

HIGH VOLTAGE SEMINAR WILL HADDEN ELECTRIC VEHICLES

PROTECTING POWER DEVICES IN
ELECTRIC VEHICLE APPLICATIONS

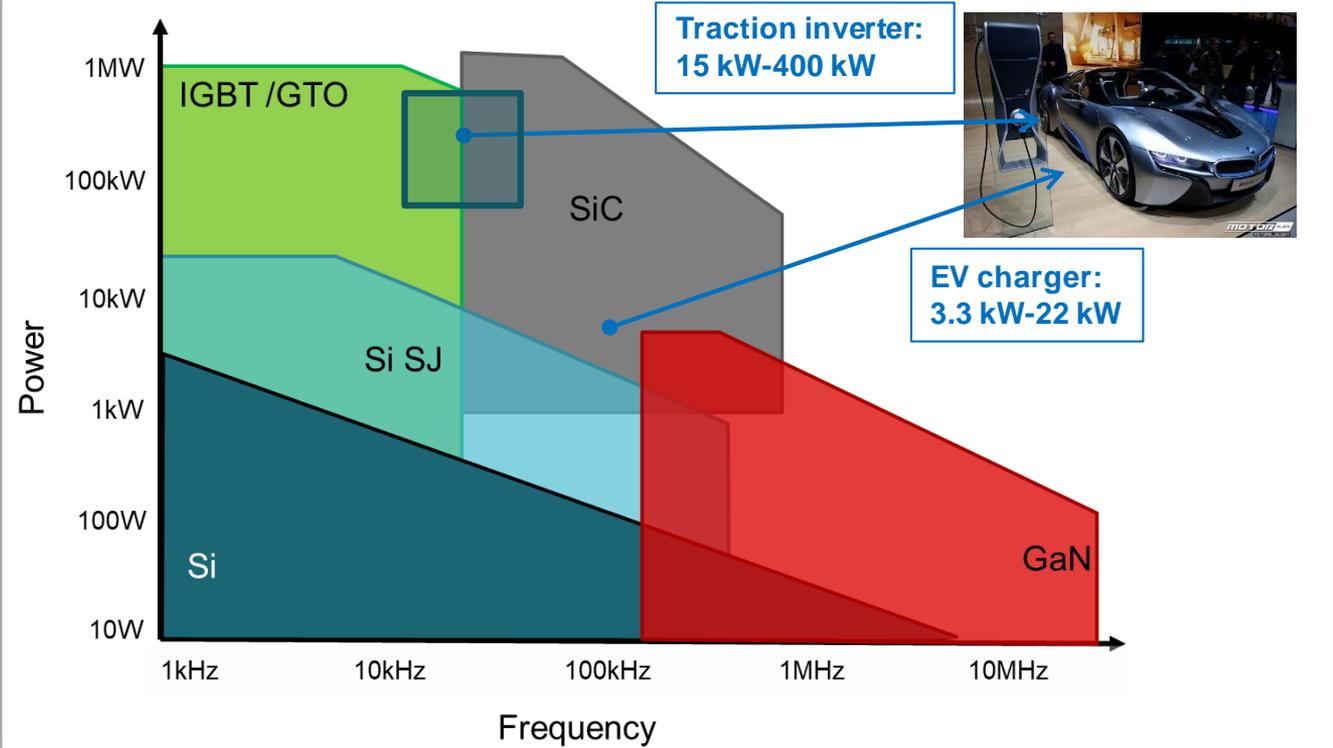


What will I get out of this session?

- Brief Description of Silicon Carbide performance
- Different short circuit current sensing and protection methods
- How to turn off MOSFET under short circuit safely
- Other diagnostics available to assist in ASIL designs

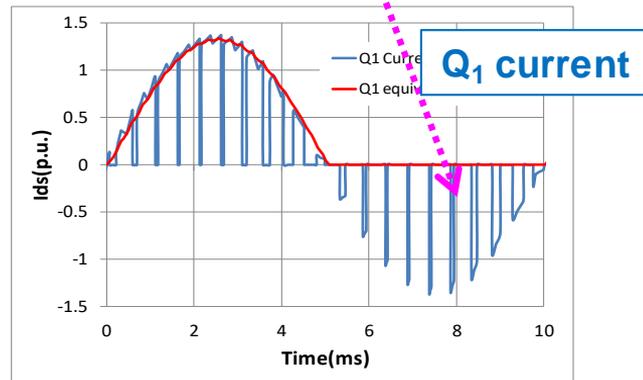
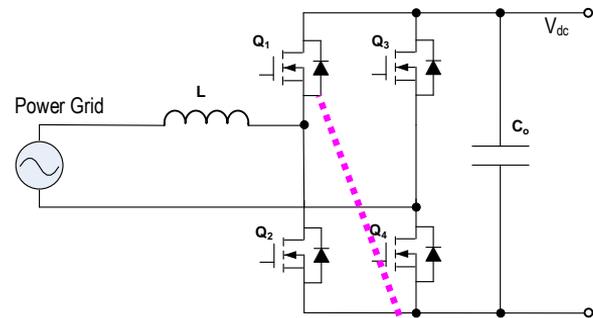
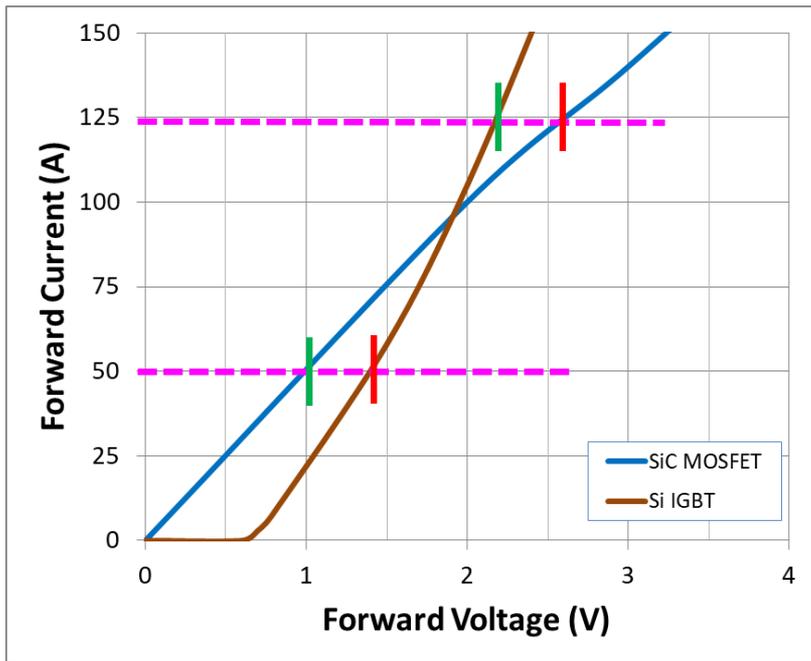
- Relevant part numbers:
 - UCC587x-Q1
 - UCC2152x
- Relevant reference designs:
 - TIDA-01605
 - TIDA-00917
- Relevant applications:
 - HEV/EV traction inverter, EV on-board charger

Power transistor application positioning



SiC MOSFET advantages over Si IGBT: conduction

1) No 0.5-1.0V knee voltage; 2) Has “body diode”; 3) 3rd quadrant operation mode



SiC MOSFET advantages over Si IGBT: switching

- **Low switching loss:**

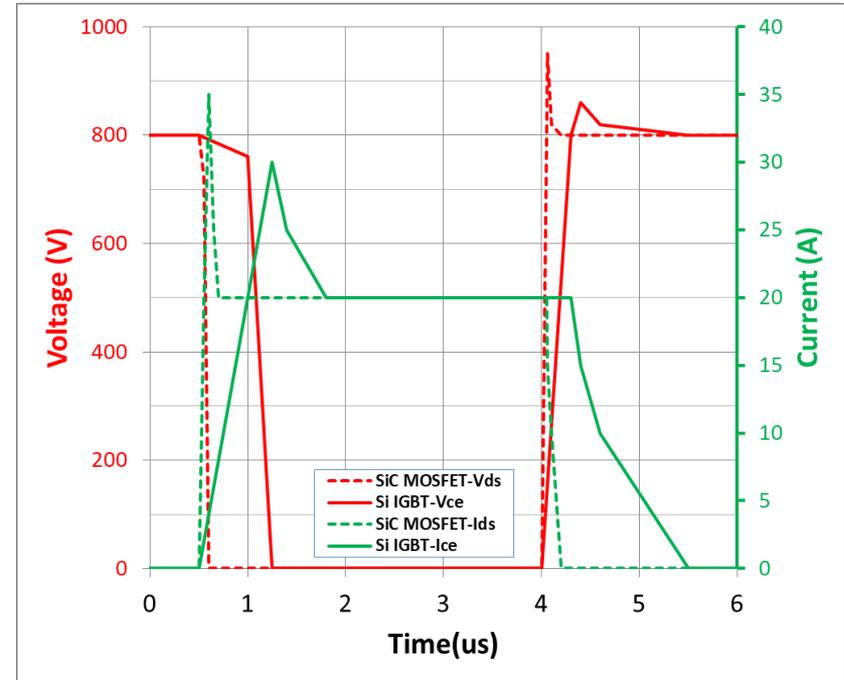
MOSFET (unipolar) vs IGBT (bipolar)-
fundamental difference

- **Switch loss less increase at elevated temperatures:**

For 1000V SiC MOSFET, $E_{sw}@25^{\circ}\text{C} = E_{sw}@150^{\circ}\text{C}$

- **Low reverse recovery for the body diode:**

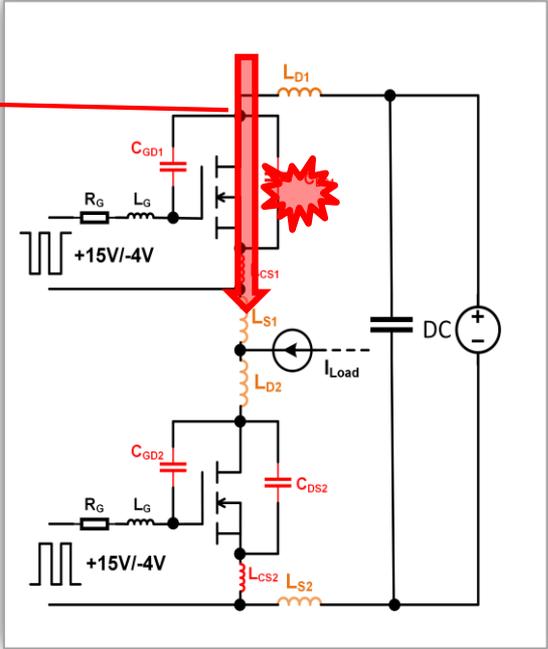
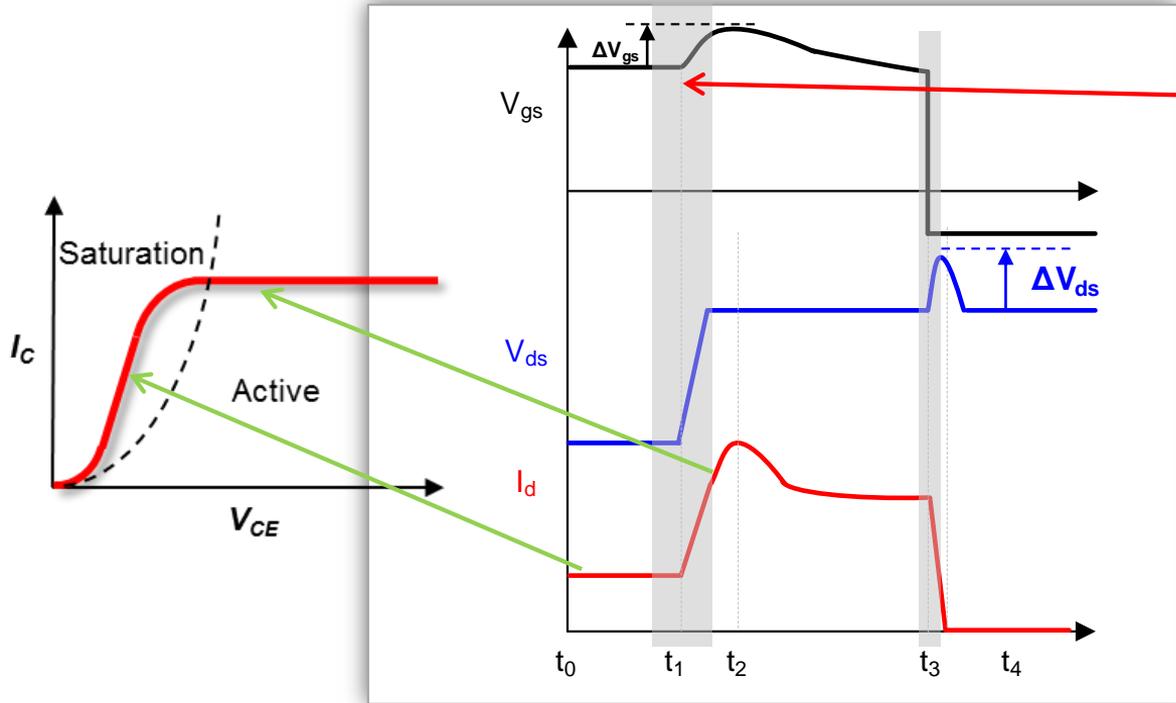
Silicon PiN diode has significant reverse recovery which has reverse recovery loss and also adds more turn on loss



Typical switching waveforms

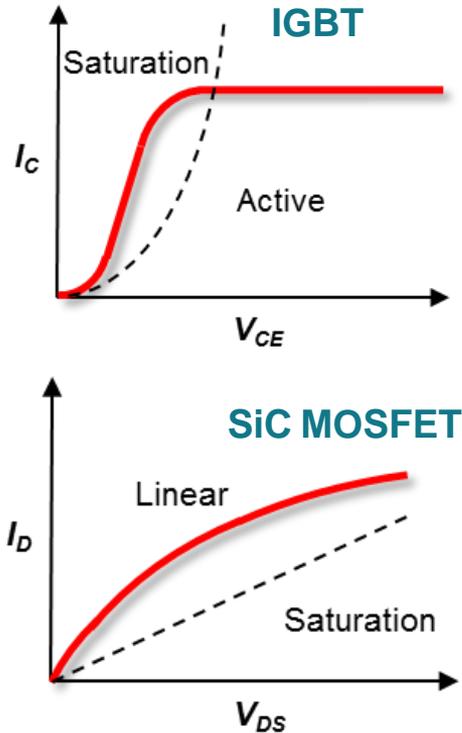
Overcurrent protection

Overcurrent/short circuit fault



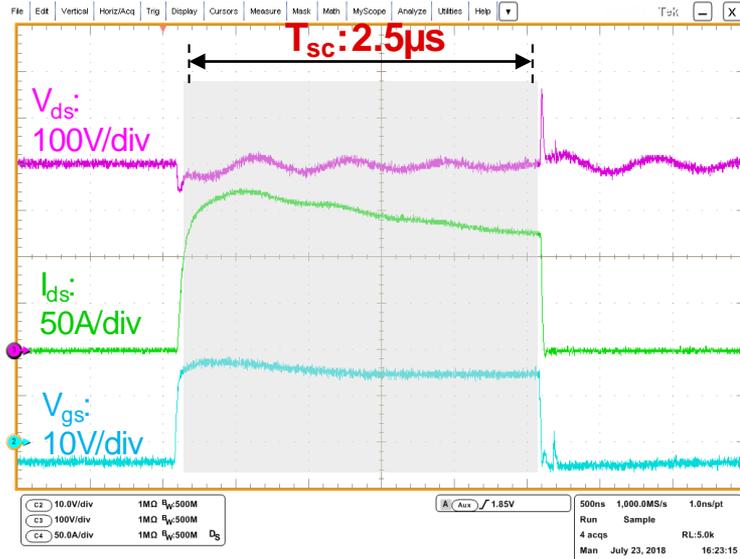
Short circuit happens at t_1

Overcurrent/short circuit fault mechanism: thermal limitation



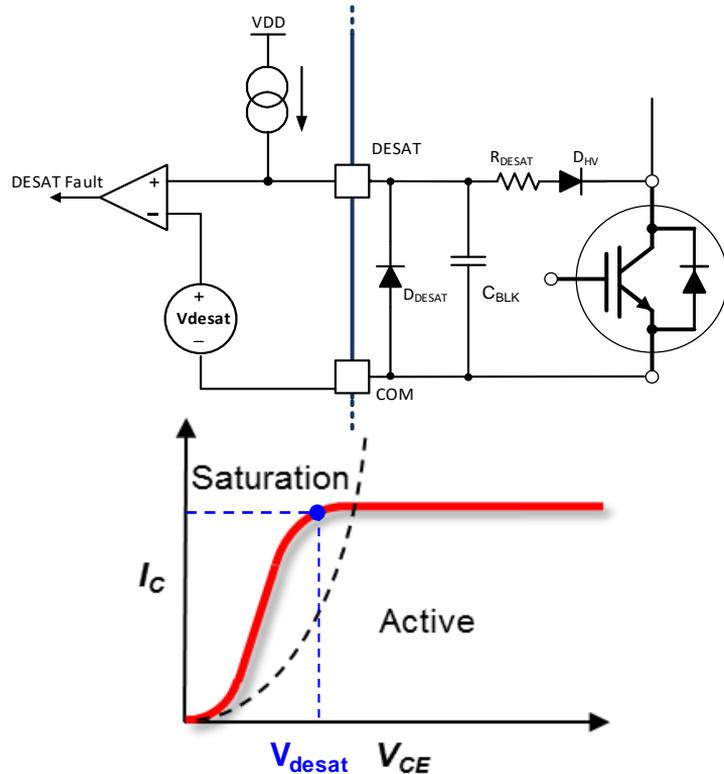
- **IGBT self-limits the current increase with lower saturation current**
 - Shape transient from saturation region to active region, collector current is limited to a constant value in active region
- **SiC MOSFET has large linear region with high saturation current**
 - In the case of SiC, I_d continues to increase with increase in V_{ds} , eventually resulting in faster breakdown
- **For same rated current & voltage, IGBT reaches active region for significantly lower V_{CE} as compared to SiC MOSFET**

Overcurrent/short circuit fault mechanism: thermal limitation



- The short circuit withstand time, t_{sc} , is determined by the critical energy
 - Minimal dissipated energy leading to device failure for one short circuit pulse
$$E_c = \int_{t_1}^{t_3} V_{ds} \cdot I_d \cdot dt$$
 - V_{ds} is the DC link voltage, I_d will be the device saturation current.
- SiC MOSFET short circuit withstand time is shorter than IGBT due to smaller chip size, less thermal capacity

Overcurrent/short circuit protection method: DESAT

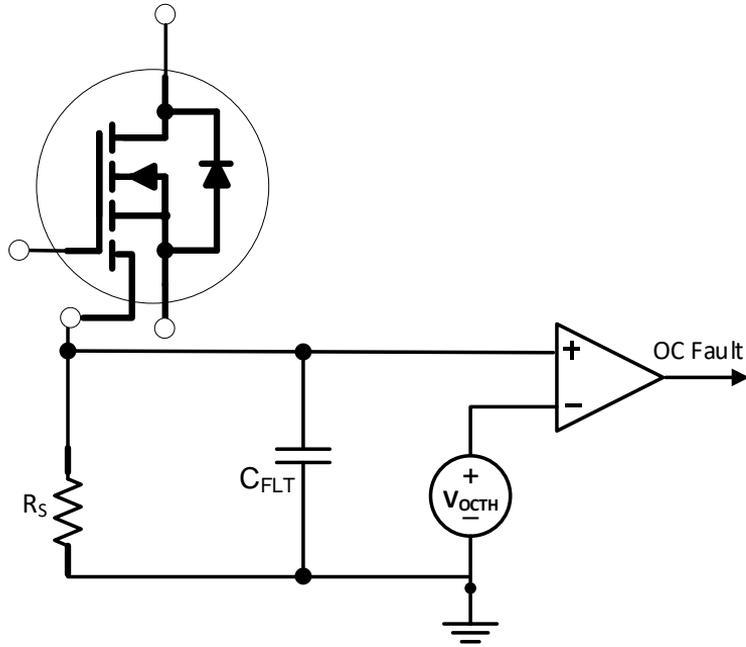


- Desaturation circuit detects the V_{ds} of MOSFET or V_{ce} of IGBT, protection is triggered when detected voltage is above pre-set reference voltage
- Blanking time is needed to prevent false trigger during switching turn on transients

$$t_{DS_BLK} = \frac{V_{DESAT} \times C_{BLK}}{I_{CHG}}$$

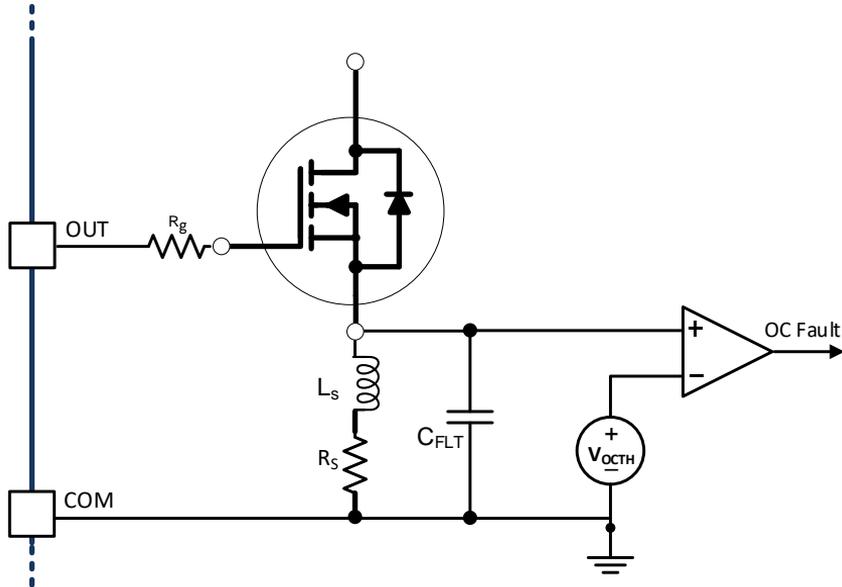
- DESAT threshold voltage varies between different devices due to the output characteristics, especially IGBT and SiC MOSFET
- Most common method for SiC where the IR drop is measured for current sense

Overcurrent/short circuit protection method: SenseFET / current mirror



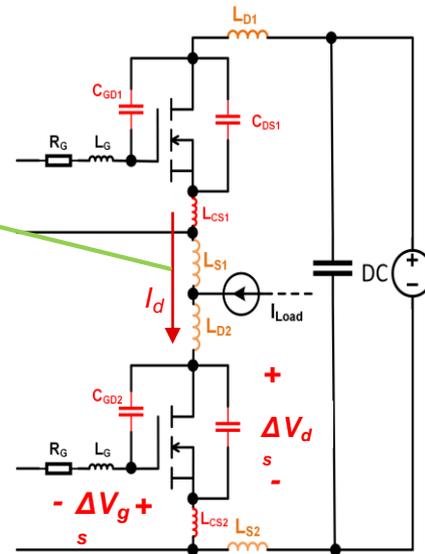
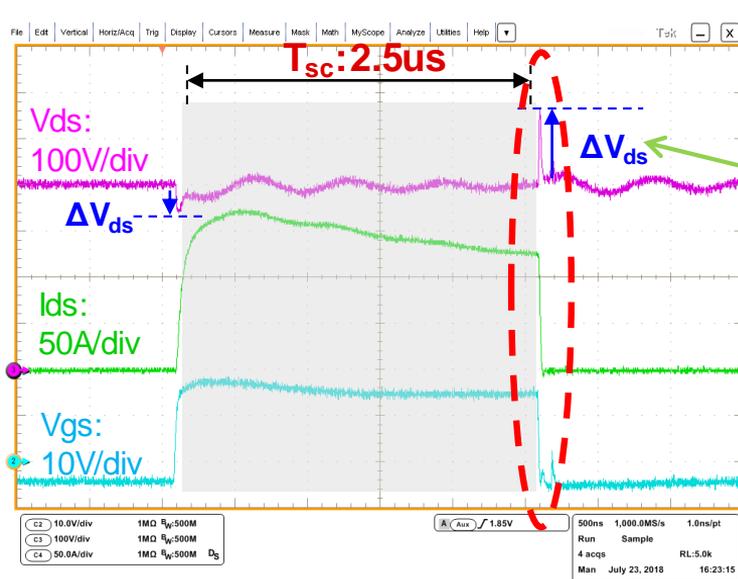
- SenseFET / current mirror is used to scale down main current, tens of mV voltage is measured on sense resistor
- Advantages
 - Fastest protection speed
 - Accurate for both AC and DC
- Challenges of SenseFET / current mirror
 - Module needs to be customized to integrate SenseFET / current mirror
 - Higher cost
 - Not currently available in SiC modules

Overcurrent/short circuit protection method: shunt resistor



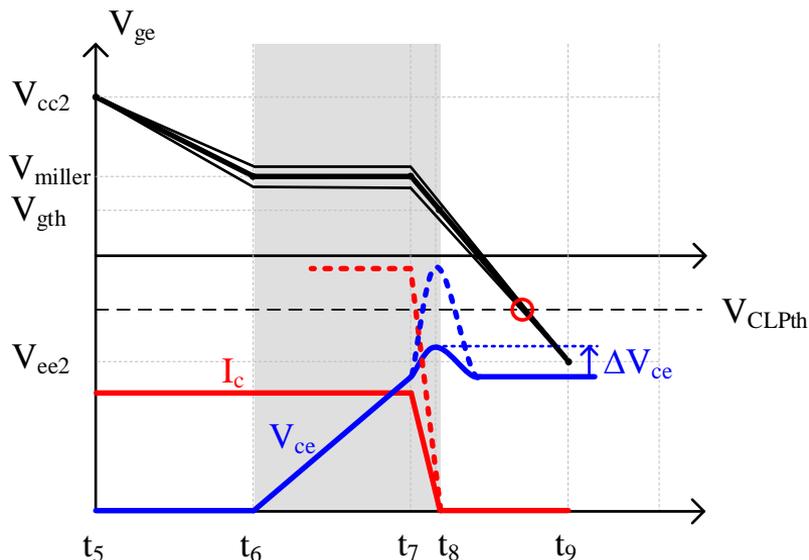
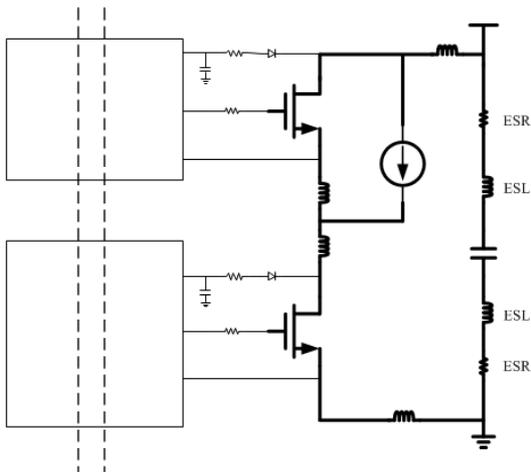
- Advantages
 - Accurate for both AC and DC
 - Fast protection speed
 - Low cost
- Challenges of shunt resistor
 - High power loss in high power applications
 - Weak noise immunity due to gate loop noise caused by parasitic inductance of shunt resistor and PCB trace

Overcurrent/short circuit safe turn-off: avalanche limitation



- **Voltage and avalanche limit:** device avalanche can be caused by the overshoot voltage on V_{ds}

SiC MOSFET protection: soft turn-off (STO) & 2L turn-off



- There are parasitic inductances in the power loop
- Parasitic inductances together with di/dt cause voltage spikes
- **The di/dt rate is much higher under short circuit fault** so a preventive turn-off measure needs to be taken in order to limit the loop inductance induced voltage spike
- Effectively, there are two ways to slow down the turn-off process: reduce di or extend dt

DESAT Timing

Gate Driver Parameters

- t_{LEB}
- V_{DESAT}
- I_{CHG}
- t_{DSOFF}
- I_{STO}

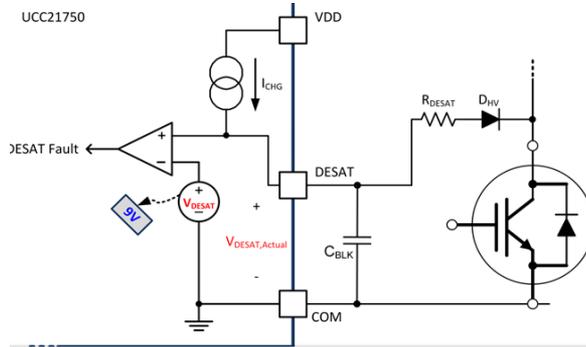
$$t_{BLK} = \frac{C_{BLK} * V_{DESAT}}{I_{CHG}}$$

- Determined by the blanking capacitor C_{BLK}

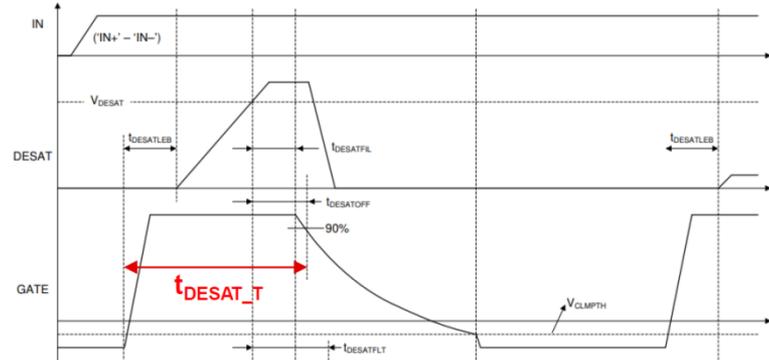
$$t_{DESAT_T} = t_{LEB} + t_{BLK} + t_{DSOFF}$$

$$t_{STO} = C_{ISS} * (V_{DD} - V_{th}) / I_{STO}$$

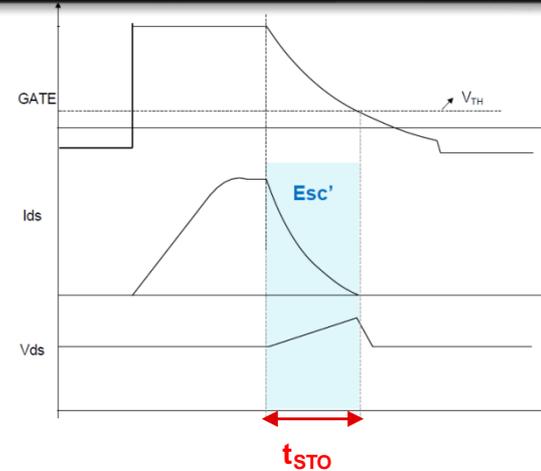
$$t_{Total} = t_{DESAT_T} + t_{STO}$$



(1) Detection time



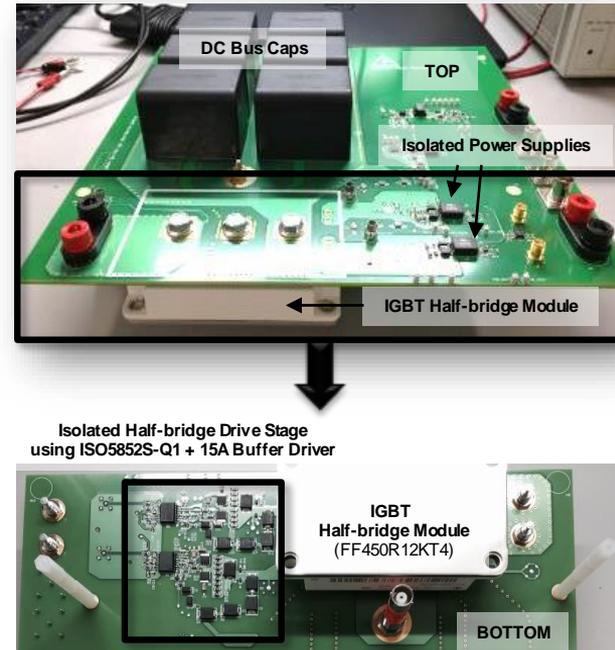
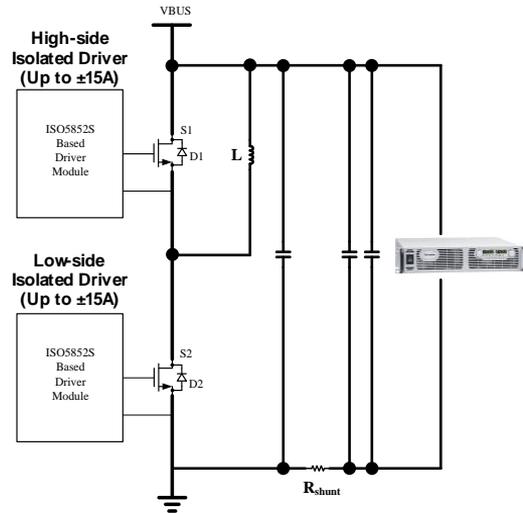
(2) Shutdown time



Short-circuit protection: hardware validation

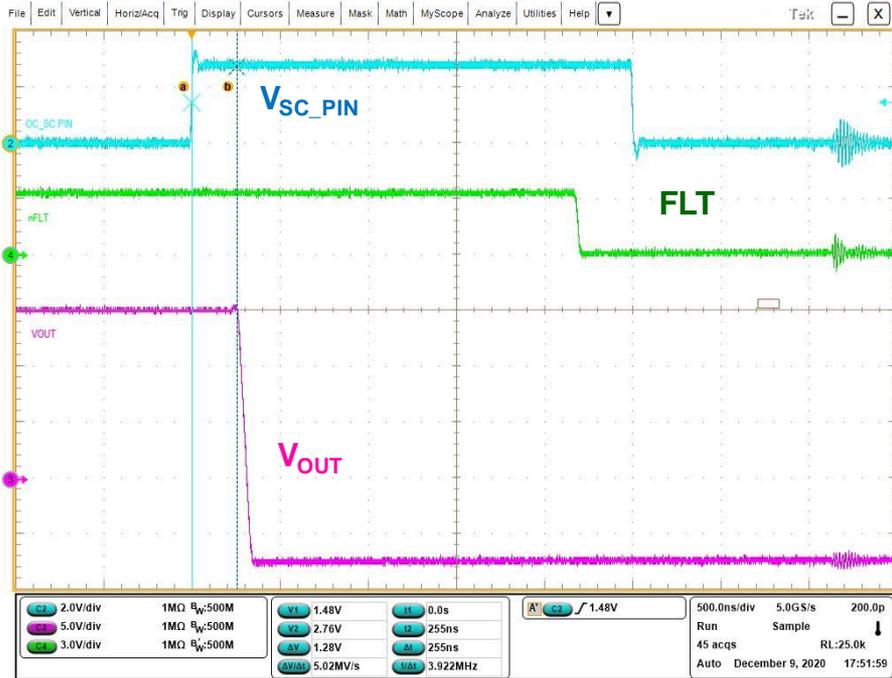
Evaluation hardware

Schematic for double-pulse test setup



SC protection response time

UCC5870-Q1

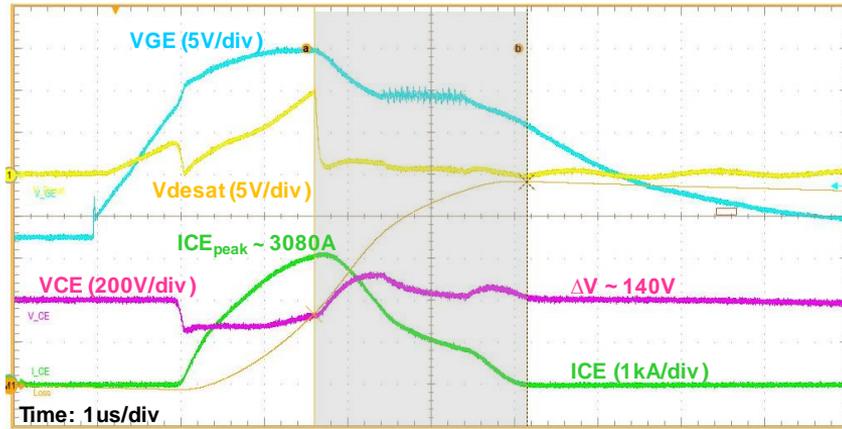


UCC5870-Q1 Response to SC event is 255ns

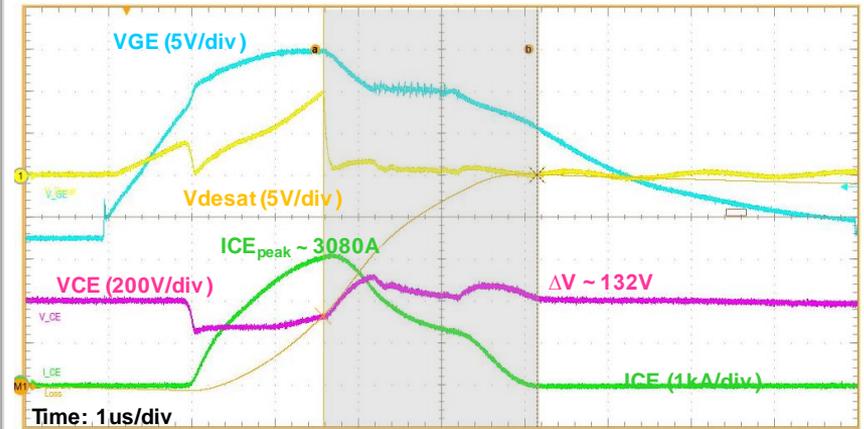
- The response time incorporates any deglitch and propagation delays in the device
- DESAT and OC features are designed to support SiC requirements, which also covers IGBT applications

DESAT with Safe Shutdown: 2LTO

Soft Turn-off (STO): $I_{STO}=0.3A$, $V_{2LOFF}=9V$, $t_{2LOFF}=1.0\mu s$



Soft Turn-off (STO): $I_{STO}=0.3A$, $V_{2LOFF}=10V$, $t_{2LOFF}=1.0\mu s$



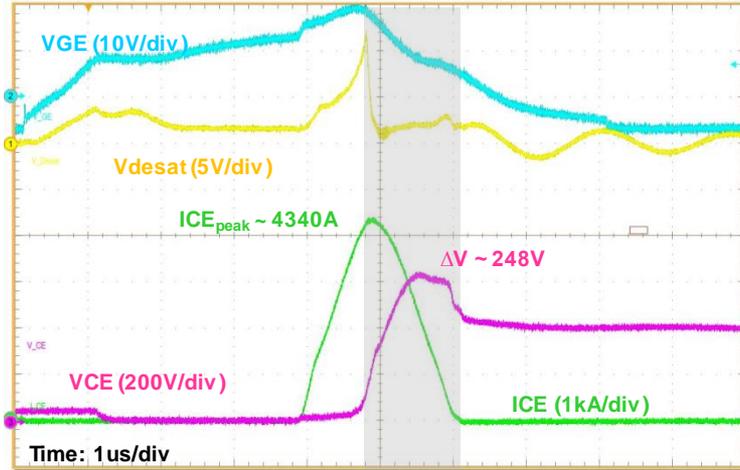
PARAMETER	MEASURED VALUE
Shutdown Energy	1.58J
V_{CE} Overshoot	140V

PARAMETER	MEASURED VALUE
Soft Turn Off Current	1.68J
V_{CE} Overshoot	132V

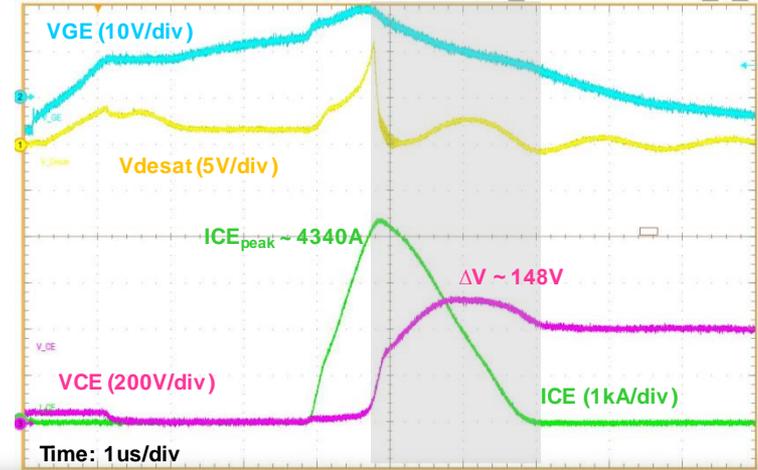
Programmable I_{STO} optimizes for the trade-off between shutdown energy and V_{CE} overshoot for IGBTs/MOSFETs

DESAT Response with Safe Shutdown: STO

Soft Turn-off (STO): $I_{STO}=0.6A$, $V_{DESAT_th}=9V$



Soft Turn-off (STO): $I_{STO}=0.3A$, $V_{DESAT_th}=9V$



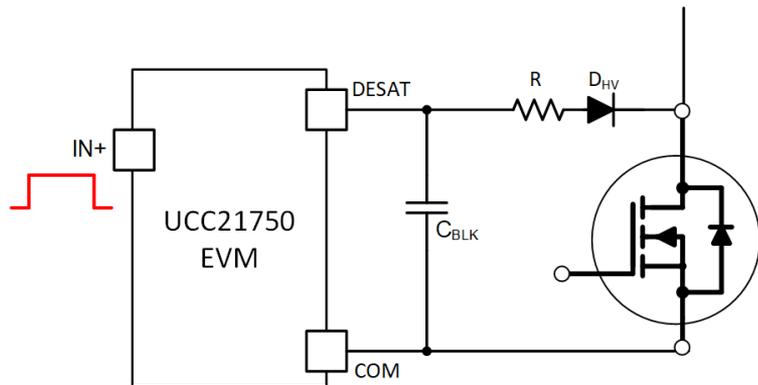
PARAMETER	MEASURED VALUE
Shutdown Energy	1.01J
V_{CE} Overshoot	248V

PARAMETER	MEASURED VALUE
Shutdown Energy	2.04J
V_{CE} Overshoot	148V

Programmable I_{STO} optimizes for the trade-off between shutdown energy and V_{CE} overshoot for IGBTs/MOSFETs

SiC with UCC21750: Soft turn-off, VDC=800 V

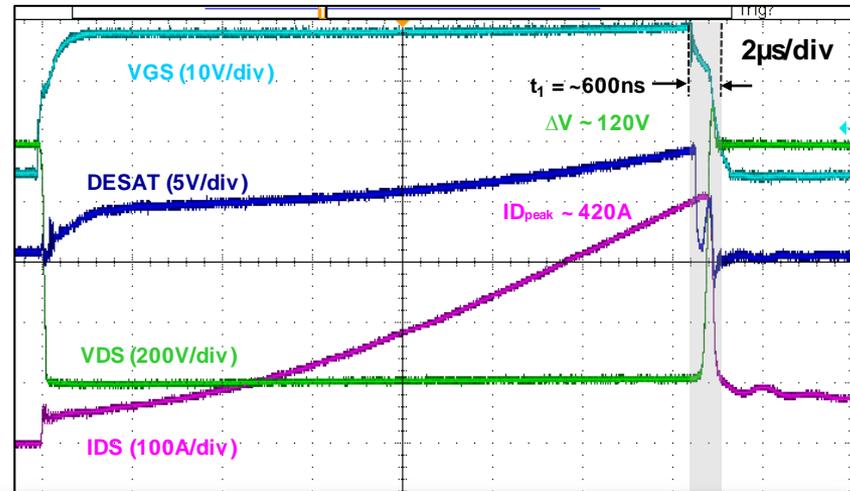
Overcurrent fault test circuit



- IN+ gives a long pulse to create Id ramp
- DESAT detects the fault and takes action

With the same *soft turn-off* characteristics, the only difference in shutdown between the UCC21750 and UCC21750 will be the detection timing

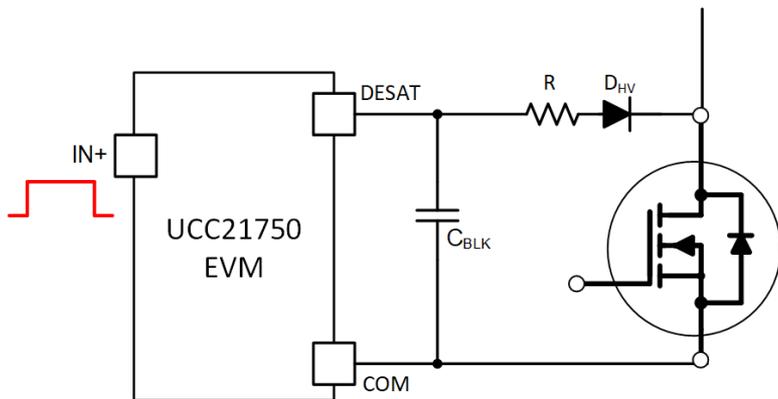
Soft turn-off (STO): UCC21750



PARAMETER	MEASURED VALUE
Shutdown Energy	~0.1J
V _{DS} Overshoot	120V

SiC with UCC21750: Soft turn-off, VDC=800 V

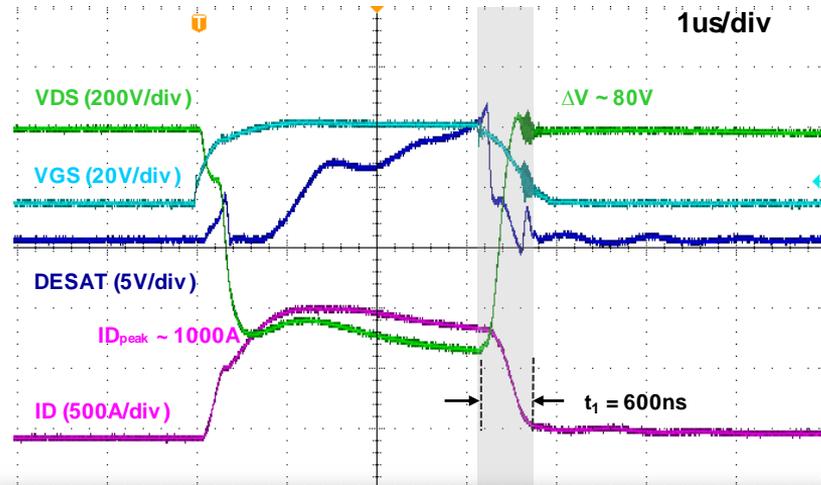
Short circuit fault test circuit



- IN+ gives a long pulse to create Id ramp
- DESAT detects the fault and takes action

With the same *soft turn-off* characteristics, the only difference in shutdown between the UCC21750 and UCC21750 will be the detection timing

Soft turn-off (STO): UCC21750



PARAMETER	MEASURED VALUE
Shutdown Energy	0.24J
V _{DS} Overshoot	80V

DESAT using OC

Gate Driver Parameters

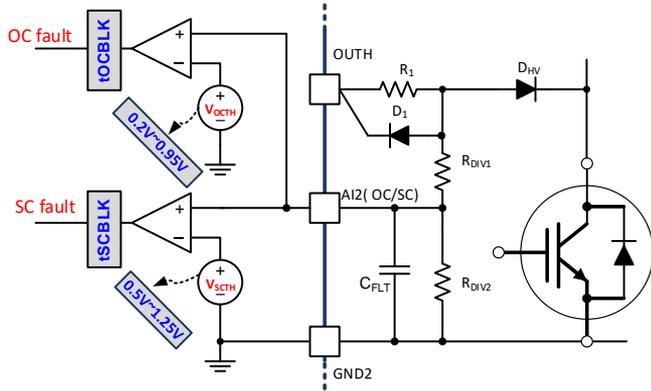
- VOCTH
- ICHG
- tOCCOFF
- ISTO

$$t_{BLK} = -\frac{R1 + R2}{R1 + R2 + R3} * R3 * C1 * \ln\left(1 - \frac{R1 + R2 + R3}{R3} * \frac{V_{OCTH}}{VDD}\right)$$

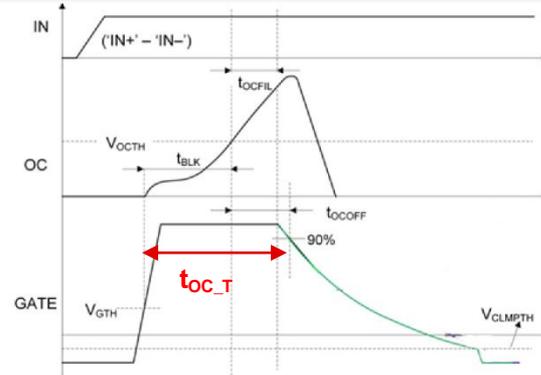
$$t_{OC_T} = t_{BLK} + t_{OCCOFF}$$

$$t_{STO} = C_{ISS} * (VDD - V_{th}) / I_{STO}$$

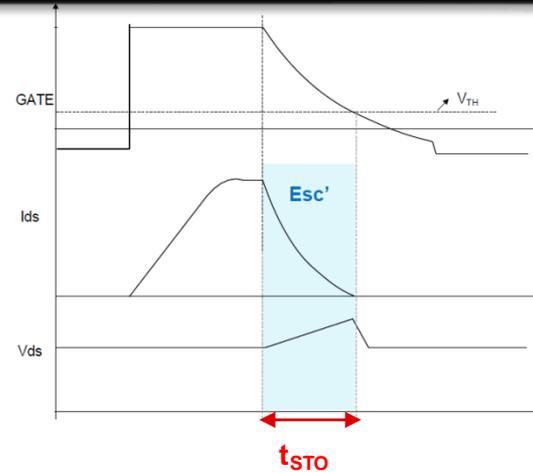
$$t_{Total} = t_{OC_T} + t_{STO}$$



(1) Detection time

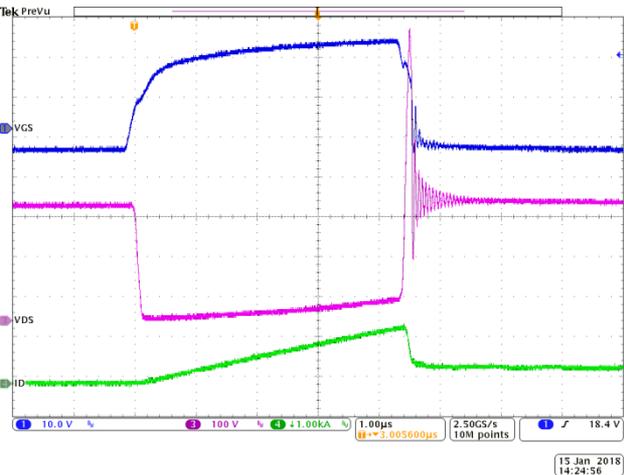


(2) Shutdown time



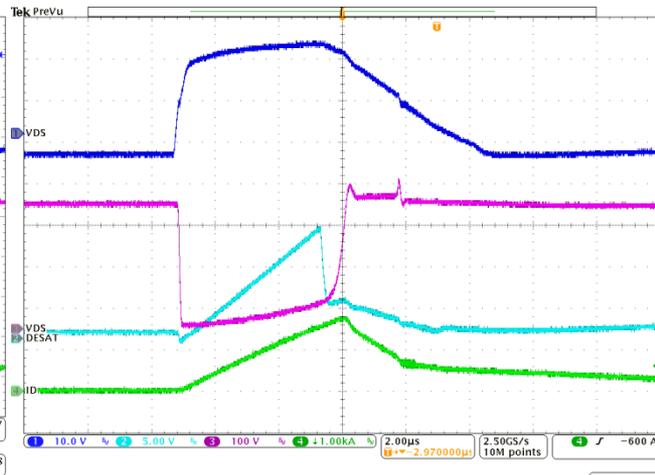
Short circuit protection for SiC MOSFETs: with or without soft turn-off

V_{DS} (100V/div), I_C (1kA/div),
 V_{GS} (10V/div)



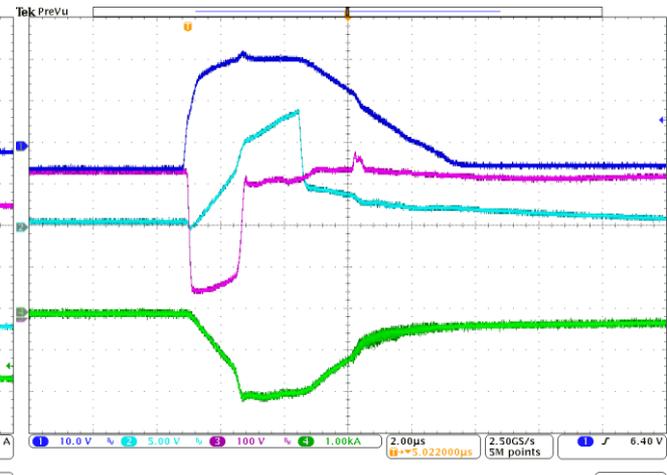
Hard turn-off

V_{DS} (100V/div), I_C (1kA/div),
 V_{GS} (10V/div), V_{DESAT} (5V/div)



Soft turn-off for hard SC

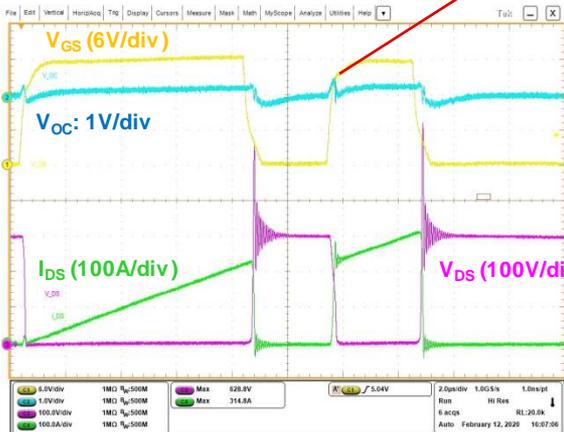
V_{DS} (100V/div), I_C (1kA/div),
 V_{GS} (10V/div), V_{DESAT} (5V/div)



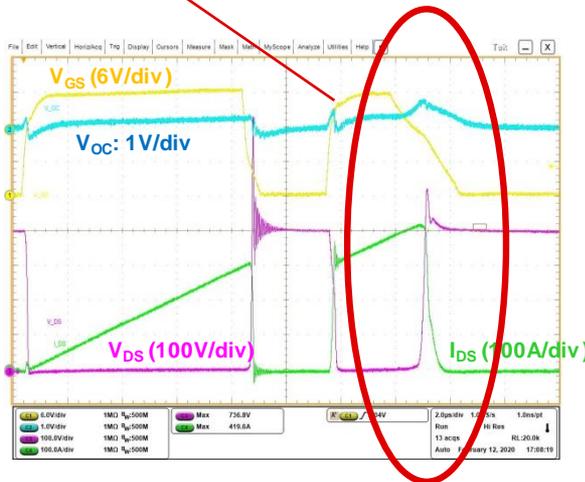
Soft turn-off for SC under load

OC pin waveform

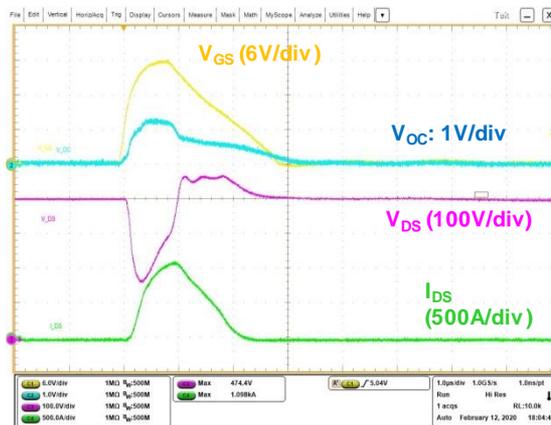
This noise must be less than the threshold



Normal switching



Soft turn-off after OC protection

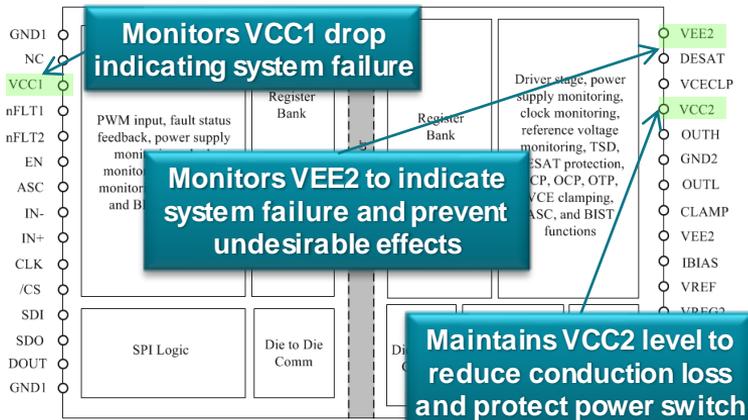


Soft turn-off after SC protection

Additional protection features

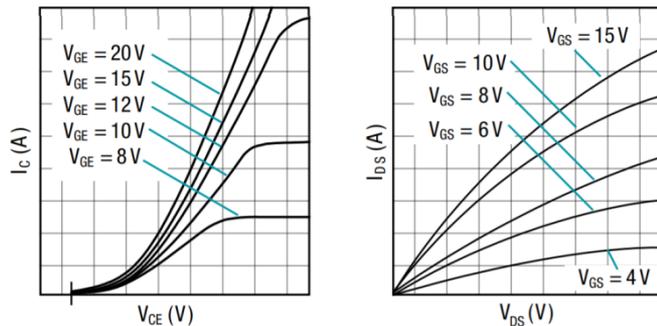
Over and undervoltage lockout: adjustability

System benefits



- ✓ **Flexible Supply Monitoring Across Platforms:** Programmable **OVLO** and **UVLO** thresholds for primary and secondary supplies
- ✓ **Adjustable to Fit Power Switch Requirements:** Monitor supply voltage to **protect IGBT and SiC MOSFETs** from undue stress
- ✓ **Gather Important Feedback:** **Configurable fault outputs** nFLT1 and nFLT2 to report various faults and warnings

Measurements & specifications



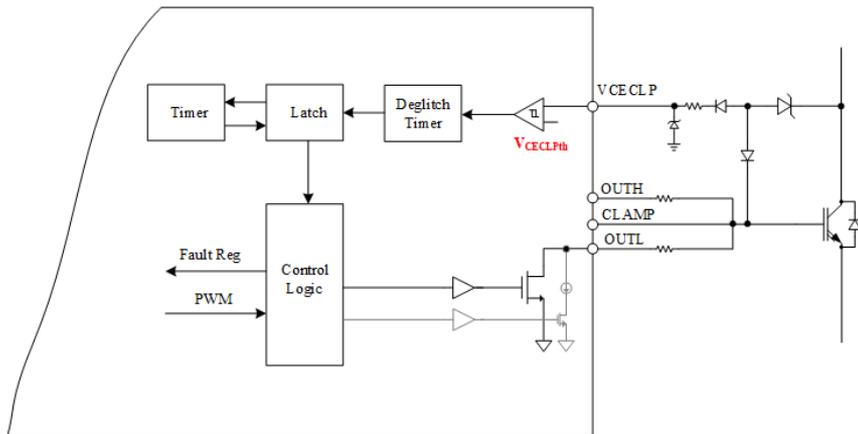
IGBT vs. SiC MOSFET I-V characteristics vary with respect to gate drive voltage (V_{CC2}) and conduction loss

Supply	UVLO Options	OVLO Options
VCC1	3.3V, 5V	3.3V, 5V
VCC2	16V, 14V, 12V, 10V	23V, 21V, 19V, 17V
VEE2	-3V, -5V, -8V, -10V	-5V, -7V, -10V, -12V

Thresholds can be optimized based on gate voltage of power switch!

V_{CE} clamp: flexibility, protection

System benefits



- ✓ **System Flexibility:** Option between **3 communication modes** (address-based, daisy-chain, regular SPI)
- ✓ **Ease of System Programmability:** Simplify design time and engineering effort
- ✓ **Address-based SPI:** Only requires **4 I/O pins** and provides **6x shorter response time** compared to daisy chain configuration

Measurements & specifications



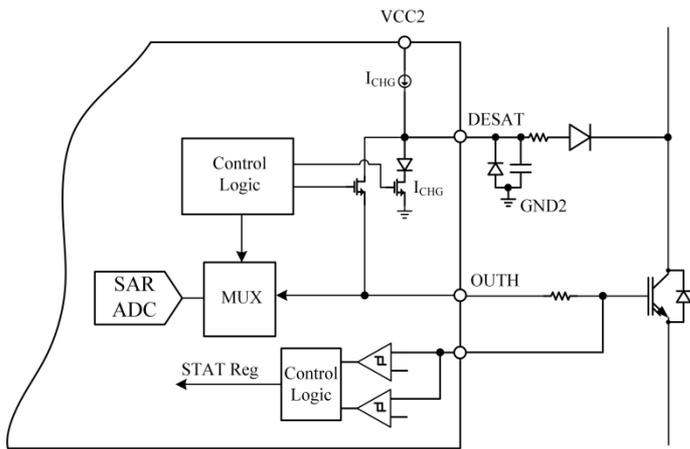
UCC5870-Q1 V_{CE} Clamp Characteristics

		Typ. Value
V _{CE} Clamp Threshold	V _{CECLPTH}	2.2 V
V _{CE} Clamp Threshold Hysteresis	V _{CECLPHYS}	200 mV
V _{CE} Clamping Intervention Time	t _{VCECLP}	20 ns
V _{CE} Clamping Hold Time	t _{VCECLP_HLD}	Adjustable

High voltage clamping only active for a short time to prevent shoot-through

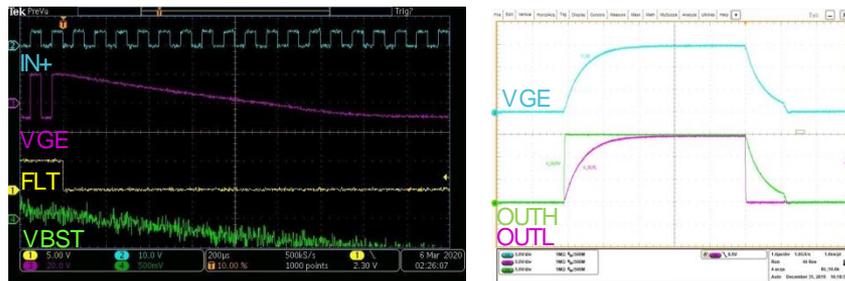
Gate voltage monitoring: flexibility, protection

System benefits



- ✓ **Enhanced System Safety:** Ensure output follows the driver input as high or low to **detect a system failure** such as open resistor at gate or gate short
- ✓ **Real-Time Fault Monitoring:** **Dedicated fault output** to feed information back to MCU
- ✓ **Size Reduction:** **Removes external components** to monitor driver output voltage

Measurements & specifications



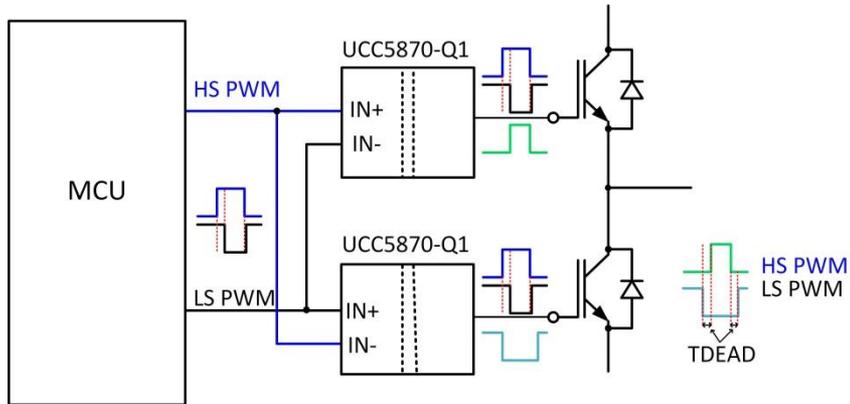
- ✓ VBST voltage drop triggers Gate Voltage Monitor Fault
- ✓ When Gate Voltage Monitor Fault is detected, the output is turned off

UCC5870-Q1 Gate Voltage Monitor Characteristics	Typ. Value
Gate Voltage Monitor Threshold Value (reference to V_{CC2})	-3 V
Gate Voltage Monitor Threshold Value (reference to V_{EE2})	3 V
Blanking Time (Programmable)	500-4000 ns
Deglintch Time	250 ns

Fault can be reported to nFLT1 or nFLT2, or report to state to nFLT2

Motor controller interface: flexibility, protection

System benefits



- ✓ **System Flexibility:** Supports **100% duty cycle** operation (min 250ns input pulse, 30 kHz max frequency with ADC)
- ✓ **Shoot-through Protection:** Interlocking with two drivers prevents shoot-through with **programmable dead time** and **fault reporting**
- ✓ **Noise Immunity:** Minimum **pulse width rejection** on IN+ and IN-

Measurements & specifications

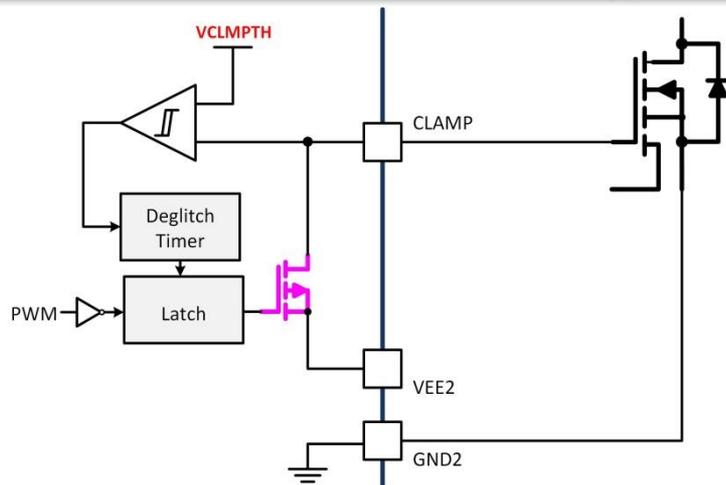


Shoot-through prevented and STP Fault triggered

- ✓ **100% Duty Cycle** possible with integrated charge pump
- ✓ **Programmable dead time:** 0ns, 105ns, 175ns...up to 4445ns in increments of 70ns
- ✓ **Noise rejection** chosen based on programming input pulse width rejection ranging from **0ns to 210ns**

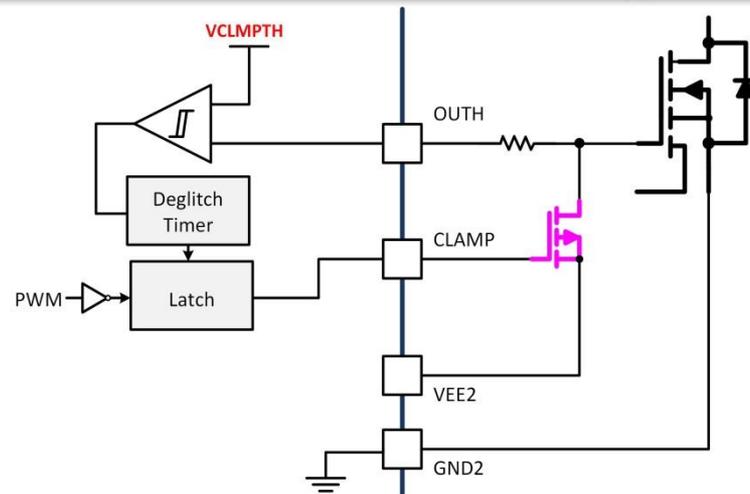
Active Miller Clamp

Internal Miller Clamp



- Miller Clamp **inside** IC
- **5A** peak pull-down current
- Adjustable VCLMPH (1.5V, 2V, 3V, or 4V)
- Suitable for close connection of CLAMP pin to gate terminal of power switch
- System challenges when connection from CLAMP pin to Gate terminal of power switch is too long (high inductance)

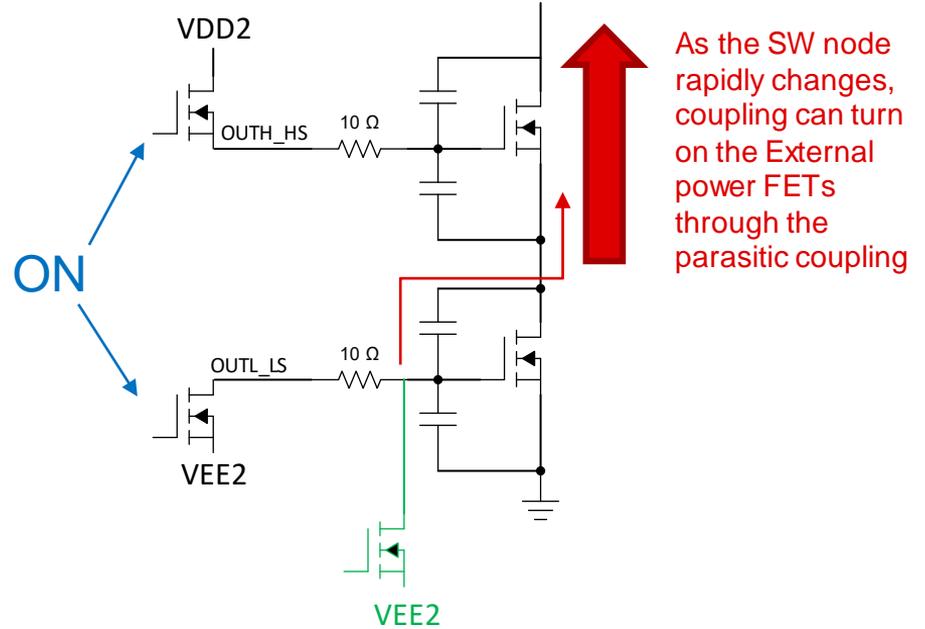
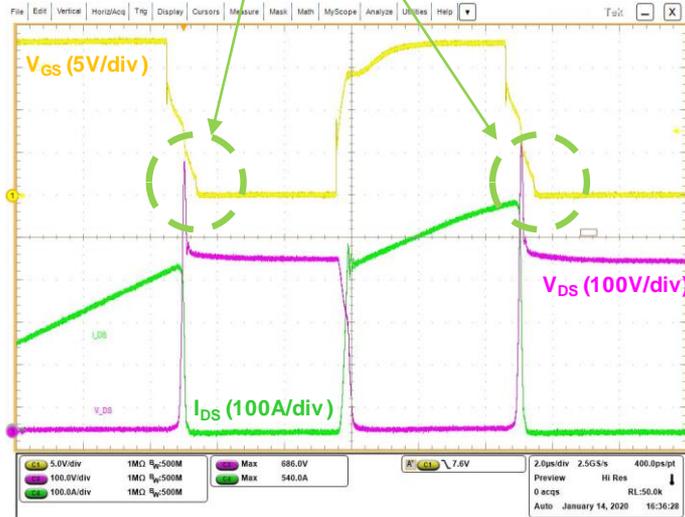
External Miller Clamp



- Miller Clamp **outside** IC: Discrete Si MOSFET switch
- **Flexible peak pull-down current +5V** signal w.r.t. VEE2 @ CLAMP pin
- Adjustable VCLMPH (1.5V, 2V, 3V, or 4V)
- Suitable when CLAMP pin is far away from gate terminal of power switch
⇒ Low inductance from clamp to gate terminal of power switch
- Better technique for power module implementation

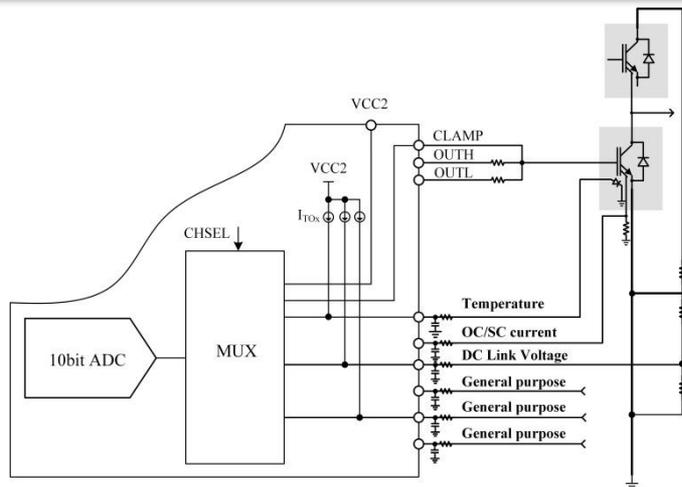
Active Miller Clamping

Miller clamp kicks in and pulls the gate to VEE with a low impedance path



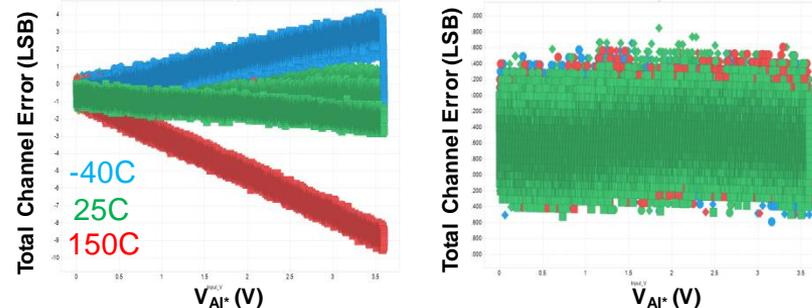
6-channel ADC: accuracy, integration, low noise

System benefits



- ✓ **System Flexibility:** Variety of configurations to measure **temperature, current, or voltage**
- ✓ **Higher Accuracy:** Integrated **10-bit** ADC and internal OR external **VREF**
- ✓ **System Cost/Size Reduction:** Save **~\$0.37** per channel by eliminating external ADC

Measurements & specifications

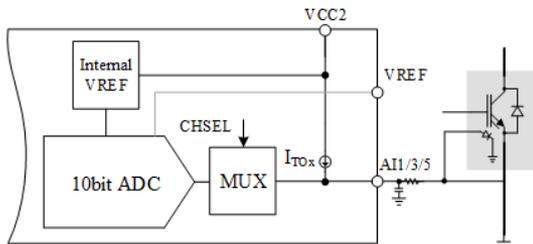


Precision reference can measure $A1^*$ to within $\pm 1/2$ LSB ($\pm 3.5/7$ mV)

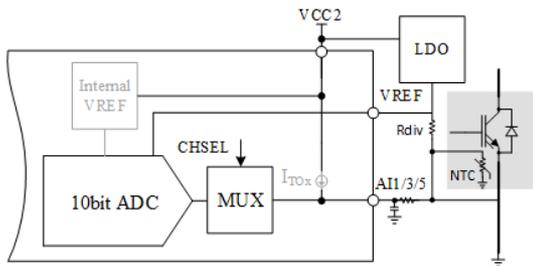
ADC	MIN	TYP	MAX
Input Voltage Range (A1 to A6)	0V	3.6V	3.636V
VREF (V)		4	
INL, external ref (LSB)	-1.2		1.2
DNL (LSB)	-0.75		0.75
External ADC Reference Turn-On Delay (us)	10		
NTC bias current (prog range, uA)	100		1000

6-channel ADC: use cases

Isolated switch temperature sensing: NTC/PTC/diode



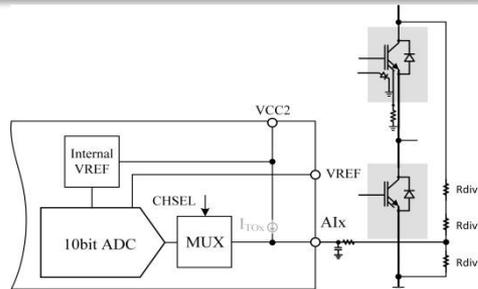
Example using thermal diode + internal VREF



Example using thermal diode + external VREF

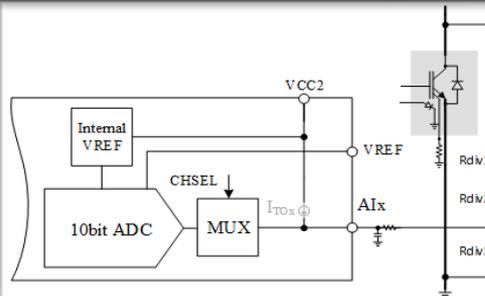
- Power switch junction temperature monitoring
- Power switch over-temperature configurable warning and shut-down
- Option for internal or external VREF to improve accuracy

Redundant Bus voltage measurement



- HV Bus Sensing
- Monitor DC link voltage
- Referenced to low-side driver common

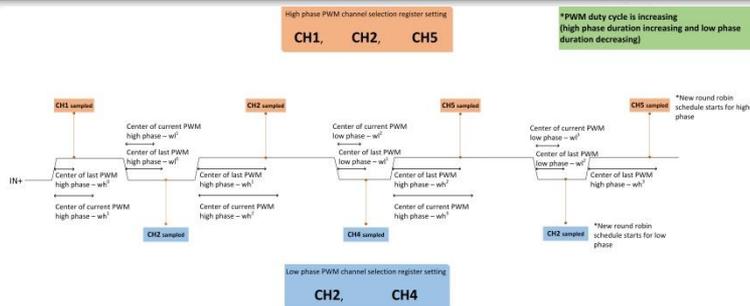
VCE voltage measurement



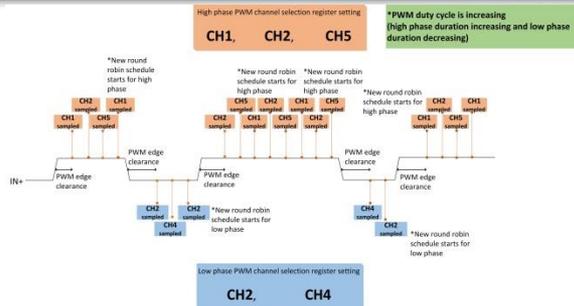
- HV VCE Sensing
- Monitor if phase open or closed
- Option to reference to high-side or low-side common

6-channel ADC: sampling modes

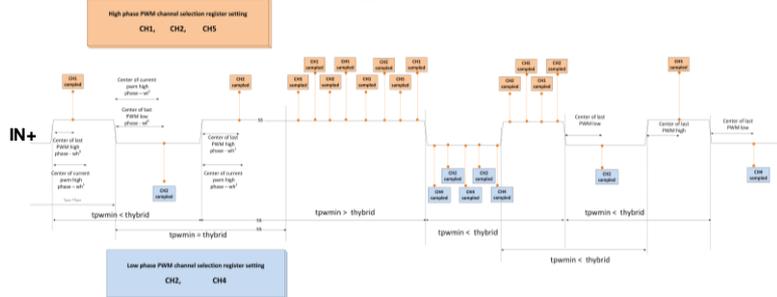
Center sampling mode



Edge sampling mode



Hybrid mode



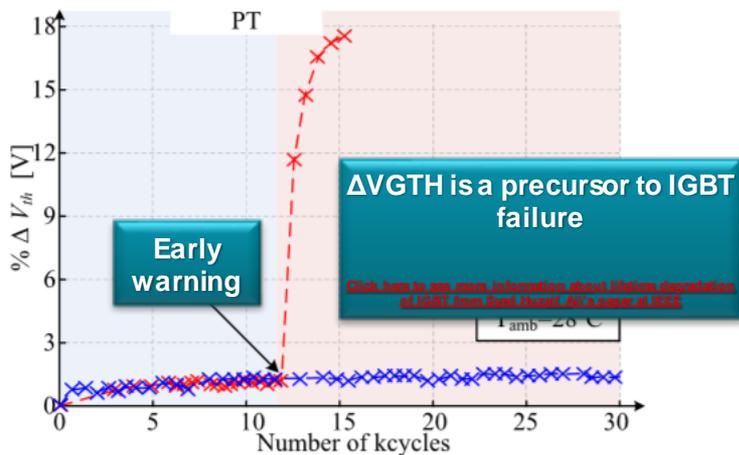
Configuring ADC modes

- Three modes available to ensure least amount of sw itching noise
- **Center sampling mode:** samples in middle of sw itching cycle
- **Edge sampling mode:** samples at start or end of each sw itching cycle
- **Hybrid sampling mode:** samples mode samples in the center until a cycle is significantly longer then the one before it

Power Switch V_{GTH} Monitor: System Reliability Feature

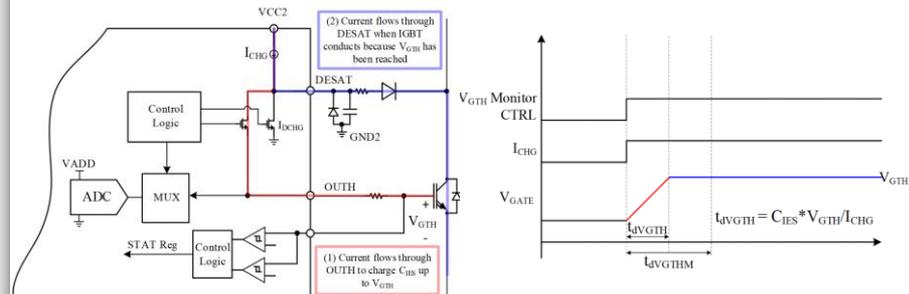


System Benefits



- ✓ **Integrated Health Monitor:** Gate driver used to perform **threshold voltage measurements** over system lifetime
- ✓ **Help System Anticipate Failure:** Works with MCU to provide **critical power switch data**

Measurements & Specifications



Operation of V_{GTH} monitor and timing diagram

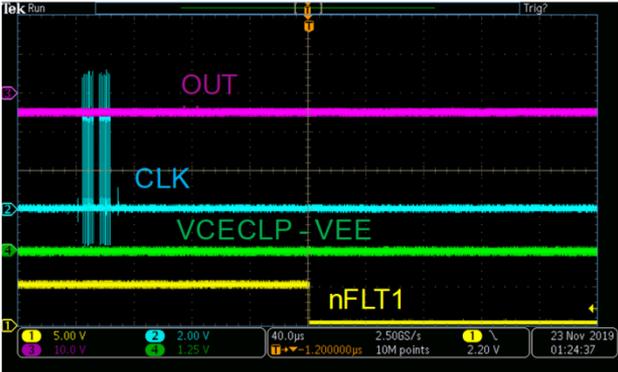


*V_{GTH} is accurate measure of health of external power switch
Test done with Cree SiC MOSFET C3M0065100K with voltage threshold of 1.8-3V*

Built-in self-test (BIST): enhanced reliability

BIST (built-in self test) mechanisms	
UV/OV Comparators of Internal Regulator	Automatic on Power Up (Latent Fail Check)
UV/OV Comparators (VCC1, VCC2, VEE2)	Automatic on Power Up (Latent Fail Check)
Main Clock	Automatic on Power Up (Latent Fail Check)
Comparator for Thermal Shutdown	Automatic on Power Up (Latent Fail Check)
DESAT diagnostic	On-Demand (with SPI Command)
OCP, SCP and OTP comparator diagnostics	On-Demand (with SPI Command)
VCE clamping detection diagnostics	On-Demand (with SPI Command)
Gate Monitoring	On-Demand (with SPI Command)
CRC	On-Demand (with SPI Command)

Example & benefits

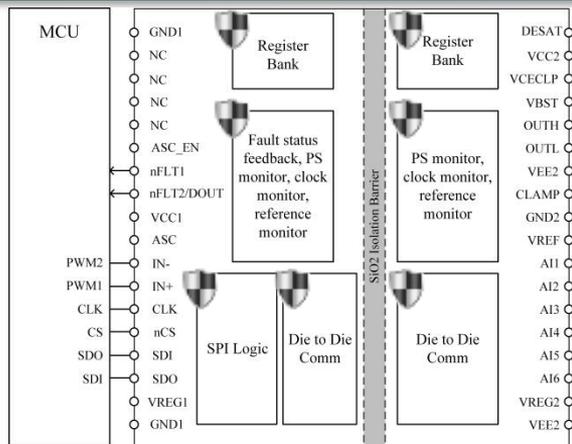


SPI command sent to check VCE Clamp and fault reported showing functioning detection mechanism

- ✓ **System and On-Demand Diagnostics:** Ensures operation of critical driver functions **at power up** and **on-demand** using SPI programming
- ✓ **Enables Easier ASIL-D Implementation:** Integrated diagnostics to **support system-level functional safety** requirements

Other diagnostics and protection: reliability and flexibility

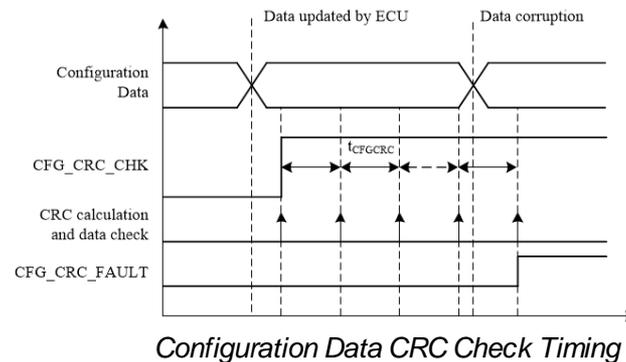
System benefits



*Block diagram only depicts certain blocks

- ✓ **Data Protection:** **CRC (cyclic redundancy check)** protects various registers against faulty data transmission when in the Active State
- ✓ **Communication Protection:** CRC checks for SPI transfer are **continuously updated** as SPI data is sent/received
- ✓ **Flexible Feedback:** **Configurable fault outputs** nFLT1 and nFLT2 to report various faults and warnings

Measurements & specifications



- ✓ When the driver is in the ACTIVE state, the configuration and control registers are protected by the CRC engine (this can be disabled)
- ✓ An error can be induced on the primary or secondary side
- ✓ CRC checks on SPI transfers are continuously updated as traffic is sent/received and is updated with every 16-bits that are received

See the datasheet Detailed Description section for more information



©2021 Texas Instruments Incorporated. All rights reserved.

The material is provided strictly "as-is" for informational purposes only and without any warranty.
Use of this material is subject to TI's **Terms of Use**, viewable at [TI.com](https://www.ti.com)

IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to TI's Terms of Sale (<https://www.ti.com/legal/termsofsale.html>) or other applicable terms available either on [ti.com](https://www.ti.com) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265
Copyright © 2021, Texas Instruments Incorporated