











TPD6S300

ZHCSG02C - SEPTEMBER 2016-REVISED JANUARY 2017

TPD6S300 USB Type-C™端口保护器: V_{BUS} 短路过压和 IEC ESD 保护

特性

- 4 通道 V_{BUS} 短路过压保护(CC1、CC2、SBU1、 SBU2): 24 V_{DC} 容差
- 6 通道 IEC 61000-4-2 ESD 保护(CC1、CC2、 SBU1、SBU2、DP、DM)
- CC1、CC2 过压保护 FET 600mA, 能够通过 V_{CONN} 电源
- 集成 CC 无电电池电阻器,可用于处理移动设备中 的无电电池用例
- 3mm × 3mm WQFN 封装

2 应用

- 笔记本电脑
- 平板电脑
- 智能手机
- 监视器和电视
- 扩展坞

3 说明

TPD6S300 是一种单芯片 USB Type-C 端口保护解决 方案,可提供 20V V_{BUS} 短路过压和 IEC ESD 保护。

自从 USB Type-C 连接器发布以来,已经发布了很多 不符合 USB Type-C 规格的 USB Type-C 的产品和配 件。其中的一个示例就是仅在 V_{BUS} 线路上布设 20V 电压的 USB Type-C 电力输送适配器。USB Type-C 的另一个问题是,由于此小型连接器中的各引脚极为靠 近,因此连接器的机械扭转和滑动可能使引脚短路。这 可能导致 20V V_{BUS} 与 CC 和 SBU 引脚短路。此外, 由于 Type-C 连接器中的各引脚极为靠近,所以存在碎 屑和水汽导致 20V V_{BUS} 引脚与 CC 和 SBU 引脚短路 的严重问题。

这些非理想的设备和机械事件使得 CC 和 SBU 引脚必 须具有 20V 容差,即使它们仅在 5V 或更低电压下工 作。TPD6S300 通过在 CC 和 SBU 引脚上提供过压保 护,可以使 CC 和 SBU 引脚实现 20V 容差,同时不 会干扰正常工作。该器件将高压 FET 串联放置在 SBU 和 CC 线路上。当在这些线路上检测到高于 OVP 阈值 的电压时,高压开关被打开,并且将系统的其余部分与 连接器上存在的高压状态隔离。

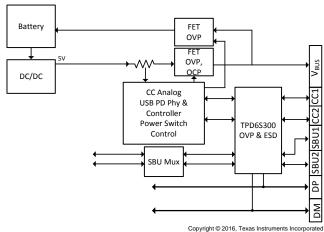
最后,大多数系统都需要为其外部引脚应用 IEC 61000-4-2 系统级 ESD 保护。TPD6S300 为 CC1、 CC2、SBU1、SBU2、DP、DM 引脚集成 IEC 61000-4-2 ESD 保护,并且无需在连接器上通过外部放置高 压 TVS 二极管。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
TPD6S300	WQFN (20)	3.00mm x 3.00mm

(1) 如需了解所有可用封装,请参阅产品说明书末尾的可订购产品 附录。

应用图表





		目录			
1 2 3 4 5 6 7	特性		9 10 11	8.3 Feature Description	
8	7.3 ESD Ratings—IEC Specification 5 7.4 Recommended Operating Conditions 5 7.5 Thermal Information 6 7.6 Electrical Characteristics 6 7.7 Timing Requirements 8 7.8 Typical Characteristics 10 Detailed Description 15 8.1 Overview 15 8.2 Functional Block Diagram 16		12	器件和文档支持 12.1 文档支持 12.2 接收文档更新通知 12.3 社区资源 12.4 商标 12.5 静电放电警告 12.6 Glossary 机械、封装和可订购信息 13.1	
-	多订历史记录 ges from Revision B (November 2016) to Revision				Page

Changes from Revision B (November 2016) to Revision C	Page
• Changed units from µs to ns for OVP_RESPONSE in the <i>Timing Requirements</i> table	8
Changes from Revision A (September 2016) to Revision B	Page
• 首次公开发布产品说明书	1

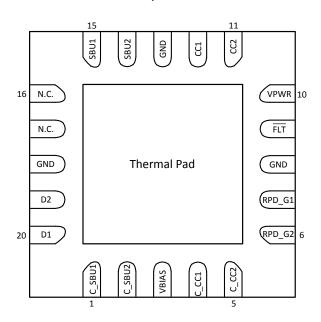


5 Device Comparison Table

Part Number	Over Voltage Protected Channels	IEC 61000-4-2 ESD Protected Channels
TPD6S300	4-Ch (CC1, CC2, SBU1, SBU2)	6-Ch (CC1, CC2, SBU1, SBU2, DP, DM)
TPD8S300	4-Ch (CC1, CC2, SBU1, SBU2)	8-Ch (CC1, CC2, SBU1, SBU2, DP_T, DM_T, DP_B, DM_B)

6 Pin Configuration and Functions

RUK Package 20 Pin WQFN Top View



Pin Functions

	PIN	TVDE	DESCRIPTION	
NO.	NAME	TYPE	DESCRIPTION	
1	C_SBU1	I/O	Connector side of the SBU1 OVP FET. Connect to either SBU pin of the USB Type-C connector	
2	C_SBU2	I/O	Connector side of the SBU2 OVP FET. Connect to either SBU pin of the USB Type-C connector	
3	VBIAS	Power	Pin for ESD support capacitor. Place a 0.1-µF capacitor on this pin to ground	
4	C_CC1	I/O	Connector side of the CC1 OVP FET. Connect to either CC pin of the USB Type-C connector	
5	C_CC2	I/O	Connector side of the CC2 OVP FET. Connect to either CC pin of the USB Type-C connector	
6	RPD_G2	I/O	Short to C_CC2 if dead battery resistors are needed. If dead battery resistors are not needed, short pin to GND	
7	RPD_G1	I/O	Short to C_CC1 if dead battery resistors are needed. If dead battery resistors are not needed, short pin to GND	
8	GND	GND	Ground	
9	FLT	0	Open drain for fault reporting	
10	V _{PWR}	Power	2.7-V-3.6-V power supply	



Pin Functions (continued)

	PIN	TVDE	DESCRIPTION	
NO.	NAME	TYPE	DESCRIPTION	
11	CC2	I/O	System side of the CC2 OVP FET. Connect to either CC pin of the CC/PD controller	
12	CC1	I/O	System side of the CC1 OVP FET. Connect to either CC pin of the CC/PD controller	
13	GND	GND	Ground	
14	SBU2	I/O	System side of the SBU2 OVP FET. Connect to either SBU pin of the SBU MUX	
15	SBU1	I/O	System side of the SBU1 OVP FET. Connect to either SBU pin of the SBU MUX	
16	N.C.	I/O	Unused pin. Connect to Ground	
17	N.C.	I/O	Unused pin. Connect to Ground	
18	GND	GND	Ground	
19	D2	I/O	USB2.0 IEC ESD protection. Connect to any of the USB2.0 pins of the USB Type-C connector	
20	D1	I/O	USB2.0 IEC ESD protection. Connect to any of the USB2.0 pins of the USB Type-C connector	
_	Thermal Pad	GND	Internally connected to GND. Used as a heatsink. Connect to the PCB GND plane	



7 Specifications

7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)

			MIN	MAX	UNIT
W	Input voltage	V_{PWR}	-0.3	5	V
VI	Input voltage	RPD_G1, RPD_G2	-0.3	24	V
V	Output valtage	FLT	-0.3	6	V
Vo	Output voltage	VBIAS	-0.3	24	V
		D1, D2	-0.3	6	V
V_{IO}	I/O voltage	CC1, CC2, SBU1, SBU2	-0.3	6	V
		C_CC1, C_CC2, C_SBU1, C_SBU2	-0.3	24	V
T _A	Operating free air temperature		-40	85	°C
T _{stg}	Storage temperature		-65	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings—JEDEC Specification

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Pins listed as ±2000 V may actually have higher performance.

7.3 ESD Ratings—IEC Specification

				VALUE	UNIT
		IEC 64000 4 2 C CC4 C CC2 D4 D2	Contact discharge	±8000	
V _(ESD) Electrostatic discharge ⁽¹⁾	F I = 1 = 1 = 1 = 1 = 1 = 1 = 1 = 1 = 1 =	IEC 61000-4-2, C_CC1, C_CC2, D1, D2	Air-gap discharge	±15000	
	JEO 04000 4 0 0 0 0 U 4 0 0 0 U 4	Contact discharge	±6000	V	
	IEC 61000-4-2, C	IEC 61000-4-2, C_SBU1, C_SBU2	Air-gap discharge	±15000	

⁽¹⁾ Tested on the TPD6S300 EVM connected to the TPS65982 EVM.

7.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
	land to alta an	V _{PWR}	2.7	3.3	4.5	V
VI	Input voltage	RPD_G1, RPD_G2	0		5.5	V
Vo	Output voltage	FLT pull-up resistor power rail	2.7		5.5	V
		D1, D2	-0.3		5.5	V
V_{IO}	I/O voltage	CC1, CC2, C_CC1, C_CC2	0		5.5	V
		SBU1, SBU2, C_SBU1, C_SBU2	0		4.3	V
I _{VCONN}	V _{CONN} current	Current flowing into CC1/2 and flowing out of C_CC1/2, VCCx − VC_CCx ≤ 250 mV			600	mA
I _{VCONN}	V _{CONN} current	Current flowing into CC1/2 and flowing out of C_CC1/2, $T_J \le 105$ °C			1.25	Α

⁽²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Pins listed as ±500 V may actually have higher performance.



Recommended Operating Conditions (continued)

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
	FLT pull-up resistance	1.7		300	kΩ
External components (1)	VBIAS capacitance ⁽²⁾		0.1		μF
	V _{PWR} capacitance	0.3	1		μF

⁽¹⁾ For recommended values for capacitors and resistors, the typical values assume a component placed on the board near the pin. Minimum and maximum values listed are inclusive of manufacturing tolerances, voltage derating, board capacitance, and temperature variation. The effective value presented must be within the minimum and maximums listed in the table.

7.5 Thermal Information

		TPD6S300	
	THERMAL METRIC ⁽¹⁾	RUK (WQFN)	UNIT
		20 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	45.2	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	48.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	17.1	°C/W
ΨЈТ	Junction-to-top characterization parameter	0.6	°C/W
ΨЈВ	Junction-to-board characterization parameter	17.1	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	3.7	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

7.6 Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN TYP		MAX	UNIT
CC OVP SWI	TCHES					
В	On resistance of CC OVP FETs, T _J ≤ 85°C	CCx = 5.5 V		278	392	mΩ
R _{ON}	On resistance of CC OVP FETs, $T_J \le 105$ °C	CCx = 5.5 V		278	415	mΩ
R _{ON(FLAT)}	On resistance flatness	Sweep CCx voltage between 0 V and 1.2 V			5	mΩ
C _{ON_CC}	Equivalent on capacitance	Capacitance from C_CCx or CCx to GND when device is powered. VC_CCx/VCCx = 0 V to 1.2 V, f = 400 kHz	60	74	120	pF
R_D	Dead battery pull-down resistance (only present when device is unpowered). Effective resistance of R _D and FET in series	V_C_CCx = 2.6 V	4.1	5.1	6.1	kΩ
VTH_DB	Threshold voltage of the pulldown FET in series with RD during dead battery	I_CC = 80 μA	0.5	0.9	1.2	V
V _{OVPCC}	OVP threshold on CC pins	Place 5.5 V on C_CCx. Step up C_CCx until the FLT pin is asserted	5.75	6	6.2	V
V _{OVPCC_HYS}	Hysteresis on CC OVP	Place 6.5 V on C_CCx. Step down the voltage on C_CCx until the FLT pin is deasserted. Measure difference between rising and falling OVP threshold for C_CCx	50			mV
BW _{ON}	On bandwidth single ended (-3 dB)	Measure the -3 -dB bandwidth from C_CCx to CCx. Single ended measurement, 50- Ω system. Vcm = 0.1 V to 1.2 V	100		MHz	

⁽²⁾ The VBIAS pin requires a minimum 35-V_{DC} rated capacitor. A 50-V_{DC} rated capacitor is recommended to reduce capacitance derating. See the *VBIAS Capacitor Selection* section for more information on selecting the VBIAS capacitor.



Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{STBUS_CC}	Short-to-VBUS tolerance on the CC pins	Hot-Plug C_CCx with a 1 meter USB Type C Cable, place a 30-Ω load on CCx			24	V
V _{STBUS_CC_CL}	Short-to-VBUS system-side clamping voltage on the CC pins (CCx)	Hot-Plug C_CCx with a 1 meter USB Type C Cable. Hot-Plug voltage C_CCx = 24 V. VPWR = 3.3 V. Place a 30-Ω load on CCx		8		V
SBU OVP SW	ITCHES					
R _{ON}	On resistance of SBU OVP FETs	SBUx = 3.6 V. -40° C $\leq T_{J} \leq +85^{\circ}$ C		4	6.5	Ω
R _{ON(FLAT)}	On resistance flatness	Sweep SBUx voltage between 0 V and 3.6 V. -40° C \leq T _J \leq +85 $^{\circ}$ C		0.7	1.5	Ω
C_{ON_SBU}	Equivalent on capacitance	Capacitance from SBUx or C_SBUx to GND when device is powered. Measure at VC_SBUx/VSBUx = 0.3 V to 3.6 V		6		pF
V _{OVPSBU}	OVP threshold on SBU pins	Place 3.6 V on C_SBUx. Step up C_SBUx until the FLT pin is asserted	4.35	4.5	4.7	V
V _{OVPSBU_} HYS	Hysteresis on SBU OVP	Place 5 V on C_CCx. Step down the voltage on C_CCx until the FLT pin is deasserted. Measure difference between rising and falling OVP threshold for C_SBUx		50		mV
BW _{ON}	On bandwidth single ended (-3 dB)	Measure the -3-dB bandwidth from C_SBUx to SBUx. Single ended measurement, 50-Ω system. Vcm = 0.1 V to 3.6 V		1000		MHz
X _{TALK}	Crosstalk	Measure crosstalk at f = 1 MHz from SBU1 to C_SBU2 or SBU2 to C_SBU1. Vcm1 = 3.6 V, Vcm2 = 0.3 V. Be sure to terminate open sides to 50 Ω		-80		dB
V _{STBUS_SBU}	Short-to-VBUS tolerance on the SBU pins	Hot-Plug C_SBUx with a 1 meter USB Type C Cable. Put a 100-nF capacitor in series with a 40-Ω resistor to GND on SBUx			24	V
V _{STBUS} SBU_C LAMP	Short-to-VBUS system-side clamping voltage on the SBU pins (SBUx)	Hot-Plug C_SBUx with a 1 meter USB Type C Cable. Hot-Plug voltage C_SBUx = 24 V. VPWR = 3.3 V. Put a 150-nF capacitor in series with a $40-\Omega$ resistor to GND on SBUx		8		V
POWER SUP	PLY and LEAKAGE CURRENTS					
V _{PWR_UVLO}	V _{PWR} under voltage lockout	Place 1 V on VPWR and raise voltage until SBU or CC FETs turnon	2.1	2.3	2.5	V
V _{PWR_UVLO_} H YS	V _{PWR} UVLO hysteresis	Place 3 V on VPWR and lower voltage until SBU or CC FETs turnoff; measure difference between rising and falling UVLO to calculate hysteresis	100	150	200	mV
I _{VPWR}	V _{PWR} supply current	VPWR = 3.3 V (typical), VPWR = 4.5 V (maximum). -40° C \leq T _J \leq +85 $^{\circ}$ C.		90	135	μΑ
I _{CC_LEAK}	Leakage current for CC pins when device is powered	VPWR = 3.3 V, VC_CCx = 3.6 V, CCx pins are floating, measure leakage into C_CCx pins. Result must be same if CCx side is biased and C_CCx is left floating			5	μΑ



Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
I _{SBU_LEAK}	Leakage current for SBU pins when device is powered	VPWR = 3.3 V, VC_SBUx = 3.6 V, SBUx pins are floating, measure leakge into C_SBUx pins. Result must be same if SBUx side is biased and C_SBUx is left floating. $-40^{\circ}\text{C} \leq \text{T}_{\text{J}} \leq +85^{\circ}\text{C}$			3	μА	
I _{C_CC_LEAK_OV}	Leakage current for CC pins when device is in OVP	VPWR = 0 V or 3.3 V, VC_CCx = 24 V, CCx pins are set to 0 V, measure leakage into C_CCx pins			1200	μΑ	
I _{C_SBU_LEAK_O} VP	Leakage current for SBU pins when device is in OVP	VPWR = 0 V or 3.3 V, VC_SBUx = 24 V, SBUx pins are set to 0 V, measure leakage into C_SBUx pins			400	μΑ	
I _{CC_LEAK_OVP}	Leakage current for CC pins when device is in OVP	VPWR = 0 V or 3.3 V, VC_CCx = 24 V, CCx pins are set to 0 V, measure leakage out of CCx pins			30	μΑ	
I _{SBU_LEAK_OVP}	Leakage current for SBU pins when device is in OVP	VPWR = 0 V or 3.3 V, VC_SBUx = 24 V, SBUx pins are set to 0 V, measure leakage out of SBUx pins	-1		1	μА	
I _{Dx_LEAK}	Leakage current for Dx pins	V_Dx = 3.6 V, measure leakage into Dx pins			1	μΑ	
FLT PIN					·		
V _{OL}	Low-level output voltage	$IOL = 3$ mA. Measure the voltage at the \overline{FLT} pin			0.4	V	
OVER TEMPE	RATURE PROTECTION						
T _{SD_RISING}	The rising over-temperature protection shutdown threshold		150	175		°C	
T _{SD_FALLING}	The falling over-temperature protection shutdown threshold		130	140		°C	
T _{SD_HYST}	The over-temperature protection shutdown threshold hysteresis			35		°C	
Dx ESD PRO	FECTION						
V _{RWM_POS}	Reverse stand-off voltage from Dx to GND	Dx to GND. I _{DX} ≤ 1 μA			5.5	V	
V _{RWM_NEG}	Reverse stand-off voltage from GND to Dx	GND to Dx			0	V	
V _{BR_POS}	Break-down voltage from Dx to GND	Dx to GND. I _{BR} = 1 mA	7			V	
V _{BR_NEG}	Break-down voltage from GND to Dx	GND to Dx. I _{BR} = 8 mA	0.6			V	
C _{IO}	Dx to GND or GND to Dx	f = 1 MHz, VIO = 2.5 V		1.7		pF	
ΔC_{IO}	Differential capacitance between two Dx pins	f = 1 MHz, VIO = 2.5 V		0.02		pF	
R _{DYN}	Dynamic on-resistance Dx IEC clamps	Dx to GND or GND to Dx		0.4		Ω	

7.7 Timing Requirements

		MINI	NOM	MAY	LINIT			
		MIN	NOM	MAX	UNIT			
POWER-ON and Off T	IMINGS							
t _{ON}	Time from crossing rising VPWR UVLO until CC and SBU OVP FETs are on			3.5	ms			
dV _{PWR_OFF} /dt	Minimum Slew rate allowed to guarantee CC and SBU FETs turnoff during a power off	-0.5			V/µs			
OVER VOLTAGE PROTECTION								
tovp_response_cc	OVP response time on the CC pins. Time from OVP asserted until OVP FETs turnoff	70		ns				

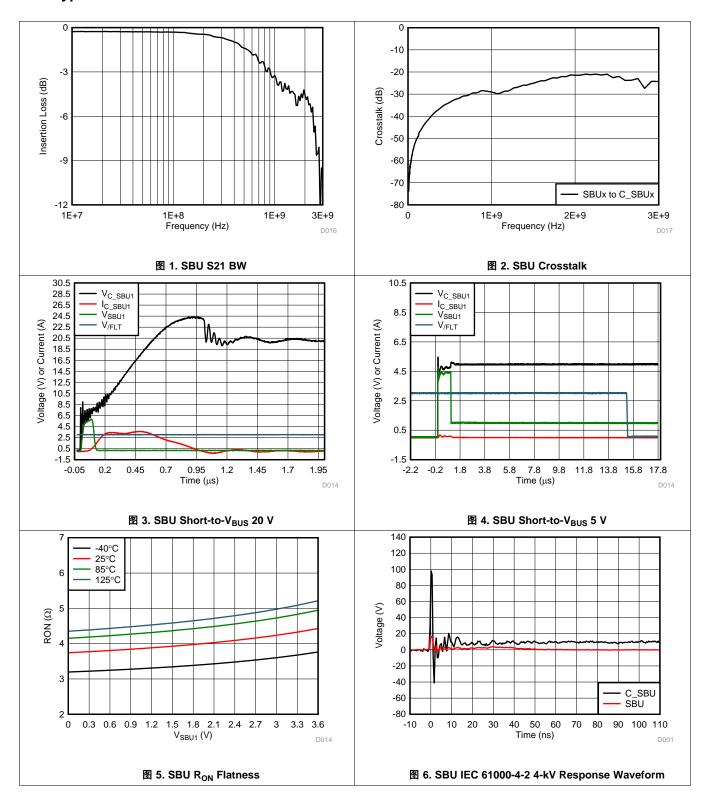


Timing Requirements (continued)

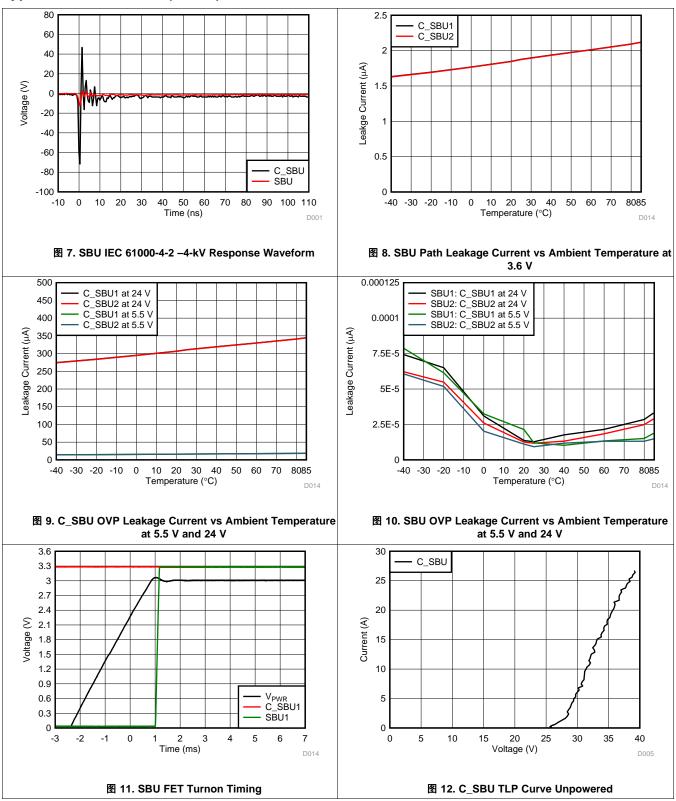
		MIN	NOM	MAX	UNIT
tovp_response_sbu	OVP response time on the SBU pins. Time from OVP asserted until OVP FETs turnoff		80		ns
tovp_recovery_cc_1	OVP recovery time on the CC pins. Once an OVP has occurred, the minimum time duration until the CC FETs turn back on. OVP must be removed for CC FETs to turn back on	21	29	39	ms
tovp_recovery_sbu_1	OVP recovery time on the SBU pins. Once an OVP has occurred, the minimum time duration until the SBU FETs turn back on. OVP must be removed for SBU FETs to turn back on	21	29	39	ms
t _{OVP_RECOVERY_CC_2}	OVP recovery time on the CC pins. Time from OVP Removal until CC FET turns back on, if device has been in OVP > 40 ms		0.5		ms
t _{OVP_RECOVERY_SBU_2}	OVP recovery time on the SBU pins. Time from OVP Removal until SBU FET turns back on, if device has been in OVP > 40 ms		0.5		ms
t _{OVP_FLT_ASSERTION}	Time from OVP asserted to FLT assertion		20		μs
tovp_flt_deassertion	Time from CC FET turnon after an OVP to FLT deassertion		5		ms

TEXAS INSTRUMENTS

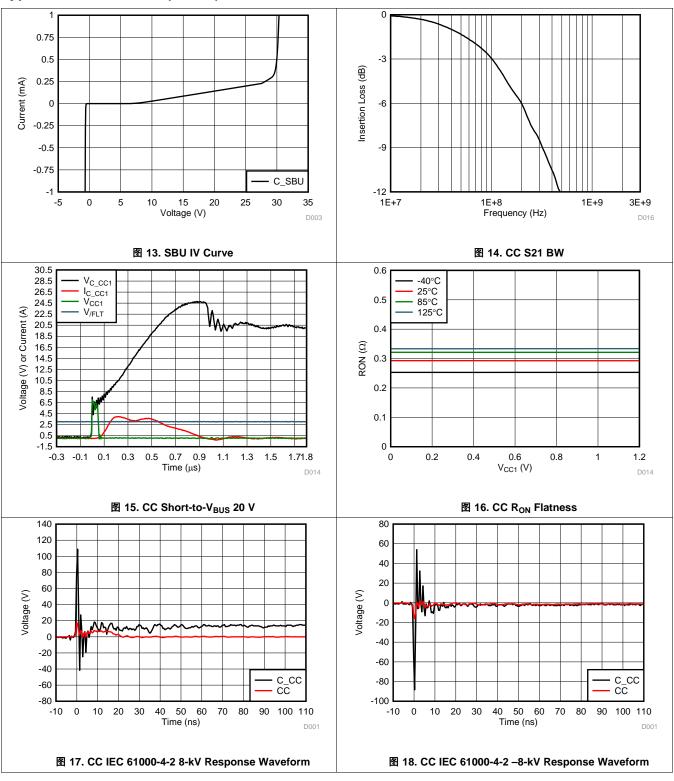
7.8 Typical Characteristics



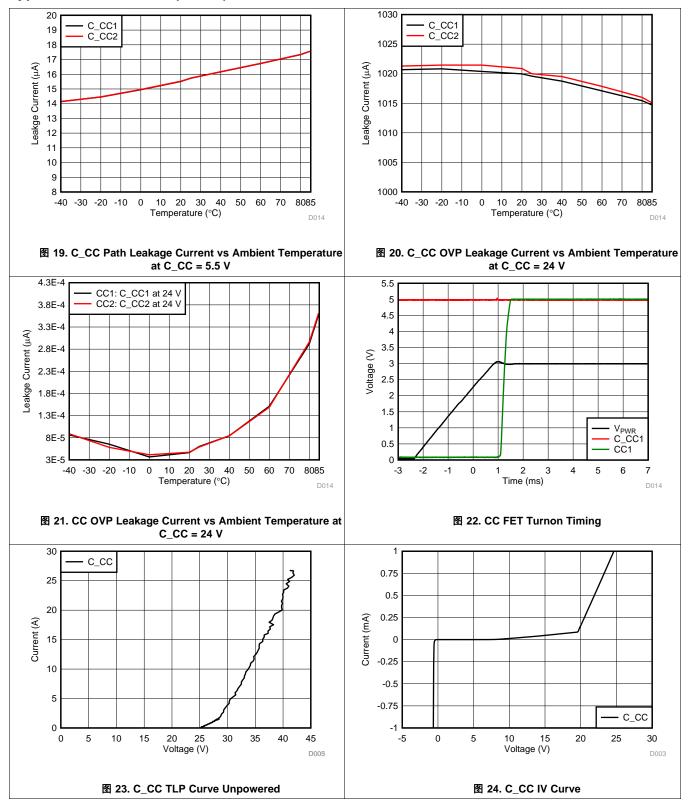




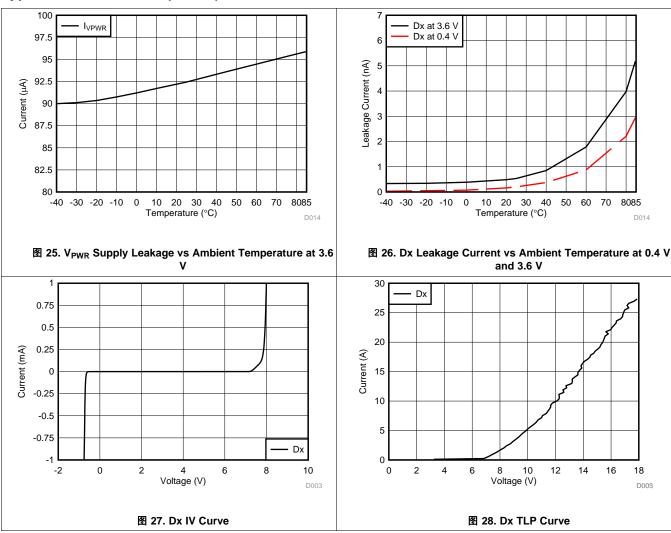
TEXAS INSTRUMENTS







TEXAS INSTRUMENTS





8 Detailed Description

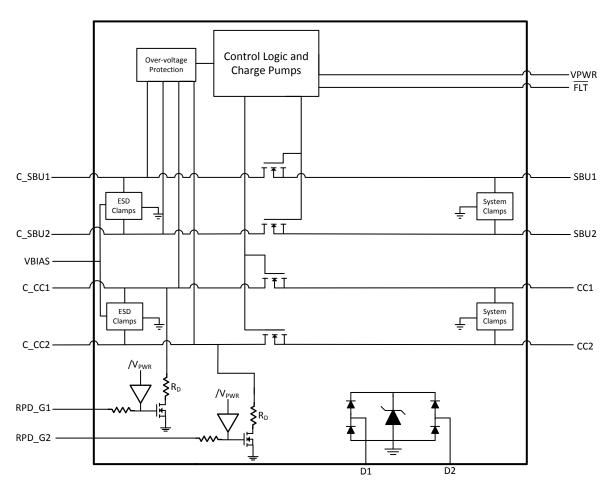
8.1 Overview

The TPD6S300 is a single chip USB Type-C port protection solution that provides 20-V Short-to- V_{BUS} overvoltage and IEC ESD protection. Due to the small pin pitch of the USB Type-C connector and non-compliant USB Type-C cables and accessories, the V_{BUS} pins can get shorted to the CC and SBU pins inside the USB Type-C connector. Because of this short-to- V_{BUS} event, the CC and SBU pins need to be 20-V tolerant, to support protection on the full USB PD voltage range. Even if a device does not support 20-V operation on V_{BUS} , non complaint adaptors can start out with 20-V V_{BUS} condition, making it necessary for any USB Type-C device to support 20 V protection. The TPD6S300 integrates four channels of 20-V Short-to- V_{BUS} overvoltage protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector.

Additionally, IEC 61000-4-2 system level ESD protection is required in order to protect a USB Type-C port from ESD strikes generated by end product users. The TPD6S300 integrates eight channels of IEC61000-4-2 ESD protection for the CC1, CC2, SBU1, SBU2, DP_T (Top side D+), DM_T (Top Side D-), DP_B (Bottom Side D+), and DM_B (Bottom Side D-) pins of the USB Type-C connector. This means IEC ESD protection is provided for all of the low-speed pins on the USB Type-C connector in a single chip in the TPD6S300. Additionally, high-voltage IEC ESD protection that is 22-V DC tolerant is required for the CC and SBU lines in order to simultaneously support IEC ESD and Short-to-V_{BUS} protection; there are not many discrete market solutions that can provide this kind of protection. This high-voltage IEC ESD diode is what the TPD6S300 integrates, specifically designed to guarantee it works in conjunction with the overvoltage protection FETs inside the device. This sort of solution is very hard to generate with discrete components.



8.2 Functional Block Diagram



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8.3 Feature Description

8.3.1 4-Channels of Short-to-V_{BUS} Overvoltage Protection (CC1, CC2, SBU1, SBU2 Pins): 24-V_{DC} Tolerant

The TPD6S300 provides 4-channels of Short-to- V_{BUS} Overvoltage Protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector. The TPD6S300 is able to handle 24- V_{DC} on its C_CC1, C_CC2, C_SBU1, and C_SBU2 pins. This is necessary because according to the USB PD specification, with V_{BUS} set for 20-V operation, the V_{BUS} voltage is allowed to legally swing up to 21 V, and 21.5 V on voltage transitions from a different USB PD V_{BUS} voltage. The TPD6S300 builds in tolerance up to 24- V_{BUS} to provide margin above this 21.5 V specification to be able to support USB PD adaptors that may break the USB PD specification.

When a short-to- V_{BUS} event occurs, ringing happens due to the RLC elements in the hot-plug event. With very low resistance in this RLC circuit, ringing up to twice the settling voltage can appear on the connector. More than 2x ringing can be generated if any capacitor on the line derates in capacitance value during the short-to- V_{BUS} event. This means that more than 44 V could be seen on a USB Type-C pin during a Short-to- V_{BUS} event. The TPD6S300 has built in circuit protection to handle this ringing. The diode clamps used for IEC ESD protection also clamp the ringing voltage during the short-to- V_{BUS} event to limit the peak ringing to around 30 V. Additionally, the overvoltage protection FETs integrated inside the TPD6S300 are 30-V tolerant, therefore being capable of supporting the high-voltage ringing waveform that is experienced during the short-to- V_{BUS} event. The well designed combination of voltage clamps and 30-V tolerant OVP FETs insures the TPD6S300 can handle Short-to- V_{BUS} hot-plug events with hot-plug voltages as high as 24- V_{DC} .



Feature Description (接下页)

The TPD6S300 has an extremely fast turnoff time of 70 ns typical. Furthermore, additional voltage clamps are placed after the OVP FET on the system side (CC1, CC2, SBU1, SBU2) pins of the TPD6S300, to further limit the voltage and current that are exposed to the USB Type-C CC/PD controller during the 70 ns interval while the OVP FET is turning off. The combination of connector side voltage clamps, OVP FETs with extremely fast turnoff time, and system side voltage clamps all work together to insure the level of stress seen on a CC1, CC2, SBU1, or SBU2 pin during a short-to-V_{BUS} event is less than or equal to an HBM event. This is done by design, as any USB Type-C CC/PD controller will have built in HBM ESD protection.

■ 29 is an example of the TPD6S300 successfully protecting the TPS65982, the world's first fully integrated, full-featured USB Type-C and PD controller.

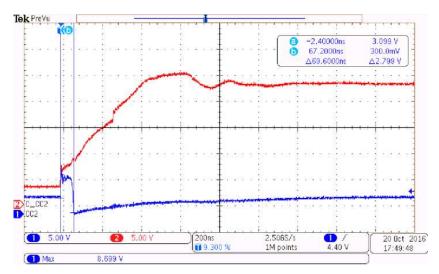


图 29. TPD6S300 Protecting the TPS65982 During a Short-to-V_{BUS} Event

8.3.2 8-Channels of IEC 61000-4-2 ESD Protection (CC1, CC2, SBU1, SBU2, DP_T, DM_T, DP_B, DM_B Pins)

The TPD6S300 integrates 8-Channels of IEC 61000-4-2 system level ESD protection for the CC1, CC2, SBU1, SBU2, DP_T (Top side D+), DM_T (Top Side D-), DP_B (Bottom Side D+), and DM_B (Bottom Side D-) pins. USB Type-C ports on end-products need system level IEC ESD protection in order to provide adequate protection for the ESD events that the connector can be exposed to from end users. The TPD6S300 integrates IEC ESD protection for all of the low-speed pins on the USB Type-C connector in a single chip. Also note, that while the RPD_Gx pins are not individually rated for IEC ESD, when they are shorted to the C_CCx pins, the C_CCx pins provide protection for both the C_CCx pins and the RPD_Gx pins. Additionally, high-voltage IEC ESD protection that is 24-V DC tolerant is required for the CC and SBU lines in order to simultaneously support IEC ESD and Short-to-V_{BUS} protection; there are not many discrete market solutions that can provide this kind of protection. The TPD6S300 integrates this type of high-voltage ESD protection so a system designer can meet both IEC ESD and Short-to-V_{BUS} protection requirements in a single device.

8.3.3 CC1, CC2 Overvoltage Protection FETs 600 mA Capable for Passing VCONN Power

The CC pins on the USB Type-C connector serve many functions; one of the functions is to be a provider of power to active cables. Active cables are required when desiring to pass greater than 3 A of current on the V_{BUS} line or when the USB Type-C port uses the super-speed lines (TX1+, TX2-, RX1+, RX1-, TX2+, TX2-, RX2+, RX2-). When CC is configured to provide power, it is called VCONN. VCONN is a DC voltage source in the range of 3 V-5.5 V. If supporting VCONN, a VCONN provider must be able to provide 1 W of power to a cable; this translates into a current range of 200 mA to 333 mA (depending on your VCONN voltage level). Additionally, if operating in a USB PD alternate mode, greater power levels are allowed on the VCONN line.



Feature Description (接下页)

When a USB Type-C port is configured for VCONN and using the TPD6S300, this VCONN current flows through the OVP FETs of the TPD6S300. Therefore, the TPD6S300 has been designed to handle these currents and have an RON low enough to provide a specification compliant VCONN voltage to the active cable. The TPD6S300 is designed to handle up to 600 mA of DC current to allow for alternate mode support in addition to the standard 1 W required by the USB Type-C specification.

8.3.4 CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices

An important feature of USB Type-C and USB PD is the ability for this connector to serve as the sole power source to mobile devices. With support up to 100 W, the USB Type-C connector supporting USB PD can be used to power a whole new range of mobile devices not previously possible with legacy USB connectors.

When the USB Type-C connector is the sole power supply for a battery powered device, the device must be able to charge from the USB Type-C connector even when its battery is dead. In order for a USB Type-C power adapter to supply power on V_{BUS}, RD pull-down resistors must be exposed on the CC pins. These RD resistors are typically included inside a USB Type-C CC/PD controller. However, when the TPD6S300 is used to protect the USB Type-C port, the OVP FETs inside the device isolate these RD resistors in the CC/PD controller when the mobile device has no power. This is because when the TPD6S300 has no power, the OVP FETs are turned off to guarantee overvoltage protection in a dead battery condition. Therefore, the TPD6S300 integrates high-voltage, dead battery RD pull-down resistors to allow dead battery charging simultaneously with high-voltage OVP protection.

If dead battery support is required, short the RPD_G1 pin to the C_CC1 pin, and short the RPD_G2 pin to the C_CC2 pin. This connects the dead battery resistors to the connector CC pins. When the TPD6S300 is unpowered, and the RP pull-up resistor is connected from a power adaptor, this RP pull-up resistor activates the RD resistor inside the TPD6S300. This enables V_{BUS} to be applied from the power adaptor even in a dead battery condition. Once power is restored back to the system and back to theTPD6S300 on its VPWR pin, the TPD6S300 removes its RD pull-down resistor and turnon its OVP FETs within 3.5 ms to guarantee the RD pull-down resistor inside the CC/PD Controller is exposed within 10 ms. This is by design, because if the RD pull-down resistor is not exposed within 10 ms, the power adaptor can legally interpret this behavior as a port disconnect and remove V_{BUS} .

If desiring to power the CC/PD controller during dead battery mode and if the CC/PD Controller is configured as a DRP, it is critical that the TPD6S300 be powered before or at the same time that the CC/PD controller is powered. It is also critical that when unpowered, the CC/PD controller also expose its dead battery resistors. When the TPD6S300 gets powered, it exposes the CC pins of the CC/PD controller within 3.5 ms. Once the TPD6S300 turns on, the RD pull-down resistors of the CC/PD controller must be present immediately, in order to guarantee the power adaptor connected to power the dead battery device keeps its V_{BUS} turned on. If the power adaptor sees any change to its CC voltage for more than 10 ms, it can disconnect V_{BUS} . This removes power from the device with its battery still not sufficiently charged, which consequently removes power from the CC/PD controller and the TPD6S300. Then the RD resistors of the TPD6S300 are exposed again, connect the power adaptor's V_{BUS} to start the cycle over. This creates an infinite loop, never or very slowly charging the mobile device.

If the CC/PD Controller is configured for DRP and has started its DRP toggle before the TPD6S300 turns on, this DRP toggle is unable to guarantee that the power adaptor does not disconnect from the port. Therefore, it is recommended if the CC/PD controller is configured for DRP, that its dead battery resistors be exposed as well, and that they remain exposed until the TPD6S300 turns on. This is typically accomplished by powering the TPD6S300 at the same time as the CC/PD controller when powering the CC/PD controller in dead battery operation.

If dead battery charging is not required in your application, connect the RPD G1 and RPD G2 pins to ground.

8.3.5 3-mm × 3-mm WQFN Package

The TPD6S300 comes in a small, 3-mm × 3-mm WQFN package, greatly reducing the size of implementing a similar protection solution discretely. The WQFN package allows support for a wider range of PCB designs. Additionally, the pin-out of the TPD6S300 was designed to optimize routing with the TPS6598x family of USB Type-C/PD controllers.



8.4 Device Functional Modes

表 1 describes all of the functional modes for the TPD6S300. The "X" in the below table are "do not care" conditions, meaning any value can be present within the absolute maximum ratings of the datasheet and maintain that functional mode. Also note the D1, D2, D3, D4 pins are not listed, because these pins have IEC ESD protection diodes that are always present, regardless of whether the device is powered and regardless of the conditions on any of the other pins.

表 1. Device Mode Table

Device Mode Table		Inputs					Outputs			
MC	DDE	VPWR	C_CCx	C_SBUx	RPD_Gx	TJ	FLT	CC FETs	SBU FETs	
Normal	Unpowered, no dead battery support	<uvlo< th=""><th>Х</th><th>Х</th><th>Grounded</th><th>Х</th><th>High-Z</th><th>OFF</th><th>OFF</th></uvlo<>	Х	Х	Grounded	Х	High-Z	OFF	OFF	
Operating Conditions	Unpowered, dead battery support	<uvlo< td=""><td>х</td><td>Х</td><td>Shorted to C_CCx</td><td>Х</td><td>High-Z</td><td>OFF</td><td>OFF</td></uvlo<>	х	Х	Shorted to C_CCx	Х	High-Z	OFF	OFF	
	Powered on	>UVLO	<ovp< td=""><td><ovp< td=""><td>X, forced OFF</td><td><tsd< td=""><td>High-Z</td><td>ON</td><td>ON</td></tsd<></td></ovp<></td></ovp<>	<ovp< td=""><td>X, forced OFF</td><td><tsd< td=""><td>High-Z</td><td>ON</td><td>ON</td></tsd<></td></ovp<>	X, forced OFF	<tsd< td=""><td>High-Z</td><td>ON</td><td>ON</td></tsd<>	High-Z	ON	ON	
	Thermal shutdown	>UVLO	Х	Х	X, forced OFF	>TSD	Low (Fault Asserted)	OFF	OFF	
Fault Conditions	CC over voltage condition	>UVLO	>OVP	Х	X, forced OFF	<tsd< td=""><td>Low (Fault Asserted)</td><td>OFF</td><td>OFF</td></tsd<>	Low (Fault Asserted)	OFF	OFF	
	SBU over voltage condition	>UVLO	х	>OVP	X, forced OFF	<tsd< td=""><td>Low (Fault Asserted)</td><td>OFF</td><td>OFF</td></tsd<>	Low (Fault Asserted)	OFF	OFF	



9 Application and Implementation

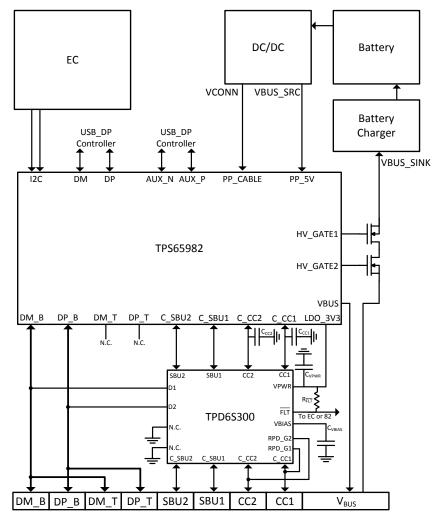
注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The TPD6S300 provides 4-channels of Short-to-VBUS overvoltage protection for the CC1, CC2, SBU1, and SBU2 pins of the USB Type-C connector, and 8-channels of IEC ESD protection for the CC1, CC2, SBU1, SBU2, DP_T, DM_T, DP_B, DM_B pins of the USB Type-C connector. Care must be taken to insure that the TPD6S300 provides adequate system protection as well as insuring that proper system operation is maintained. The following application example explains how to properly design the TPD6S300 into a USB Type-C system.

9.2 Typical Application

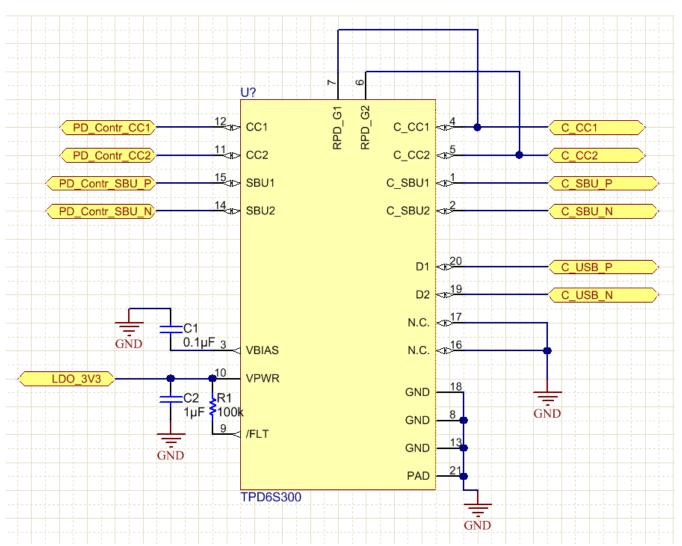


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图 30. TPD6S300 Typical Application Diagram



Typical Application (接下页)



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图 31. TPD6S300 Reference Schematic

9.2.1 Design Requirements

In this application example we study the protection requirements for a full-featured USB Type-C DRP Port, fully equipped with USB-PD, USB2.0, USB3.0, Display Port, and 100 W charging. The TPS65982 is used to easily enable a full-featured port with a single chip solution. In this application, all the pins of the USB Type-C connector are utilized. Both the CC and SBU pins are susceptible to shorting to the V_{BUS} pin. With 100 W charging, V_{BUS} operates at 20 V, requiring the CC and SBU pins to tolerate 20- V_{DC} . Additionally, the CC, SBU, and USB2.0 pins require IEC system level ESD protection. With these protection requirements present for the USB Type-C connector, the TPD6S300 is utilized. The TPD6S300 is a single chip solution that provides all the required protection for the low speed and USB2.0 pins in the USB Type-C connector.

表 2 lists the TPD6S300 design parameters.

表 2. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
V _{BUS} nominal operating voltage	20 V
Short-to-V _{BUS} tolerance for the CC and SBU pins	24 V



表 2. Design Parameters (接下页)

DESIGN PARAMETER	EXAMPLE VALUE
VBIAS nominal capacitance	0.1 μF
Dead battery charging	100 W
Maximum ambient temperature requirement	85°C

9.2.2 Detailed Design Procedure

9.2.2.1 VBIAS Capacitor Selection

As noted in the *Recommended Operating Conditions* table, a minimum of 35-V_{BUS} rated capacitor is required for the VBIAS pin, and a 50-V_{BUS} capacitor is recommended. The VBIAS capacitor is in parallel with the central IEC diode clamp integrated inside the TPD6S300. A forward biased hiding diode connects the VBIAS pin to the C_CCx and C_SBUx pins. Therefore, when a Short-to-V_{BUS} event occurs at 20 V, 20-V_{BUS} minus a forward biased diode drop is exposed to the VBIAS pin. Additionally, during the short-to-V_{BUS} event, ringing can occur almost double the settling voltage of 20 V, allowing a potential 40 V to be exposed to the C_CCx and C_SBUx pins. However, the internal IEC clamps limit the voltage exposed to the C_CCx and C_SBUx pins to around 30 V. Therefore, at least 35-V_{BUS} capacitor is required to insure the VBIAS capacitor does not get destroyed during Short-to-V_{BUS} events.

A 50-V, X7R capacitor is recommended, however. This is to further improve the derating performance of the capacitors. When the voltage across a real capacitor is increased, its capacitance value derates. The more the capacitor derates, the greater than 2x ringing can occur in the short-to-V_{BUS} RLC circuit. 50-V X7R capacitors have great derating performance, allowing for the best short-to-V_{BUS} performance of the TPD6S300.

Additionally, the VBIAS capacitor helps pass IEC 61000-4-2 ESD strikes. The more capacitance present, the better the IEC performance. So the less the VBIAS capacitor derates, the better the IEC performance. 表 3 shows real capacitors recommended to achieve the best performance with the TPD6S300.

表 3. Design Parameters

CAPACITOR SIZE	PART NUMBER
0402	CC0402KRX7R9BB104
0603	GRM188R71H104KA93D

9.2.2.2 Dead Battery Operation

For this application, we want to support 100-W dead battery operation; when the laptop is out of battery, we still want to charge the laptop at 20 V and 5 A. This means that the USB PD Controller must receive power in dead battery mode. The TPS65982 has its own built in LDO in order to supply the TPS65982 power from V_{BUS} in a dead battery condition. The TPS65982 can also provide power to its flash during this condition through its LDO 3V3 pin.

The TPD6S300s OVP FETs remain OFF when it is unpowered in order to insure in a dead battery situation proper protection is still provided to the PD controller in the system, in this case the TPS65982. However, when the OVP FETs are OFF, this isolates the TPS65982s dead battery resistors from the USB Type-C ports CC pins. A USB Type-C power adaptor must see the RD pull-down dead battery resistors on the CC pins or it does not provide power on V_{BUS} . Since the TPS65982s dead battery resistors are isolated from the USB Type-C connector's CC pins, the TPD6S300s built in dead battery resistors must be connected. Short the RPD_G1 pin to the C_CC1 pin, and short the RPD_G2 pin to the C_CC2 pin.

Once the power adaptor sees the TPD6S300s dead battery resistors, it applies 5 V on the V_{BUS} pin. This provides power to the TPS65982, turning the PD controller on, and allowing the battery to begin to charge. However, this application requires 100 W charging in dead battery mode, so V_{BUS} at 20 V and 5 A is required. USB PD negotiation is required to accomplish this, so the TPS65982 needs to be able to communicate on the CC pins. This means the TPD6S3000 needs to be turned on in dead battery mode as well so the TPS65982s PD controller can be exposed to the CC lines. To accomplish this, it is critical that the TPD6S300 is powered by the TPS65982s internal LDO, the LDO_3V3 pin. This way, when the TPS65982 receives power on V_{BUS} , the TPD6S300 is turned on simultaneously.



It is critical that the TPS65982s dead battery resistors are also connected to its CC pins for dead battery operation. Short the TPS65982s RPD_G1 pin to its C_CC1 pin, and its RPD_G2 pin to its C_CC2 pin. It is critical that the TPS65982s dead battery resistors are present; once the TPD6S300 receives power, removes its dead battery resistors and turns on its OVP FETs, RD pull-down resistors must be present on the CC line in order to guarantee the power adaptor stays connected. If RD is not present and the voltage on CC changes for more than 10 ms, the power adaptor interprets this as a disconnect and remove V_{BUS} .

Also, it is important that the TPS65982s dead battery resistors are present so it properly boots up in dead battery operation with the correct voltages on its CC pins.

Once this process has occurred, the TPS65982 can start negotiating with the power adaptor through USB PD for higher power levels, allowing 100-W operation in dead battery mode.

For more information on the TPD6S300 dead battery operation, see the *CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices* section of the datasheet. Also, see 图 32 for a waveform of the CC line when the TPD6S300 is turning on and exposing the TPS65982s dead battery resistors to the USB Type-C connector.

9.2.2.3 CC Line Capacitance

USB PD has a specification for the total amount of capacitance that is required for proper USB PD BMC operation on the CC lines. The specification from section 5.8.6 of the *USB PD Specification* is given below in 表 4.

表 4. USB PD cReceiver Specification

NAME	DESCRIPTION	MIN	MAX	UNIT	COMMENT
cReceiver	CC receiver capacitance	200	600	pF	The DFP or UFP system shall have capacitance within this range when not transmitting on the line

Therefore, the capacitance on the CC lines must stay in between 200 pF and 600 pF when USB PD is being used. Therefore, the combination of capacitances added to the system by the TPS65982, the TPD6S300, and any external capacitor must fall within these limits. 表 5 shows the analysis involved in choosing the correct external CC capacitor for this system, and shows that an external CC capacitor is required.

表 5. CC Line Capacitor Calculation

CC Capacitance	MIN	MAX	UNIT	COMMENT
CC line target capacitance	200	600	pF	From the USB PD Specification section
TPS65982 capacitance	70	120	pF	From the TPS65982 Datasheet
TPD6S300 capacitance	60	120	pF	From the table.
Proposed capacitor GRM033R71E221KA01D	110	330	pF	CAP, CERM, 220 pF, 25 V, ±10%, X7R, 0201 (For min and max, assume ±50% capacitance change with temperature and voltage derating to be overly conservative)
TPS65982 + TPD6S300 + GRM033R71E221KA01D	240	570	pF	Meets USB PD cReceiver specification

9.2.2.4 Additional ESD Protection on CC and SBU Lines

If additional IEC ESD protection is desired to be placed on either the CC or SBU lines, it is important that high-voltage ESD protection diodes be used. The maximum DC voltage that can be seen in USB PD is 21-V_{BUS} , with 21.5 V allowed during voltage transitions. Therefore, an ESD protection diode must have a reverse stand off voltage higher than 21.5 V in order to guarantee the diode does not breakdown during a short-to-V_{BUS} event and have large amounts of current flowing through it indefinitely, destroying the diode. A reverse stand off voltage of 24 V is recommended to give margin above 21.5 V in case USB Type-C power adaptors are released in the market which break the USB Type-C specification.



Furthermore, due to the fact that the Short-to- V_{BUS} event applies a DC voltage to the CC and SBU pins, a deep-snap back diode cannot be used unless its minimum trigger voltage is above 42 V. During a Short-to- V_{BUS} event, RLC ringing of up to 2x the settling voltage can be exposed to CC and SBU, allowing for up to 42 V to be exposed. Furthermore, if any capacitor derates on the CC or SBU line, greater than 2x ringing can occur. Since this ringing is hard to bound, it is recommended to not use deep-snap back diodes. If the diode triggers during the short-to- V_{BUS} hot-plug event, it begins to operate in is conduction region. With a $20-V_{BUS}$ source present on the CC or SBU line, this allows the diode to conduct indefinitely, destroying the diode.

9.2.2.5 FLT Pin Operation

The FLT and OVP FET have specific timing parameters to allow different benefits depending on how the system designer desires the system to respond to a Short-to-V_{BUS} event.

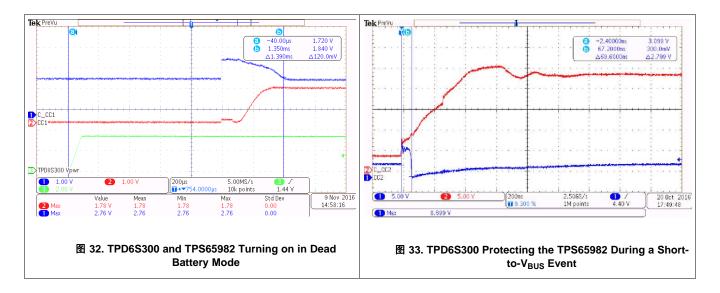
Once a Short-to- V_{BUS} occurs on the C_CCx or C_SBUx pins, the \overline{FLT} pin is asserted in 20 µs (typical) so the PD controller can be notified quickly. If V_{BUS} is being shorted to CC or SBU, it is recommended to respond to the event by forcing a detach in the USB PD controller to remove V_{BUS} from the port. Although the USB Type-C port using the TPD6S300 is not damaged, as the TPD6S300 provides protection from these events, the other device connected through the USB Type-C Cable or any active circuitry in the cable can be damaged. Although shutting the V_{BUS} off through a detach does not guarantee it stops the other device or cable from being damaged, it can mitigate any high current paths from causing further damage after the initial damage takes place. Additionally, even if the active cable or other device does have proper protection, the short-to- V_{BUS} event may corrupt a configuration in an active cable or in the other PD controller, so it is best to detach and reconfigure the port.

For UFP's, the TPD6S300 automatically forces a detach, removing the need to use the FLT pin if the only response required by your system during a short-to- V_{BUS} event is forcing a detach on the port. The TPD6S300 keeps its CC OVP FET OFF for at least 21 ms after a Short-to- V_{BUS} event occurs, causing the CC line voltage to change from is configuration value for more than 20 ms, forcing the PD controllers to detach. For DFPs, this operation cannot be guaranteed because of the <u>parasitic diode</u> in the OVP FET from CCx to C_CCx and from SBUx to C_SBUx. Therefore for DFPs, using the FLT pin recommended. For our application using the TPS65982 as a DRP, using the FLT pin is recommended.

9.2.2.6 How to Connect Unused Pins

If either the RPD Gx pins or any of the Dx pins are unused in a design, they must be connected to GND.

9.2.3 Application Curves





10 Power Supply Recommendations

The VPWR pin provides power to all the circuitry in the TPD6S300. It is recommended a 1-µF decoupling capacitor is placed as close as possible to the VPWR pin. If USB PD is desired to be operated in dead battery conditions, it is critical that the TPD6S300 share the same power supply as the PD controller in dead battery boot-up (such as sharing the same dead battery LDO). See the CC Dead Battery Resistors Integrated for Handling the Dead Battery Use Case in Mobile Devices section for more details.



11 Layout

11.1 Layout Guidelines

Proper routing and placement is important to maintain the signal integrity the USB2.0, SBU, CC line signals. The following guidelines apply to the TPD6S300:

- Place the bypass capacitors as close as possible to the V_{PWR} pin, and ESD protection capacitor as close as
 possible to the V_{BIAS} pin. Capacitors must be attached to a solid ground. This minimizes voltage disturbances
 during transient events such as short-to-V_{BUS} and ESD strikes.
- The USB2.0 and SBU lines must be routed as straight as possible and any sharp bends must be minimized.

Standard ESD recommendations apply to the C_CC1, C_CC2, C_SBU1, C_SBU2, D1, D2, D3, and D4 pins as well:

- The optimum placement for the device is as close to the connector as possible:
 - EMI during an ESD event can couple from the trace being struck to other nearby unprotected traces, resulting in early system failures.
 - The PCB designer must minimize the possibility of EMI coupling by keeping any unprotected traces away from the protected traces which are between the TPD6S300 and the connector.
- · Route the protected traces as straight as possible.
- Eliminate any sharp corners on the protected traces between the TVS and the connector by using rounded corners with the largest radii possible.
 - Electric fields tend to build up on corners, increasing EMI coupling.
- It is best practice to not via up to the D1, D2, D3, and D4 pins from a trace routed on another layer. Rather, it is better to via the trace to the layer with the Dx pin, and to continue that trace on that same layer. See the ESD Protection Layout Guide application report, section 1.3 for more details.

11.2 Layout Example

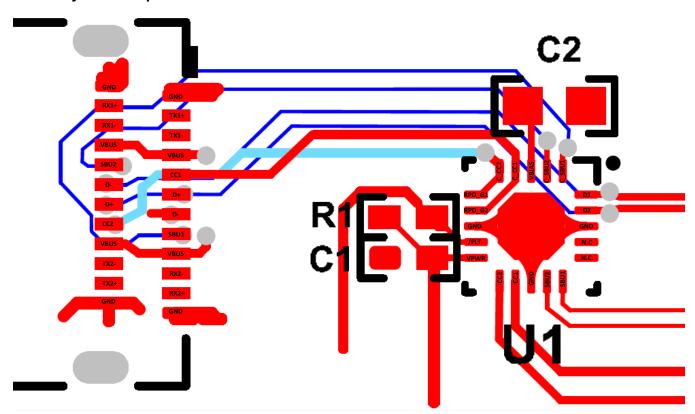


图 34. TPD6S300 Typical Layout



12 器件和文档支持

12.1 文档支持

12.1.1 相关文档

请参阅如下相关文档:

《TPD6S300 评估模块用户指南》

12.2 接收文档更新通知

如需接收文档更新通知,请访问 ti.com 上的器件产品文件夹。请单击右上角的通知我 进行注册,即可收到任意产品信息更改每周摘要。有关更改的详细信息,请查看任意已修订文档中包含的修订历史记录。

12.3 社区资源

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Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

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12.5 静电放电警告



这些装置包含有限的内置 ESD 保护。 存储或装卸时,应将导线一起截短或将装置放置于导电泡棉中,以防止 MOS 门极遭受静电损伤。

12.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

13 机械、封装和可订购信息

以下页面包括机械、封装和可订购信息。这些信息是指定器件的最新可用数据。这些数据发生变化时,我们可能不会另行通知或修订此文档。如欲获取此产品说明书的浏览器版本,请参见左侧的导航栏。

13.1

www.ti.com 10-Nov-2025

PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
						(4)	(5)		
TPD6S300RUKR	Active	Production	WQFN (RUK) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	6S30
TPD6S300RUKR.A	Active	Production	WQFN (RUK) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	6S30
TPD6S300RUKRG4	Active	Production	WQFN (RUK) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	6S30
TPD6S300RUKRG4.A	Active	Production	WQFN (RUK) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	6S30

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

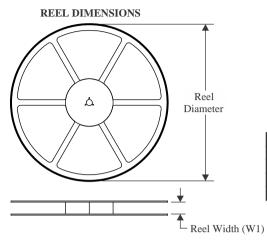
⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

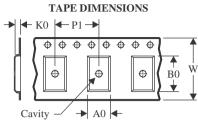
⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

PACKAGE MATERIALS INFORMATION

www.ti.com 18-Jun-2025

TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

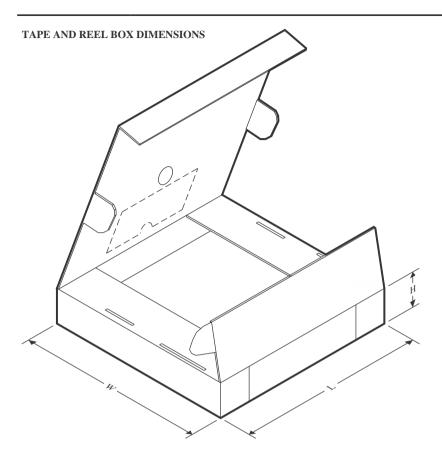
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPD6S300RUKR	WQFN	RUK	20	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPD6S300RUKRG4	WQFN	RUK	20	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

www.ti.com 18-Jun-2025



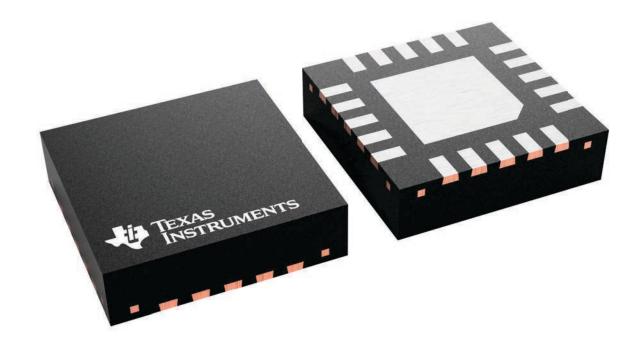
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPD6S300RUKR	WQFN	RUK	20	3000	346.0	346.0	33.0
TPD6S300RUKRG4	WQFN	RUK	20	3000	346.0	346.0	33.0

3 x 3, 0.4 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

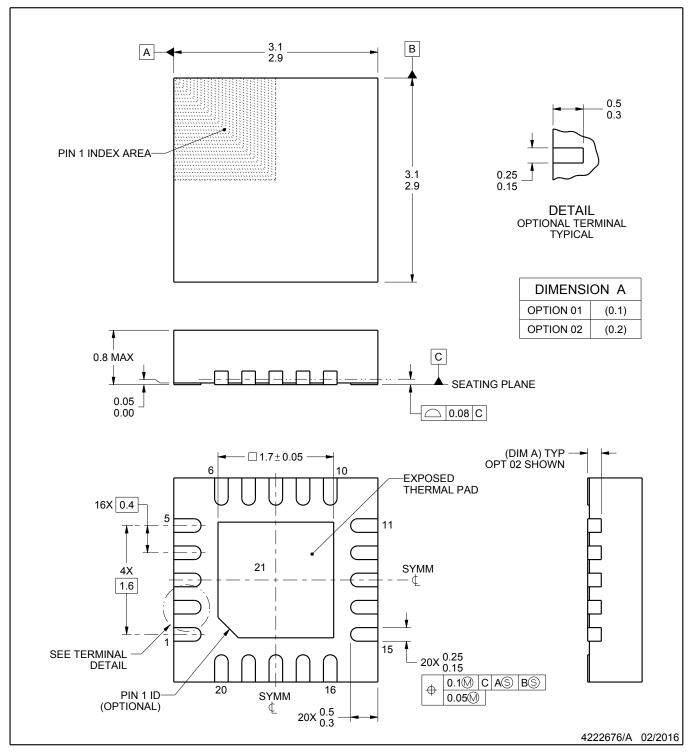
This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



INSTRUMENTS www.ti.com



PLASTIC QUAD FLATPACK - NO LEAD



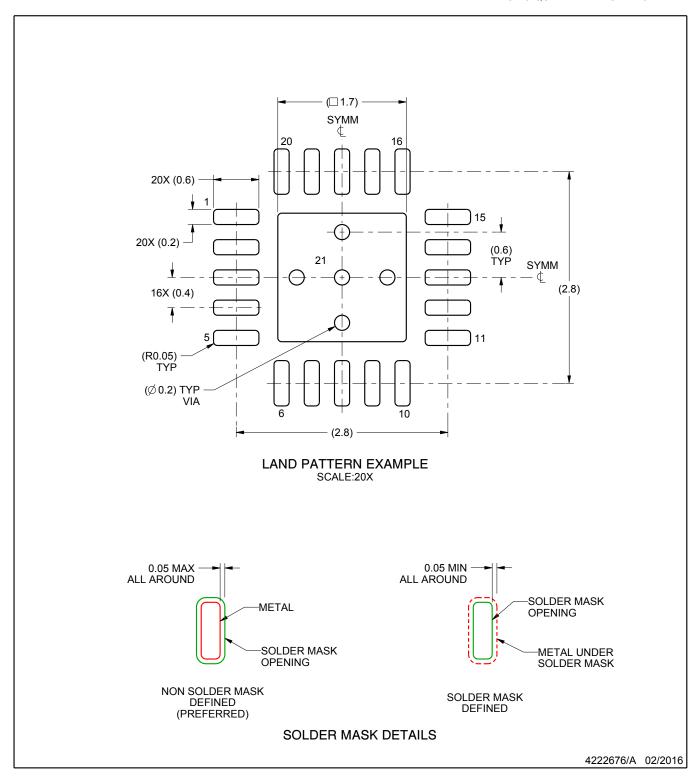
NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

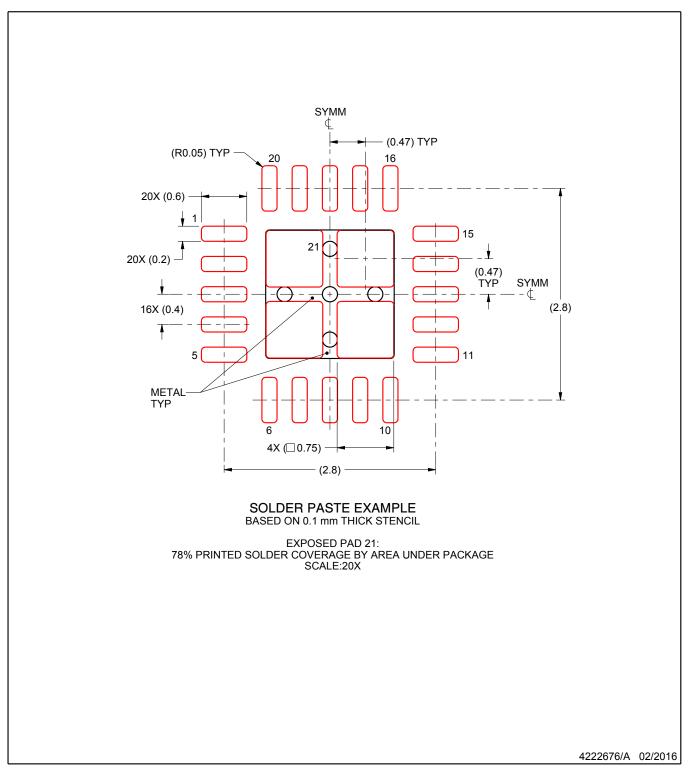


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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最后更新日期: 2025 年 10 月