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- Trimmed Offset Voltage: TLC277 ... 500 μV Max at 25°C, V_{DD} = 5 V
- Input Offset Voltage Drift . . . Typically 0.1 μV/Month, Including the First 30 Days
- Wide Range of Supply Voltages Over Specified Temperature Range: 0°C to 70°C ... 3 V to 16 V -40°C to 85°C ... 4 V to 16 V -55°C to 125°C ... 4 V to 16 V
- Single-Supply Operation
- Common-Mode Input Voltage Range Extends Below the Negative Rail (C-Suffix, I-Suffix types)
- Low Noise . . . Typically 25 nV/\Hz at f = 1 kHz
- Output Voltage Range Includes Negative Rail
- High Input impedance . . . $10^{12} \Omega$ Typ
- ESD-Protection Circuitry
- Small-Outline Package Option Also Available in Tape and Reel
- Designed-In Latch-Up Immunity

description

The TLC272 and TLC277 precision dual operational amplifiers combine a wide range of input offset voltage grades with low offset voltage drift, high input impedance, low noise, and speeds approaching those of general-purpose BiFET devices.

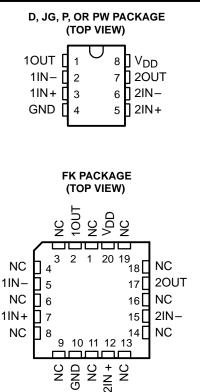
These devices use Texas Instruments silicongate LinCMOS[™] technology, which provides offset voltage stability far exceeding the stability available with conventional metal-gate processes.

The extremely high input impedance, low bias currents, and high slew rates make these costeffective devices ideal for applications previously reserved for BiFET and NFET products. Four offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC272 (10 mV) to the high-precision TLC277 (500 μ V). These advantages, in combination with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

LinCMOS is a trademark of Texas Instruments.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of Texas instruments standard warranty. Production processing does not necessarily include testing of all parameters.





NC - No internal connection

DISTRIBUTION OF TLC277 INPUT OFFSET VOLTAGE 30 473 Units Tested From 2 Wafer Lots $V_{DD} = 5 V$ T_A = 25°C 25 P Package Percentage of Units – % 20 15 10 5 n -800 -4000 400 800 VIO – Input Offset Voltage – µV

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description (continued)

			AVAILABLE	E OPTIONS			
			PAC	KAGED DEVIC	CES		0,115
TA	V _{IO} max AT 25°C	SMALL OUTLINE (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	PLASTIC DIP (P)	TSSOP (PW)	CHIP FORM (Y)
0°C to 70°c	500 μV 2 mV 5 mV 10mV	TLC277CD TLC272BCD TLC272ACD TLC272CD			TLC277CP TLC272BCP TLC272ACP TLC272CP	— — — TLC272CPW	— — — TLC272Y
−40°C to 85°C	500 μV 2 mV 5 mV 10 mV	TLC277ID TLC272BID TLC272AID TLC272ID			TLC277IP TLC272BIP TLC272AIP TLC272IP		

AVAILARIE ORTIONS

The D package is available taped and reeled. Add R suffix to the device type (e.g., TLC277CDR).

In general, many features associated with bipolar technology are available on LinCMOSTM operational amplifiers without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are easily designed with the TLC272 and TLC277. The devices also exhibit low voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip carrier versions for high-density system applications.

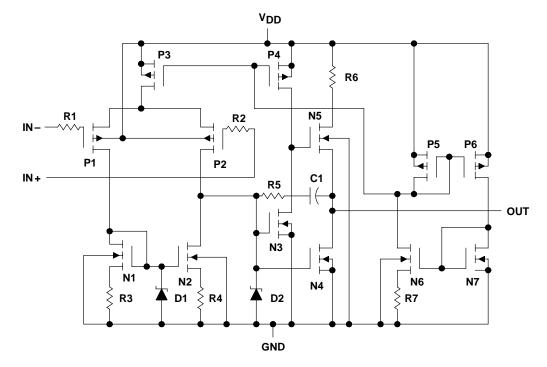
The device inputs and outputs are designed to withstand -100-mA surge currents without sustaining latch-up.

The TLC272 and TLC277 incorporate internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from -40°C to 85°C. The M-suffix devices are characterized for operation over the full military temperature range of -55°C to 125°C.



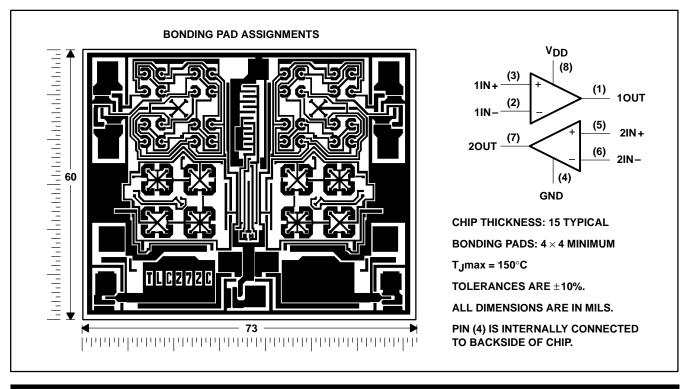
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equivalent schematic (each amplifier)

TLC272Y chip information

This chip, when properly assembled, displays characteristics similar to the TLC272C. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. Chips may be mounted with conductive epoxy or a gold-silicon preform.



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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)[†]

Supply voltage, V _{DD} (see Note 1) Differential input voltage, V _{ID} (see Note 2) Input voltage range, V _I (any input)	$\dots \dots \pm V_{DD}$
Input current, I	
output current, I_O (each output)	
Total current into V _{DD}	
Total current out of GND	45 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total dissipation See I	Dissipation Rating Table
Operating free-air temperature, T _A : C suffix	0°C to 70°C
l suffix	40°C to 85°C
M suffix	
Storage temperature range	−65°C to 150°C
Case temperature for 60 seconds: FK package	
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D, P, or PW packag Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. All voltage values, except differential voltages, are with respect to network ground.

2. Differential voltages are at IN+ with respect to IN-.

3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

		DISSIPATION RA	ATING TABLE		
PACKAGE	T _A ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T _A = 25°C	T _A = 70°C POWER RATING	T _A = 85°C POWER RATING	T _A = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	N/A
FK	1375 mW	11 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
Р	1000 mW	8.0 mW/°C	640 mW	520 mW	N/A
PW	525 mW	4.2 mW/°C	336 mW	N/A	N/A

recommended operating conditions

		C SU	FFIX	I SUF	FIX	M SU	FFIX	
		MIN	MAX	MIN	MAX	MIN	MAX	UNIT
Supply voltage, VDD		3	16	4	16	4	16	V
	$V_{DD} = 5 V$	-0.2	3.5	-0.2	3.5	0	3.5	V
Common-mode input voltage, V_{IC}	V _{DD} = 10 V	-0.2	8.5	-0.2	8.5	0	8.5	V
Operating free-air temperature, T_A		0	70	-40	85	-55	125	°C



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electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

	PARAMETER		TEST COND	TIONS	TAT	TLC272 TLC272	C, TLC2 BC, TLC		UNIT
						MIN	TYP	MAX	
		TI 00700	V _O = 1.4 V,	V _{IC} = 0,	25°C		1.1	10	
		TLC272C	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			12	
		TI 00704.0	V _O = 1.4 V,	V _{IC} = 0,	25°C		0.9	5	mV
	land offerstandle as	TLC272AC	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			6.5	
VIO	Input offset voltage	TI 0070D0	V _O = 1.4 V,	V _{IC} = 0,	25°C		230	2000	
		TLC272BC	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			3000	
		TI 00770	V _O = 1.4 V,	V _{IC} = 0,	25°C		200	500	μV
		TLC277C	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			1500	
α_{VIO}	Temperature coefficient of inpu	t offset voltage			25°C to 70°C		1.8		μV/°C
					25°C		0.1	60	
ΙΟ	Input offset current (see Note 4)			70°C		7	300	pА
			V _O = 2.5 V,	V _{IC} = 2.5 V	25°C		0.6	60	
I _{IB}	Input bias current (see Note 4)				70°C		40	600	pА
						-0.2	-0.3		
					25°C	to	to		V
VICR	Common-mode input voltage ra	ange				4	4.2		
·ICK	(see Note 5)				Full range	-0.2 to		V	
					Fuillange	3.5			v
					25°C	3.2	3.8		
√он	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	0°C	3	3.8		V
011				-	70°C	3	3.8		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	IOT = 0	0°C		0	50	mV
02				01	70°C		0	50	
					25°C	5	23		
AVD	Large-signal differential voltage	amplification	$V_{O} = 0.25 V \text{ to } 2 V,$	R _I = 10 kΩ	0°C	4	27		V/mV
		·		-	70°C	4	20		
					25°C	65	80		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICR}min$		0°C	60	84		dB
	•				70°C	60	85		
					25°C	65	95		
k SVR		upply-voltage rejection ratio	$V_{DD} = 5 V \text{ to } 10 V,$	V _O = 1.4 V	0°C	60	94		dB
	$(\Delta V_{DD}/\Delta V_{IO})$			-	70°C	60	96		
					25°C		1.4	3.2	
IDD	Supply current (two amplifiers)		$V_O = 2.5 V$, No load	V _{IC} = 2.5 V,	0°C		1.6	3.6	mA
			INU IUAU		70°C		1.2	2.6	

[†]Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

	PARAMETER		TEST CONDI	TIONS	т _А †	TLC272 TLC272			UNIT
					^	MIN	TYP	MAX	-
		-	V _O = 1.4 V,	VIC = 0,	25°C		1.1	10	
		TLC272C	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			12	
			V _O = 1.4 V,	V _{IC} = 0,	25°C		0.9	5	mV
	· · · · · ·	TLC272AC	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			6.5	
VIO	Input offset voltage	-	V _O = 1.4 V,	VIC = 0,	25°C		290	2000	
		TLC272BC	$R_{S} = 50 \Omega,$	RL = 10 kΩ	Full range			3000	
		-	V _O = 1.4 V,	V _{IC} = 0,	25°C		250	800	μV
		TLC277C	R _S = 50 Ω,	RL = 10 kΩ	Full range			1900	
0	Tomporature coefficient of inp	ut offect veltage			25°C to		2		μV/°C
αVIO	Temperature coefficient of inp	ut onset voltage			70°C		Z		μv/C
lia	Input offset current (see Note	1)			25°C		0.1	60	pА
IIO	input onset current (see Note	4)	V _O = 5 V,	$V_{10} = 5 V_{10}$	70°C		7	300	рА
l	Innut hing ourrent (and Note 4	\	$v_0 = 5 v$,	V _{IC} = 5 V	25°C		0.7	60	-
IΒ	Input bias current (see Note 4)			70°C		50	600	рА
						-0.2	-0.3		
					25°C	to 9	to 9.2		V
VICR	Common-mode input voltage (see Note 5)	range				-0.2	5.2		
					Full range	-0.2 to			V
					5	8.5			
					25°C	8	8.5		
Vон	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	0°C	7.8	8.5		V
					70°C	7.8	8.4		
					25°C		0	50	
VOL	Low-level output voltage		V _{ID} = -100 mV,	IOT = 0	0°C		0	50	mV
					70°C		0	50	
					25°C	10	36		
AVD	Large-signal differential voltag	e amplification	$V_{O} = 1 V \text{ to } 6 V,$	$R_L = 10 \ k\Omega$	0°C	7.5	42		V/mV
					70°C	7.5	32		
					25°C	65	85		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICR}min$		0°C	60	88		dB
					70°C	60	88		
					25°C	65	95		
k SVR	Supply-voltage rejection ratio $(\Delta V_{DD}/\Delta V_{IO})$		$V_{DD} = 5 V \text{ to } 10 V,$	V _O = 1.4 V	0°C	60	94		dB
					70°C	60	96		
					25°C		1.9	4	
IDD	Supply current (two amplifiers)	V _O = 5 V, No load	V _{IC} = 5 V,	0°C		2.3	4.4	mA
			NU IUdu		70°C		1.6	3.4	

[†] Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

	PARAMETER		TEST COND	ITIONS	т _А †		2I, TLC2 2BI, TL(UNIT
						MIN	TYP	MAX	
			V _O = 1.4 V,	V _{IC} = 0,	25°C		1.1	10	
		TLC272I	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			13	.,
		TI 007041	V _O = 1.4 V,	V _{IC} = 0,	25°C		0.9	5	mV
.,		TLC272AI	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			7	
VIO	Input offset voltage		V _O = 1.4 V,	VIC = 0,	25°C		230	2000	
		TLC272BI	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			3500	
		TI 00771	V _O = 1.4 V,	V _{IC} = 0,	25°C		200	500	μV
		TLC2771	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			2000	
ανιο	Temperature coefficient of input	It offset voltage			25°C to		1.8		μV/°C
μγlΟ	Temperature coefficient of mp	it onset voltage			85°C				μν/Ο
IIO	Input offset current (see Note	1)			25°C		0.1	60	pА
U	input onset current (see Note	+)	V _O = 2.5 V,	V _{IC} = 2.5 V	85°C		24	15	рл
lin	Input bias current (see Note 4		VO = 2.5 V,	VIC - 2.5 V	25°C		0.6	60	pА
ΙB					85°C		200	35	μл
						-0.2	-0.3		.,
	0				25°C	to 4	to 4.2		V
VICR	Common-mode input voltage ((see Note 5)	ange				-0.2	7.2		
	Full range to		V						
					0	3.5			
					25°C	3.2	3.8		
∨он	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	-40°C	3	3.8		V
					85°C	3	3.8		
					25°C		0	50	
VOL	Low-level output voltage		V _{ID} = -100 mV,	$I_{OL} = 0$	-40°C		0	50	mV
					85°C		0	50	
					25°C	5	23		
AVD	Large-signal differential voltag	e amplification	$V_{O} = 1 V \text{ to } 6 V,$	$R_L = 10 \ k\Omega$	-40°C	3.5	32		V/mV
					85°C	3.5	19		
					25°C	65	80		
CMRR	Common-mode rejection ratio		VIC = VICR ^{min}		-40°C	60	81		dB
					85°C	60	86		
					25°C	65	95		
k SVR	Supply-voltage rejection ratio		$V_{DD} = 5 V \text{ to } 10 V,$	V _O = 1.4 V	-40°C	60	92		dB
	$(\Delta V_{DD}/\Delta V_{IO})$			-	85°C	60	96		
					25°C		1.4	3.2	
IDD	Supply current (two amplifiers)	1	$V_{O} = 2.5 V$,	V _{IC} = 2.5 V,	-40°C		1.9	4.4	mA
			No load		85°C		1.1	2.4	

[†] Full range is -40° C to 85° C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

	PARAMETER		TEST COND	TIONS	т _А †		2I, TLC2 2BI, TL(UNIT
						MIN	TYP	MAX	•••••
		-	V _O = 1.4 V,	VIC = 0,	25°C		1.1	10	
		TLC272I	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			13	
		TI 007041	V _O = 1.4 V,	VIC = 0,	25°C		0.9	5	mV
v	lanut affectualters	TLC272AI	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			7	
VIO	Input offset voltage		V _O = 1.4 V,	VIC = 0,	25°C		290	2000	
		TLC272BI	R _S = 50 Ω,	$R_L = 10 \ k\Omega$	Full range			3500	
		TI 00771	V _O = 1.4 V,	V _{IC} = 0,	25°C		250	800	μV
		TLC277I	R _S = 50 Ω,	$R_L = 10 \ k\Omega$	Full range			2900	
α _{VIO}	Temperature coefficient of input	it offset voltage			25°C to		2		μV/°C
~00		at onset voltage			85°C				μν/Ο
IIO	Input offset current (see Note	4)			25°C		0.1	60	pА
U			V _O = 5 V,	V _{IC} = 5 V	85°C		26	1000	prv
I _{IB}	Input bias current (see Note 4	h	v0 = 0 v,		25°C		0.7	60	pА
UD		/			85°C		220	2000	ρ/ (
					0500	-0.2	-0.3		
	Common mode innut velkome				25°C	to 9	to 9.2		V
VICR	Common-mode input voltage range (see Note 5)		-0.2	0.2					
	()				Full range	to			V
					Ū	8.5			
					25°C	8	8.5		
Vон	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	-40°C	7.8	8.5		V
					85°C	7.8	8.5		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	-40°C		0	50	mV
					85°C		0	50	
					25°C	10	36		
Avd	Large-signal differential voltag	e amplification	$V_{O} = 1 V \text{ to } 6 V,$	$R_L = 10 \ k\Omega$	-40°C	7	46		V/mV
					85°C	7	31		
					25°C	65	85		
CMRR	Common-mode rejection ratio		VIC = VICRmin		-40°C	60	87		dB
					85°C	60	88		
					25°C	65	95		
k SVR	Supply-voltage rejection ratio		V _{DD} = 5 V to 10 V,	V _O = 1.4 V	−40°C	60	92		dB
	$(\Delta V_{DD}/\Delta V_{IO})$			-	85°C	60	96		
					25°C		1.4	4	
IDD	Supply current (two amplifiers))	$V_{O} = 5 V$,	VIC = 5 V,	−40°C		2.8	5	mA
			No load		85°C		1.5	3.2	

[†] Full range is -40° C to 85° C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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			TEAT AAND		T . †	TLC27	2M, TLC	277M	
	PARAMETER		TEST COND	ITIONS	TA†	MIN	TYP	MAX	UNIT
			V _O = 1.4 V,	V _{IC} = 0,	25°C		1.1	10	
.,		TLC272M	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			12	mV
VIO	Input offset voltage	TI 007714	V _O = 1.4 V,	V _{IC} = 0,	25°C		200	500	.,
		TLC277M	R _S = 50 Ω,	$R_L = 10 k\Omega$	Full range			3750	μV
αΛΙΟ	Temperature coefficient of input ovoltage	offset			25°C to 125°C		2.1		μV/° C
	lanut affaat aumant (aan Nata 4)				25°C		0.1	60	pА
IIO	Input offset current (see Note 4)		N- 05V		125°C		1.4	15	nA
	Input biog ourrent (and Note 4)		V _O = 2.5 V	V _{IC} = 2.5 V	25°C		0.6	60	pА
IВ	Input bias current (see Note 4)				125°C		9	35	nA
.,	Common-mode input voltage ran	ge			25°C	0 to 4	-0.3 to 4.2		V
VICR	(see Note 5)	-			Full range	0 to 3.5			V
					25°C	3.2	3.8		
∨он	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	−55°C	3	3.8		V
					125°C	3	3.8		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	−55°C		0	50	mV
					125°C		0	50	
					25°C	5	23		
Avd	Large-signal differential voltage a	mplification	V_{O} = 0.25 V to 2 V	$R_L = 10 \ k\Omega$	−55°C	3.5	35		۷/m
					125°C	3.5	16		
					25°C	65	80		
CMRR	Common-mode rejection ratio		$V_{IC} = V_{ICR}min$		−55°C	60	81		dB
					125°C	60	84		
	Quantu value as asisetian as the				25°C	65	95		
^k SVR	Supply-voltage rejection ratio $(\Delta V_{DD}/\Delta V_{IO})$		V_{DD} = 5 V to 10 V,	V _O = 1.4 V	−55°C	60	90		dB
					125°C	60	97		
					25°C		1.4	3.2	
IDD	Supply current (two amplifiers)		$V_{O} = 2.5 V$, No load	V _{IC} = 2.5 V,	−55°C		2	5	mA
					125°C		1	2.2	

electrical characteristics at specified free-air temperature, V_{DD} = 5 V (unless otherwise noted)

[†]Full range is –55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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electrical characteristics at specified free-air temperature, V_{DD} = 10 V (unless otherwise noted)

			TEOT OON		T . †	TLC27	2M, TLC	277M	
	PARAMETER		TEST CONE	MINONS	TA†	MIN	TYP	MAX	UNIT
		TI 007014	V _O = 1.4 V,	$V_{IC} = 0,$	25°C		1.1	10	
.,	1 . 1 . 1	TLC272M	R _S = 50 Ω,	$R_L = 10 \ k\Omega$	Full range			12	mV
VIO	Input offset voltage	TI 007714	V _O = 1.4 V,	$V_{IC} = 0,$	25°C		250	800	
		TLC277M	$R_S = 50 \Omega$,	$R_L = 10 \ k\Omega$	Full range			4300	μV
αΛΙΟ	Temperature coefficient of input voltage	offset			25°C to 125°C		2.2		μV/°C
	lonut effect coment (e.e. Nete 4)				25°C		0.1	60	pА
10	Input offset current (see Note 4)				125°C		1.8	15	nA
L	lenut hier summert (see Note 4)		V _O = 5 V,	V _{IC} = 5 V	25°C		0.7	60	pА
IВ	Input bias current (see Note 4)				125°C		10	35	nA
	Common-mode input voltage ra	nae			25°C	0 to 9	-0.3 to 9.2		V
VICR	(see Note 5)				Full range	0 to 8.5			V
					25°C	8	8.5		
∨он	High-level output voltage		V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	−55°C	7.8	8.5		V
					125°C	7.8	8.4		
					25°C		0	50	
VOL	Low-level output voltage		$V_{ID} = -100 \text{ mV},$	$I_{OL} = 0$	−55°C		0	50	mV
					125°C		0	50	
	the second stress of stress of stress to the second				25°C	10	36		
Avd	Large-signal differential voltage amplification		$V_{O} = 1 V \text{ to } 6 V,$	$R_L = 10 \ k\Omega$	−55°C	7	50		V/m\
					125°C	7	27		
					25°C	65	85		
CMRR	Common-mode rejection ratio		VIC = VICRmin		−55°C	60	87		dB
					125°C	60	86		
					25°C	65	95		
^k SVR	Supply-voltage rejection ratio (ΔVDD/ΔVIO)		V_{DD} = 5 V to 10 V,	V _O = 1.4 V	−55°C	60	90		dB
	· ····································				125°C	60	97		
					25°C		1.9	4	
IDD	Supply current (two amplifiers)		V _O = 5 V, No load	V _{IC} = 5 V,	−55°C		3	6	mA
					125°C		1.3	2.8	

[†] Full range is -55° C to 125° C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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		TEAT OON		Т	LC272Y		
	PARAMETER	TEST CONI	JIIONS	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	V _O = 1.4 V, R _S = 50 Ω,	V _{IC} = 0, R _L = 10 kΩ		1.1	10	mV
α_{VIO}	Temperature coefficient of input offset voltage				1.8		μV/°C
١O	Input offset current (see Note 4)				0.1		pА
IIB	Input bias current (see Note 4)	V _O = 2.5 V,	V _{IC} = 2.5 V		0.6		pА
VICR	Common-mode input voltage range (see Note 5)			-0.2 to 4	-0.3 to 4.2		V
Vон	High-level output voltage	V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	3.2	3.8		V
VOL	Low-level output voltage	$V_{ID} = -100 \text{ mV},$	IOT = 0		0	50	mV
AVD	Large-signal differential voltage amplification	$V_{O} = 0.25 V \text{ to } 2 V$	$R_L = 10 \ k\Omega$	5	23		V/mV
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}min$		65	80		dB
ksvr	Supply-voltage rejection ratio ($\Delta V_{DD} / \Delta V_{IO}$)	V _{DD} = 5 V to 10 V,	V _O = 1.4 V	65	95		dB
IDD	Supply current (two amplifiers)	V _O = 2.5 V, No load	V _{IC} = 2.5 V,		1.4	3.2	mA

electrical characteristics, V_{DD} = 5 V, T_A = 25°C (unless otherwise noted)

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically. 5. This range also applies to each input individually.

electrical characteristics, V_{DD} = 10 V, T_A = 25°C (unless otherwise noted)

	DADAMETED	TEST CON		Т	LC272Y		
	PARAMETER	TEST CONI	JIIONS	MIN	TYP	MAX	UNIT
V _{IO}	Input offset voltage	$V_{O} = 1.4 V,$ R _S = 50 Ω ,	V _{IC} = 0, R _L = 10 kΩ		1.1	10	mV
α_{VIO}	Temperature coefficient of input offset voltage				1.8		μV/°C
IIO	Input offset current (see Note 4)				0.1		pА
I _{IB}	Input bias current (see Note 4)	V _O = 5 V,	V _{IC} = 5 V		0.7		pА
VICR	Common-mode input voltage range (see Note 5)			-0.2 to 9	-0.3 to 9.2		V
VOH	High-level output voltage	V _{ID} = 100 mV,	$R_L = 10 \ k\Omega$	8	8.5		V
VOL	Low-level output voltage	$V_{ID} = -100 \text{ mV},$	IOT = 0		0	50	mV
AVD	Large-signal differential voltage amplification	$V_{O} = 1 V \text{ to } 6 V,$	$R_L = 10 \ k\Omega$	10	36		V/mV
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}min$		65	85		dB
k SVR	Supply-voltage rejection ratio $(\Delta V_{DD} / \Delta V_{IO})$	V _{DD} = 5 V to 10 V,	V _O = 1.4 V	65	95		dB
IDD	Supply current (two amplifiers)	V _O = 5 V, No load	V _{IC} = 5 V,		1.9	4	mA

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.



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operating characteristics at specified free-air temperature, V_{DD} = 5 V

	PARAMETER	TEST CO	NDITIONS	TA	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
					MIN	TYP	MAX	
				25°C		3.6		
			VIPP = 1 V	0°C		4		
0.0		$R_L = 10 k\Omega$,		70°C		3		Mar
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		2.9		V/µs
		g	VIPP = 2.5 V	0°C		3.1		
				70°C		2.5		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√ Hz
		$V_{O} = V_{OH},$	_	25°C		320		kHz
Вом	Maximum output-swing bandwidth		C _L = 20 pF, See Figure 1	0°C		340		
		INC = 10 K32,	See l'igure i	70°C		260		
				25°C		1.7		
В ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	C _L = 20 pF,	0°C		2		MHz
		Occ right o		70°C		1.3		
			(-	25°C		46°		
φm	Phase margin	V _I = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	0°C		47°		
		ο _L = 20 pi ,	occ rigule 5	70°C		43°		

operating characteristics at specified free-air temperature, $V_{\mbox{DD}}$ = 10 V

	PARAMETER		NDITIONS	TA	TLC272C, TLC272AC, TLC272BC, TLC277C			UNIT
					MIN	TYP	MAX	_
				25°C		5.3		
			V _{IPP} = 1 V	0°C		5.9		
0.0		$R_{L} = 10 k\Omega$,		70°C		4.3		V/µs
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		4.6		
			VIPP = 5.5 V	0°C		5.1		
				70°C		3.8		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√Hz
	Maximum output-swing bandwidth	$V_{O} = V_{OH}$	C _L = 20 pF, See Figure 1	25°C		200		kHz
вом				0°C		220		
		$R_{L} = 10 \text{ Ksz},$		70°C		140		
				25°C		2.2		
B ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	C _L = 20 pF,	0°C		2.5		MHz
		See Figure 5		70°C		1.8		
				25°C		49°		
φm	Phase margin	$V_{I} = 10 \text{ mV},$	f = B ₁ , See Figure 3	0°C		50°		
		$O_{L} = 20 \text{pr}$	occ rigule 5	70°C		46°		



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	PARAMETER	TEST CO	NDITIONS	TA	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
						TYP	MAX	
				25°C		3.6		
SR			VIPP = 1 V	-40°C		4.5		
	Class rate at write rain	$R_{L} = 10 k\Omega$		85°C		2.8		V/µs
	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		2.9		
		g	VIPP = 2.5 V	-40°C		3.5		
				85°C		2.3		
Vn	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√Hz
				25°C		320		kHz
Вом	Maximum output-swing bandwidth	$V_{O} = V_{OH},$ $R_{I} = 10 k\Omega,$	C _L = 20 pF, See Figure 1	-40°C		380		
		$R_{L} = 10 R_{22}$		85°C		250		
				25°C		1.7		
B ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	C _L = 20 pF,	-40°C		2.6		MHz
		See Figure 5		85°C		1.2		
				25°C		46°		
φm	Phase margin	V _I = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	−40°C		49°		1
		$C_{L} = 20 \text{pr},$	See Figure 5	85°C	1	43°		1

operating characteristics at specified free-air temperature, V_{DD} = 5 V

operating characteristics at specified free-air temperature, $V_{\mbox{DD}}$ = 10 V

	PARAMETER		NDITIONS	TA	TLC272I, TLC272AI, TLC272BI, TLC277I			UNIT
					MIN	TYP	MAX	_
				25°C		5.3		
			VIPP = 1 V	-40°C		6.8		
0.0		$R_L = 10 k\Omega$,		85°C		4		N// -
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		4.6		V/µs
		eee rigale r	V _{IPP} = 5.5 V	-40°C		5.8		-
				85°C		3.5		
۷ _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√Hz
		$V_{O} = V_{OH}$		25°C		200		kHz
Вом	Maximum output-swing bandwidth		C _L = 20 pF, See Figure 1	-40°C		260		
		NL = 10 K32,		85°C		130		
				25°C		2.2		
B ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	C _L = 20 pF,	-40°C		3.1		MHz
		See Figure 5		85°C		1.7		
				25°C		49°		
φm	Phase margin	V _I = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	-40°C		52°		
		ο <u> </u>		85°C		46°		



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operating characteristics at specified free-air temperature, $V_{DD} = 5 V$

		TEOT OO		-	TLC272	2M, TLC	277M	
	PARAMETER	TEST CO	NDITIONS	Τ _Α	MIN	TYP	MAX	UNIT
				25°C		3.6		
			V _{IPP} = 1 V	−55°C		4.7		
0.0	Slew rate at unity gain	$R_L = 10 k\Omega$,		125°C		2.3		V/µs
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		2.9		
			VIPP = 2.5 V	−55°C		3.7		
				125°C		2		
v _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√Hz
	Maximum output-swing bandwidth	$V_{O} = V_{OH},$ R _L = 10 kΩ,		25°C		320		kHz
вом			CL = 20 pF, See Figure 1	−55°C		400		
				125°C		230		
				25°C		1.7		
^B 1	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	C _L = 20 pF,	−55°C		2.9		MHz
		See Figure 5		125°C		1.1		
				25°C		46°		
φm	Phase margin	$V_{I} = 10 \text{ mV},$	f = B ₁ , See Figure 3	−55°C		49°		
		$O_{L} = 20 \text{pr}$	occ rigure o	125°C		41°		

operating characteristics at specified free-air temperature, V_{DD} = 10 V

	DADAMETED	TEST CO		-	TLC27	2M, TLC	277M	
	PARAMETER	1251 CO	NDITIONS	Τ _Α	MIN	TYP	MAX	UNIT
				25°C		5.3		
			VIPP = 1 V	−55°C		7.1		
00	Slow rate at upity gain	$R_L = 10 k\Omega$,		125°C		3.1		
SR	Slew rate at unity gain	C _L = 20 pF, See Figure 1		25°C		4.6		V/μs
			VIPP = 5.5 V	−55°C		6.1		
				125°C		2.7		
V _n	Equivalent input noise voltage	f = 1 kHz, See Figure 2	R _S = 20 Ω,	25°C		25		nV/√Hz
	Maximum output-swing bandwidth	$V_{O} = V_{OH},$ R _L = 10 kΩ,		25°C		200		kHz
Вом			C _L = 20 pF, See Figure 1	−55°C		280		
			See l'igure i	125°C		110		
				25°C		2.2		
B ₁	Unity-gain bandwidth	V _I = 10 mV, See Figure 3	CL = 20 pF,	−55°C		3.4		MHz
		Gee rigure 3		125°C		1.6		
				25°C		49°		
φm	Phase margin	V _I = 10 mV, C _L = 20 pF,	f = B ₁ , See Figure 3	−55°C		52°		
		ο <u></u> - 20 μ,		125°C		44°		



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operating characteristics, V_DD = 5 V, T_A = 25^{\circ}C

	PARAMETER	т	EST CONDITIO	Ne	TLC272Y			UNIT
	PARAMETER	I	TEST CONDITIONS				MAX	UNIT
0.0		R _L = 10 kΩ,	C _L = 20 pF,	V _{IPP} = 1 V		3.6		Mar
SR	Slew rate at unity gain	See Figure 1		VIPP = 2.5 V		2.9		V/µs
Vn	Equivalent input noise voltage	f = 1 kHz,	R _S = 20 Ω,	See Figure 2		25		nV/√Hz
BOM	Maximum output-swing bandwidth	V _O = V _{OH} , See Figure 1	C _L = 20 pF,	R _L = 10 kΩ,		320		kHz
B ₁	Unity-gain bandwidth	Vj = 10 mV,	C _L = 20 pF,	See Figure 3		1.7		MHz
φm	Phase margin	V _I = 10 mV, See Figure 3	f = B ₁ ,	C _L = 20 pF,		46°		

operating characteristics, V_{DD} = 10 V, T_A = 25°C

	PARAMETER	-	EST CONDITIO	Ne	TLC272Y			UNIT
	FARAMETER						MAX	UNIT
0.0	Slew rate at unity gain	RL = 10 kΩ,	C _L = 20 pF,	VIPP = 1 V		5.3		Mure
SR		See Figure 1		VIPP = 5.5 V		4.6		V/µs
Vn	Equivalent input noise voltage	f = 1 kHz,	R _S = 20 Ω,	See Figure 2		25		nV/√Hz
BOM	Maximum output-swing bandwidth	V _O = V _{OH} , See Figure 1	C _L = 20 pF,	R _L = 10 kΩ,		200		kHz
B ₁	Unity-gain bandwidth	Vj = 10 mV,	C _L = 20 pF,	See Figure 3		2.2		MHz
[¢] m	Phase margin	V _I = 10 mV, See Figure 3	f = B ₁ ,	C _L = 20 pF,		49°		

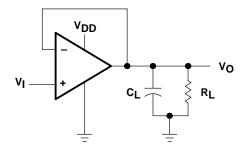


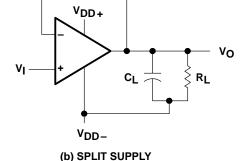
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PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

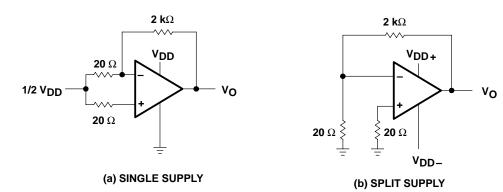
Because the TLC272 and TLC277 are optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.



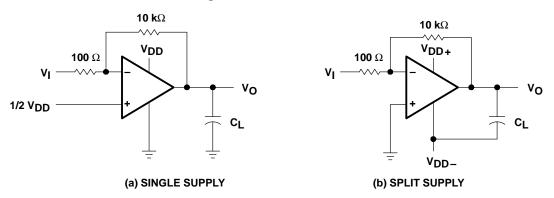
















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PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC272 and TLC277 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

- 1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 4). Leakages that would otherwise flow to the inputs are shunted away.
- 2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

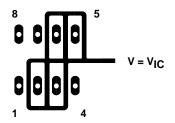


Figure 4. Isolation Metal Around Device Inputs (JG and P packages)

low-level output voltage

To obtain low-supply-voltage operation, some compromise was necessary in the input stage. This compromise results in the device low-level output being dependent on the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to Figures 14 through 19 in the Typical Characteristics of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance, which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.



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PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the operational amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 1. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 5). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

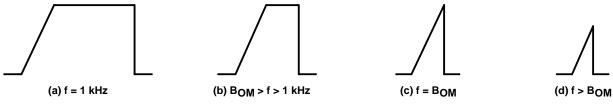


Figure 5. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.



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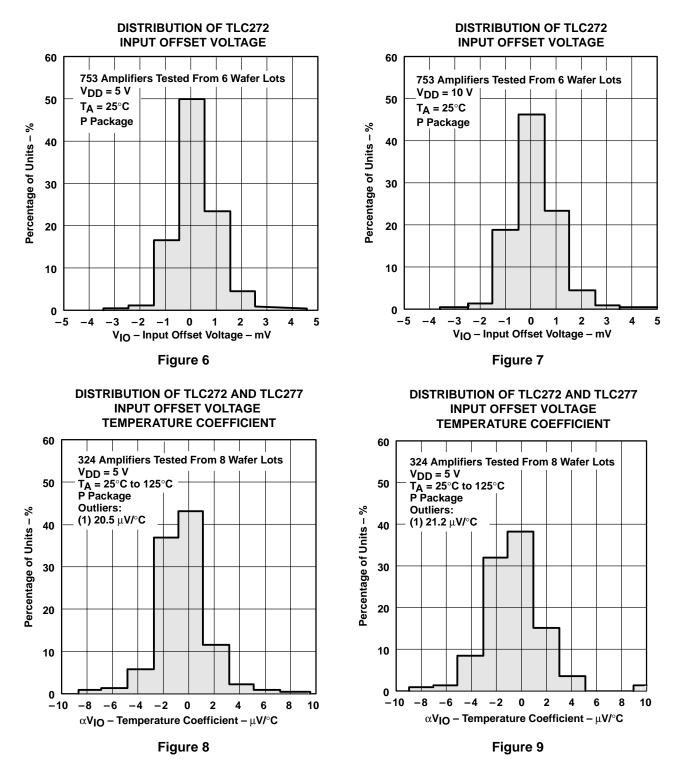
TYPICAL CHARACTERISTICS

Table of Graphs

			FIGURE
VIO	Input offset voltage	Distribution	6, 7
ανιο	Temperature coefficient of input offset voltage	Distribution	8, 9
∨он	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	10, 11 12 13
V _{OL}	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	14, 15 16 17 18, 19
Avd	Large-signal differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	20 21 32, 33
I _{IB}	Input bias current	vs Free-air temperature	22
IIO	Input offset current	vs Free-air temperature	22
VIC	Common-mode input voltage	vs Supply voltage	23
I _{DD}	Supply current	vs Supply voltage vs Free-air temperature	24 25
SR	Slew rate	vs Supply voltage vs Free-air temperature	26 27
	Normalized slew rate	vs Free-air temperature	28
V _{O(PP)}	Maximum peak-to-peak output voltage	vs Frequency	29
B ₁	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	30 31
φm	Phase margin	vs Supply voltage vs Free-air temperature vs Load capacitance	34 35 36
Vn	Equivalent input noise voltage	vs Frequency	37
	Phase shift	vs Frequency	32, 33



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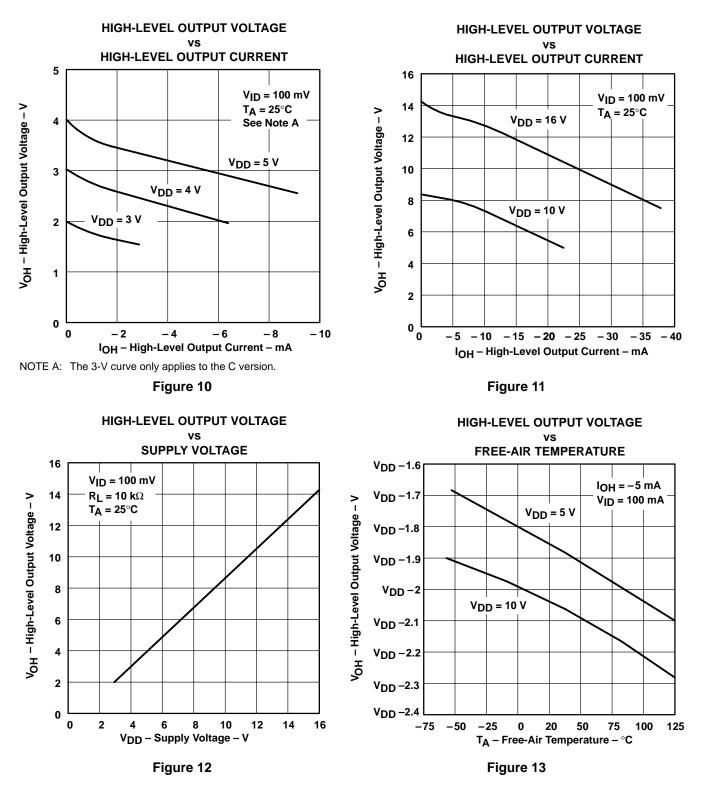


TYPICAL CHARACTERISTICS



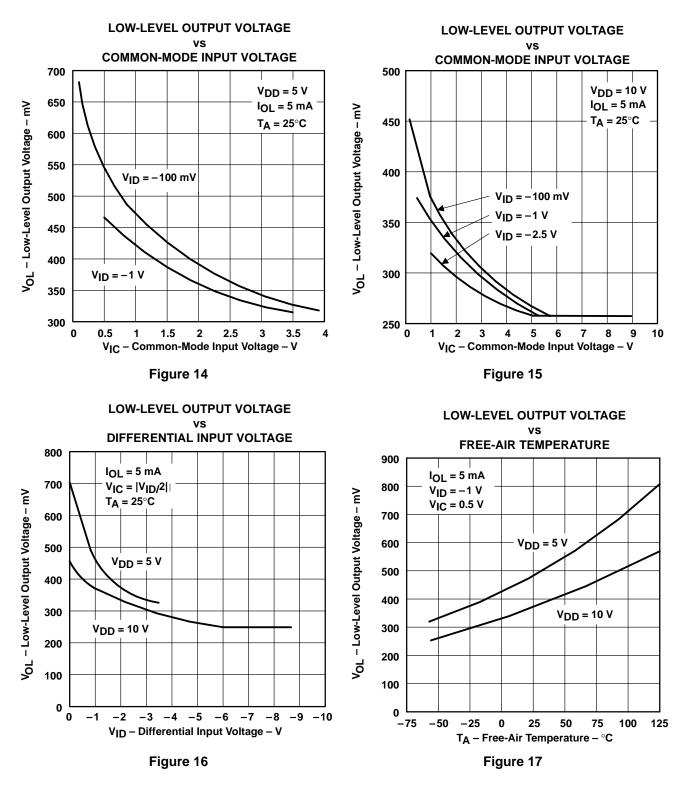
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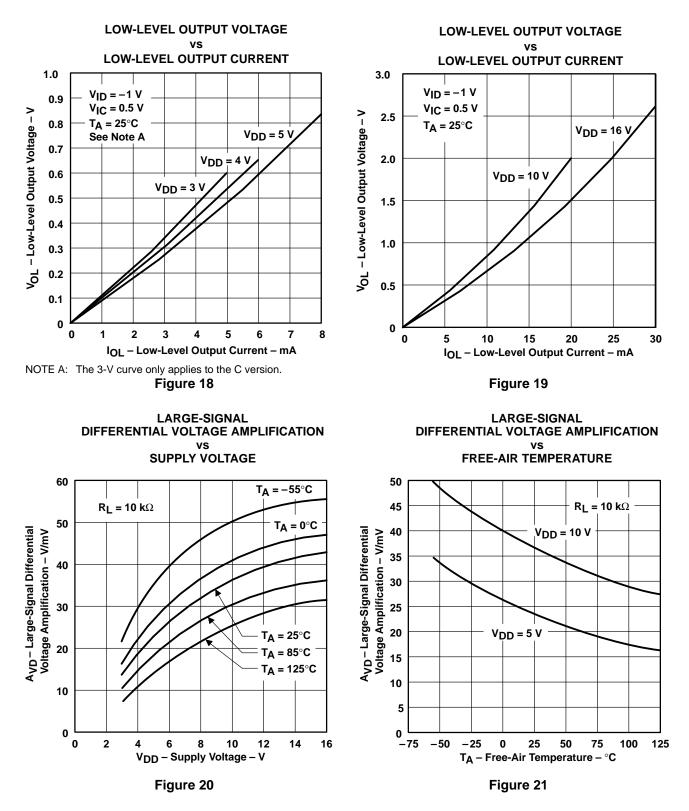
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TYPICAL CHARACTERISTICS[†]



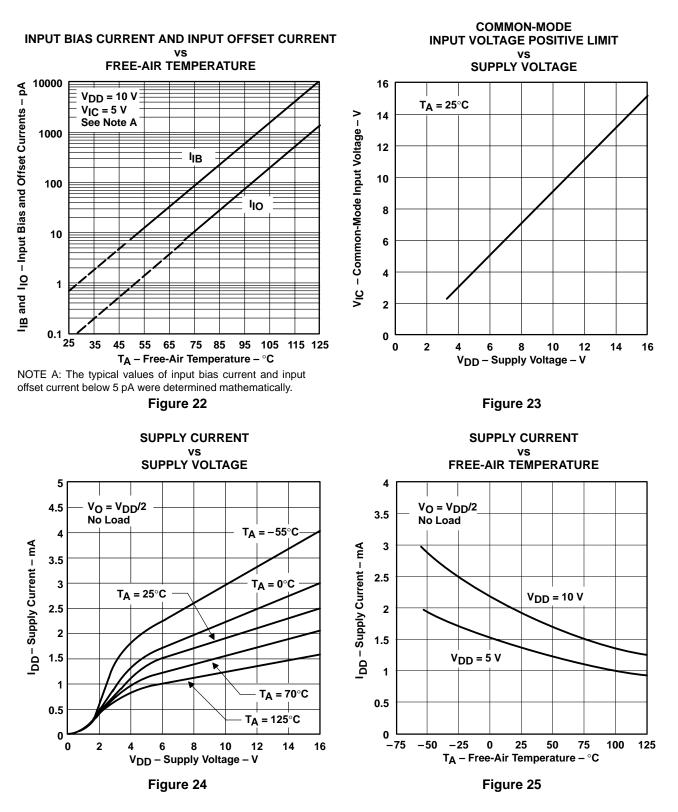
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TYPICAL CHARACTERISTICS[†]



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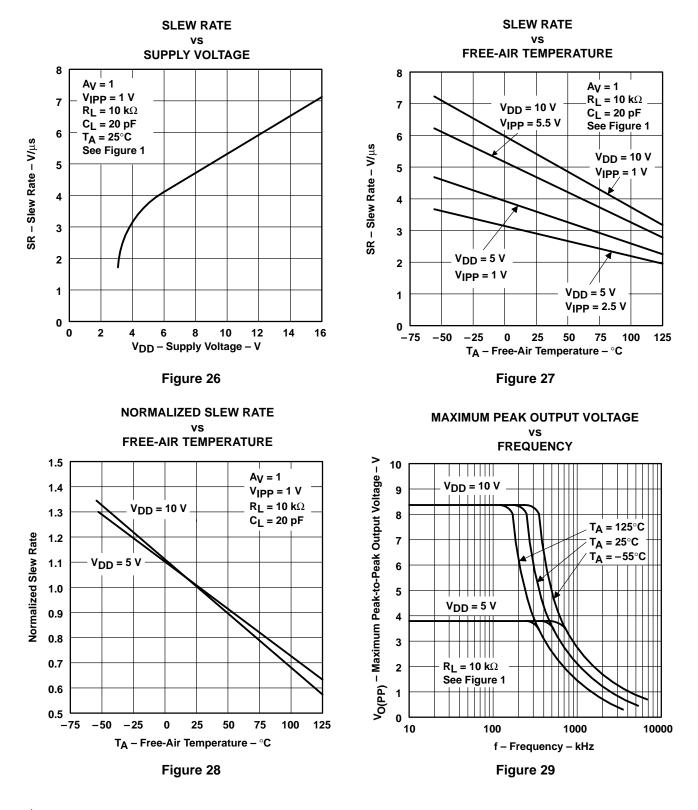


TYPICAL CHARACTERISTICS[†]



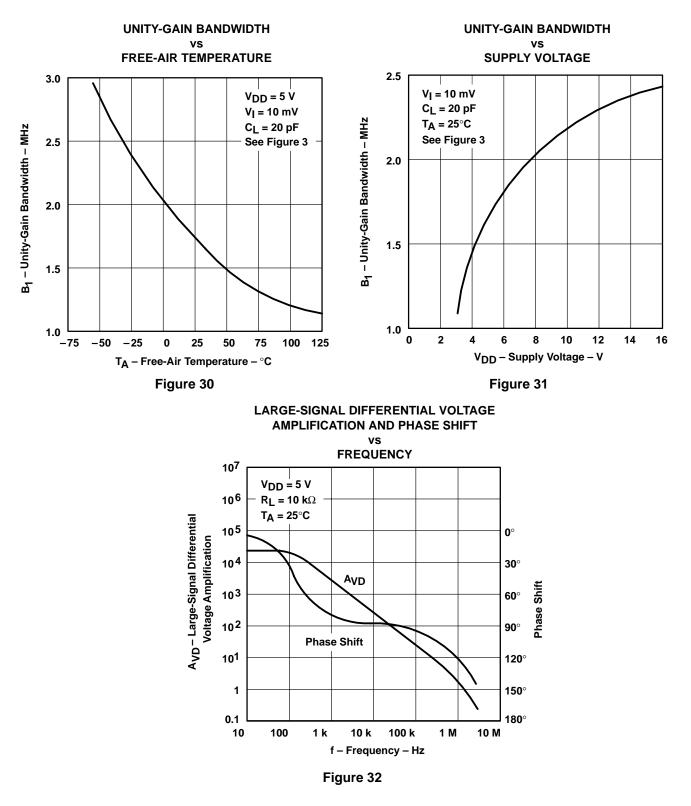
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TYPICAL CHARACTERISTICS[†]





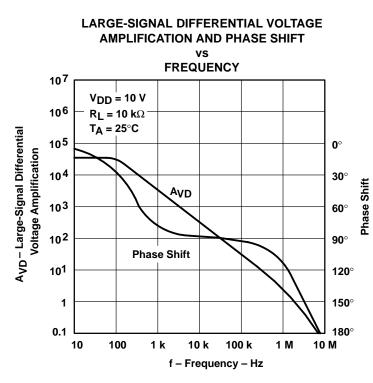
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TYPICAL CHARACTERISTICS[†]

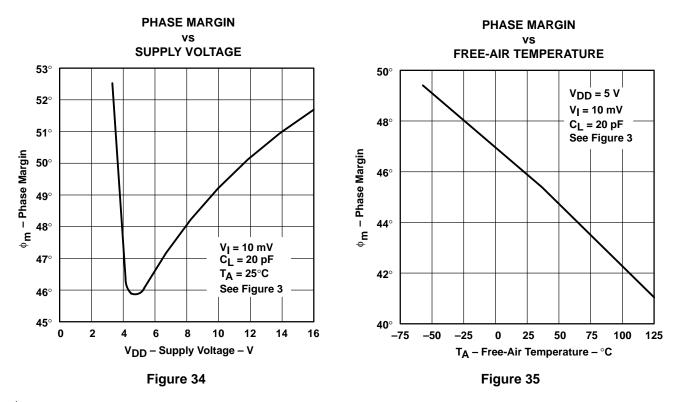


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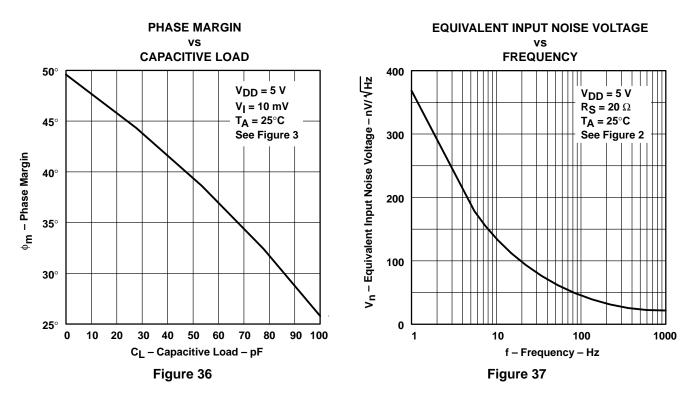
TYPICAL CHARACTERISTICS[†]







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TYPICAL CHARACTERISTICS



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APPLICATION INFORMATION

single-supply operation

While the TLC272 and TLC277 perform well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This design includes an input common-mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 38). The low input bias current of the TLC272 and TLC277 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC272 and TLC277 work well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

- 1. Power the linear devices from separate bypassed supply lines (see Figure 39); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
- 2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, high-frequency applications may require RC decoupling.

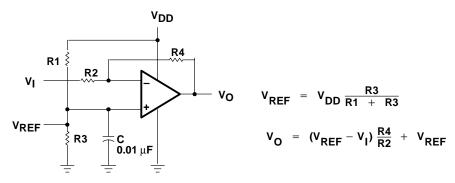
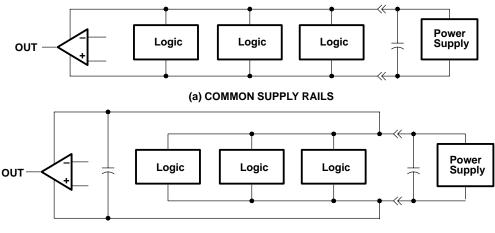


Figure 38. Inverting Amplifier With Voltage Reference



(b) SEPARATE BYPASSED SUPPLY RAILS (preferred)

Figure 39. Common vs Separate Supply Rails



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APPLICATION INFORMATION

input characteristics

The TLC272 and TLC277 are specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1$ V at $T_A = 25^{\circ}$ C and at $V_{DD} - 1.5$ V at all other temperatures.

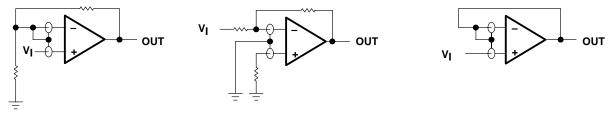
The use of the polysilicon-gate process and the careful input circuit design gives the TLC272 and TLC277 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically 0.1 μ V/month, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC272 and TLC277 are well suited for low-level signal processing; however, leakage currents on printed-circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 4 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 40).

Unused amplifiers should be connected as grounded unity-gain followers to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC272 and TLC277 result in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than 50 k Ω , since bipolar devices exhibit greater noise currents.



(a) NONINVERTING AMPLIFIER

(b) INVERTING AMPLIFIER

(c) UNITY-GAIN AMPLIFIER

Figure 40. Guard-Ring Schemes

output characteristics

The output stage of the TLC272 and TLC277 is designed to sink and source relatively high amounts of current (see typical characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

All operating characteristics of the TLC272 and TLC277 are measured using a 20-pF load. The devices can drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figure 41). In many cases, adding a small amount of resistance in series with the load capacitance alleviates the problem.



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APPLICATION INFORMATION

output characteristics (continued)

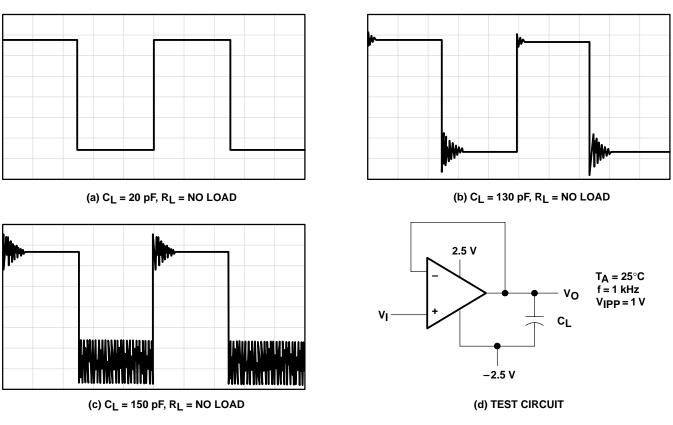


Figure 41. Effect of Capacitive Loads and Test Circuit

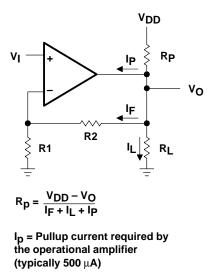
Although the TLC272 and TLC277 possess excellent high-level output voltage and current capability, methods for boosting this capability are available, if needed. The simplest method involves the use of a pullup resistor (R_P) connected from the output to the positive supply rail (see Figure 42). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on resistance between approximately 60 Ω and 180 Ω , depending on how hard the operational amplifier input is driven. With very low values of R_P , a voltage offset from 0 V at the output occurs. Second, pullup resistor R_P acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.



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APPLICATION INFORMATION

output characteristics (continued)



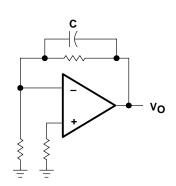


Figure 42. Resistive Pullup to Increase VOH



feedback

Operational amplifier circuits almost always employ feedback, and since feedback is the first prerequisite for oscillation, some caution is appropriate. Most oscillation problems result from driving capacitive loads (discussed previously) and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 43). The value of this capacitor is optimized empirically.

electrostatic discharge protection

The TLC272 and TLC277 incorporate an internal electrostatic discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

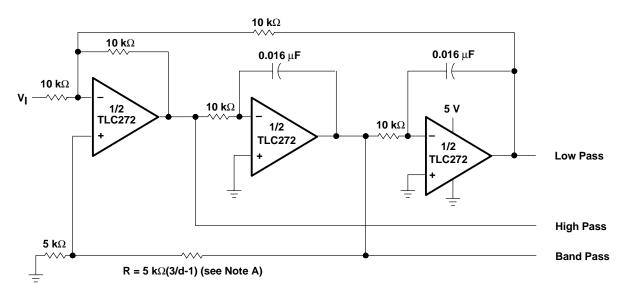
latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC272 and TLC277 inputs and outputs were designed to withstand –100-mA surge currents without sustaining latch-up; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not, by design, be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors (0.1 μ F typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.



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APPLICATION INFORMATION

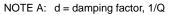


Figure 44. State-Variable Filter

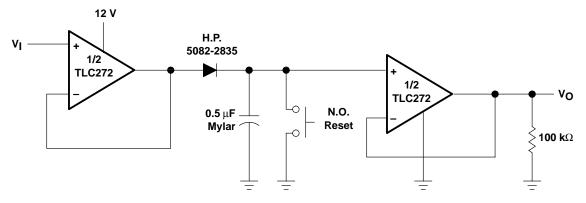
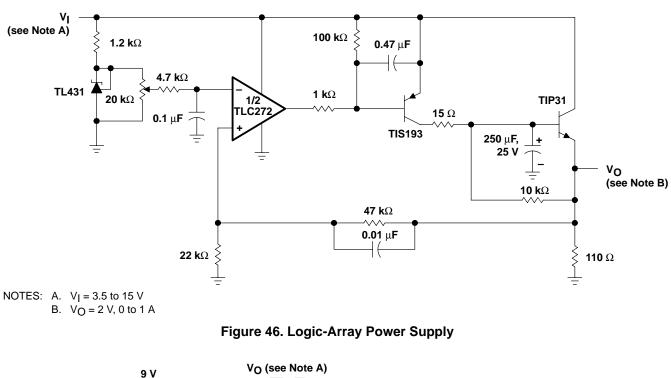


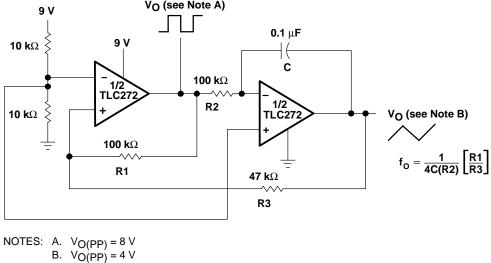
Figure 45. Positive-Peak Detector



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APPLICATION INFORMATION

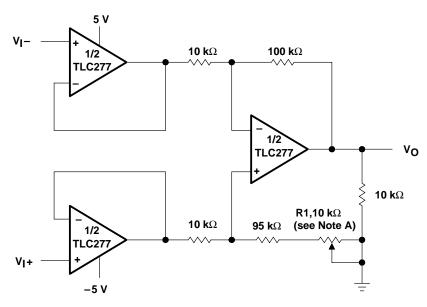






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APPLICATION INFORMATION



NOTE B: CMRR adjustment must be noninductive.

Figure 48. Low-Power Instrumentation Amplifier

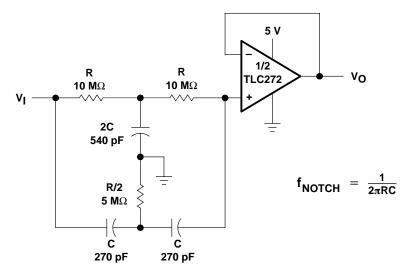


Figure 49. Single-Supply Twin-T Notch Filter





PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TLC272ACD	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	272AC
TLC272ACDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272AC
TLC272ACDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272AC
TLC272ACP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272ACP
TLC272ACP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272ACP
TLC272ACPE4	Active	Production	PDIP (P) 8	50 TUBE	-	Call TI	Call TI	0 to 70	
TLC272AID	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	272AI
TLC272AIDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272AI
TLC272AIDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272AI
TLC272AIDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272AI
TLC272AIDRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272AI
TLC272AIP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272AIP
TLC272AIP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272AIP
TLC272BCD	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	272BC
TLC272BCDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272BC
TLC272BCDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272BC
TLC272BCP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272BCP
TLC272BCP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272BCP
TLC272BCPE4	Active	Production	PDIP (P) 8	50 TUBE	-	Call TI	Call TI	0 to 70	
TLC272BCPS	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272B
TLC272BCPS.A	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272B
TLC272BID	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	272BI
TLC272BIDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272BI
TLC272BIDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272BI
TLC272BIDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272BI
TLC272BIDRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272BI
TLC272BIP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272BIP
TLC272BIP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272BIP
TLC272CD	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	0 to 70	272C



20-Aug-2025

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TLC272CDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272C
TLC272CDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	272C
TLC272CDR1G4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272C
TLC272CDR1G4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	272C
TLC272CDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	-	Call TI	Call TI	0 to 70	
TLC272CP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272CP
TLC272CP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC272CP
TLC272CPE4	Active	Production	PDIP (P) 8	50 TUBE	-	Call TI	Call TI	0 to 70	
TLC272CPS	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272
TLC272CPS.A	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272
TLC272CPSR	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272
TLC272CPSR.A	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272
TLC272CPWR	Active	Production	TSSOP (PW) 8	2000 LARGE T&R	Yes	NIPDAU NIPDAU	Level-1-260C-UNLIM	0 to 70	P272C
TLC272CPWR.A	Active	Production	TSSOP (PW) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P272C
TLC272ID	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	2721
TLC272IDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2721
TLC272IDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2721
TLC272IDR1G4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2721
TLC272IDR1G4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2721
TLC272IDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	-	Call TI	Call TI	-40 to 85	
TLC272IP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272IP
TLC272IP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC272IP
TLC272IPE4	Active	Production	PDIP (P) 8	50 TUBE	-	Call TI	Call TI	-40 to 85	
TLC277CD	Obsolete	Production	SOIC (D) 8		-	Call TI	Call TI	0 to 70	277C
TLC277CDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	277C
TLC277CDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	277C
TLC277CP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC277CP
TLC277CP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	0 to 70	TLC277CP
TLC277CPS	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P277
TLC277CPS.A	Active	Production	SO (PS) 8	80 TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P277
TLC277CPSR	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P277



Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking
	(1)	(2)			(3)	(4)	(5)		(6)
TLC277CPSR.A	Active	Production	SO (PS) 8	2000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	0 to 70	P277
TLC277ID	Obsolete	Production	SOIC (D) 8	-	-	Call TI	Call TI	-40 to 85	2771
TLC277IDR	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2771
TLC277IDR.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2771
TLC277IDRG4	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2771
TLC277IDRG4.A	Active	Production	SOIC (D) 8	2500 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2771
TLC277IP	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC277IP
TLC277IP.A	Active	Production	PDIP (P) 8	50 TUBE	Yes	NIPDAU	N/A for Pkg Type	-40 to 85	TLC277IP

⁽¹⁾ Status: For more details on status, see our product life cycle.

⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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PACKAGE OPTION ADDENDUM

20-Aug-2025

Texas

STRUMENTS

TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



All dimensions are nominal Device	- <u>r</u>	Package	Pine	SPQ	Reel	Reel	A0	B0	К0	P1	w	Pin1
Device	Туре	Drawing			Diameter		(mm)	(mm)	(mm)	(mm)	(mm)	Quadrant
	,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	J			(mm)	W1 (mm)	` '	· /	、 ,	` '	· ,	
TLC272ACDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272AIDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272BCDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272BIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272BIDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272CDR1G4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.4	12.0	16.0	Q1
TLC272CPWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TLC272IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC272IDR1G4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC277CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC277CPSR	SO	PS	8	2000	330.0	16.4	8.35	6.6	2.4	12.0	16.0	Q1
TLC277IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC277IDRG4	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



PACKAGE MATERIALS INFORMATION

30-Jul-2025



Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC272ACDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC272AIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC272AIDRG4	SOIC	D	8	2500	353.0	353.0	32.0
TLC272BCDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC272BIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC272BIDRG4	SOIC	D	8	2500	353.0	353.0	32.0
TLC272CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC272CDR1G4	SOIC	D	8	2500	353.0	353.0	32.0
TLC272CPSR	SO	PS	8	2000	353.0	353.0	32.0
TLC272CPWR	TSSOP	PW	8	2000	353.0	353.0	32.0
TLC272IDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC272IDR1G4	SOIC	D	8	2500	353.0	353.0	32.0
TLC277CDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC277CPSR	SO	PS	8	2000	353.0	353.0	32.0
TLC277IDR	SOIC	D	8	2500	353.0	353.0	32.0
TLC277IDRG4	SOIC	D	8	2500	353.0	353.0	32.0

TEXAS INSTRUMENTS

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TUBE



- B - Alignment groove width

*All dimensions are nomina	
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Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
TLC272ACP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272ACP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272AIP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272AIP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272BCP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272BCP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272BCPS	PS	SOP	8	80	530	10.5	4000	4.1
TLC272BCPS.A	PS	SOP	8	80	530	10.5	4000	4.1
TLC272BIP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272BIP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272CP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272CP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272CPS	PS	SOP	8	80	530	10.5	4000	4.1
TLC272CPS.A	PS	SOP	8	80	530	10.5	4000	4.1
TLC272IP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC272IP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC277CP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC277CP.A	Р	PDIP	8	50	506	13.97	11230	4.32
TLC277CPS	PS	SOP	8	80	530	10.5	4000	4.1
TLC277CPS.A	PS	SOP	8	80	530	10.5	4000	4.1
TLC277IP	Р	PDIP	8	50	506	13.97	11230	4.32
TLC277IP.A	Р	PDIP	8	50	506	13.97	11230	4.32

D0008A



PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.

- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



D0008A

EXAMPLE BOARD LAYOUT

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



D0008A

EXAMPLE STENCIL DESIGN

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.



MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.





NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



P(R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- A. All linear dimensions are in inches (millimeters).B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



PW0008A



PACKAGE OUTLINE

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice. 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153, variation AA.



PW0008A

EXAMPLE BOARD LAYOUT

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



PW0008A

EXAMPLE STENCIL DESIGN

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

9. Board assembly site may have different recommendations for stencil design.



^{8.} Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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