shunt voltage references are versatile, easy-to-use references suitable for a wide array of applications. The advanced design of the LM4041-N eliminates the need for an external stabilizing capacitor while ensuring stability with any capacitive load, thus making the LM4041-N easy to use. Further reducing design effort is the availability of a fixed (1.225 V) and adjustable reverse breakdown voltage.

Additionally, the reference offers low dynamic impedance, low noise, and a low temperature coefficient to ensure a stable output voltage over a wide range of operating currents and temperatures. The LM4041 uses fuse and Zener-zap reverse breakdown voltage trim during wafer sort to ensure that the prime parts have an accuracy of better than  $\pm 0.1\%$  (A grade) at 25°C. Thus, a great deal of flexibility is offered to designers in choosing the best cost-to-performance ratio for their applications. The LM4041 is available in a fixed (1.225 V nominal) or an adjustable version, which requires an external resistor divider to set the output to a value between 1.225 V and 10 V.

### Features

- Small packages: SOT23, SC70 and TO-92
- No output capacitor required
- Output voltage tolerance  $\pm 0.1\%$  (max) at 25°C
- Low output noise: 20  $\mu\text{V}_{\text{rms}}$  (10Hz to 10kHz)
- Wide operating current range: 60  $\mu\text{A}$  to 12 mA
- Extended temperature range: -40°C to +125°C
- Low temperature coefficient: 100 ppm/°C (max)

### 3.7 Single-Pole, Double-Throw Analog Switch - SN74LVC1G3157

This single-pole double-throw (SPDT) analog switch is designed for 1.65-V to 5.5-V VCC operation. The SN74LVC1G3157 device can handle both analog and digital signals. The SN74LVC1G3157 device permits signals with amplitudes of up to VCC (peak) to be transmitted in either direction.

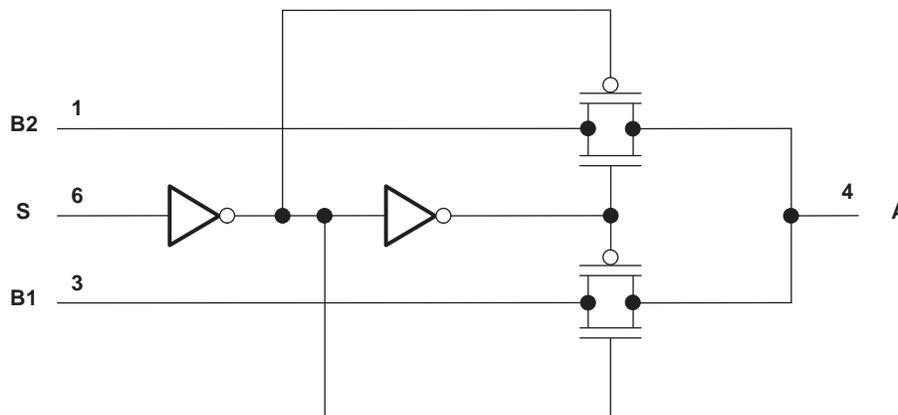


图 4. Logic Diagram (Positive Logic)

The 1.65-V to 5.5-V supply operation allows the device to function in many different systems comprised of different logic levels (see 图 4), allowing rail-to-rail signal switching. Either the B1 channel or the B2 channel is activated depending upon the control input. If the control input is low, B1 channel is selected. If the control input is high, B2 channel is selected.

Applications include signal gating, chopping, modulation or demodulation (modem), and signal multiplexing for analog-to-digital converter (ADC) and digital-to-analog conversion systems.

### Features

- 1.65-V to 5.5-V VCC Operation
- Useful for Both Analog and Digital Applications

- Rail-to-Rail Signal Handling
- Operating Frequency Up to 300 MHz
- High Degree of Linearity
- High Speed, Typically 0.5 ns
  - ( $V_{CC} = 3\text{ V}$ ,  $CL = 50\text{ pF}$ )
- Low ON-State Resistance, Typically  $\approx 6\ \Omega$ 
  - ( $V_{CC} = 4.5\text{ V}$ )
- ESD Protection Exceeds JESD 22

### 3.8 10-Bit DAC With I2C-Compatible Interface - DAC101C081

The DAC101C081 device is a 10-bit, single channel, voltage-output DAC that operates from a 2.7 V to 5.5 V supply. The output amplifier allows rail-to-rail output swing and has an 6- $\mu\text{sec}$  settling time. The DAC101C081 uses the supply voltage as the reference to provide the widest dynamic output range and typically consumes 132  $\mu\text{A}$  while operating at 5.0 V. It is available in 6-lead SOT and WSON packages and provides three address options (pin selectable).

The DAC101C081 use a 2-wire, I2C-compatible serial interface that operates in all three speed modes, including high speed mode (3.4 MHz). An external address selection pin allows up to three DAC101C081 devices per 2-wire bus. Pin compatible alternatives to the DAC101C081 are available that provide additional address options.

The DAC101C081 each have a 16-bit register that controls the mode of operation, the power-down condition, and the output voltage. A power-on reset circuit ensures that the DAC output powers up to zero volts. A power-down feature reduces power consumption to less than a microWatt. Their low power consumption and small packages make these DACs an excellent choice for use in battery operated equipment. Each DAC operates over the extended industrial temperature range of  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .

#### Features

- Ensured Monotonicity to 10-bits
- Low Power Operation: 156- $\mu\text{A}$  maximum at 3.3 V
- Extended Power Supply Range (2.7 V to 5.5 V)
- I2C-Compatible 2-Wire Interface Which Supports Standard (100-kHz), Fast (400-kHz), and High-Speed (3.4-MHz) Modes
- Rail-to-Rail Voltage Output
- Very Small Package
- Resolution: 10 Bits
- INL:  $\pm 2$  LSB (Maximum)
- DNL:  $+0.3/-0.2$  LSB (Maximum)
- Setting Time: 6- $\mu\text{s}$  (Maximum)
- Zero Code Error:  $+10\text{-mV}$  (Maximum)
- Full-Scale Error:  $-0.7\ \%\text{FS}$  (Maximum)

### 3.9 Low Power Digital-Switch Hall Effect Sensor – DRV5032

The DRV5032 is an automotive-qualified low-power digital-switch Hall effect sensor.

When the applied magnetic flux density exceeds the BOP threshold, the device's open-drain output drives a low voltage. When the flux density decreases below the BRP threshold, the output becomes high-impedance. Integrated hysteresis separates BOP and BRP to provide robust crossovers.

By incorporating an internal oscillator, the DRV5032 samples the magnetic field and updates the output every 3 ms, using just 14  $\mu\text{A}$  when powered from 5 V. This low-latency low-current draw sensor is suitable for automotive systems that need continuous monitoring when the engine is off.

The DRV5032 operates from a VCC range of 1.65 to 5.5 V, and supports a Grade 2 ambient temperature range of  $-40^{\circ}\text{C}$  to  $105^{\circ}\text{C}$ .

### 3.10 Single General Purpose Low-Voltage Comparator – LMV331

The LMV393 and LMV339 devices are low-voltage (2.7 V to 5.5 V) versions of the dual and quad comparators, LM393 and LM339, which operate from 5 V to 30 V. The LMV331 is the single-comparator version.

The LMV331, LMV339, and LMV393 are the most cost-effective solutions for applications where low-voltage operation, low power, and space saving are the primary specifications in circuit design for portable consumer products. These devices offer specifications that meet or exceed the familiar LM339 and LM393 devices at a fraction of the supply current.

#### Features

- 2.7-V and 5-V Performance
- Low Supply Current
  - LMV331 130  $\mu\text{A}$  Typ
  - LMV393 210  $\mu\text{A}$  Typ
  - LMV339 410  $\mu\text{A}$  Typ
- Input Common-Mode Voltage Range Includes Ground
- Low Output Saturation Voltage 200 mV Typical
- Open-Collector Output for Maximum Flexibility

### 3.11 Reference Voltage

When AC input is applied, the ADC input has to be level shifted to  $V_{CC}/2$  to measure the AC waveform.

The DC offset is level shifted for both Voltage and Current measurement to  $V_{CC}/2$ , and the Op-Amp is used for both voltage and current reference.

## 4 Software

This section describes the software flow to process analog inputs, check for over current faults and indicate the occurrence of faults. Figure 5 shows a high-level software flow of the application.

The software implements the modes listed below:

- **Active Mode:** MCU Timer configured to sample analog inputs every sec and place the MCU in Low power mode after sampling. The duty cycle can be varied in firmware to achieve low power consumption.
- **Standby Mode:** The MCU is placed in Low-power standby mode. A comparator remains active to detect a fault greater than a user-defined threshold level and wake the MCU upon any fault activity.
- **Self-Test Mode:** If a magnet is detected by the hall switch (when the flux density of the magnet exceeds the operating point threshold (7.5 mT)), the hall switch devices open-drain drives a low voltage. The software then enters Self-test mechanism mode to generate a test waveform to validate the sensing circuitry and software computation.

The firmware is setup to monitor comparator input, magnet detection and follow the flow as shown in the Figure 5 based on the activity decision. If no activity is detected on comparator input (High Current inputs > 90 A) and Hall switch (magnet detection) the firmware enters Active Mode.

In Active mode MCU timer is configured at 1 sec interval to wake up MCU, sample 64 samples of data conditioned by the Op-Amp on current and voltage lines. The captured samples are stored in a buffer, processed and analyzed in software to check the line status and detect if a fault is detected. If a fault is detected a LED on the board is toggled to indicate occurrence of fault. If a fault is not detected, the firmware places MCU in Low power mode and wakes up the MCU on the next 1 sec timer interrupt to capture data samples and process the data. The benefit of varying the duty cycle and placing the devices in LPM mode helps in achieving low system power consumption numbers. The duty cycle can be varied in software based on the application requirement. To achieve system low power consumption numbers the MCU is placed in LPM mode and programmable DC converter disconnects the load switch rail. The current software is configured to 16% duty cycle and the average current consumption at 8MHz CPU frequency is < 500  $\mu$ A (Active Mode + (MCU in LPM mode and LDO ON)).

If a magnet is detected by the hall switch (when the flux density of the magnet exceeds the operating point threshold (7.5 mT)), the hall switch devices open-drain drives a low voltage. The software then enters Self-test mechanism mode to generate a test waveform to validate the sensing circuitry and software computation.

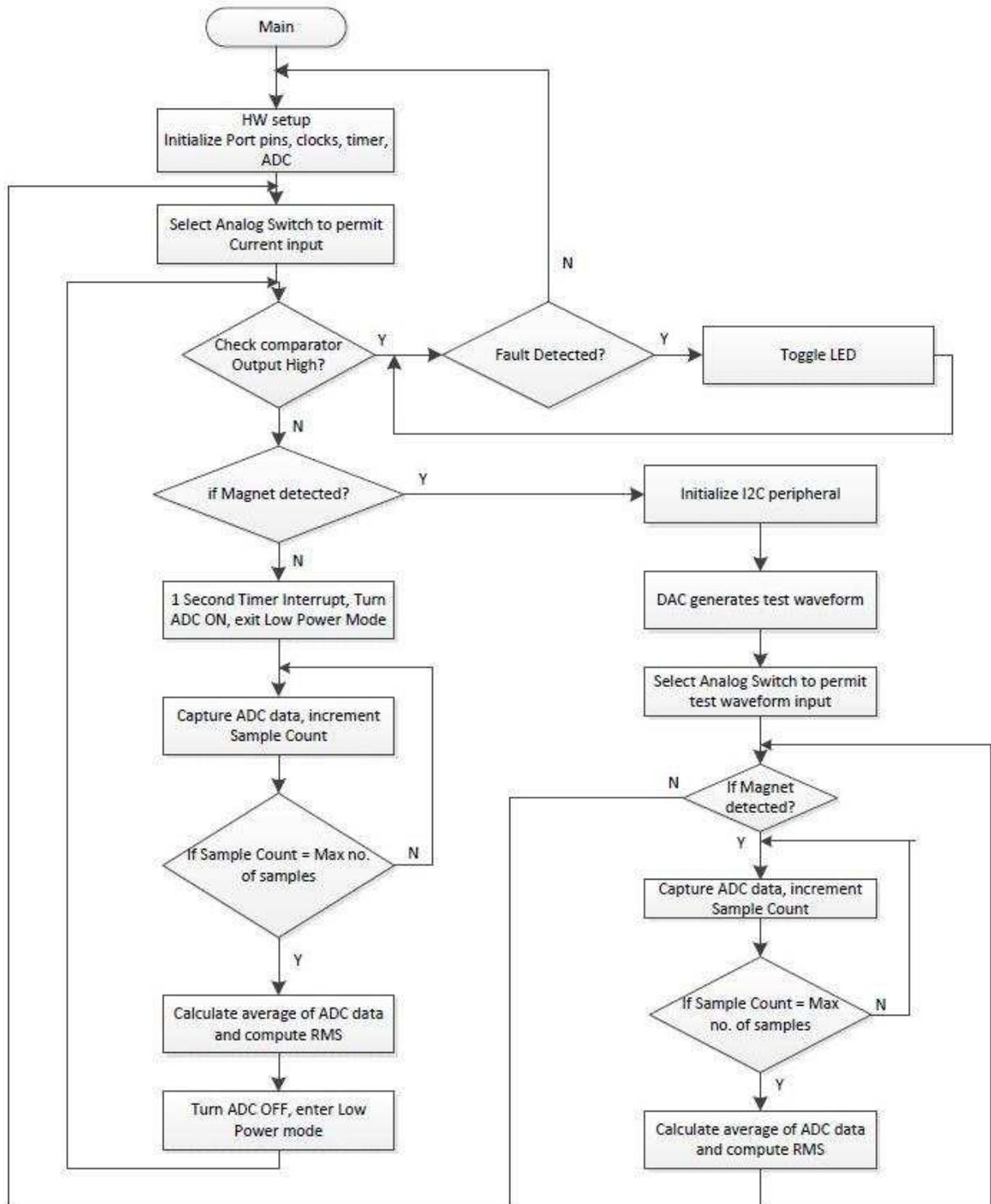


图 5. High Level Software Flow

## 4.1 Initialization Routine

The program execution for the initialization routine is shown below during power-up:

- Watch dog
  - Disable the watchdog timer
- System Clock
  - Initialize the system clock to 8 Mhz DCO.
- Load switch control pin for TPS62740
- Input pin selected to read comparator output
- Analog switch control pin selected for current input
- LED pins
  - Configure the LED port pins as GPIO and initialize the LED ports as output.
- ADC10
  - Initialize timer to trigger ADC conversions
  - Configure the ADC10 port pins to peripheral mode
  - Configure ADC DMA and enable DMA interrupts

## 4.2 Self Test

When a magnet (<http://www.digikey.com/product-detail/en/radial-magnet-inc/8031/469-1014-ND/2810356>) is placed near the hall switch and the detected magnetic flux exceeds 7 mT, an output low signal is generated by the Hall switch. In the software a test mode flag is set to indicate magnet detection during which the following peripherals are set:

- Initialize I2C communication with DAC
- Analog switch control pin selected for test waveform input

For magnet detection case, the program flow can be followed by referring to [图 5](#).

## 4.3 RMS Calculation

The ADC samples for the selected channels are read and stored in a buffer. The sampled data is corrected for offset by subtracting the DC offset (1.65 V) from each ADC samples for the corresponding channel. This process is repeated for all channels. The offset corrected samples are copied into a buffer until the required number of samples (64 samples) is reached. The RMS is calculated and multiplied with the gain factor (High and Low gain factors).

## 5 Test Setup

Follow the steps provided below to setup the hardware for testing:

1. Power Supply - A DC input of 3.3 V DC can be applied to connector J5. The TPS62740 DC-DC converter on the hardware provides an output of 3.3V to power the MCU.
2. Jumper JP2 - Connect Jumper JP2 to ensure 3.3V DC is fed to power the Microcontroller.
3. Current Inputs - connect an external current transformer with a turns ratio of 2000:1 to connector J3 to feed current inputs to ADC channels of MCU for processing.
4. Voltage Inputs - Connector J4 is provided on the board to provide a provision to apply AC voltage inputs.
5. AC Source: To run the tests a programmable AC source with current adjustable up to 30A has been

used. The currents were measured with a Metering reference meter of error  $<\pm 0.1\%$ . Current inputs from programmable AC source are applied to connector J3 and AC Voltages from programmable AC source are applied to connector J4.

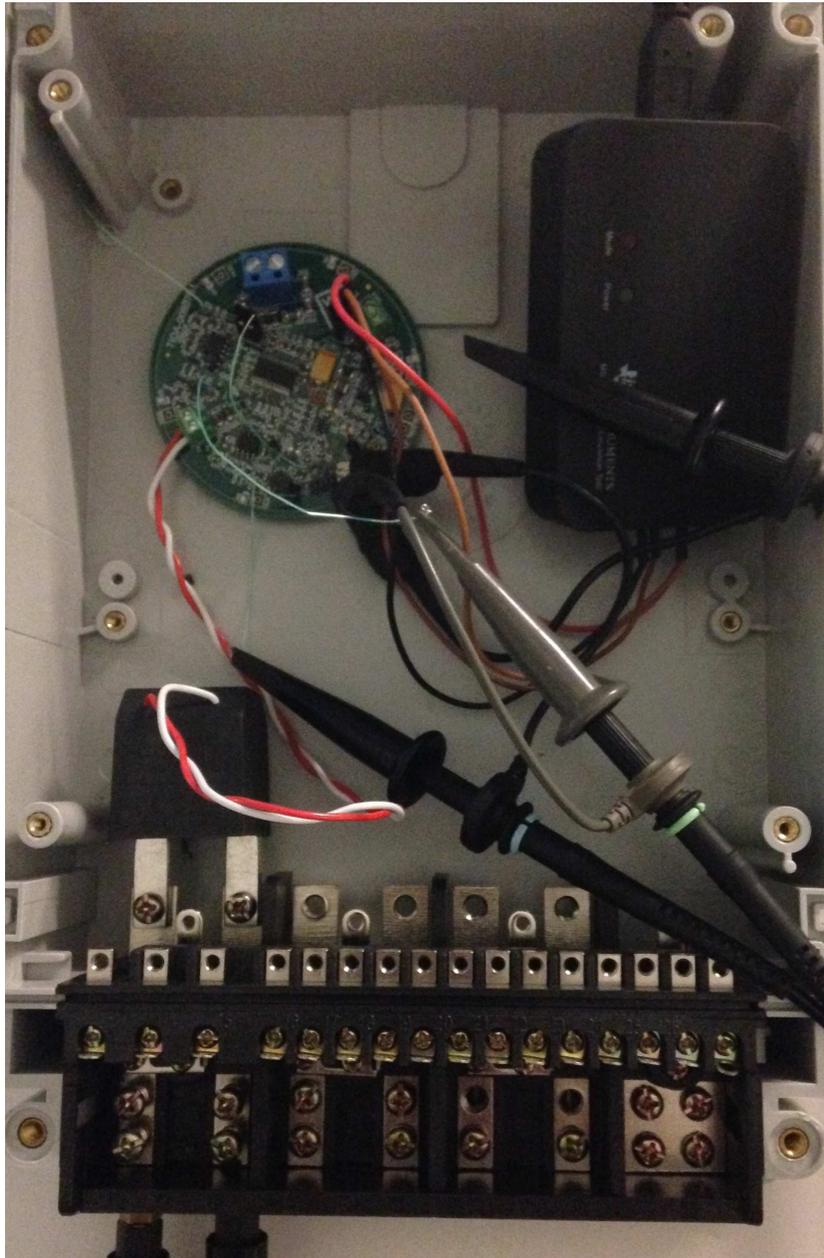


图 6. Test Setup

## 6 Test Data

The focus of this design is to test the performance of the system, achieve Accuracy measurements of  $\pm 2\%$ , showcase Self-Test Mechanism and highlight the various methods implemented in hardware and software to achieve low-power consumption.

### 6.1 Current Accuracy

表 4 provides power measurement accuracy results for Current Transformer inputs with Nominal frequency of 50 Hz and 60 Hz. The accuracy results achieved is  $< +2\%$ .

The highlighted yellow rows in 表 5 showcase the overlapping current inputs between CUR\_L and CUR\_H gain paths. 表 5 indicates that CUR\_L gain path is optimal for current inputs  $> 20A$  and CUR\_H gain path for current inputs  $\leq 20A$ .

注: CUR\_L represents the low gain path accuracy results and CUR\_H represents the High gain path accuracy results.

表 4. Current Accuracy

CURRENT (A)	60 Hz FREQUENCY		50 Hz FREQUENCY	
	CUR_L	CUR_H	CUR_L	CUR_H
0.5		-2.00		-1.40
1		-1.00		-1.00
1.5		-1.33		-0.66
3		0.66		0.26
5		0.20		0.40
10	-1.50	0.20	-1.70	0.70
20	-1.10	0.30	-1.05	1.50
30	-0.80		-0.66	
40	-0.12		-0.05	
50	0.08		0.10	
60	0.16		0.28	
70	0.28		0.71	
80	0.42		0.75	
90	0.55		0.77	
95	1.05		1.45	

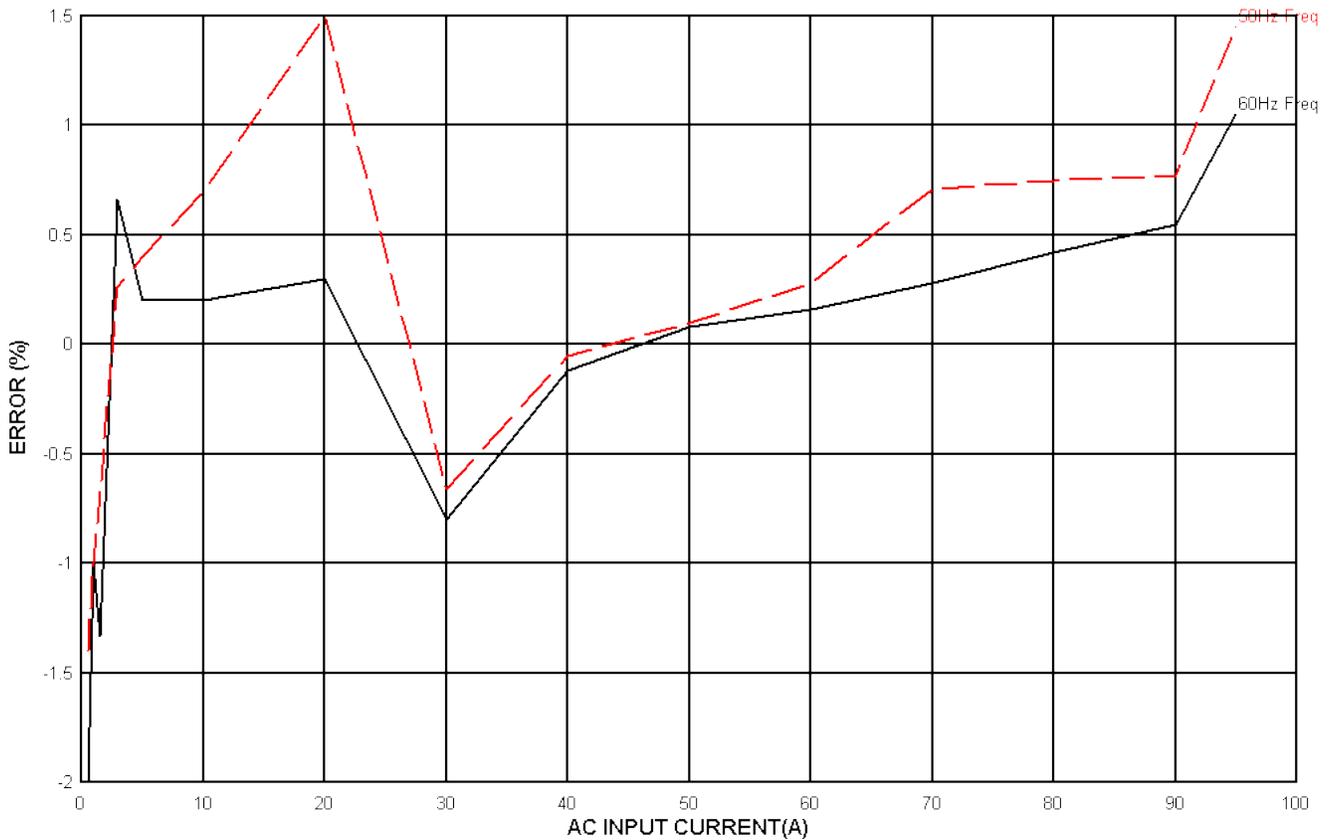


图 7. Current Accuracy Plot

## 6.2 Voltage Accuracy

表 5 provides the voltage accuracy measurement results.

表 5. Voltage Accuracy

VOLTAGE (V)	60 Hz FREQUENCY	50 Hz FREQUENCY
	ACCURACY	ACCURACY
50	-1.60	-1.60
75	-0.80	-0.80
100	-0.60	-0.40
125	-1.12	-1.12
150	-1.33	-1.60
175	-1.04	-1.14
200	-1.00	-1.50
225	-1.33	-1.33
250	-0.16	-0.64
275	-0.36	-0.36

## 6.3 Current Consumption

Table 6 provides a list of condition/modes with device state (Powered/Unpowered) to highlight the modes available to achieve low system power consumption.

The current consumption test was measured by measuring the current across jumper JP2.

注: Time to wakeup MCU from LPM3 mode to Active mode is between 78-120  $\mu$ Sec.

表 6. Current Consumption

CONDITION/MODE	DEVICE STATE	CURRENT	CLOCKS	DETAILS
Continuous Sampling of input Data (no fault detected)	All devices powered	1.75 mA	8 MHz	
		700 $\mu$ A	1 MHz	
Continuous Sampling of input Data (fault detected)	All devices powered	3.75 mA	8 MHz	LED ON indicating fault occurrence
		2.70 mA	1 MHz	
Standby Mode (MCU in LPM3 mode)	All devices powered	500 $\mu$ A	8 MHz	No sampling of Data
		500 $\mu$ A	1 MHz	
Standby Mode (MCU in LPM3 mode)	Comparator and MCU powered with all other devices unpowered	< 15 $\mu$ A	8 MHz	No sampling of Data
		< 15 $\mu$ A	1 MHz	
Active Mode (Duty cycle 2%)	Active Time: All devices powered Standby Time: Only comparator powered with MCU in LPM mode	< 100 $\mu$ A	8 MHz	
		< 50 $\mu$ A	1 MHz	
Active Mode (Duty cycle 10%)	Active Time: All devices powered Standby Time: Only comparator powered with MCU in LPM mode	< 200 $\mu$ A	8 MHz	
		< 100 $\mu$ A	1 MHz	
Active Mode (Duty cycle 50%)	Active Time: All devices powered Standby Time: Only comparator powered with MCU in LPM mode	< 900 $\mu$ A	8 MHz	
		< 450 $\mu$ A	1 MHz	
Active Mode (Duty cycle 90%)	Active Time: All devices powered Standby Time: Only comparator powered with MCU in LPM mode	< 1.6 mA	8 MHz	

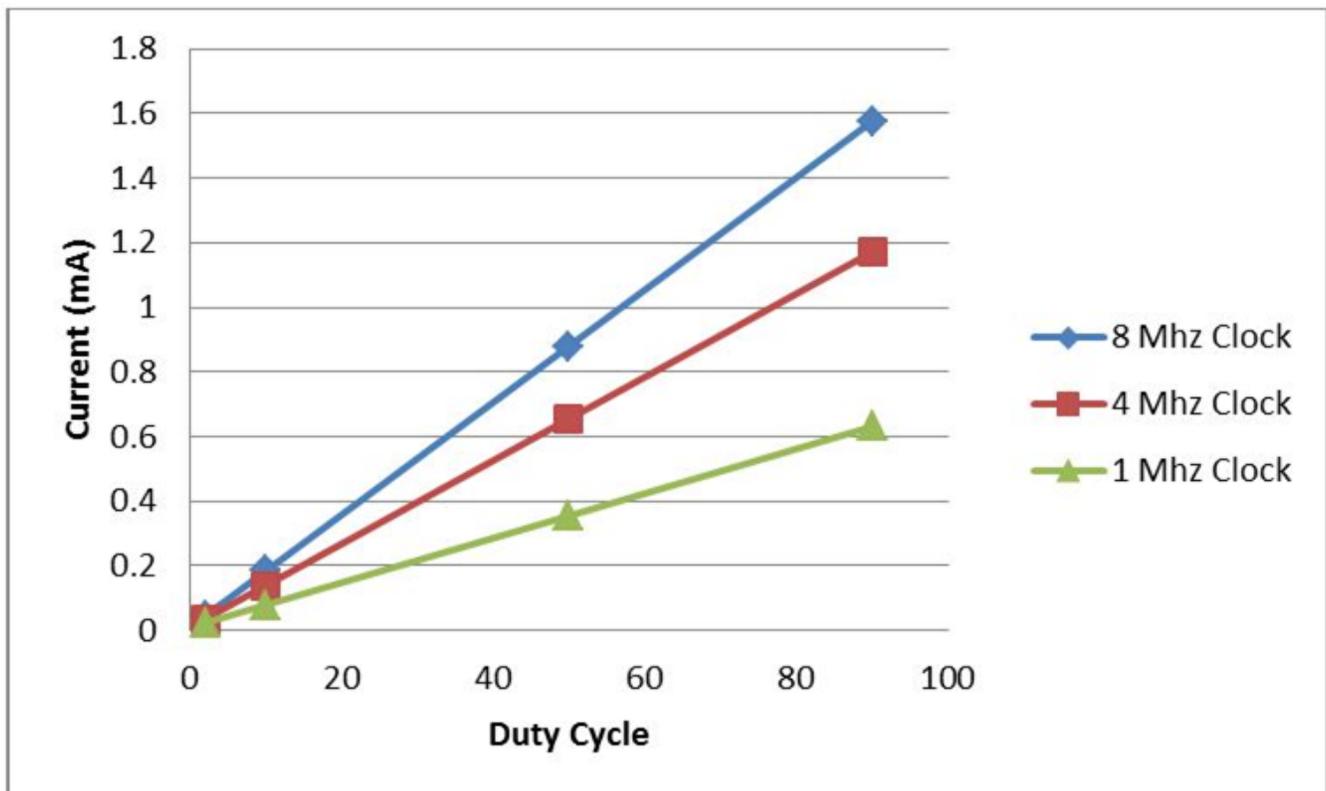


图 8. Current Consumption Plot

## 6.4 Self-Test Mode

Fault Indicator designs require in-field testing to ensure the device is working as expected and identify defective devices. The Self-Test mode was validated by placing a magnet (<http://www.digikey.com/product-detail/en/radial-magnet-inc/8031/469-1014-ND/2810356>) at a distance of 5 mm over the hall switch. An output low signal is generated by Hall switch, then the software enters Self-Test mechanism mode. In this mode the test waveform (10A) generated by DAC is fed to sensing circuitry to validate the Analog front end and software computation. The LED is toggled to indicate successful validation of device.

## 6.5 DAC Measurement

表 7 shows the current accuracy results for test waveforms generated by DAC.

表 7. DAC measurement – Current Accuracy

CURRENT (A)	CUR_L	CUR_H
1		0.50
5		0.80
10		1.00
20		1.40
30	0.32	
50	0.65	
90	1.20	

## 6.6 Temperature Sensor Measurement

The output voltage of temperature sensor was measured at room temperature, and the corresponding operating temperature was obtained from the *LMT87/LMT87-Q1 Transfer* table in the *LLMT87/LMT87-Q1 SC70/TO-92 Analog Temperature Sensors with Class-AB Output Data Sheet (SNIS170)*. The obtained temperature measurement was within  $\pm 0.3^{\circ}\text{C}$  accuracy of reference room temperature.

## 7 Design Files

### 7.1 Schematics

To download the schematics, see the design files at [TIDA-00807](#).

### 7.2 Bill of Materials

To download the bill of materials (BOM), see the design files at [TIDA-00807](#).

### 7.3 PCB Layout Recommendations

#### 7.3.1 OPA2317 Layout Guidelines

Attention to good layout practice is always recommended. Keep traces short and, when possible, use a printed circuit board (PCB) ground plane with surface-mount components placed as close to the device pins as possible.

Place a 0.1- $\mu\text{F}$  capacitor closely across the supply pins. Apply these guidelines throughout the analog circuit to improve performance and provide benefits, such as reducing the electromagnetic interference (EMI) susceptibility.

Optimize circuit layout and mechanical conditions for lowest offset voltage and precision performance. Avoid temperature gradients that create thermoelectric (Seebeck) effects in the thermocouple junctions formed from connecting dissimilar conductors. These thermally-generated potentials can be made to cancel by assuring they are equal on both input terminals. Other layout and design considerations include:

- Use low thermoelectric-coefficient conditions (avoid dissimilar metals).
- Thermally isolate components from power supplies or other heat sources.
- Shield op amp and input circuitry from air currents, such as cooling fans.

Following these guidelines reduces the likelihood of junctions being at different temperatures, which can cause thermoelectric voltages of  $0.1 \mu\text{V}/^\circ\text{C}$  or higher, depending on the materials used.

### 7.3.2 SN74LVC1G3157 Layout Guidelines

TI recommends keeping signal lines as short as possible. TI also recommends incorporating microstrip or stripline techniques when signal lines are greater than 1 inch in length. These traces must be designed with a characteristic impedance of either  $50 \Omega$  or  $75 \Omega$ , as required by the application. Do not place this device too close to high-voltage switching components, as they may interfere with the device.

### 7.3.3 TPS62740 Layout Guidelines

As for all switching power supplies, the layout is an important step in the design. Care must be taken in board layout to get the specified performance. If the layout is not carefully done, the regulator could show poor line and/or load regulation, stability issues as well as EMI problems and interference with RF circuits. It is critical to provide a low inductance, impedance ground path. Therefore, use wide and short traces for the main current paths. The input capacitor should be placed as close as possible to the IC pins VIN and GND. The output capacitor should be placed close between VOUT and GND pins. The VOUT line should be connected to the output capacitor and routed away from noisy components and traces (for example, SW line) or other noise sources. The exposed thermal pad of the package and the GND pin should be connected. See Figure 6 for the recommended PCB layout.

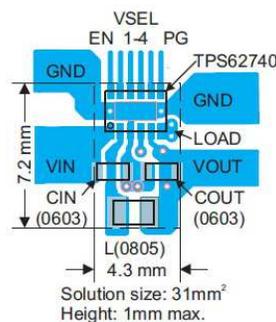


图 9. Recommended PCB Layout

### 7.3.4 Layout Prints

To download the layer plots, see the design files at [TIDA-00807](http://www.ti.com/design-files/TIDA-00807).

## 7.4 Altium Project

To download the Altium project files, see the design files at [TIDA-00807](http://www.ti.com/design-files/TIDA-00807).

## 7.5 Layout Guidelines

To download the layout guidelines, see the design files at [TIDA-00807](#).

## 7.6 Gerber Files

To download the Gerber files, see the design files at [TIDA-00807](#).

## 7.7 Assembly Drawings

To download the assembly drawings, see the design files at [TIDA-00807](#).

## 8 Software Files

To download the software files, see the design files at [TIDA-00807](#).

## 9 Related Documentation

1. [MSP430FR572x Mixed-Signal Microcontrollers Data Manual](#)
2. [OPA317, OPA2317, OPA4317 Low Offset, Rail-to-Rail I/O Operational Amplifier Precision Catalog Data Sheet](#)
3. [SN74LVC1G3157 Single-Pole Double-Throw Analog Switch Data Sheet](#)
4. [TPS6274x 360nA IQ Step Down Converter For Low Power Applications Data Sheet](#)
5. [DAC101C08xx 10-Bit Micro Power Digital-to-Analog Converter With an I2C-Compatible Interface Data Sheet](#)
6. [LM4041-N-xx Precision Micropower Shunt Voltage Reference Data Sheet](#)

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## 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

Changes from Original (December 2016) to A Revision	Page
• 已添加 在“设计资源”中添加了“DRV5032” <a href="#">设计资源</a> .....	1

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