







TMUX7234

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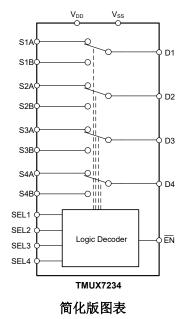
TMUX7234 具有闩锁效应抑制和 1.8V 逻辑电平的 44V、低 RON、2:1、4 通道精密开关

1 特性

- 闩锁效应抑制
- 双电源电压范围: ±4.5 V 至 ±22 V
- 单电源电压范围:4.5 V 至 44 V
- 低导通电阻:3Ω
- 低电荷注入: 3pC
- 高电流支持:400 mA(最大值)
- - 40°C 至 +125°C 工作温度
- 1.8V 逻辑兼容输入
- 失效防护逻辑
- 轨到轨运行
- 双向信号路径
- 先断后合开关

2 应用

- 工厂自动化和控制
- 流量变送器
- 可编程逻辑控制器 (PLC)
- 模拟输入模块
- 数据采集系统 (DAQ)
- 半导体测试设备
- 电池测试设备
- 超声波扫描仪
- 患者监护和诊断
- 光纤网络
- 光学测试设备
- 有线网络
- 远程射频单元 (RRU)
- 有源天线系统 mMIMO (AAS)



3 说明

TMUX7234 是支持闩锁效应抑制的互补金属氧化物半导体 (CMOS) 多路复用器。TMUX7234 包含四个独立 控制的 SPDT 开关和一个用于启用或禁用全部四个通 道的 \overline{EN} 引脚。该器件支持单电源(4.5V 至 44V)、 双电源(±4.5V 至 ±22V)或非对称电源(例如,V_{DD} = 12V, V_{SS} = -5V)。TMUX7234 可在源极 (Sx) 和 漏极 (D) 引脚上支持从 V_{SS} 到 V_{DD} 范围的双向模拟和 数字信号。

所有逻辑控制输入均支持 1.8 V 到 V_{DD} 的逻辑电平, 因此,当器件在有效电源电压范围内运行时,可确保 TTL 和 CMOS 逻辑兼容性。失效防护逻辑电路允许先 在控制引脚上施加电压,然后在电源引脚上施加电压, 从而保护器件免受潜在的损害。

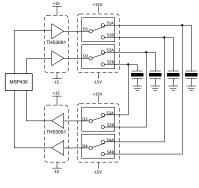
TMUX72xx 系列具有闩锁效应抑制特性,可防止器件 内寄生结构之间通常由过压事件引起的大电流不良事 件。闩锁状态通常会一直持续到电源轨关闭为止,并可 能导致器件失效。闩锁效应抑制特性使得 TMUX72xx 系列开关和多路复用器能够在恶劣的环境中使用。

封装信息

器件型号 封装 ⁽¹⁾		封装尺寸 ⁽²⁾
TMUX7234	PW(TSSOP,20)	6.5 mm × 6.4 mm
	RRQ (WQFN , 20)	4 mm × 4 mm

(1) 如需了解所有可用封装,请参阅数据表末尾的封装选项附录。

(2) 封装尺寸(长×宽)为标称值,并包括引脚(如适用)。



应用示意图

▲ 本资源的原文使用英文撰写。为方便起见,TI提供了译文;由于翻译过程中可能使用了自动化工具,TI不保证译文的准确性。为确认 准确性,请务必访问 ti.com 参考最新的英文版本(控制文档)。



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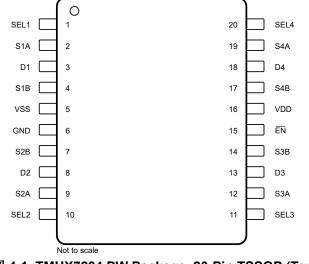
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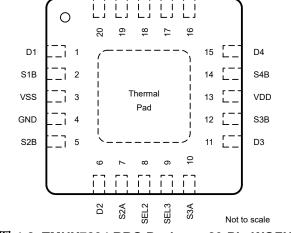
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S4A

4 Pin Configuration and Functions





SEL1

N N N

S1A

SEL4





	PIN		TYPE ⁽¹⁾	DESCRIPTION ⁽²⁾		
NAME			ITPE	DESCRIPTION		
SEL1	1	19	I	Logic control input 1; has internal pull-down resistor. Controls switch 1 (see # 7.5).		
S1A	2	20	I/O	Source pin 1A. Can be an input or output.		
D1	3	1	I/O	Drain pin 1. Can be an input or output.		
S1B	4	2	I/O	Source pin 1B. Can be an input or output.		
VSS	5	3	Р	Negative power supply. This pin has the most negative power-supply potential. This pin can be connected to ground in single supply applications. Connect a decoupling capacitor ranging from 0.1 μ F to 10 μ F between VSS and GND for reliable operation.		
GND	6	4	Р	Ground (0 V) reference.		
S2B	7	5	I/O	Source pin 2B. Can be an input or output.		
D2	8	6	I/O	Drain pin 2. Can be an input or output.		
S2A	9	7	I/O	Source pin 2A. Can be an input or output.		
SEL2	10	8	I	Logic control input 2; has internal pull-down resistor. Controls switch 2 (see # 7.5).		
SEL3	11	9	I	Logic control input 3; has internal pull-down resistor. Controls switch 3 (see # 7.5).		
S3A	12	10	I/O	Source pin 3A. Can be an input or output.		
D3	13	11	I/O	Drain pin 3. Can be an input or output.		
S3B	14	12	I/O	Source pin 3B. Can be an input or output.		
EN	15	18	I	Active low logic enable; has internal pull-down resistor. The SELx logic inputs determine switch connections when this pin is low (see \ddagger 7.5).		
VDD	16	13	Р	Positive power supply. This pin has the most positive power-supply potential. For reliable operation, connect a decoupling capacitor ranging from 0.1 μ F to 10 μ F between VDD and GND.		
S4B	17	14	I/O	Source pin 4B. Can be an input or output.		
D4	18	15	I/O	Drain pin 4. Can be input or output		
S4A	19	16	I/O	Source pin 4A. Can be an input or output.		
SEL4	20	17	I	Logic control input 4, has internal pull-down resistor. Controls switch 4 (see # 7.5).		
Thermal P	Pad .		_	The thermal pad is not connected internally. There is no requirement to solder this pad. If connected, it is recommended to leave the pad floating or tied to GND.		

表 4-1. Pin Functions TMUX7234

(1) I = input, O = output, I/O = input and output, P = power.

(2) Refer to \ddagger 7.4 for what to do with unused pins.

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5 Specifications

5.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾ (2)

		MIN	MAX	UNIT
V _{DD} - V _{SS}			48	V
V _{DD}	Supply voltage	- 0.5	48	V
V _{SS}		- 48	0.5	V
V_{SEL} or V_{EN}	Logic control input pin voltage (SELx, EN)	- 0.5	48	V
I _{SEL} or I _{EN}	Logic control input pin current (SELx, EN)	- 30	30	mA
V_{S} or V_{D}	Source or drain voltage (SxA, SxB, Dx)	V _{SS} - 0.5	V _{DD} +0.5	V
I _{IK}	Diode clamp current ⁽³⁾	- 30	30	mA
I _S or I _{D (CONT)}	Source or drain continuous current (SxA, SxB, Dx)		I _{DC} ± 10 % ⁽⁴⁾	mA
T _A	Ambient temperature	- 55	150	°C
T _{stg}	Storage temperature	- 65	150	°C
TJ	Junction temperature		150	°C
P _{tot}	Total power dissipation (QFN package) ⁽⁵⁾		1680	mW

(1) Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages are with respect to ground, unless otherwise specified.

(3) Pins are diode-clamped to the power-supply rails. Over voltage signals must be voltage and current limited to maximum ratings.

(4) Refer to *Source or Drain Continuous Current* table for I_{DC} specifications.

(5) For QFN package: P_{tot} derates linearily above $T_A = 70^{\circ}C$ by 24.8mW/°C.

5.2 ESD Ratings

			VALUE	UNIT
V _(ESD) Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ⁽¹⁾	±2000	V	
V(ESD)		Charged device model (CDM), per ANSI/ESDA/ JEDEC JS-002, all pins ⁽²⁾		v

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

5.3 Thermal Information

		TMU	X7234	
	THERMAL METRIC ⁽¹⁾	PW (TSSOP)	RRQ (WQFN)	UNIT
		20 PINS	20 PINS	
R _{0 JA}	Junction-to-ambient thermal resistance	74.7	40.5	°C/W
R ₀ JC(top)	Junction-to-case (top) thermal resistance	19.9	24.2	°C/W
R _{0 JB}	Junction-to-board thermal resistance	32.3	16.4	°C/W
ΨJT	Junction-to-top characterization parameter	0.7	0.2	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	31.7	16.4	°C/W
R ₀ JC(bot)	Junction-to-case (bottom) thermal resistance	N/A	2.8	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.



5.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM MAX	UNIT
$V_{DD} - V_{SS}$ ⁽¹⁾	Power supply voltage differential	4.5	44	V
V _{DD}	Positive power supply voltage	4.5	44	V
V_{S} or V_{D}	Signal path input/output voltage (source or drain pin) (SxA, SxB, Dx)	V _{SS}	V _{DD}	V
V_{SEL} or V_{EN}	Address or enable pin voltage	0	44	V
I _S or I _{D (CONT)}	Source or drain continuous current (SxA, SxB, Dx)		I _{DC} ⁽²⁾	mA
T _A	Ambient temperature	- 40	125	°C

5.5 Source or Drain Continuous Current

at supply voltage of $V_{DD} \pm 10\%$, $V_{SS} \pm 10\%$ (unless otherwise noted)

CONT	INUOUS CURRENT PER CHANNEL	T 25°C	T₄ = 85°C	T₄ = 125°C	UNIT
PACKAGE	TEST CONDITIONS		TA = 85 C	1 _A = 125 C	UNIT
	+44 V Single Supply ⁽¹⁾	350	230	129	mA
	±15 V Dual Supply	360	235	130	mA
PW (TSSOP)	+12 V Single Supply	260	177	108	mA
	±5 V Dual Supply	255	175	105	mA
	+5 V Single Supply	170	129	80	mA
	+44 V Single Supply ⁽¹⁾	400	230	120	mA
	±15 V Dual Supply	400	230	120	mA
RRQ (WQFN)	+12 V Single Supply	300	180	100	mA
	±5 V Dual Supply	300	180	100	mA
	+5 V Single Supply	240	150	85	mA

(1) Specified for nominal supply voltage only.



5.6 ±15 V Dual Supply: Electrical Characteristics

 $V_{DD} = +15 \text{ V} \pm 10\%, V_{SS} = -15 \text{ V} \pm 10\%, \text{ GND} = 0 \text{ V} \text{ (unless otherwise noted)}$ Typical at $V_{DD} = +15 \text{ V}, V_{SS} = -15 \text{ V}, T_A = 25 ^{\circ}\mathbb{C}$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	TA	MIN	TYP	MAX	UNIT
ANALOG	SWITCH						
		V _S = - 10 V to +10 V	25°C		3.6	5.5	Ω
R _{ON}	On-resistance	$I_{\rm D} = -10 {\rm mA}$	- 40°C to +85°C			7.1	Ω
		Refer to On-Resistance	- 40°C to +125°C			8.4	Ω
		V _S = - 10 V to +10 V	25°C		0.2	0.7	Ω
ΔR_{ON}	On-resistance mismatch between channels	I _D = - 10 mA	- 40°C to +85°C			0.8	Ω
	Refer to On-Resistance	- 40°C to +125°C			0.9	Ω	
		V _S = - 10 V to +10 V	25°C		0.4	1.5	Ω
R _{ON FLAT}	On-resistance flatness	I _S = - 10 mA	- 40°C to +85°C			1.7	Ω
		Refer to On-Resistance	- 40°C to +125°C		3.6 5.5 7.1 8.4 0.2 0.7 0.8 0.9 0.4 1.5 1.7 1.9 0.015 0.01 0.4 1 8 0.02 0.5 4 12 0.02 0.5 4 8 0.02 0.5 4 12 0.02 0.5 4 12 0.02 0.5 4 0.8 0.02 0.5 4 0.8 0.6 2	Ω	
R _{ON DRIFT}	On-resistance drift	V _S = 0 V, I _S = - 10 mA Refer to On-Resistance	- 40°C to +125°C		0.015		Ω/°C
		V _{DD} = 16.5 V, V _{SS} = - 16.5 V	25°C	- 0.4	0.01	0.4	nA
la (a)	Source off leakage current ⁽¹⁾	Switch state is off V _S = +10 V / - 10 V	- 40°C to +85°C	- 1		1	nA
I _{S(OFF)}	Source on leakage current	$V_D = -10 V / + 10 V$ Refer to $\# 6.2$	- 40°C to +125°C	- 8	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	nA	
		V _{DD} = 16.5 V, V _{SS} = - 16.5 V	25°C	- 0.5	0.02	0.5	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	Switch state is off V _S = +10 V / - 10 V V _D = - 10 V / + 10 V Refer to 节 6.2	- 40°C to +85°C	- 4		4	nA
			- 40°C to +125°C	- 12		12	nA
		V _{DD} = 16.5 V, V _{SS} = - 16.5 V	25°C	- 0.5	0.02	0.5	nA
S(ON)	Channel on leakage current ⁽²⁾	Switch state is on	- 40°C to +85°C	- 4	$\begin{array}{c c c c c c c c } & & & & & & & & & & & & & & & & & & &$	nA	
I _{D(ON)}		V _S = V _D = ±10 V Refer to 节 6.3	- 40°C to +125°C	- 8		8	nA
LOGIC IN	PUTS (SEL / EN pins)						
VIH	Logic voltage high		- 40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		- 40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		- 40°C to +125°C		0.6	2	μA
IIL	Input leakage current		- 40°C to +125°C	- 0.1	- 0.005		μA
C _{IN}	Logic input capacitance		- 40°C to +125°C		3		pF
POWER	SUPPLY					I	
			25°C		45	70	μA
DD	V _{DD} supply current	V_{DD} = 16.5 V, V_{SS} = - 16.5 V Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +85°C			80	μA
			- 40°C to +125°C			95	μA
			25°C		8	25	μA
I _{SS}	V _{SS} supply current	V_{DD} = 16.5 V, V_{SS} = - 16.5 V Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +85°C			30	μA
			- 40°C to +125°C			40	μA

(1) When V_S is positive, V_D is negative. Or when V_S is negative, V_D is positive.

(2) When V_S is at a voltage potential, V_D is floating. Or when V_D is at a voltage potential, V_S is floating.



5.7 ±15 V Dual Supply: Switching Characteristics

 $\label{eq:VDD} V_{DD} = +15 \ V \pm 10\%, \ V_{SS} = -15 \ V \pm 10\%, \ GND = 0 \ V \ (unless \ otherwise \ noted)$ Typical at $V_{DD} = +15 \ V, \ V_{SS} = -15 \ V, \ T_A = 25 \ ^{\circ}C \ \ (unless \ otherwise \ noted)$

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
		V _S = 10 V	25°C		90	180	ns
t _{TRAN}	Transition time from control input	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			190	ns
		Refer to Transition Time	- 40°C to +125°C			0 180 190 200 0 180 190 210 0 140 150 160 0 160 0 160 0 160 0 160 0 160 0 160 0 160 0 160 0 160 160 160 0 160 3 2 5 160 3 2 5 160 3 2 4 160 6 8	ns
		V _S = 10 V	25°C		110	180 190 200 180 190 210 140 150	ns
t _{ON (EN)}	Turn-on time from enable	$R_{L} = 300 \ \Omega$, $C_{L} = 35 \text{ pF}$	- 40°C to +85°C			190	ns
		Refer to ^{††} 6.5	- 40°C to +125°C	90 180 190 200 110 180 190 210 80 140 100 150 100 50 1 0.16 0.16 0.16 0.16 0.16 0.16 3 - 82 - - 62 - 100 - - 105 100 - 0.3 - - 48 0.0004 16 28	ns		
		V _S = 10 V	25°C		$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	ns	
t _{OFF (EN)}	Turn-off time from enable	$R_{L} = 300 \ \Omega$, $C_{L} = 35 \text{ pF}$	- 40°C to +85°C			150	ns
		Refer to ^{††} 6.5	- 40°C to +125°C			160	ns
		V _S = 10 V,	25°C		50		ns
t _{BBM}	Break-before-make time delay	$R_{L} = 300 \ \Omega$, $C_{L} = 35 \text{ pF}$	- 40°C to +85°C	1			ns
		Refer to Break-Before-Make	- 40°C to +125°C	1			ns
		$V_{\rm rise}$ time = 1.0	25°C		0.16		ms
T _{ON (VDD)}	Device turn on time	V_{DD} rise time = 1µs R _L = 300 Ω , C _L = 35 pF	- 40°C to +85°C		0.16		ms
()	(V _{DD} to output)	Refer to Turn-on (VDD) Time	- 40°C to +125°C		0.16		ms
t _{PD}	Propagation delay	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ Refer to $\frac{1}{10} \frac{6.8}{100}$	25°C		450		ps
Q _{INJ}	Charge injection	V _D = 0 V, C _L = 100 pF Refer to 节 6.9	25°C		3		рС
O _{ISO}	Off-isolation	$R_L = 50 \ \Omega$, $C_L = 5 pF$ $V_S = 0 V$, $f = 100 kHz$ Refer to Off Isolation	25°C		- 82		dB
O _{ISO}	Off-isolation	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 0 \ V$, $f = 1 \ MHz$ Refer to Off Isolation	25°C		- 62		dB
X _{TALK}	Crosstalk	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 0 \ V$, $f = 1 \ MHz$ Refer to Crosstalk	25°C		- 105		dB
BW	- 3dB Bandwidth	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 0 \ V$ Refer to Bandwidth	25°C		100		MHz
IL	Insertion loss	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 0 \ V$, f = 1 MHz	25°C		- 0.3		dB
ACPSRR	AC Power Supply Rejection Ratio	V_{PP} = 0.62 V on V _{DD} and V _{SS} R _L = 10 M Ω , C _L = 5 pF, f = 1 MHz Refer to 节 6.14	25°C		- 48		dB
THD+N	Total Harmonic Distortion + Noise	V_{PP} = 15 V, V_{BIAS} = 0 V R_L = 10 k Ω , C_L = 5 pF, f = 20 Hz to 20 kHz Refer to $\#$ 6.13	25°C		0.0004		%
C _{S(OFF)}	Source off capacitance	V _S = 0 V, f = 1 MHz	25°C		16		pF
C _{D(OFF)}	Drain off capacitance	V _S = 0 V, f = 1 MHz	25°C		28		pF
C _{S(ON),} C _{D(ON)}	On capacitance	V _S = 0 V, f = 1 MHz	25°C		77		pF



5.8 ±20 V Dual Supply: Electrical Characteristics

 $V_{DD} = +20 \text{ V} \pm 10\%, V_{SS} = -20 \text{ V} \pm 10\%, \text{ GND} = 0 \text{ V} \text{ (unless otherwise noted)}$ Typical at $V_{DD} = +20 \text{ V}, V_{SS} = -20 \text{ V}, T_A = 25 ^{\circ}\text{C}$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG	SWITCH						
		V _S = - 15 V to +15 V	25°C		3.2	5.4	Ω
R _{ON}	On-resistance	I _D = - 10 mA	- 40°C to +85°C			6.7	Ω
		Refer to On-Resistance	- 40°C to +125°C			7.9	Ω
		V _S = - 15 V to +15 V	25°C		0.2	0.7	Ω
ΔR_{ON}	On-resistance mismatch between channels	I _D = - 10 mA	- 40°C to +85°C			0.8	Ω
		Refer to On-Resistance	- 40°C to +125°C			0.9	Ω
		V _S = - 15 V to +15 V	25°C		0.6	1.5	Ω
R _{ON FLAT}	On-resistance flatness	I _S = − 10 mA	- 40°C to +85°C			1.7	Ω
		Refer to On-Resistance	- 40°C to +125°C	3.2 5°C $5°C$ 0.2 $25°C$ 0.2 $5°C$ 0.2 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 0.014 -1 0.02 $5°C$ -1 $25°C$ -12 $25°C$ -12 $25°C$ -12 -1 0.04 $5°C$ -4 $25°C$ -30 -1 0.04 $5°C$ -4 $25°C$ -30 $25°C$ 0.06 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 0.6 $25°C$ 3 $50°C$ 3 $5°C$ 1.3 $5°C$ 10	1.9	Ω	
R _{ON DRIFT}	On-resistance drift	V _S = 0 V, I _S = - 10 mA Refer to On-Resistance	- 40°C to +125°C		0.014		Ω/°C
		V _{DD} = 22 V, V _{SS} = - 22 V	25°C	- 1	0.02	1	nA
I	Source off leakage current ⁽¹⁾	Switch state is off V _S = +15 V / - 15 V	- 40°C to +85°C	- 2		2 5.4 6.7 7.9 2 0.7 0.8 0.9 5 1.5 1.7 1.9 4 1 2 1 2 1 2 1 4 30 4 1 4 30 4 1 4 30 4 30 4 30 4 30 5 2 5 3 0 80 95 110	nA
'S(OFF)	Source on leakage current	$V_D = -15 V / + 15 V$ Refer to $\ddagger 6.2$	- 40°C to +125°C	- 12	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	12	nA
		V _{DD} = 22 V, V _{SS} = - 22 V	25°C	- 1	0.04	1	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	Switch state is off V _S = +15 V / - 15 V V _D = - 15 V / + 15 V Refer to 节 6.2	- 40°C to +85°C	- 4		4	nA
			- 40°C to +125°C	- 30		30	nA
		$V_{DD} = 22 V, V_{SS} = -22 V$ Switch state is on $V_S = V_D = \pm 15 V$ Refer to $\ddagger 6.3$	25°C	- 1	0.04	1	nA
I _{S(OFF)} Sc I _{D(OFF)} Dr I _{S(ON)} Cr I _{D(ON)} Cr LOGIC INPU V _{IH} Lo V _{IL} Lo	Channel on leakage current ⁽²⁾		- 40°C to +85°C	- 4		4	nA
			- 40°C to +125°C	- 30		30	nA
LOGIC IN	IPUTS (SEL / EN pins)						
V _{IH}	Logic voltage high		- 40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		- 40°C to +125°C	0		0.8	V
IIH	Input leakage current		- 40°C to +125°C		0.6	2	μA
IIL	Input leakage current		- 40°C to +125°C	- 0.1	- 0.005		μA
C _{IN}	Logic input capacitance		- 40°C to +125°C		3		pF
POWER	SUPPLY						
			25°C		50	80	μA
DD	V _{DD} supply current	V_{DD} = 22 V, V_{SS} = -22 V Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +85°C			95	μA
			- 40°C to +125°C			110	μA
			25°C		10	30	μA
I _{SS}	V _{SS} supply current	$V_{DD} = 22 \text{ V}, V_{SS} = -22 \text{ V}$	- 40°C to +85°C			1 30 1 4 30 4 4 4 4 30 4 4 4 30 5 5 110 30 35	μA
		Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +125°C			45	μA

When V_S is positive, V_D is negative. Or when V_S is negative, V_D is positive. (1)

(2) When V_S is at a voltage potential, V_D is floating. Or when V_D is at a voltage potential, V_S is floating.



5.9 ±20 V Dual Supply: Switching Characteristics

 V_{DD} = +20 V ± 10%, V_{SS} = - 20 V ±10%, GND = 0 V (unless otherwise noted) Typical at V_{DD} = +20 V, V_{SS} = - 20 V, T_A = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
		V _S = 10 V	25°C		90	190	ns
t _{TRAN}	Transition time from control input	$R_{L} = 300 \ \Omega$, $C_{L} = 35 \ pF$	- 40°C to +85°C			200	ns
		Refer to Transition Time	- 40°C to +125°C			210	ns
		V _S = 10 V	25°C		110	190	ns
t _{ON (EN)}	Turn-on time from enable	$R_{L} = 300 \ \Omega$, $C_{L} = 35 \text{ pF}$	- 40°C to +85°C			200	ns
()		Refer to 节 6.5	- 40°C to +125°C			210	ns
		V _S = 10 V	25°C		75	140	ns
t _{OFF (EN)}	Turn-off time from enable	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			150	ns
		Refer to 节 6.5	- 40°C to +125°C			160	ns
		V _S = 10 V,	25°C		50		ns
t _{BBM}	Break-before-make time delay	$R_{L} = 300 \Omega$, $C_{L} = 35 pF$	- 40°C to +85°C	1			ns
		Refer to Break-Before-Make	- 40°C to +125°C	1			ns
		V _{DD} rise time = 1µs	25°C		0.16		ms
T _{ON (VDD)}	Device turn on time (V _{DD} to output)	$R_L = 300 \ \Omega$, $C_L = 35 \text{ pF}$	- 40°C to +85°C		0.16		ms
. ,	Refer to $\neg 6.5$ -40° C to $+125^{\circ}$ CPropagation delay $R_L = 50 \ \Omega$, $C_L = 5 \ pF$ Refer to $\neg 6.8$ 25° CCharge injection $V_D = 0 \ V, \ C_L = 100 \ pF$ 25° C		0.16		ms		
t _{PD}	Propagation delay		25°C		470		ps
Q _{INJ}	Charge injection	V _D = 0 V, C _L = 100 pF Refer to 节 6.9	25°C		3		рС
O _{ISO}	Off-isolation	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 0 V$, f = 100 kHz Refer to Off Isolation	25°C		- 82		dB
O _{ISO}	Off-isolation	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 0 V$, $f = 1 MHz$ Refer to Off Isolation	25°C		- 62		dB
X _{TALK}	Crosstalk	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 0 V$, $f = 1MHz$ Refer to Crosstalk	25°C		- 105		dB
BW	- 3dB Bandwidth	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 0 V$ Refer to Bandwidth	25°C		95		MHz
IL	Insertion loss	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 0 \ V$, f = 1 MHz	25°C		- 0.25		dB
ACPSRR	AC Power Supply Rejection Ratio	V_{PP} = 0.62 V on V_{DD} and V_{SS} R _L = 10 M Ω , C _L = 5 pF, f = 1 MHz Refer to \ddagger 6.14	25°C		- 48		dB
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 20 V, V_{BIAS} = 0 V$ $R_L = 10 k \Omega$, $C_L = 5 pF$, f = 20 Hz to 20 kHz Refer to $\# 6.13$	25°C		0.002		%
C _{S(OFF)}	Source off capacitance	V _S = 0 V, f = 1 MHz	25°C		16		pF
C _{D(OFF)}	Drain off capacitance	V _S = 0 V, f = 1 MHz	25°C		26		pF
C _{S(ON),} C _{D(ON)}	On capacitance	V _S = 0 V, f = 1 MHz	25°C		77		pF



5.10 44 V Single Supply: Electrical Characteristics

V_{DD} = +44 V, V_{SS} = 0 V, GND = 0 V (unless otherwise noted)

Typical at V_{DD} = +44 V, V_{SS} = 0 V, T_A = 25°C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT
ANALOG	SWITCH						
		$V_{\rm S} = 0$ V to 40 V	25°C		3	5.8	Ω
R _{ON}	On-resistance	I _D = - 10 mA	- 40°C to +85°C			7.2	Ω
		Refer to On-Resistance	- 40°C to +125°C			8.9	Ω
		$V_{S} = 0 V \text{ to } 40 V$	25°C		0.2	0.7	Ω
ΔR_{ON}	On-resistance mismatch between channels	I _D = - 10 mA	- 40°C to +85°C			0.8	Ω
		Refer to On-Resistance	- 40°C to +125°C			0.9	Ω
		V _S = 0 V to 40 V	25°C		1.5	2	Ω
R _{ON FLAT}	On-resistance flatness	I _D = - 10 mA	- 40°C to +85°C			2.5	Ω
		Refer to On-Resistance	- 40°C to +125°C			3.3	Ω
R _{ON DRIFT}	On-resistance drift	V _S = 22 V, I _S = − 10 mA Refer to On-Resistance	- 40°C to +125°C		0.012		Ω/°C
		V _{DD} = 44 V, V _{SS} = 0 V	25°C	- 1	0.02	1	nA
I _{S(OFF)}	Source off leakage current ⁽¹⁾	Switch state is off $V_S = 40 V / 1 V$	- 40°C to +85°C	- 4		4	nA
.5(011)		$V_{\rm D} = 1 \text{ V} / 40 \text{ V}$ Refer to $\frac{1}{7} 6.2$	- 40°C to +125°C	- 20		20	nA
		V _{DD} = 44 V, V _{SS} = 0 V	25°C	- 1	0.04	1	nA
I _{D(OFF)}	Drain off leakage current ⁽¹⁾	Switch state is off $V_S = 40 V / 1 V$	- 40°C to +85°C	- 8		8	nA
D(OFF)	gg_	V _D = 1 V / 40 V Refer to 节 6.2	- 40°C to +125°C	- 40		40	nA
		V _{DD} = 44 V, V _{SS} = 0 V	25°C	- 1	0.04	1	nA
I _{S(ON)} I _{D(ON)}	Channel on leakage current ⁽²⁾	Switch state is on $V_S = V_D = 40 V \text{ or } 1 V$	- 40°C to +85°C	- 8		8	nA
D(ON)		Refer to $\# 6.3$	- 40°C to +125°C	- 40		40	nA
LOGIC IN	PUTS (SEL / EN pins)						
V _{IH}	Logic voltage high		- 40°C to +125°C	1.3		44	V
V _{IL}	Logic voltage low		- 40°C to +125°C	0		0.8	V
I _{IH}	Input leakage current		- 40°C to +125°C		0.6	2	μA
IIL	Input leakage current		- 40°C to +125°C	- 0.1	- 0.005		μA
C _{IN}	Logic input capacitance		- 40°C to +125°C		3		pF
POWER S	UPPLY	1					
			25°C		70	110	μA
I _{DD}	V _{DD} supply current	V_{DD} = 44 V, V_{SS} = 0 V Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +85°C			118	μA
			- 40°C to +125°C			140	μA

When V_S is 40 V, V_D is 1 V. Or when V_S is 1 V, V_D is 40 V.
 When V_S is at a voltage potential, V_D is floating. Or when V_D is at a voltage potential, V_S is floating.



5.11 44 V Single Supply: Switching Characteristics

 $\label{eq:VDD} \begin{array}{l} V_{DD} = +44 \ \text{V}, \ \text{V}_{SS} = 0 \ \text{V}, \ \text{GND} = 0 \ \text{V} \ (\text{unless otherwise noted}) \\ \hline \text{Typical at } V_{DD} = +44 \ \text{V}, \ \text{V}_{SS} = 0 \ \text{V}, \ \text{T}_{A} = 25 \ ^{\circ}\text{C} \ \ (\text{unless otherwise noted}) \end{array}$

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT			
		V _S = 18 V	25°C		90	200	ns			
t _{TRAN}	Transition time from control input	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			220	ns			
		Refer to Transition Time	- 40°C to +125°C			240	ns			
t _{TRAN} Transition time from control input t _{ON (EN)} Turn-on time from enable		V _S = 18 V	25°C		100	200	ns			
t _{ON (EN)}	Turn-on time from enable	$R_L = 300 \ \Omega$, $C_L = 35 \text{ pF}$	- 40°C to +85°C			220	ns			
		Refer to $\# 6.5$	- 40°C to +125°C			240	ns			
		V _S = 18 V	25°C		90	180	ns			
t _{OFF (EN)}	Turn-off time from enable	$R_L = 300 \ \Omega$, $C_L = 35 \text{ pF}$	- 40°C to +85°C			200	200 ns 220 ns 240 ns 180 ns 200 ns 300 ms 300 ms 300 pc 300 dB 300 dB 300 dB			
		Refer to $\# 6.5$	- 40°C to +125°C			220	ns			
		V _S = 18 V,	25°C		45		ns			
t _{BBM}	Break-before-make time delay	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C	1			ns			
		Refer to Break-Before-Make	- 40°C to +125°C	1			ns			
		V _{DD} rise time = 1µs	25°C		0.13		ms			
T _{ON (VDD)}	Device turn on time (V _{DD} to output)	$R_L = 300 \ \Omega$, $C_L = 35 \text{ pF}$	- 40°C to +85°C		0.13	ms				
		Refer to Turn-on (VDD) Time	- 40°C to +125°C							
t _{PD}	Propagation delay	R_L = 50 Ω , C_L = 5 pF Refer to $\frac{1}{7}$ 6.8	25°C		570		ps			
Q _{INJ}	Charge injection	V_{D} = 22 V, C _L = 100 pF Refer to $\frac{1}{10}$ 6.9	25°C		3		рС			
O _{ISO}	Off-isolation	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 6 V$, $f = 100 kHz$ Refer to Off Isolation	25°C		- 82		dB			
O _{ISO}	Off-isolation	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 6 \ V$, $f = 1 \ MHz$ Refer to Off Isolation	25°C		- 62		dB			
X _{TALK}	Crosstalk	$R_L = 50 \ \Omega$, $C_L = 5 pF$ $V_S = 6 V$, $f = 1MHz$ Refer to Crosstalk	25°C		- 105		dB			
BW	- 3dB Bandwidth	$R_L = 50 \Omega$, $C_L = 5 pF$ $V_S = 6 V$ Refer to Bandwidth	25°C		92		MHz			
IL	Insertion loss	$R_L = 50 \ \Omega$, $C_L = 5 \text{ pF}$ $V_S = 6 \text{ V, f} = 1 \text{ MHz}$	25°C		- 0.3		dB			
ACPSRR	AC Power Supply Rejection Ratio	V_{PP} = 0.62 V on V _{DD} and V _{SS} R _L = 10 M Ω , C _L = 5 pF, f = 1 MHz Refer to 节 6.14	25°C		- 45		dB			
C _{S(OFF)}	Source off capacitance	V _S = 22 V, f = 1 MHz	25°C		16		pF			
C _{D(OFF)}	Drain off capacitance	V _S = 22 V, f = 1 MHz	25°C		28		pF			
C _{S(ON),} C _{D(ON)}	On capacitance	V _S = 22 V, f = 1 MHz	25°C		77		pF			



5.12 12 V Single Supply: Electrical Characteristics

 V_{DD} = +12 V ± 10%, V_{SS} = 0 V, GND = 0 V (unless otherwise noted)

Typical at V_{DD} = +12 V, V_{SS} = 0 V, T_A = 25 °C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT			
$ \begin{array}{ c c c c c c } \hline \text{ANALOG SWITCH} \\ \hline \text{R}_{\text{R}_{\text{ON}}} & On-resistance & V_{\text{S}} = 0 \ V \ to \ 10 \ V \\ h_{\text{D}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{A}_{\text{R}_{\text{ON}}} & On-resistance \ \text{mismatch between} \\ \hline \text{channels} & V_{\text{D}} = 0 \ \text{to } 10 \ V \\ h_{\text{D}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{channels} & V_{\text{D}} = 0 \ \text{to } 10 \ V \\ h_{\text{D}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{channels} & V_{\text{S}} = 0 \ V \ to \ 10 \ V \\ h_{\text{D}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{channels} & V_{\text{S}} = 0 \ V \ to \ 10 \ V \\ h_{\text{D}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{channels} & V_{\text{S}} = 0 \ V \ to \ 10 \ V \\ h_{\text{S}} = -10 \ \text{mA} \\ \hline \text{Refer to } On-\text{Resistance} \\ \hline \text{chance} & -40^{\circ}\text{C} \ to \ +85^{\circ}\text{C} & 0.2 \ 0.7 \ 0.8 \ 0.9 \\ \hline \text{chance} & -40^{\circ}\text{C} \ to \ +85^{\circ}\text{C} & 0.2 \ 0.9 \ 0.9 \ 0.0 \\ \hline \text{chance} & -40^{\circ}\text{C} \ to \ +125^{\circ}\text{C} & 0.9 \ 0.0 \\ \hline \text{chance} & -40^{\circ}\text{C} \ to \ +125^{\circ}\text{C} & 0.025 \ 0.025 \ 0.02 \ 0.5 \ 0.02 \ 0$										
		$V_{0} = 0 V t_{0} 10 V$	25°C		6.2	12	Ω			
R _{ON}	On-resistance	-	- 40°C to +85°C			15	Ω			
		Refer to On-Resistance	- 40°C to +125°C			18	Ω			
		$V_{c} = 0 V to 10 V$	25°C		0.2	0.7	Ω			
ΔR_{ON}	-	5	- 40°C to +85°C			0.8	Ω			
		Refer to On-Resistance	- 40°C to +125°C			0.9	Ω			
		$V_{s} = 0 V \text{ to } 10 V$	25°C		2.4	3.6	Ω			
R _{ON FLAT}	On-resistance flatness	I _S = - 10 mA	- 40°C to +85°C			3.9	Ω			
		Refer to On-Resistance	- 40°C to +125°C			4.8	Ω			
R _{ON DRIFT}	On-resistance drift		- 40°C to +125°C		0.025		Ω/°C			
			25°C	- 0.4	0.01	0.4	nA			
S(OFF)	Source off leakage current ⁽¹⁾		- 40°C to +85°C	- 1		15 18 0.7 0.8 0.9 3.6 3.9 4.8 0.4 1 8 0.5 4 12 0.5 4 8 0.5 4 0.5 4 0.5 4 0.5 4 0.5 4 0.5 4 8 0.5 4 0.8	nA			
		V _D = 1 V / 10 V	- 40°C to +125°C	- 8	0.02	8	nA			
			25°C	- 0.5	0.02	0.5	nA			
	Drain off leakage current ⁽¹⁾		- 40°C to +85°C	- 4		4	nA			
-D(011)		V _D = 1 V / 10 V	- 40°C to +125°C	- 12		12	nA			
			25°C	- 0.5	0.02	0.5	nA			
. ` ′	Channel on leakage current ⁽²⁾		- 40°C to +85°C	- 4		4	nA			
-D(ON)			- 40°C to +125°C	- 8		8	nA			
LOGIC IN	PUTS (SEL / EN pins)									
V _{IH}	Logic voltage high		- 40°C to +125°C	1.3		44	V			
V _{IL}	Logic voltage low		- 40°C to +125°C	0		0.8	V			
I _{IH}	Input leakage current		- 40°C to +125°C		0.6	2	μA			
IIL	Input leakage current		- 40°C to +125°C	- 0.1	- 0.005		μA			
C _{IN}	Logic input capacitance		- 40°C to +125°C		3		pF			
POWER S	SUPPLY					I				
			25°C		36	55	μA			
I _{DD}	V _{DD} supply current	V_{DD} = 13.2 V, V_{SS} = 0 V Logic inputs = 0 V, 5 V, or V_{DD}	- 40°C to +85°C			65	μA			
			- 40°C to +125°C			75	μA			

5.13 12 V Single Supply: Switching Characteristics

 V_{DD} = +12 V ± 10%, V_{SS} = 0 V, GND = 0 V (unless otherwise noted)

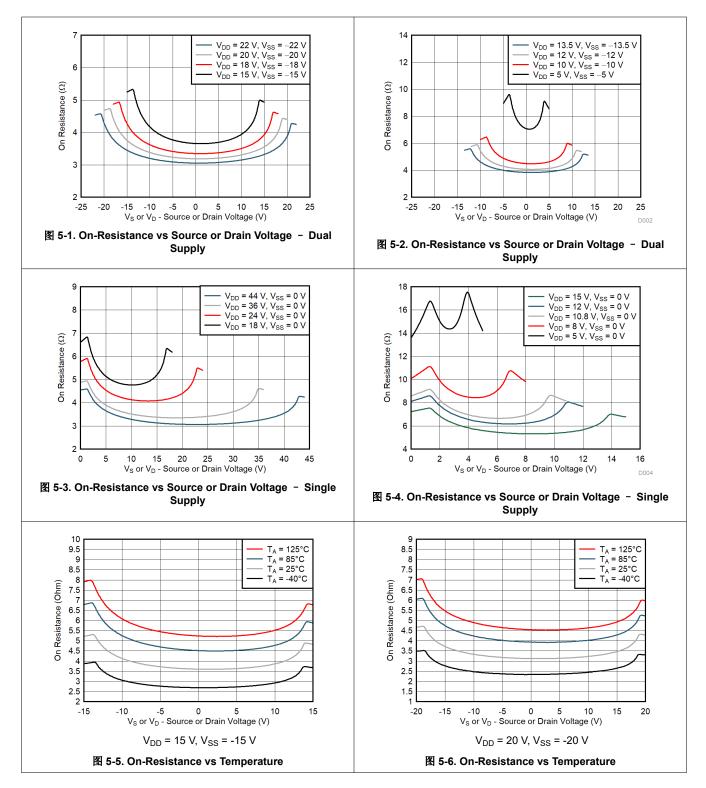
Typical at V_{DD} = +12 V, V_{SS} = 0 V, T_A = 25 °C (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T _A	MIN	TYP	MAX	UNIT	
		V _S = 8 V	25°C		105	200	ns	
t _{TRAN}	Transition time from control input	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			220	ns	
		Refer to Transition Time	- 40°C to +125°C			250	ns	
		V _S = 8 V	25°C		110	200	ns	
t _{ON (EN)}	Turn-on time from enable	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			220	ns	
		Refer to ^{††} 6.5	- 40°C to +125°C			250	ns	
		V _S = 8 V	25°C		105	190	ns	
t _{OFF (EN)}	Turn-off time from enable	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C			210	ns	
		Refer to ^{††} 6.5	- 40°C to +125°C			240	ns	
		V _S = 8 V,	25°C		60		ns	
t _{BBM}	Break-before-make time delay	R _L = 300 Ω, C _L = 35 pF	- 40°C to +85°C	1			ns	
		Refer to Break-Before-Make	- 40°C to +125°C	1			ns	
		V _{DD} rise time = 1µs	25°C		0.16		ms	
T _{ON (VDD)}	Device turn on time (V _{DD} to output)	$R_L = 300 \ \Omega$, $C_L = 35 \text{ pF}$	- 40°C to +85°C		0.16		ms	
		Refer to Turn-on (VDD) Time	- 40°C to +125°C		0.16			
t _{PD}	Propagation delay	$R_L = 50 \Omega$, $C_L = 5 pF$ Refer to $\frac{1}{7}$ 6.8	25°C		490		ps	
Q _{INJ}	Charge injection	V _D = 6 V, C _L = 100 pF Refer to 节 6.9	25°C		1		рС	
O _{ISO}	Off-isolation	R_L = 50 Ω , C_L = 5 pF V _S = 6 V, f = 100 kHz	25°C		- 82		dB	
O _{ISO}	Off-isolation	$R_L = 50 \ \Omega$, $C_L = 5 pF$ $V_S = 6 V$, f = 1 MHz Refer to Off Isolation	25°C		- 62		dB	
X _{TALK}	Crosstalk	$R_L = 50 \ \Omega$, $C_L = 5 pF$ $V_S = 6 V$, f = 1MHz Refer to Crosstalk	25°C		- 105		dB	
BW	- 3dB Bandwidth	$R_L = 50 \ \Omega$, $C_L = 5 \ pF$ $V_S = 6 \ V$ Refer to Bandwidth	25°C		130		MHz	
IL	Insertion loss	R_L = 50 Ω , C_L = 5 pF V _S = 6 V, f = 1 MHz	25°C		- 0.5		dB	
ACPSRR	AC Power Supply Rejection Ratio	V_{PP} = 0.62 V on V_{DD} and V_{SS} R _L = 10 M Ω , C _L = 5 pF, f = 1 MHz Refer to \ddagger 6.14	25°C		- 50		dB	
THD+N	Total Harmonic Distortion + Noise	$V_{PP} = 6 V, V_{BIAS} = 6 V$ $R_L = 10 k \Omega, C_L = 5 pF,$ f = 20 Hz to 20 kHz Refer to $\# 6.13$	25°C		0.0016		%	
C _{S(OFF)}	Source off capacitance	V _S = 6 V, f = 1 MHz	25°C		19		pF	
C _{D(OFF)}	Drain off capacitance	V _S = 6 V, f = 1 MHz	25°C		33		pF	
C _{S(ON),} C _{D(ON)}	On capacitance	V _S = 6 V, f = 1 MHz	25°C		78		pF	



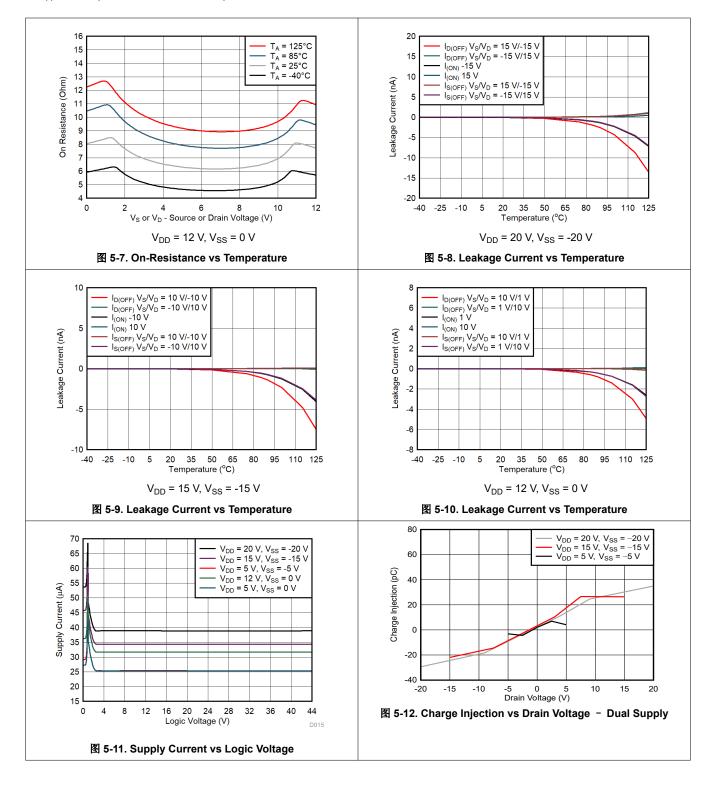
5.14 Typical Characteristics

at T_A = 25°C (unless otherwise noted)



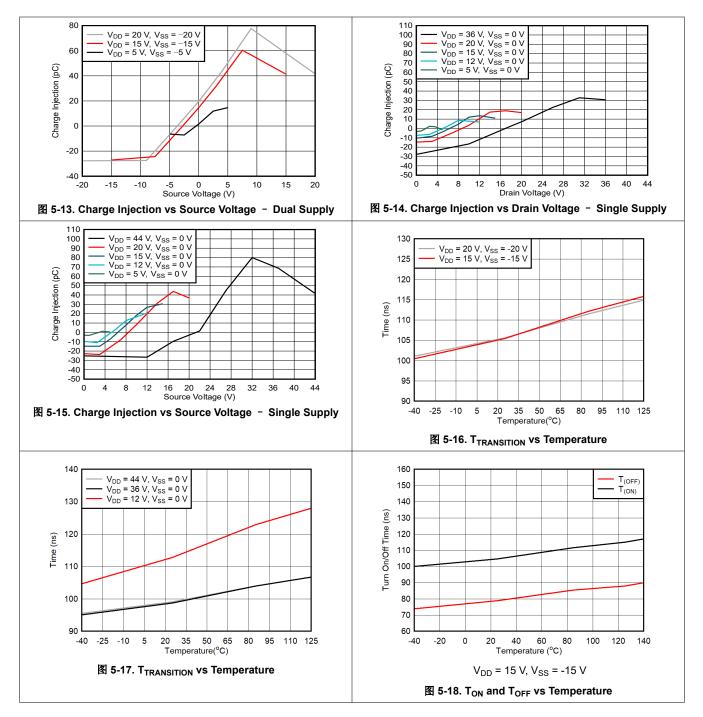


at $T_A = 25^{\circ}C$ (unless otherwise noted)



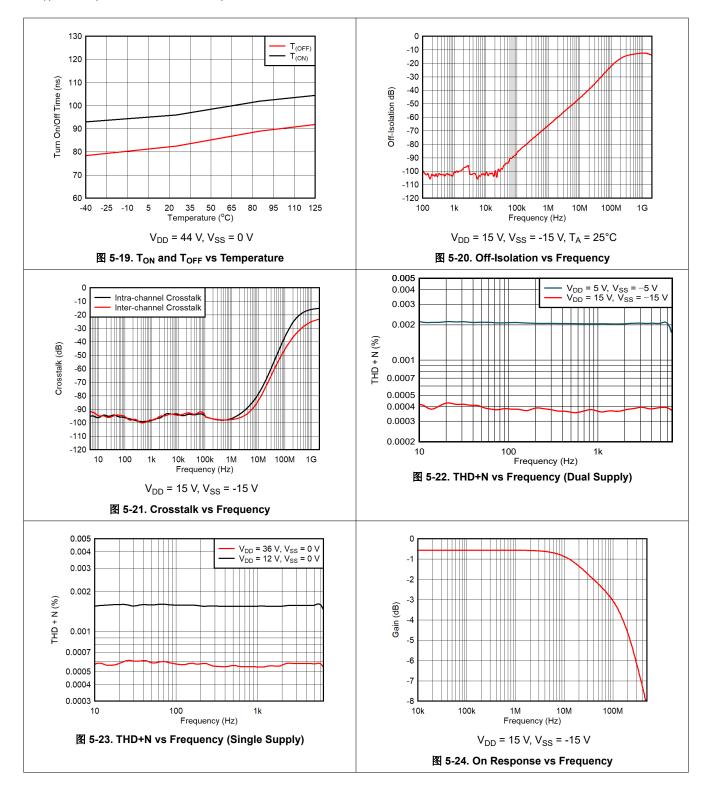


at T_A = 25°C (unless otherwise noted)



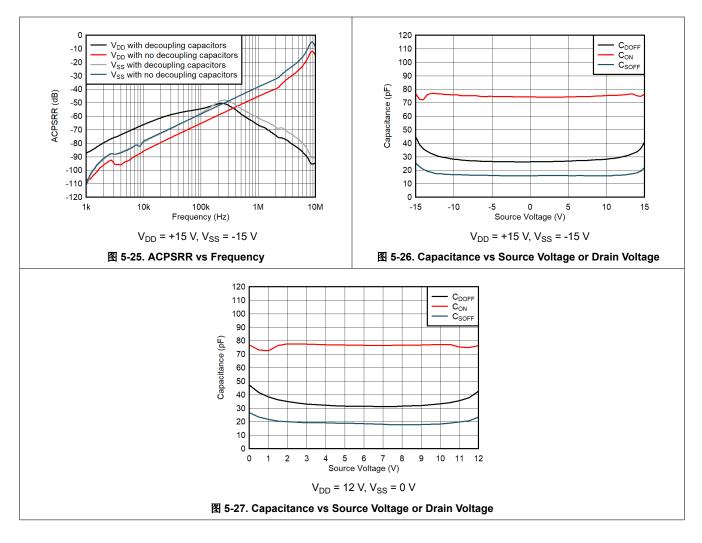


at $T_A = 25^{\circ}C$ (unless otherwise noted)





at T_A = 25°C (unless otherwise noted)





6 Parameter Measurement Information

6.1 On-Resistance

The on-resistance of a device is the ohmic resistance between the source (Sx) and drain (D) pins of the device. The on-resistance varies with input voltage and supply voltage. The symbol R_{ON} is used to denote on-resistance. 6-1 shows the measurement setup used to measure R_{ON} . Voltage (V) and current (I_{SD}) are measured using this setup, and R_{ON} is computed with $R_{ON} = V / I_{SD}$.

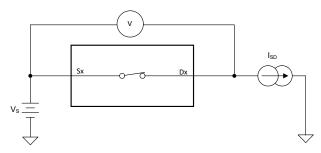


图 6-1. On-Resistance Measurement Setup

6.2 Off-Leakage Current

There are two types of leakage currents associated with a switch during the off state:

- · Source off-leakage current
- Drain off-leakage current

Source leakage current is defined as the leakage current flowing into or out of the source pin when the switch is off. This current is denoted by the symbol $I_{S(OFF)}$.

Drain leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is off. This current is denoted by the symbol $I_{D(OFF)}$.

图 6-2 shows the setup used to measure both off-leakage currents.

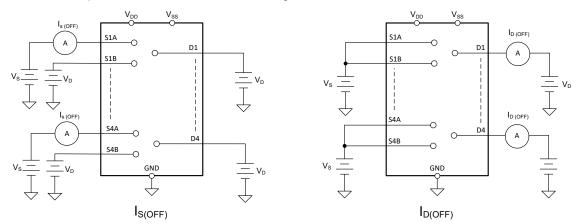


图 6-2. Off-Leakage Measurement Setup



6.3 On-Leakage Current

Source on-leakage current is defined as the leakage current flowing into or out of the source pin when the switch is on. This current is denoted by the symbol $I_{S(ON)}$.

Drain on-leakage current is defined as the leakage current flowing into or out of the drain pin when the switch is on. This current is denoted by the symbol $I_{D(ON)}$.

Either the source pin or drain pin is left floating during the measurement. [4] 6-3 shows the circuit used for measuring the on-leakage current, denoted by $I_{S(ON)}$ or $I_{D(ON)}$.

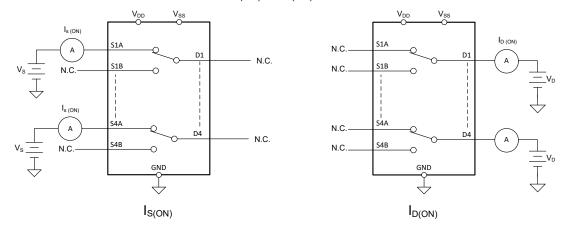


图 6-3. On-Leakage Measurement Setup

6.4 Transition Time

Transition time is defined as the time taken by the output of the device to rise or fall 90% after the address signal has risen or fallen past the logic threshold. The 90% transition measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. \mathbb{K} 6-4 shows the setup used to measure transition time, denoted by the symbol t_{TRANSITION}.

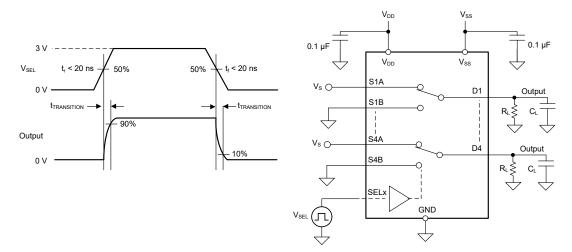


图 6-4. Transition-Time Measurement Setup



6.5 t_{ON(EN)} and t_{OFF(EN)}

Turn-on time is defined as the time taken by the output of the device to rise to 90% after the enable has risen past the logic threshold. The 90% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. \mathbb{E} 6-5 shows the setup used to measure turn-on time, denoted by the symbol $t_{ON(FN)}$.

Turn-off time is defined as the time taken by the output of the device to fall to 10% after the enable has fallen past the logic threshold. The 10% measurement is utilized to provide the timing of the device. System level timing can then account for the time constant added from the load resistance and load capacitance. $\boxed{8}$ 6-5 shows the setup used to measure turn-off time, denoted by the symbol t_{OFF(EN)}.

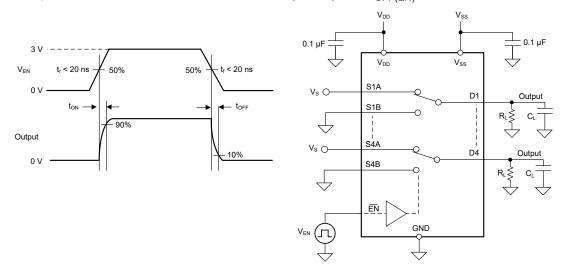
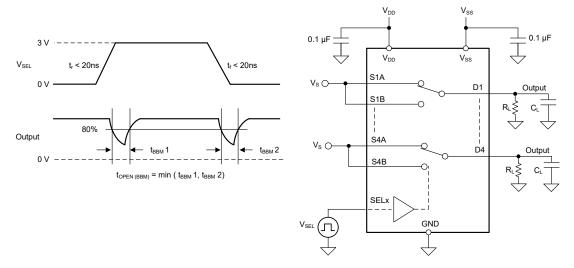


图 6-5. Turn-On and Turn-Off Time Measurement Setup

6.6 Break-Before-Make

Break-before-make delay is a safety feature that prevents two inputs from connecting when the device is switching. The output first breaks from the on-state switch before making the connection with the next on-state switch. The time delay between the *break* and the *make* is known as break-before-make delay. \boxtimes 6-6 shows the setup used to measure break-before-make delay, denoted by the symbol t_{OPEN(BBM)}.







6.7 t_{ON (VDD)} Time

The $t_{ON (VDD)}$ time is defined as the time taken by the output of the device to rise to 90% after the supply has risen past the supply threshold. The 90% measurement is used to provide the timing of the device turning on in the system. 🔀 6-7 shows the setup used to measure turn on time, denoted by the symbol $t_{ON (VDD)}$.

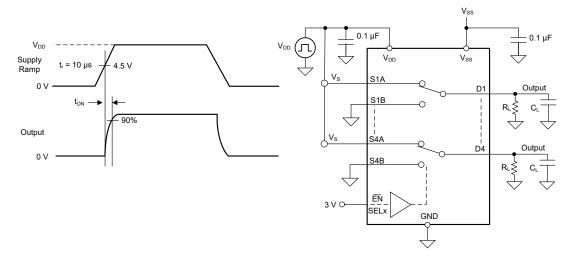


图 6-7. t_{ON (VDD)} Time Measurement Setup

6.8 Propagation Delay

Propagation delay is defined as the time taken by the output of the device to rise or fall 50% after the input signal has risen or fallen past the 50% threshold. \boxtimes 6-8 shows the setup used to measure propagation delay, denoted by the symbol t_{PD}.

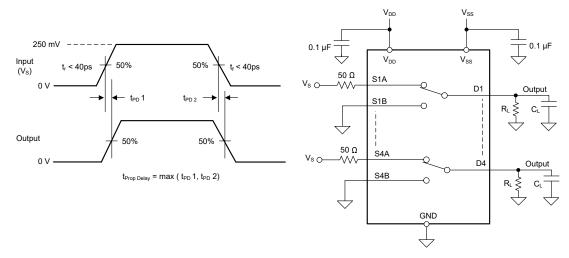


图 6-8. Propagation Delay Measurement Setup



6.9 Charge Injection

The TMUX7234 has a transmission-gate topology. Any mismatch in capacitance between the NMOS and PMOS transistors results in a charge injected into the drain or source during the falling or rising edge of the gate signal. The amount of charge injected into the source or drain of the device is known as charge injection, and is denoted by the symbol Q_{INJ} . \bigotimes 6-9 shows the setup used to measure charge injection from source (Sx) to drain (D).

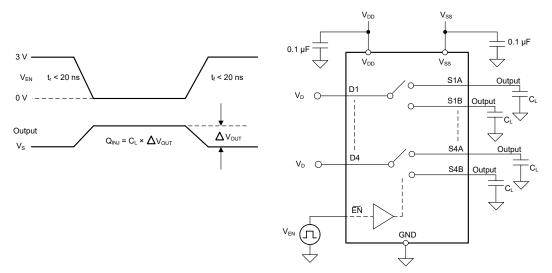


图 6-9. Charge-Injection Measurement Setup

6.10 Off Isolation

Off isolation is defined as the ratio of the signal at the drain pin (D) of the device when a signal is applied to the source pin (Sx) of an off-channel. \mathbb{E} 6-10 shows the setup used to measure, and the equation used to calculate off isolation.

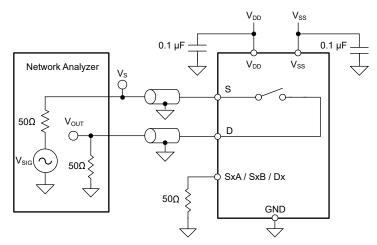


图 6-10. Off Isolation Measurement Setup



6.11 Crosstalk

Crosstalk is defined as the ratio of the signal at the drain pin (D) of a different channel, when a signal is applied at the source pin (Sx) of an on-channel. \boxtimes 6-11 shows the setup used to measure and the equation used to calculate crosstalk.

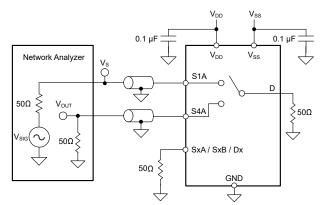


图 6-11. Crosstalk Measurement Setup

6.12 Bandwidth

Bandwidth is defined as the range of frequencies that are attenuated by less than 3 dB when the input is applied to the source pin (Sx) of an on-channel, and the output is measured at the drain pin (D) of the device. \boxtimes 6-12 shows the setup used to measure bandwidth.

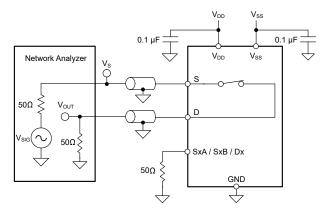


图 6-12. Bandwidth Measurement Setup



6.13 THD + Noise

The total harmonic distortion (THD) of a signal is a measurement of the harmonic distortion, and is defined as the ratio of the sum of the powers of all harmonic components to the power of the fundamental frequency at the mux output. The on-resistance of the device varies with the amplitude of the input signal and results in distortion when the drain pin is connected to a low-impedance load. Total harmonic distortion plus noise is denoted as THD.

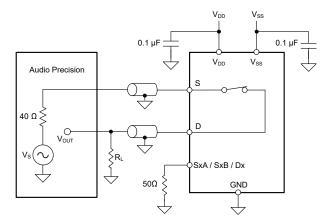


图 6-13. THD Measurement Setup

6.14 Power Supply Rejection Ratio (PSRR)

PSRR measures the ability of a device to prevent noise and spurious signals that appear on the supply voltage pin from coupling to the output of the switch. The DC voltage on the device supply is modulated by a sine wave of 620mVPP. The ratio of the amplitude of signal on the output to the amplitude of the modulated signal is the ACPSRR. A high ratio represents a high degree of tolerance to supply rail variation.

The below shows how the decoupling capacitors reduce high frequency noise on the supply pins. This helps stabilize the supply and immediately filter as much of the supply noise as possible.

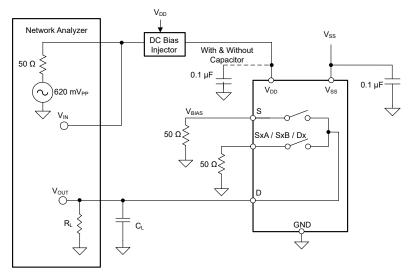


图 6-14. ACPSRR Measurement Setup

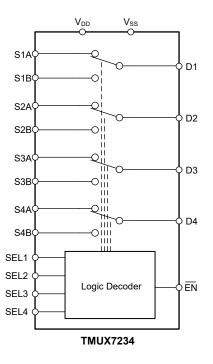


7 Detailed Description

7.1 Overview

The TMUX7234 contains four independently controlled SPDT switches with an \overline{EN} pin to enable or disable all four switches.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Bidirectional Operation

The TMUX7234 conducts equally well from source (Sx) to drain (Dx) or from drain (Dx) to source (Sx). Each channel has very similar characteristics in both directions and supports both analog and digital signals.

7.3.2 Rail-to-Rail Operation

The valid signal path input or output voltage for the TMUX7234 ranges from V_{SS} to V_{DD} .

7.3.3 1.8 V Logic Compatible Inputs

The TMUX7234 has 1.8 V logic compatible control for all logic control inputs. 1.8 V logic level inputs allows the switch to interface with processors that have lower logic I/O rails and eliminates the need for an external translator, which saves both space and BOM cost. For more information on 1.8 V logic implementations refer to *Simplifying Design with 1.8 V logic Muxes and Switches*.



7.3.4 Fail-Safe Logic

TMUX7234 supports Fail-Safe Logic on the control input pins (EN and SELx) allowing it to operate up to 44 V, regardless of the state of the supply pins. This feature allows voltages on the control pins to be applied before the supply pin, protecting the device from potential damage. Fail-Safe Logic minimizes system complexity by removing the need for power supply sequencing on the logic control pins. For example, the Fail-Safe Logic feature allows the TMUX7234 logic input pins to ramp up to +44 V while V_{DD} and V_{SS} = 0 V. The logic control inputs are protected against positive faults of up to +44 V in powered-off condition, but do not offer protection against negative overvoltage conditions.

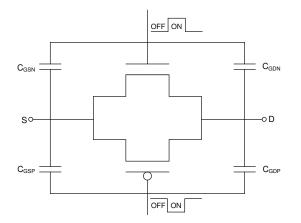
7.3.5 Latch-Up Immune

Latch-Up is a condition where a low impedance path is created between a supply pin and ground. This condition is caused by a trigger (current injection or overvoltage), but once activated, the low impedance path remains even after the trigger is no longer present. This low impedance path may cause system upset or catastrophic damage due to excessive current levels. The Latch-Up condition typically requires a power cycle to eliminate the low impedance path.

The TMUX72xx family of devices are constructed on Silicon on Insulator (SOI) based process where an oxide layer is added between the PMOS and NMOS transistor of each CMOS switch to prevent parasitic structures from forming. The oxide layer is also known as an insulating trench and prevents triggering of latch up events due to overvoltage or current injections. The latch-up immunity feature allows the TMUX72xx family of switches and multiplexers to be used in harsh environments. For more information on latch-up immunity refer to *Using Latch Up Immune Multiplexers to Help Improve System Reliability*.

7.3.6 Ultra-Low Charge Injection

The TMUX7234 has a transmission gate topology, as shown in 🖄 7-1. Any mismatch in the stray capacitance associated with the NMOS and PMOS causes an output level change whenever the switch is opened or closed.





The TMUX7234 contains specialized architecture to reduce charge injection on the source (Sx). To further reduce charge injection in a sensitive application, a compensation capacitor (Cp) can be added on the drain (D). This will ensure that excess charge from the switch transition will be pushed into the compensation capacitor on the drain (D) instead of the source (Sx). As a general rule of thumb, Cp should be 20x larger than the equivalent load capacitance on the source (Sx). \boxtimes 7-2 shows charge injection variation with different compensation capacitors on the drain side. This plot was captured on the TMUX7219 as part of the TMUX72xx family with a 100pF load capacitance.



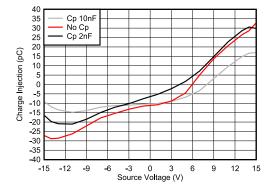


图 7-2. Charge Injection Compensation

7.4 Device Functional Modes

The enable \overline{EN} pin is an active-low logic pin that controls the connection between the source (SxA and SxB) and drain (Dx) pins of the device. The TMUX7234 SELx logic control inputs determine which source pin is connected to the drain pin for each channel. When the \overline{EN} pin of the TMUX7234 is pulled low, the SELx logic control inputs determine which source input is selected. When the \overline{EN} pin is pulled high, all of the switches are in an open state regardless of the state of the SELx logic control inputs. The control pins can be as high as 44 V.

The TMUX7234 can be operated without any external components except for the supply decoupling capacitors. The \overline{EN} and SELx pins have internal pull-down resistors of 4 M Ω . If unused, \overline{EN} and SELx pins should be tied to GND in order to ensure the device does not consume additional current as highlighted in Implications of Slow or Floating CMOS Inputs. Unused signal path inputs (Sx or Dx) should be connected to GND.

7.5 Truth Tables

		表 7-1.	TMUX7234 Trut	h Table	
EN	SEL1	SEL2	SEL3	SEL4	Selected Source Pins Connected to Drain Pins
0	0	X ⁽¹⁾	Х	Х	S1B to D1
0	1	Х	Х	X	S1A to D1
0	Х	0	Х	X	S2B to D2
0	Х	1	Х	X	S2A to D2
0	Х	Х	0	Х	S3B to D3
0	Х	Х	1	Х	S3A to D3
0	Х	Х	Х	0	S4B to D4
0	Х	Х	Х	1	S4A to D4
1	Х	Х	Х	Х	Hi-Z (OFF)

表 7-1 shows the truth tables for the TMUX7234.

(1) X means do not care.



8 Application and Implementation

备注

以下应用部分中的信息不属于 TI 器件规格的范围, TI 不担保其准确性和完整性。TI 的客 户应负责确定器件是否适用于其应用。客户应验证并测试其设计,以确保系统功能。

8.1 Application Information

The TMUX7234 is part of the precision switches and multiplexers family of devices. This device operates with dual supplies (\pm 4.5 V to \pm 22 V), a single supply (4.5 V and 44 V), or asymmetric supplies (such as, V_{DD} = 12 V and V_{SS} = -5 V), and offers rail-to-rail input and output. The TMUX7234 offers low R_{ON}, low on and off leakage currents and ultra-low charge injection performance. These features makes the TMUX7234 a precision, robust, high-performance analog multiplexer for high-voltage, industrial applications.

8.2 Typical Application

One key application of the TMUX7234 is in the ultrasonic water flow measurement system. Ultrasonic flow meters use time of flight (ToF) of an ultrasonic wave and its dependency and behavior in the medium using two transducer pairs for upstream and downstream paths. The signal waveforms are transmitted between two adjacent transducers. One transducer transmits an upstream path signal and the other transducer receives a downstream signal path. The flight time for the signal can be calculated using the known velocity of sound and length between the transducers. The upstream and downstream waveforms are processed on the main MCU to obtain the volume. A shows a circuit example utilizing the MSP430FR66047 MCU, high voltage low distortion operational amplifiers (THS3091), along with TMUX7234, 2:1, 4 channel precision switches. The TMUX7234 is used to select the Rx and Tx path of the transducer. The TMUX7234 offers low on-state resistance, flat capacitance performance, and low propagation delay which leads to very low signal distortion. The break-before-make feature allows transferring of a signal from one port to another, without shorting the inputs together. This device also offers low charge injection which makes this device suitable for high precision data acquisition systems.

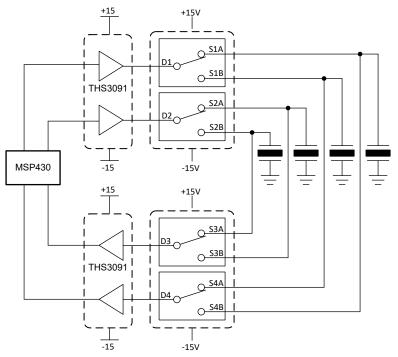


图 8-1. Ultrasonic Water Flow Measurement System

8.2.1 Design Requirements

For this design example, use the parameters listed in $\frac{1}{8}$ 8-1.

PARAMETERS	VALUES								
Supply (V _{DD})	15 V								
Supply (V _{SS})	-15 V								
MUX I/O signal range	-15 V to 15 V (Rail-to-Rail)								
Control logic thresholds	1.8 V compatiable (up to V _{DD})								
EN	EN pulled low to enable the switch								

表 8-1. Design Parameters

8.2.2 Detailed Design Procedure

The TMUX7234 can operate without any external components except for the supply decoupling capacitors. All inputs passing through the switch must fall within the recommended operating conditions of the TMUX7234, including signal range and continuous current. $\ddagger 5.4$ shows how the signal range for this design can be -15 V to +15 V and the maximum continuous current can be up to 400 mA for wide-range current measurement with a positive supply of 15 V on V_{DD} and negative supply of -15 V on V_{SS}. The TMUX7234 device are bidirectional, single-pole double-throw (SPDT) switches that offer low on-resistance, low leakage, and low power. These features make these devices suitable for portable and power sensitive applications such as ultrasonic water metering systems. For a more detailed analysis of the ultrasonic water flow measurement system refer to the reference design.

8.2.3 Application Curve

The low on and off leakage currents of TMUX7234 and ultra-low charge injection performance make this device ideal for implementing high precision industrial systems. The TMUX7234 contains specialized architecture to reduce charge injection on the Source side (Sx) (For more details, see \ddagger 7.3.6). 🛛 8-2 shows the plot for the charge injection versus drain voltage for the TMUX7234.

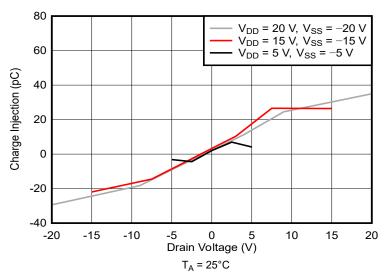


图 8-2. Charge Injection vs Drain Voltage

9 Power Supply Recommendations

TMUX7234 operates across a wide supply range of ± 4.5 V to ± 22 V (4.5 V to 44 V in single-supply mode). TMUX7234 also perform well with asymmetrical supplies such as V_{DD} = 12 V and V_{SS} = -5 V.



Power-supply bypassing improves noise margin and prevents switching noise propagation from the supply rails to other components. Good power-supply decoupling is important to achieve optimum performance. Use a supply decoupling capacitor ranging from 0.1 μ F to 10 μ F at both the V_{DD} and V_{SS} pins to ground for an improved supply noise immunity. Place the bypass capacitors as close to the power supply pins of the device as possible using low-impedance connections. TI recommends using multi-layer ceramic chip capacitors (MLCCs) that offer low equivalent series resistance (ESR) and inductance (ESL) characteristics for power-supply decoupling purposes. For very sensitive systems, or for systems in harsh noise environments, avoiding the use of vias for connecting the capacitors to the device pins may offer superior noise immunity. The use of multiple vias in parallel lowers the overall inductance and is beneficial for connections to ground and power planes. Always ensure the ground (GND) connection is established before supplies are ramped.

10 Layout

10.1 Layout Guidelines

A reflection can occur when a PCB trace turns a corner at a 90° angle. A reflection occurs primarily because of the change of width of the trace. The trace width increases to 1.414 times the width at the apex of the turn. This increase upsets the transmission-line characteristics, especially the distributed capacitance and self – inductance of the trace which results in the reflection. Not all PCB traces can be straight and therefore some traces must turn corners. [4] 10-1 shows progressively better techniques of rounding corners. Only the last example (BEST) maintains constant trace width and minimizes reflections.

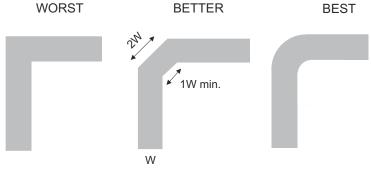


图 10-1. Trace Example

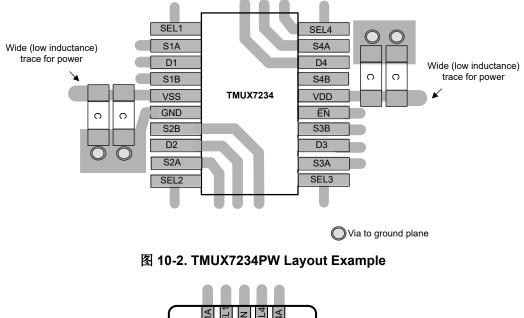
Route high-speed signals using a minimum of vias and corners which reduces signal reflections and impedance changes. When a via must be used, increase the clearance size around it to minimize its capacitance. Each via introduces discontinuities in the signal's transmission line and increases the chance of picking up interference from the other layers of the board. Be careful when designing test points, through-hole pins are not recommended at high frequencies.

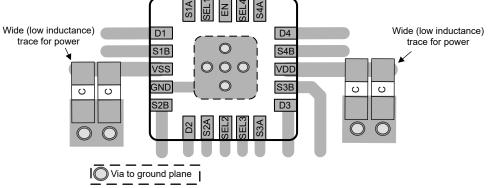
图 10-2 and 图 10-3 illustrate an example of a PCB layout with the TMUX7234. Some key considerations are:

- Decouple the supply pins with a 0.1 µF and 1 µF capacitor, placed lowest value capacitor as close to the pin as possible. Make sure that the capacitor voltage rating is sufficient for the supply voltage.
- Keep the input lines as short as possible.
- Use a solid ground plane to help reduce electromagnetic interference (EMI) noise pickup.
- Do not run sensitive analog traces in parallel with digital traces. Avoid crossing digital and analog traces if possible, and only make perpendicular crossings when necessary.
- Using multiple vias in parallel will lower the overall inductance and is beneficial for connection to ground planes.

10.2 Layout Example

图 10-2 shows an example board layout for the TMUX7234.









11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

- Texas Instruments, Using Latch Up Immune Multiplexers to Help Improve System Reliability application
 report
- · Texas Instruments, Improve Stability Issues with Low CON Multiplexers application brief
- Texas Instruments, Improving Signal Measurement Accuracy in Automated Test Equipment application brief
- Texas Instruments, Sample & Hold Glitch Reduction for Precision Outputs Reference Design reference guide
- Texas Instruments, Simplifying Design with 1.8 V logic Muxes and Switches application brief
- Texas Instruments, System-Level Protection for High-Voltage Analog Multiplexers application report
- Texas Instruments, True Differential, 4 x 2 MUX, Analog Front End, Simultaneous-Sampling ADC Circuit application report
- Texas Instruments, QFN/SON PCB Attachment application report
- Texas Instruments, Quad Flatpack No-Lead Logic Packages application report

11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 支持资源

TI E2E[™] 中文支持论坛是工程师的重要参考资料,可直接从专家处获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题,获得所需的快速设计帮助。

链接的内容由各个贡献者"按原样"提供。这些内容并不构成 TI 技术规范,并且不一定反映 TI 的观点;请参阅 TI 的使用条款。

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11.5 静电放电警告



静电放电 (ESD) 会损坏这个集成电路。德州仪器 (TI) 建议通过适当的预防措施处理所有集成电路。如果不遵守正确的处理和安装程序,可能会损坏集成电路。

ESD 的损坏小至导致微小的性能降级,大至整个器件故障。精密的集成电路可能更容易受到损坏,这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

11.6 术语表

TI术语表 本术语表列出并解释了术语、首字母缩略词和定义。

12 Revision History

注:以前版本的页码可能与当前版本的页码不同

С	hanges from Revision F (November 2022) to Revision G (July 2024)	Page
•	Updated ESD HBM value	4
•	Updated IIH max specification	6



Page

Changes from Revision E (August 2021) to Revision F (November 2022)

• 将 PW 封装状态从 预发布 更改为 正在供货1

С	hanges from Revision D (August 2021) to Revision E (August 2021)	Page
•	Updated ESD HBM spec	4

С	hanges from Revision C (June 2021) to Revision D (August 2021)	Page
•	将状态从: <i>预告信息</i> 更改为 <i>量产数据</i>	1

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
TMUX7234PWR	Active	Production	TSSOP (PW) 20	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	T234
TMUX7234PWR.B	Active	Production	TSSOP (PW) 20	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	T234
TMUX7234RRQR	Active	Production	WQFN (RRQ) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TMUX
									X234
TMUX7234RRQR.B	Active	Production	WQFN (RRQ) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	TMUX
									X234

⁽¹⁾ **Status:** For more details on status, see our product life cycle.

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*/	All dimensions are nominal												
ſ	Device	-	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
Γ	TMUX7234PWR	TSSOP	PW	20	3000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
	TMUX7234RRQR	WQFN	RRQ	20	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2



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PACKAGE MATERIALS INFORMATION

25-Sep-2024



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TMUX7234PWR	TSSOP	PW	20	3000	356.0	356.0	35.0
TMUX7234RRQR	WQFN	RRQ	20	3000	367.0	367.0	35.0

PW0020A



PACKAGE OUTLINE

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice. 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



PW0020A

EXAMPLE BOARD LAYOUT

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PW0020A

EXAMPLE STENCIL DESIGN

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



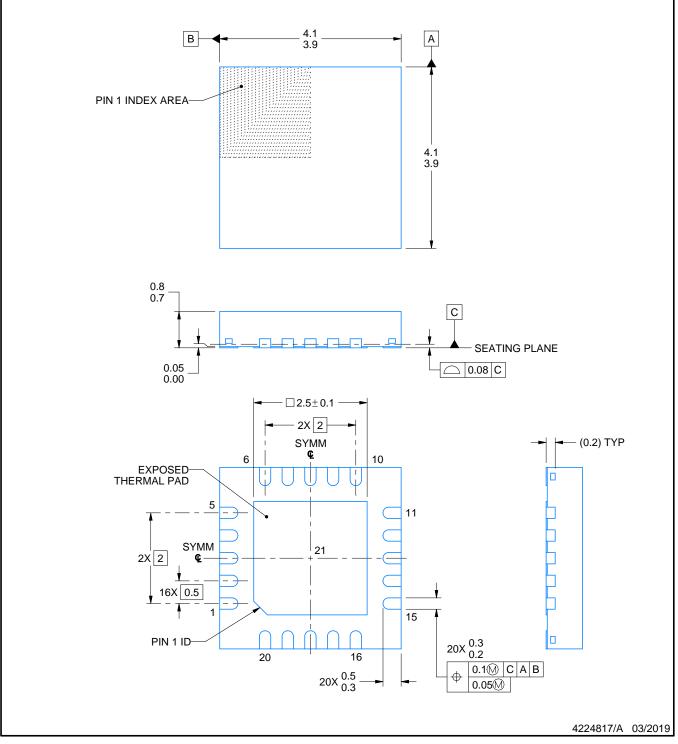
RRQ0020A



PACKAGE OUTLINE

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

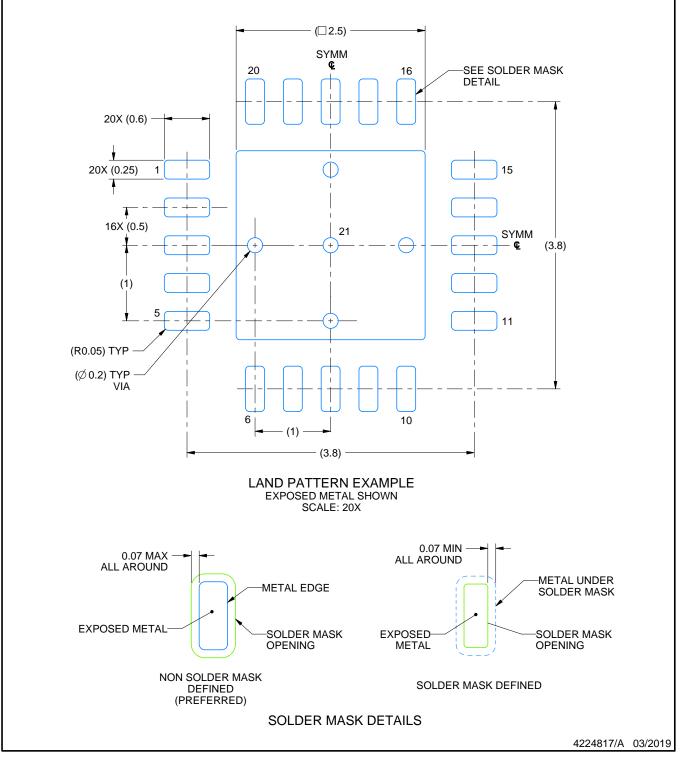


RRQ0020A

EXAMPLE BOARD LAYOUT

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

 This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).

5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

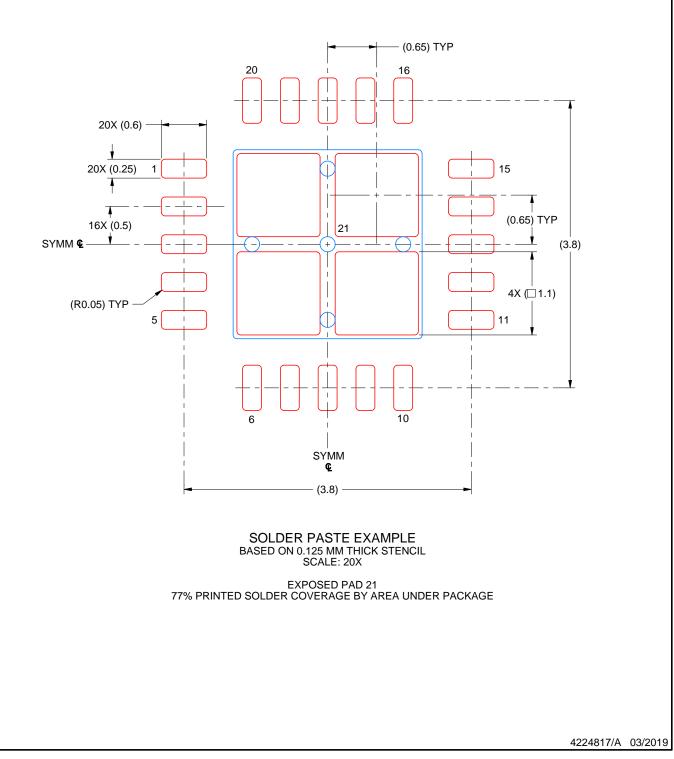


RRQ0020A

EXAMPLE STENCIL DESIGN

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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