











**TPS65980** 

ZHCSCE0A - APRIL 2014 - REVISED APRIL 2014

# TPS65980 Thunderbolt™ 总线电源降压/升压

## 特性

- 由 Thunderbolt™ 总线供电
- 2.5V 至 15.75V 输入
- 3.3V 输出
- 电缆电源输出电流限制
- 热关断

## 应用范围

- Thunderbolt™/Thunderbolt™ 2 系统
- 总线供电系统
- 电源管理系统

## 3 说明

TPS65980 是一款直流/直流开关稳压器,此稳压器由 电压范围介于 2.5V 至 15.75V 之间的 Thunderbolt™ 或 Thunderbolt™ 2 电源总线供电,并且生成 3 个独立 3.3V 电源输出。

TBT OUT 电源为本地外设 Thunderbolt™ 控制器和支 持电路供电。 CBL OUT 电源将电能输送回 Thunderbolt™ 电缆,并且具有可调电流限值。 DEV\_OUT 电源为器件中的所有其他电路供电来执行 其设计的功能。

TPS65980 采用 24 引脚

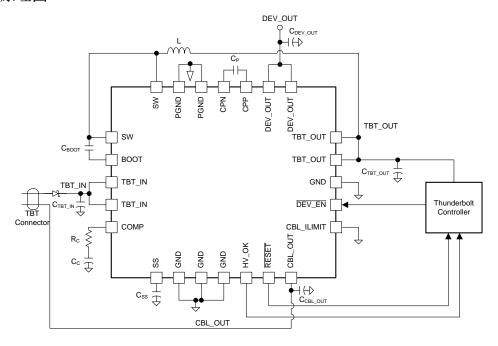
5mm x 4mm x 0.9mm 超薄四方平面无引线 (VQFN) 封装。

#### 器件信息(1)

器件名称	封装	封装尺寸
TPS65980	VQFN (24)	5mm x 4mm

(1) 要了解所有可用封装,请见数据表末尾的可订购产品附录。

# 简化电路原理图

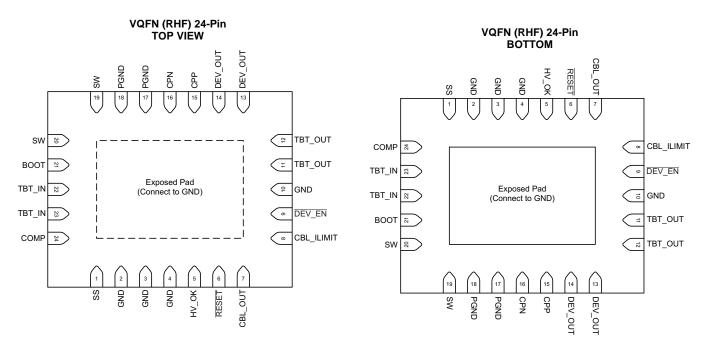




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# 6 Pin Configuration and Functions



#### **Pin Functions**

	PIN	1/0	DECORPTION
NO.	NAME	I/O	DESCRIPTION
1	SS	ANALOG	Soft Start Capacitance. This pin sets the soft start ramp rate when the TBT_IN voltage ramps from 0V to high voltage.
2, 3, 4	GND	GND	Device Ground
5	HV_OK	OUTPUT	High Voltage Present Indicator. This pin indicates that a high voltage is present on TBT_IN. The output asserts high when the TBT_IN pin is above the V <sub>HVT</sub> voltage and the RESET output is asserting high.
6	RESET	OUTPUT	Reset output indicator. This pin asserts low when TBT_OUT is in under-voltage.
7	CBL_OUT	PWROUT	Current Limited Power Output to Thunderbolt™ Cable. This pin supplies power to the Thunderbolt™ cable. The current limit of this pin is set by the CBL_ILIMIT pin.
8	CBL_ILIMIT	INPUT	Current Limit Set. Logic input that sets the current limit state on the CBL_OUT pin. Tie pin to TBT_OUT for a logic high input.
9	DEV_EN	INPUT	Device Enable Input. When input pin is high, DEV_OUT is high impedance. When input pin low, DEV_OUT is connected to TBT_OUT.
10	GND	ANALOG	Device Ground
11, 12	TBT_OUT	PWROUT	Power Output to Thunderbolt™ circuitry. This pin supplies power to the Thunderbolt™ controller.
13, 14	DEV_OUT	PWROUT	Power Output to peripheral device. This pin supplies power to circuitry not associated with the Thunderbolt™ controller or the Thunderbolt™ cable. It is intended to supply power to the peripheral device main function.
15	CPP	ANALOG	Charge Pump Capacitance Positive Output
16	CPN	ANALOG	Charge Pump Capacitance Negative Output
17, 18	PGND	GND	Buck Controller Power Ground
19, 20	SW	ANALOG	Buck Controller Switch Output
21	BOOT	ANALOG	Buck Controller Bootstrap
22, 23	TBT_IN	PWRIN	Power Input from Thunderbolt™ Cable. This pin is the power supply to the device.
24	COMP	ANALOG	Buck Converter Compensation. This pin provides compensation to the buck converter feedback loop.



## **Specifications**

## 7.1 Absolute Maximum Ratings

over operating free-air temperature (unless otherwise noted) (1)

			MIN	MAX	UNIT
		TBT_IN	-0.3	18	
		DEV_EN	-0.3	3.6	
		BOOT	-0.3	25	
		BOOT (10 ns transient)	-0.3	27	
		BOOT (vs SW)	-0.3	7	
	Input voltage range (2)	SW	-0.6	18	V
	input voltage range V	SW (10 ns transient)	-2	20	V
		COMP	-0.3	3.6	
		SS	-0.3	3.6	
		CBL_ILIMIT	-0.3	3.6	
		CPP	-0.3	7.2	
		CPN	-0.3	3.6	
	Output voltage range (2)	TBT_OUT, CBL_OUT, DEV_OUT	-0.3	3.6	V
	Output voltage range	RESET, HV_OK	-0.3	3.6	V
V <sub>diff</sub>	Voltage from GND to Thermal Pad		-0.2	0.2	V
	Voltage from PGND to GND		-0.2	0.2	V
T <sub>A</sub>	Operating ambient temperature		-40	85	°C
TJ	Operating junction temper	ature	-40	125	°C

Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

All voltage values are with respect to network ground pin.

## 7.2 Handling Ratings

			MIN	MAX	UNIT
T <sub>stg</sub>	Storage temperature range		-55	150	°C
V	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	0	2	kV
V <sub>(ESD)</sub>		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins (2)	0	500	V

JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



# 7.3 Recommended Operating Conditions

over operating free-air temperature (unless otherwise noted)

			MIN	MAX	UNIT
TBT_IN	Supply input volta	upply input voltage range		15.75	
	DEV_EN	-0.1	3.6		
		BOOT	-0.1	25	
		SW	-0.6	16.5	
	Input voltage range	COMP	-0.1	3.6	V
VI		SS	-0.1	3.6	
		CBL_ILIMIT	-0.1	3.6	
		CPP	-0.1	7.2	
		CPN	-0.1	3.6	
	Output voltage	TBT_OUT, CBL_OUT, DEV_OUT	-0.1	3.6	
Vo	range	RESET, HV_OK	-0.1	3.6	V
T <sub>A</sub>	Operating free-air temperature		-40	85	°C
T <sub>J</sub>	Operating junction temperature		-40	125	°C

## 7.4 Thermal Information

		TPS65980	
	THERMAL METRIC <sup>(1)</sup>	RHF	UNIT
		24 PIN	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	30.1	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	26.9	
$R_{\theta JB}$	Junction-to-board thermal resistance	8.2	90044
ΨЈТ	Junction-to-top characterization parameter	0.3	°C/W
ΨЈВ	Junction-to-board characterization parameter	8.2	
R <sub>0JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	1.5	

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.



#### 7.5 Electrical Characteristics

Unless otherwise noted all specifications applies over the  $V_{TBT\_IN}$  range and operating ambient temperature of  $-40^{\circ}\text{C} \le T_{A} \le 85^{\circ}\text{C}$ ,  $C_{TBT\_IN} = 22~\mu\text{F}$ ,  $C_{TBT\_OUT} = 10~\mu\text{F}$ ,  $C_{CBL\_OUT} = 1~\mu\text{F}$ , CSS = 10 nF, and 33 V/ $\mu$ s logic input transitions. Typical values are for  $V_{TBT\_IN} = 12~V$  and  $T_{A} = 25^{\circ}\text{C}$ .

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER SUPPI	LIES AND CURRENTS	·				
V <sub>TBT_IN</sub>	TBT_IN Input voltage range		2.5	12	15.75	V
	TBT_OUT to RESET clear high	TBT_OUT rising	3	3.1	3.2	
$V_{REF\_RSTN}$	TBT_OUT to RESET assert low	TBT_OUT falling	2.5	2.6	2.7	V
V <sub>HVTR</sub>	TBT_IN to HV_OK assert	TBT_IN rising	4.36	4.5	4.64	V
V <sub>HVTHYST</sub>	TBT_IN to HV_OK clear	TBT_IN Falling hysteresis		100		mV
SR <sub>02L</sub>	TBT_IN Input slew rate	TBT_IN transition from 0 V to 3.3 V	0.1		30	kV/s
SR <sub>L2H</sub>	TBT_IN Input slew rate	TBT_IN transition from 3.3 V to 15 V	0.1		30	kV/s
I <sub>RAMP</sub>	Combined output di/dt <sup>(1)</sup>				5	kA/s
	Buck converter efficiency	I <sub>LOADTOTAL</sub> = 3 A, V <sub>TBT IN</sub> = 12 V		87%		
Efficiency	Charge pump efficiency	$V_{TBT_{LIN}} = 3.3 \text{ V, } I_{LOADTOTAL} = 25 \text{ mA}$		47%		
POWER OUTP	UT PINS (LOW VOLTAGE INPUT)(2)	TOT_IN PEOPLETAL				
V <sub>TBT_IN</sub>	TBT_IN Input voltage range		2.5	3.3	3.4	V
V <sub>TBT_OUT</sub>	TBT_OUT Output voltage range <sup>(3)</sup>		3.135	3.25	3.4	V
- 181_001		RESET high	5		50	mA
I <sub>TBT_OUT</sub>	TBT_OUT Load current <sup>(4)(5)</sup>	RESET low			100	μA
POWER OUTP	UT PINS (HIGH VOLTAGE INPUT) (6)	1,12021.1011			.00	
V <sub>TBT_IN</sub>	TBT_IN Input voltage range		10	12	15.75	V
* IBI_IN	121_iit input voitago rango	I LOADTOTAL = 1 A to 3.5 A	3.221	3.27	3.319	
$V_{TBT\_OUT}$	TBT_OUT Output voltage range (3)	I_LOADTOTAL = 0.235 A to 3.5 A	3.221	3.27	3.42	V
	TBT_OUT Load current <sup>(4)</sup>	1_LOAD TO TAL = 0.233 A to 3.3 A	235	5.21	1000	mA
I <sub>TBT_OUT</sub>	TBT_OOT LOAD CUITETIE	ILIMIT = 0, I <sub>CBL_OUT</sub> = 0 to 720 mA	3.171	3.27	3.319	IIIA
V <sub>CBL_OUT</sub>	CBL_OUT Output voltage range (3)		3.17	3.27	3.319	V
V	DEV OUT Output Voltage Bange	ILIMIT = 1, I <sub>CBL_OUT</sub> = 0 to 1.44 A	3.12	3.27	3.319	V
V <sub>DEV_OUT</sub>	DEV_OUT Output Voltage Range	I <sub>DEV_OUT</sub> = 0 to 2500 mA	3	3.21	3.319	v
	UT PINS (HIGH VOLTAGE INPUT DURING	STSTEM SLEEP)	5.0	40	45.75	V
V <sub>TBT_IN</sub>	TBT_IN Input voltage range		5.2	12	15.75	V
V <sub>TBT OUT</sub>	TBT_OUT DC Output voltage range	I_LOADTOTAL = 1 A to 3.5 A	3.221	3.27	3.319	V
		I_LOADTOTAL = 0.235 A to 3.5 A	3.221	3.27	3.42	
I <sub>TBT_OUT</sub>	TBT_OUT Load current		5		31	mA
V <sub>CBL_OUT</sub>	CBL_OUT Output voltage range (3)	I <sub>CBL_OUT</sub> = 0 to 235 mA	3.171	3.27	3.319	V
V <sub>DEV_OUT</sub>	DEV_OUT Output voltage range	$I_{DEV\_OUT} = 0$ to 700 mA	3	3.3	3.319	V
CABLE OUTPU	JT (HIGH VOLTAGE INPUT & HIGH VOLTA	•				
V <sub>CBL_OUT_MON</sub>	CBL_OUT Ramp-up monotonicity <sup>(7)</sup>	CBL_OUT ramp from off to on			0	mV
V <sub>CBL_OUT_RIP</sub>	CBL_OUT Voltage ripple	After settling All output combined Load > 1 mA			2	% <sub>P-P</sub>
		All output combined Load < 1 mA			40	$mV_{P=P}$
	CPI OUT Current limit	ILIMIT = 0	0.8	1.1	1.4	
I <sub>LIM_CBLOUT</sub>	CBL_OUT Current limit	ILIMIT = 1	1.6	2.2	2.8	Α
		RCBL_OUT = $0.5 \Omega$ to GND, ILIMIT = $0$			500	
t <sub>LIM_CBLOUT</sub>	Short circuit response time	RCBL_OUT = 0.01 $\Omega$ to GND, ILIMIT = 0			8	μs

- (1) The three voltage outputs (TBT\_OUT, CBL\_OUT, DEV\_OUT) all pull current from a single node. Therefore, the total combined current cannot exceed the maximum di/dt.
- (2) CBL\_OUT and DEV\_OUT are open (high impedance) for this input voltage range.
- (3) During light load conditions, the average output voltage may reach 3.5 V with peaks not exceeding 3.42 V.
- (4) TBT\_OUT load current flows from the TBT\_OUT pin when the device is in charge pump mode and pulls the buck converter inductor when the device is in buck mode.
- (5) TBT\_OUT load current will not go higher than 50mA until after the device asserts HV\_OK.
- (6) The maximum current supplied by the TPS65980 to all outputs is limited to 3.5 A. Max power depends on the Thunderbolt™ system and how much power is supplied to the input.
- (7) A monotonicity of 0 mV means that the output does not have a negative going ramp at anytime during its power up ramp. A ripple of up to 62 mV from the DC/DC will occur.



## **Electrical Characteristics (continued)**

Unless otherwise noted all specifications applies over the  $V_{TBT\_IN}$  range and operating ambient temperature of  $-40^{\circ}\text{C} \leq T_{A} \leq 85^{\circ}\text{C}$ ,  $C_{TBT\_IN} = 22~\mu\text{F}$ ,  $C_{TBT\_OUT} = 10~\mu\text{F}$ ,  $C_{CBL\_OUT} = 1~\mu\text{F}$ , CSS = 10 nF, and 33 V/ $\mu$ s logic input transitions. Typical values are for  $V_{TBT\_IN} = 12~V$  and  $T_{A} = 25^{\circ}\text{C}$ .

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT			
DEV_EN AN	DEV_EN AND ILIMIT INPUT LOGIC								
V <sub>IH</sub>	High-level input voltage		2.6			V			
V <sub>IL</sub>	Low-level input voltage				0.6	V			
I <sub>IN</sub>	Input leakage to GND	$V_{\overline{DEV}_{\overline{EN}}} = 3.3V$			1	mA			
RESET AND	HV_OK OUTPUT LOGIC								
V <sub>OH</sub>	High-level output voltage	$I_L = -1.5$ mA, Referenced to $V_{TBT\_OUT}$	-250		0	mV			
V <sub>OL</sub>	Low-level output voltage	I <sub>L</sub> = 1.5 mA	0		250	mV			
SOFT STAR	T <sup>(8)</sup>								
I <sub>INRUSH</sub>	Inrush current di/dt				250	kA/s			
THERMAL S	SHUTDOWN								
T <sub>SD</sub>	Shutdown temperature		120	135	150	°C			
T <sub>SDHYST</sub>	Shutdown hysteresis			10		°C			

<sup>(8)</sup> The charge pump will limit the normal ramp of current. Soft start will control the inrush current when the input ramps from 0 V to high voltage (not a normal operating condition). See recommended components section for required soft-start cap.

## 7.6 Timing Requirements

	ng Roquiromonto		MIN	TYP	MAX	UNIT
t <sub>IN2OR</sub>	TBT_IN to TBT_OUT On Time	$V_{TBT\_IN} \ge 0.9 \times V_{TBT\_IN(min)}$ to $V_{TBT\_OUT} \ge 0.99 \times V_{TBT\_OUT(min)}$ $R_{TBT\_OUT} = 100 \Omega$			20	ms
t <sub>IN2OF</sub>	TBT_IN to TBT_OUT Off Time	$V_{TBT\_IN} \le 0.9 \times V_{TBT\_IN(min)}$ to $V_{TBT\_OUT} \le 0.1 \times V_{TBT\_OUT(min)}$ $R_{TBT\_OUT} = 100 \Omega$		2.4	4	ms
t <sub>OUT2RR</sub>	TBT_OUT to RESETZ High time	$V_{TBT\_OUT} \ge V_{REF\_RSTN(max)}$ rising to $V_{RESET} = 0.9 \times VOH$ , $C_{RESETN} = 100 pF$			20	μs
t <sub>IN2RF</sub>	TBT_IN to RESETZ Low time	$V_{TBT\_IN} \le 0.9 \times V_{TBT\_IN(min)}$ to $V_{RESET} = 0.1 \times V_{OH}$ , $C_{RESETN} = 100 \text{ pF}$			20	ms
t <sub>HV2OKR</sub>	TBT_IN Rise to HV_OK	$V_{TBT\_IN} \ge V_{HVTR}$ to $V_{HV\_OK} = 0.9 \times V_{OH}$ $C_{HV\_OK} = 100 \text{ pF}$			10	μs
t <sub>HV2OKF</sub>	TBT_IN Fall to HV_OK	$V_{TBT\_IN} \le V_{HVTR} - V_{HVTHYST}$ to $V_{HV\_OK} = 0.1 \times V_{OH}$ , $C_{HV\_OK} = 100 \text{ pF}$			10	μs
t <sub>HV2CR</sub> (1)(2)	HV_OK to CBL_OUT On time	$V_{HV_OK} \ge 1.65 \text{ V to } V_{CBL_OUT} = 2.95 \text{ V}$ $R_{CB_OUT} = 100 \Omega$ , $C_{HV_OK} = 100 \text{ pF}$	0.1		10	ms
t <sub>HV2CF</sub>	HV_OK to CBL_OUT Off time	$V_{HV_OK} \le 1.65 \text{ V to } V_{CBL_OUT} = 2.95 \text{ V}$ $R_{CB_OUT} = 100 \Omega$ , $C_{HV_OK} = 100 \text{ pF}$			40	μs
t <sub>RCBL</sub>	CABLE_OUT Ramp time	$V_{CBL\_OUT}$ ramp 10% to 90% $C_{CBL\_OUT}$ = 0 to 52 $\mu$ F	0.1		10	ms
t <sub>DEVEN</sub>	DEV_EN to DEV_OUT On time	$V_{\overline{DEV\_EN}} \le 1.65 \text{ V to } V_{\overline{DEV\_OUT}} = 2.7 \text{ V}$ $R_{\overline{DEV\_OUT}} = 100 \Omega$	0.1		10	ms
t <sub>DEVDIS</sub>	DEV_EN to DEV_OUT Off time	$V_{\overline{DEV\_EN}} \ge 1.65 \text{V to } V_{\overline{DEV\_OUT}} = 2.7 \text{ V}$ RDEV_OUT = 100 $\Omega$			50	ms
t <sub>HV2DEVEN</sub>	Wait time from HV_OK High before DEV_EN can be asserted low <sup>(2)</sup>	$V_{HV\_OK} \ge 1.65 \text{ V to } V_{\overline{DEV\_EN}} \le 1.65 \text{ V}$ $C_{HV\_OK} = 100 \text{ pF}$	2			ms

<sup>(1)</sup> TBT\_IN must transition from 3.3 V to high voltage, not from 0 V to high voltage

<sup>(2)</sup> During the transition from low voltage input to high voltage input, the total load of all outputs combined can not exceed 85 mA until 2 ms after HV\_OK asserts high.

#### 7.7 Timing Diagrams

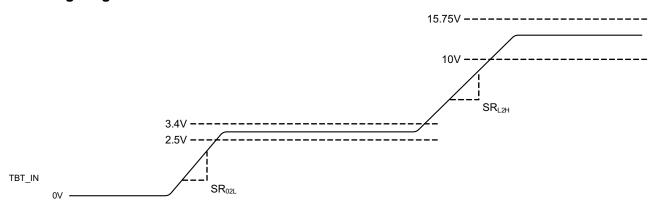


Figure 1. TBT\_IN Slew Rates

The TPS65980 has two normal operating regions. The first region is when  $2.5 \text{ V} \leq V_{TBT\_IN} \leq 3.4 \text{ V}$ . This is the normal power-up state and is termed the low-voltage state. When the input transitions to this range, the input slew rate must meet the  $SR_{02L}$  limits. In this voltage range, the TPS65980 operates with a charge pump to generate the nominally 3.3 V output. When the input voltage moves to the higher end of this range, the buck converter takes over to produce the 3.3 V. In normal operation, the TPS65980 input voltage will transition from the low-voltage range to a high-voltage range where  $10 \text{ V} \leq V_{TBT\_IN} \leq 15.75 \text{ V}$ . This is the high-voltage state and is the state where the TPS65980 will operate most of the time. In this state, the device operates as a buck converter providing a nominally 3.3 V output. Figure 1 shows the input voltage transitions and states.

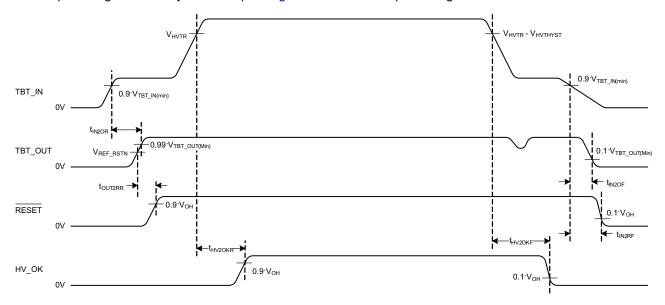


Figure 2. Timing Diagram

Figure 2 shows normal operating timing diagram for the TBT\_OUT output voltage and the RESET and HV\_OK output indicator signals. When TBT\_IN transitions to the low-voltage range, TBT\_OUT will power up a short time later. Once TBT\_OUT reaches the normal output range, RESET will transition high. However, timing for RESET is measured from the input TBT\_IN transitioning high. When TBT\_IN transitions from the low-voltage input range to the high-voltage input range, HV\_OK will transition high. RESET is an active-high output indicating that the TBT\_OUT voltage is valid and. HV\_OK is an active-high output indicating that the TBT\_IN voltage is in the high-voltage range. When in the high-voltage state, the TPS65980 can provide much higher output current than when in the low-voltage state.



## **Timing Diagrams (continued)**

When the TBT\_IN input transitions from high-voltage to low-voltage, HV\_OK will de-assert to a logic low. When the TBT\_IN input voltage falls below the minimum operating voltage, the RESET output will de-assert low.

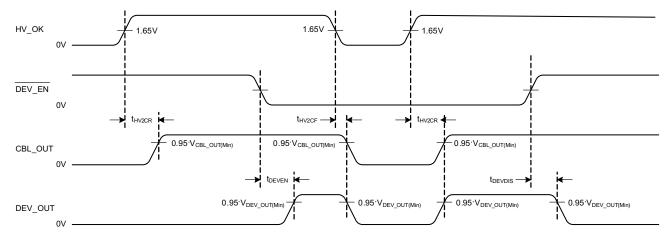
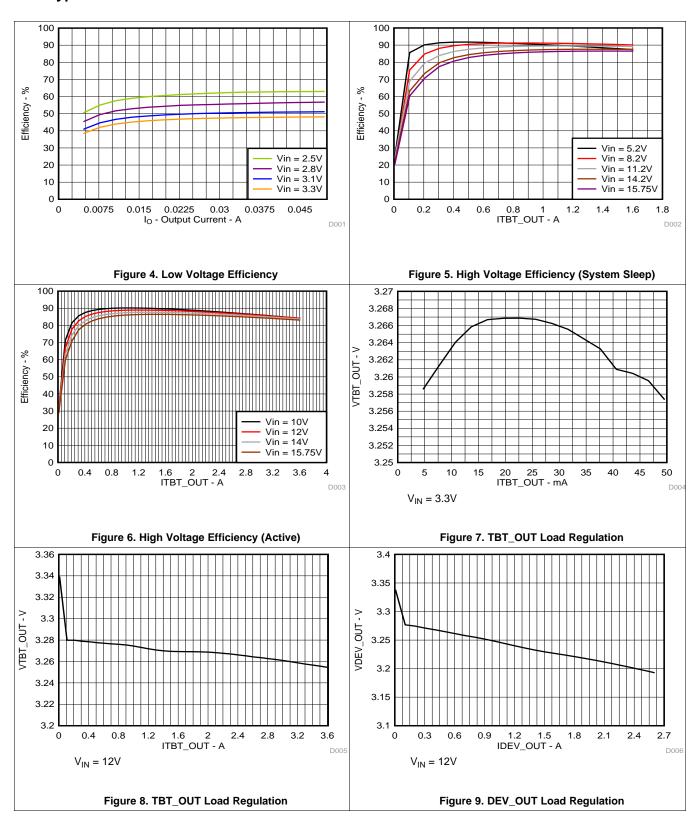


Figure 3. Timing Diagram

Figure 3 shows the CBL\_OUT and DEV\_OUT outputs and timing based on the HV\_OK signal and the DEV\_ENZ input. The CBL\_OUT output will be connected to TBT\_OUT and supplying 3.3V when HV\_OK is asserting high. The DEV\_OUT output will be connected to TBT\_OUT and supplying 3.3 V when HV\_OK is asserting high and the DEV\_ENZ input is low.

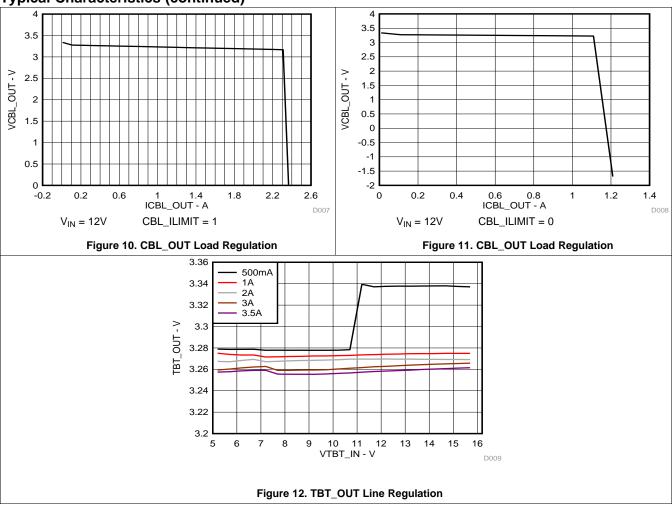
# TEXAS INSTRUMENTS

## 7.8 Typical Characteristics





## **Typical Characteristics (continued)**





#### 8 Detailed Description

#### 8.1 Overview

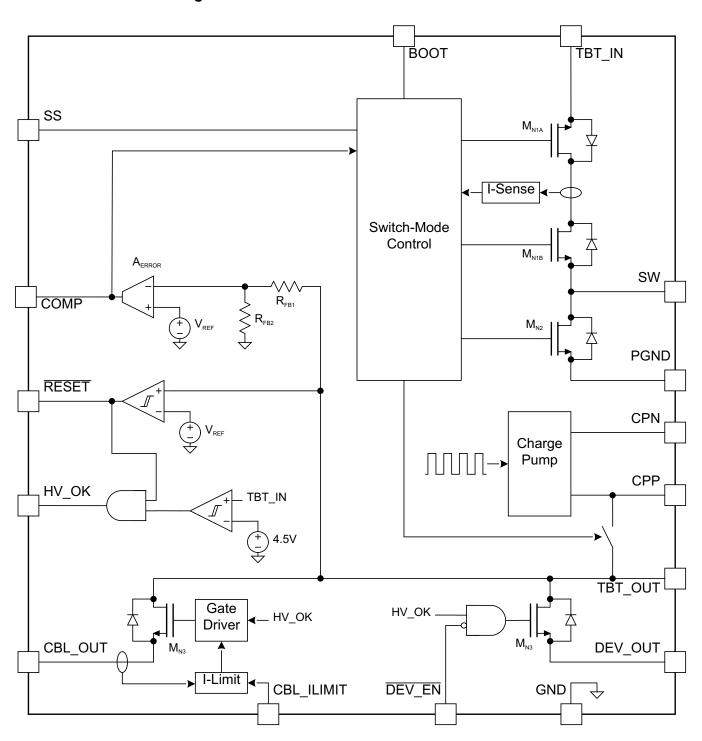
The TPS65980 is a switching regulator designed for Thunderbolt™ and Thunderbolt™ 2 bus-powered systems. The TPS65980 receives power from a Thunderbolt™ host in the range of 2.5 V to 15.75 V and produces three separate 3.3 V outputs. TBT\_OUT is the main output from the regulator. This output is generated from a switched-cap charge pump when the input is in the low-voltage range. The output is generated from a switching buck converter when the input voltage is in the high-voltage range. The TBT\_OUT output powers the local Thunderbolt™ controller and any additional Thunderbolt™ circuitry. Once in the input has settled in the high-voltage range, the other two outputs can be powered from the TBT\_OUT output. When the TBT\_OUT is supplying 3.3 V, the RESET output asserts high. When the TBT\_OUT voltage is below the valid output range, RESET asserts low. When TBT\_IN is in the high-voltage input range and RESET is asserting high (valid output), HV\_OK will assert high indicating that high-voltage has been received.

The CBL\_OUT output supplies power back to the Thunderbolt™ cable for powering the active cable circuitry. This output is connected to the TBT OUT output with a FET switch and is current limited.

The CBL\_ILIMIT logic input pin sets the current limit level. The DEV\_OUT output provides power to all other circuitry in the system. This output is not current limited and is enabled/disabled by the DEV\_EN logic input.



# 8.2 Functional Block Diagram





#### 8.3 Feature Description

#### 8.3.1 2.5-V to 15.75-V Input

The TPS65980 is powered from a Thunderbolt<sup>TM</sup> Bus. This is typically an input to a port from Thunderbolt<sup>TM</sup> cable. This input will start at 3.3 V (2.5 V  $\leq$  V<sub>TBT\_IN</sub>  $\leq$  3.4 V) until a link is established between a host and the peripheral device containing the TPS65980. Once the link is established, the voltage at the input can transition to a higher operating voltage (10 V  $\leq$  V<sub>TBT\_IN</sub>  $\leq$  15.75 V).

#### 8.3.2 3.3-V Outputs

The TPS65980 has three separate 3.3 V outputs. One output, TBT\_OUT, is the output from the buck/boost and the other outputs, CBL\_OUT and DEV\_OUT, are outputs that through load switches from TBT\_OUT.

The TBT\_OUT supply provides power to the local peripheral Thunderbolt™ controller and support circuitry. The CBL\_OUT supply provides power back to the Thunderbolt™ cable and has adjustable current limit. The DEV\_OUT supply provides power to all other circuitry in the device to perform its designed function.

#### 8.3.3 Thermal Shutdown

The TPS65980 as a thermal shutdown feature preventing the device from over heating during current limiting situations. The thermal shutdown occurs at a 135°C junction temperature typically. A 10°C hysteresis occurs before the thermal shutdown is cleared.

#### 8.3.4 Cable Power Out Current Limit

The CBL\_OUT output is current limited internally. The current limit has two values which are set by the CBL\_ILIMIT logic input. When CBL\_ILIMIT = 0, the current limit will bet set to 1.1 A typically. When CBL\_ILIMIT = 1, the current limit will be set to 2.2 A typically.

#### 8.4 Device Functional Modes

#### 8.4.1 Operation with 2.5 V $\leq$ V<sub>TBT IN</sub> $\leq$ 3.4 V

The TPS65980 has two normal operating regions. The first region is when  $2.5 \text{ V} \leq \text{V}_{\text{TBT\_IN}} \leq 3.4 \text{ V}$ . This is the normal power-up state and is termed the low-voltage state. When the input transitions to this range, the input slew rate must meet the SR02L limits. In this voltage range, the TPS65980 operates with a charge pump to generate the nominally 3.3 V output. When the input voltage moves to the higher end of this range, the buck converter takes over to produce the 3.3 V.

## 8.4.2 Operation with 10 V ≤ V<sub>TBT IN</sub> ≤ 15.75 V

In normal operation, the TPS65980 input voltage will transition from the low-voltage range to a high-voltage range where 10 V  $\leq$  V<sub>TBT\_IN</sub>  $\leq$  15.75 V. This is the high-voltage state and is the state where the TPS65980 will operate most of the time. In this state, the device operates as a buck converter providing a nominally 3.3 V output.



## 9 Application and Implementation

#### 9.1 Application Information

The TPS65980 DC/DC switching regulator that receives power from a Thunderbolt<sup>™</sup> or Thunderbolt<sup>™</sup> 2 power bus ranging from 2.5 V to 15.75 V and generates three separate 3.3-V supply outputs.

## 9.2 Typical Application

#### 9.2.1 Single-Port Bus-Powered Thunderbolt™ Device

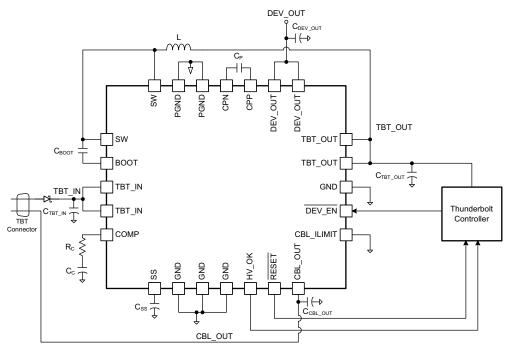


Figure 13. Typical Application (Single-Port Bus-Powered Thunderbolt™ Device)

## 9.2.1.1 Design Requirements

**Table 1. Recommended Component Values** 

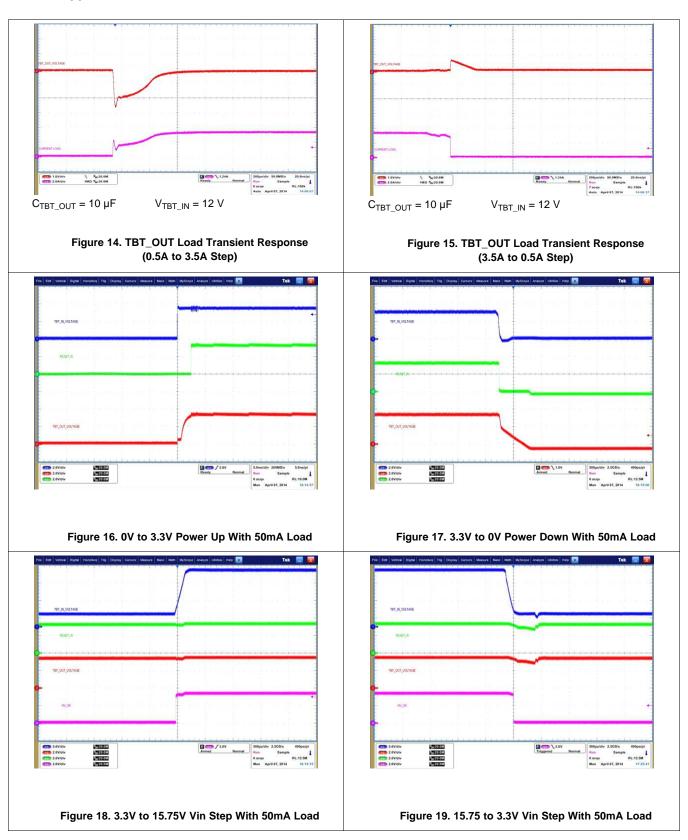
COMPONENT	DESCRIPTION	MIN	TYP	MAX	UNIT
C <sub>IN</sub>	TBT_IN Input Capacitance	17.6	22	52	μF
C <sub>BOOT</sub>	Converter Bootstrap Capacitance	8	10	12	nF
C <sub>CP</sub>	Charge Pump Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
C <sub>SS</sub>	Soft Start Capacitance	8	10	12	nF
C <sub>TBT</sub>	TBT_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	16	20	24	μF
C <sub>CBL</sub>	CBL_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
C <sub>DEV</sub>	DEV_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
C <sub>C</sub>	Compensation Capacitance	8	10	12	nF
R <sub>C</sub>	Compensation Resistance	8	10	12	kΩ
L	Inductor SRR1280 (ESR ≤ 20 mΩ)	8	10	12	μH

#### 9.2.1.2 Detailed Design Procedure

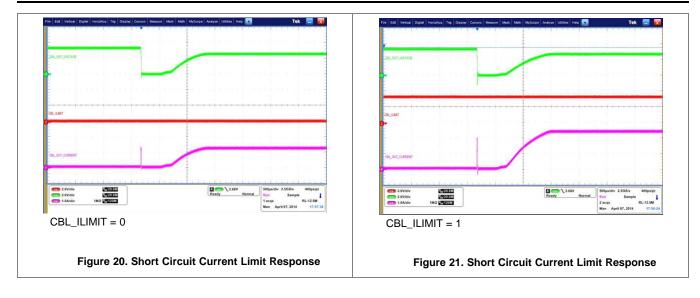
The TPS65980 should use the recommended component values in Table 1. The device is designed to fit the needs of a Thunderbolt<sup>TM</sup> bus powered peripheral and the recommended component values are chosen to satisfy those conditions. The input capacitance  $C_{IN}$  can be as high as 52  $\mu$ F, but this maximum capacitance must include all capacitances seen at the input to the Thunderbolt<sup>TM</sup> port.

# TEXAS INSTRUMENTS

#### 9.2.1.3 Application Performance Plots









#### 9.2.2 Dual-Port Bus-Powered Thunderbolt™ Device

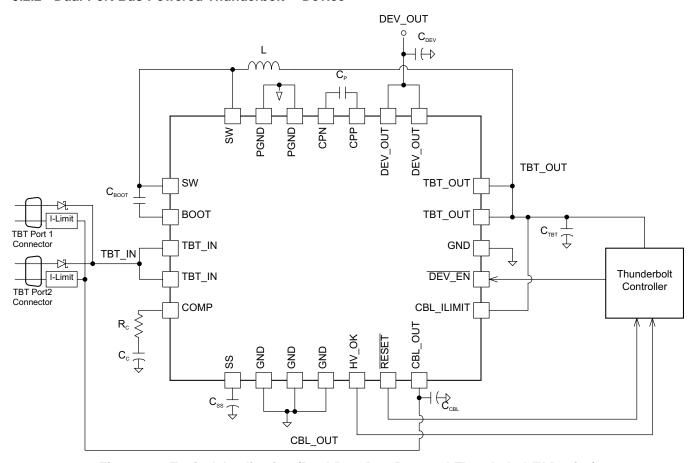


Figure 22. Typical Application (Dual-Port Bus-Powered Thunderbolt™ Device)

#### 9.2.2.1 Design Requirements

In a dual-port application, the TBT\_IN input voltage will be selected from either port. A simple diode-or function will produce TBT\_IN from the higher of the two inputs. The diode-or selection will allow the high-voltage supply to be at TBT\_IN.

In a dual port system, the TPS65980 must provide cable power to both ports. In this case, a second current limiting device (TPS22920) connected between TBT\_OUT and the port is recommended. The CBL\_OUT pin can also supply current to both ports. In this case, tying CBL\_ILIMIT to TBT\_OUT will double the amount of current that can be supplied before current limiting. When using this method, the voltage drop to the CBL\_OUT pin will increase and care must be taken that other systems resistance do not cause the cable voltage to drop below the allowed pin voltage specified in the Thunderbolt™ specification. To avoid issues with voltage drop in the system, it is recommended that the second port be powered from TBT\_OUT as shown in Figure 22. This relieves the voltage drop due to extra current through the CBL\_OUT load switch.



## **Table 2. Recommended Component Values**

COMPONENT	DESCRIPTION	MIN	TYP	MAX	UNIT
C <sub>IN</sub>	TBT_IN Input Capacitance	17.6	22	52	μF
C <sub>BOOT</sub>	Converter Bootstrap Capacitance	8	10	12	nF
$C_{CP}$	Charge Pump Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
$C_{SS}$	Soft Start Capacitance	8	10	12	nF
C <sub>TBT</sub>	TBT_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	16	20	24	μF
C <sub>CBL</sub>	CBL_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
$C_{DEV}$	DEV_OUT Output Capacitance (ceramic with ESR ≤ 10 mΩ)	0.8	1	1.2	μF
$C_{C}$	Compensation Capacitance	8	10	12	nF
R <sub>C</sub>	Compensation Resistance	8	10	12	kΩ
L	Inductor SRR1280 (ESR ≤ 20 mΩ)	8	10	12	μH

## 9.2.2.2 Detailed Design Procedure

Refer to Detailed Design Procedure in the Single-Port Bus-Powered Thunderbolt™ Device section.

## 9.2.2.3 Application Performance Plots

Refer to Application Performance Plots in the Single-Port Bus-Powered Thunderbolt™ Device section.



## 10 Power Supply Recommendations

The TPS65980 is designed to operate from a Thunderbolt<sup>™</sup> bus. The input will range from 2.5 V to 15.75 V. The input should be placed as near to the port connector as possible.

## 11 Layout

## 11.1 Layout Guidelines

Proper placement and routing will maximize the performance of the TPS65980. Follow Figure 23 for optimized layout and routing (hashed planes indicate bottom layer).

## 11.2 Layout Example

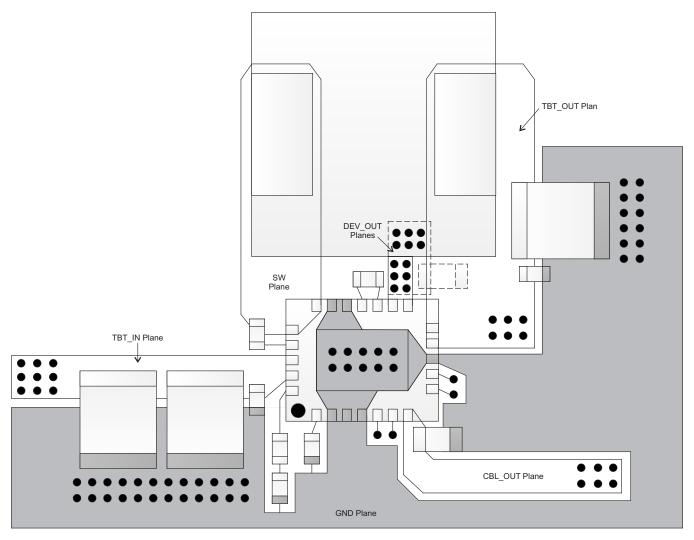


Figure 23. Top View Board Layout



#### Layout Example (continued)

For TBT\_IN, the input capacitors must be placed close to the device with an inductance less than 1nH from input capacitors to the TBT\_IN pins. Layout tools and calculators are available to approximate the inductance. The input capacitors must have their GND side area via stitched to the GND plane. The GND side of the input cap should also share the same polygon as the PGND/PowerPad on the top layer. PowerPad should be connected to the GND plane through multiple vias.

Inductor placement should be above the TPS65980, slightly to the left of the device. The SW pins to the inductor must be connected though a plane as shown in Figure 23. The TBT\_OUT pins also have to be connected to the other side of the inductor with a plane. This plane should be wide to overlap the output capacitors. The GND side of the output capacitors should be stitched to the GND plane.

The CBL\_OUT output capacitor should be placed close to the device on the top layer with an inductance less than 1 nH from the capacitor to the CBL\_OUT pin. The DEV\_OUT capacitor is best placed on the bottom side of the board with two planes (top and bottom) connect through a set of vias. The number of vias placed should be able to carry at least 3 A (DEV\_OUT = 2.5 A max) for margin and inductance path less than 1nH from DEV\_OUT pins to capacitor. When routing DEV\_OUT to an internal power plane, follow Figure 24 for via paths.

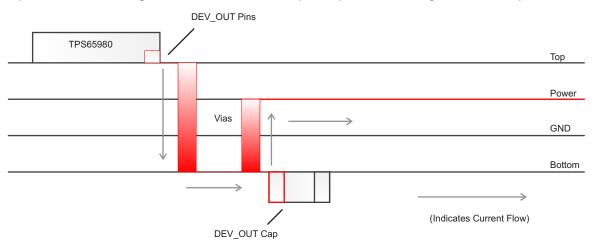


Figure 24. DEV\_OUT Recommended Routing

The charge pump capacitor must be placed on the top layer close to the CPP and CPN pins. The inductance paths from capacitor to the pins must be less than 1 nH. SS and Compensation components should be placed on the top layer close to the device.



## 12 器件和文档支持

## 12.1 Trademarks

Thunderbolt is a trademark of Intel Corporation.

## 12.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## 12.3 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms and definitions.

## 13 机械封装和可订购信息

以下页中包括机械封装和可订购信息。 这些信息是针对指定器件可提供的最新数据。 这些数据会在无通知且不对 本文档进行修订的情况下发生改变。 欲获得该数据表的浏览器版本,请查阅左侧的导航栏。

www.ti.com 23-May-2025

#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking
	(1)	(2)			(3)	(4)	(5)		(6)
TPS65980RHFR	Active	Production	VQFN (RHF)   24	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TPS 65980
TPS65980RHFR.A	Active	Production	VQFN (RHF)   24	3000   LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TPS 65980
TPS65980RHFT	Active	Production	VQFN (RHF)   24	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TPS 65980
TPS65980RHFT.A	Active	Production	VQFN (RHF)   24	250   SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TPS 65980

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



# **PACKAGE OPTION ADDENDUM**

www.ti.com 23-May-2025

# **PACKAGE MATERIALS INFORMATION**

www.ti.com 20-Apr-2023

## TAPE AND REEL INFORMATION





	· · · · · · · · · · · · · · · · · · ·
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS65980RHFR	VQFN	RHF	24	3000	330.0	12.4	4.3	5.3	1.3	8.0	12.0	Q1
TPS65980RHFT	VQFN	RHF	24	250	180.0	12.4	4.3	5.3	1.3	8.0	12.0	Q1

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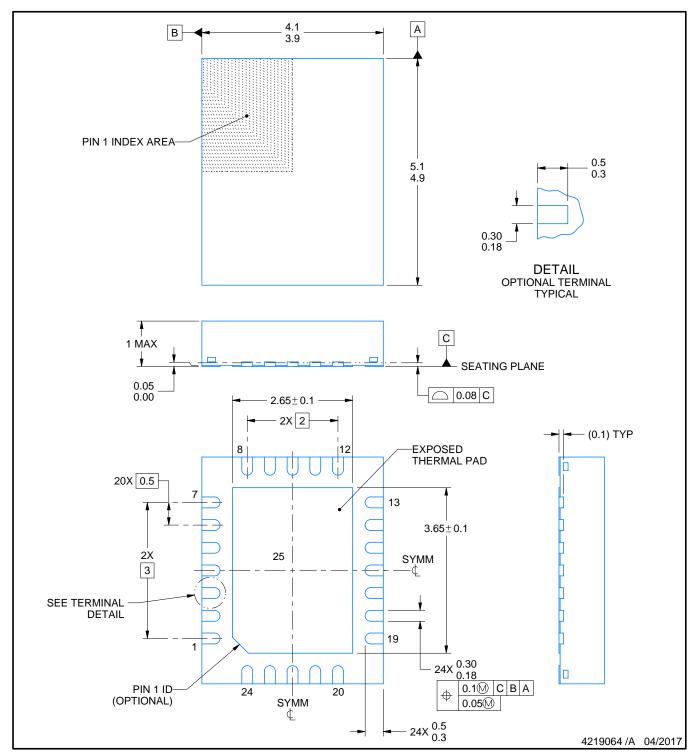


#### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS65980RHFR	VQFN	RHF	24	3000	346.0	346.0	33.0
TPS65980RHFT	VQFN	RHF	24	250	210.0	185.0	35.0



PLASTIC QUAD FLATPACK - NO LEAD



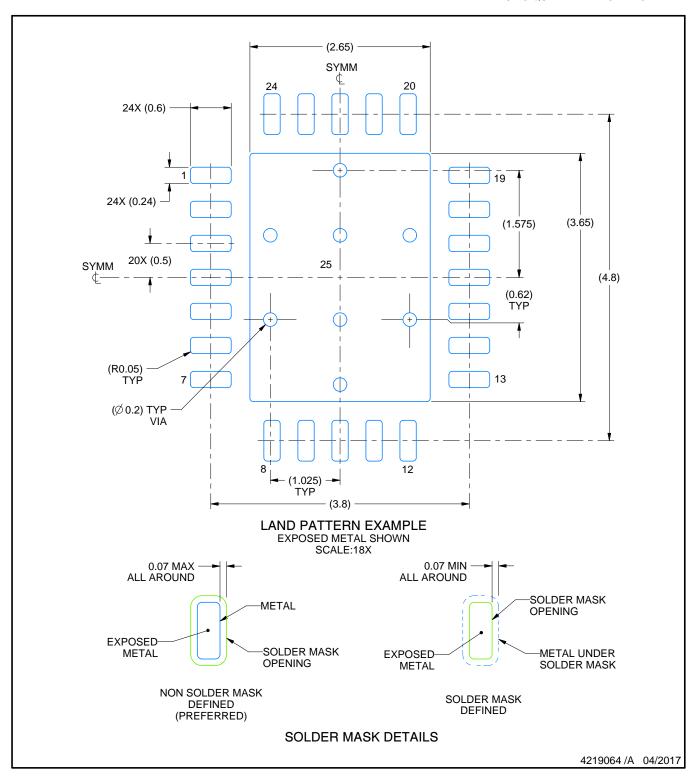
#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

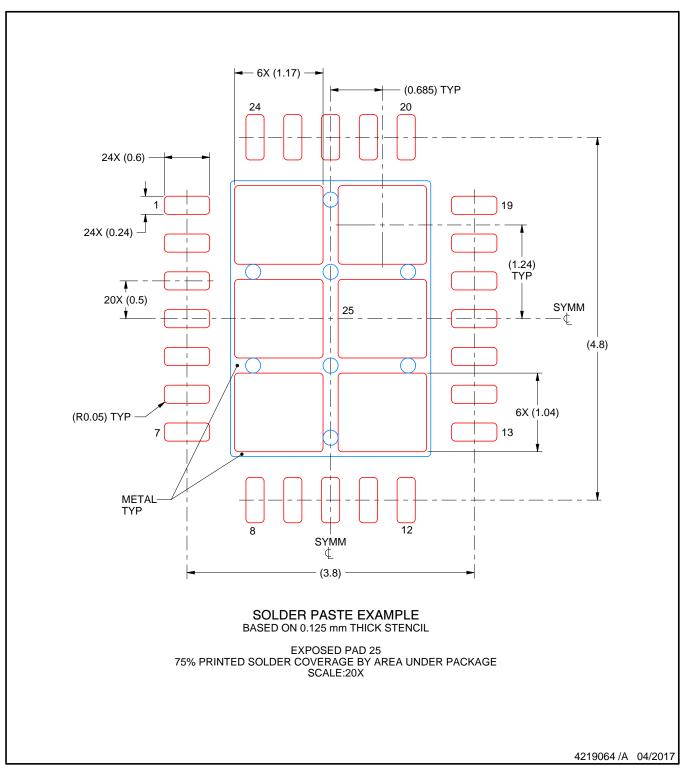


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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