

具有平衡-非平衡变压器的 LMH9226 单端至差分 2.3GHz 至 2.9GHz 射频放大器

1 特性

- 单通道、单端输入至差分输出射频增益块放大器
- 支持 2.6GHz 中心频率，具有 400MHz、1dB 带宽
- 在 1dB 带宽下、匹配 $Z_{LOAD} = 50\Omega$ 时，具有 17dB 典型增益
- 在 1dB 带宽下噪声系数 $< 3\text{dB}$
- 在 2dBm 双音输出功率、匹配 $Z_{LOAD} = 50\Omega$ 时，具有 35dBm OIP3
- 匹配 $Z_{LOAD} = 50\Omega$ 时，具有 17.5dBm 输出 P1dB
- 3.3V 单电源供电，具有 275mW 功耗
- 工作温度高达 $T_A = 105^\circ\text{C}$

2 应用

- 5G m-MIMO 基站
- 有源天线系统，m-MIMO (AAS)
- 小型蜂窝基站
- TDD/FDD 蜂窝基站
- 无线基础设施
- 低成本无线电设备
- 单端至差分转换
- 平衡-非平衡变压器替代产品
- 射频增益块
- 适用于 GPS ADC 的差分驱动器

3 说明

LMH9226 是一款高性能、单通道、单端 50Ω 输入至差分 50Ω 或 100Ω 输出的射频增益块放大器，支持 2.3GHz 至 2.9GHz 频段。该器件非常符合 5G m-MIMO 或小型蜂窝基站 应用的要求。该器件集成了具有无源平衡-非平衡变压器的单端输入和输出射频增益块功能，它主要用于接收器信号链末级，以驱动模拟至差分转换器 (ADC) 差分输入的满量程电压。

LMH9226 提供 17dB 的典型增益，且在 2.6GHz 下具有 35dBm 输出 IP3 的出色线性性能，而且噪声系数在整个 400MHz 1dB 带宽范围内低于 3dB。该器件的单端输入内部匹配 50Ω 阻抗。差分输出可轻松连接 50Ω 阻抗，无需任何外部匹配电路。如需匹配 100Ω 阻抗，则需要外部匹配电路，这通常会在 2.6GHz 下产生 0.3dB 增益损失。

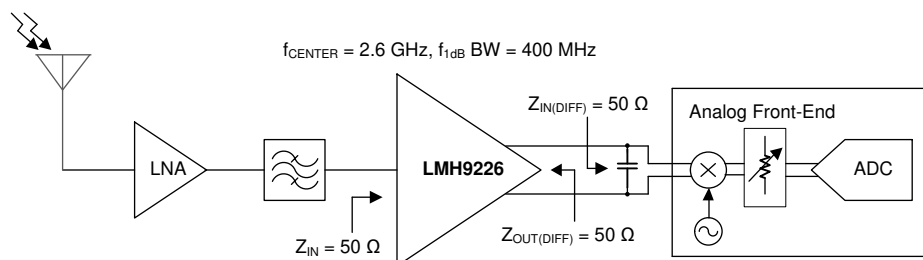
该器件使用 3.3V 单电源供电，产生旁路功耗约 275mW，因此适用于高密度 5G 大规模多输入多输出 (MIMO) 应用。此外，该器件采用节省空间的 2mm × 2mm、12 引脚 WQFN 封装。该器件的额定工作温度高达 105°C ，可提供稳健的系统设计。符合 JEDEC 标准的 1.8V 断电引脚可为该器件快速断电和上电，适用于时分双工 (TDD) 系统。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
LMH9226	WQFN (12)	2.00mm × 2.00mm

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。

LMH9226: 2.3GHz 至 2.9GHz 单端输入至差分输出射频增益块放大器



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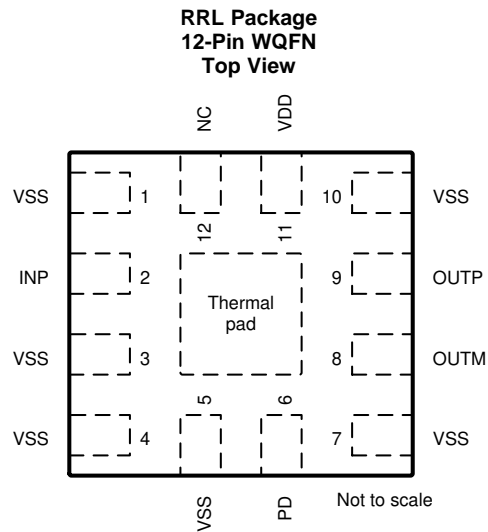
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4 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

日期	修订版本	说明
2019 年 12 月	*	初始发行版。

5 Pin Configuration and Functions



Pin Functions

PIN		I/O	DESCRIPTION
NO.	NAME		
1	VSS	Power	Analog ground
2	INP	Input	RF single-ended input into amplifier
3	VSS	Power	Analog ground
4	VSS	Power	Analog ground
5	VSS	Power	Analog ground
6	PD	Input	Power-down connection. PD = 0 V, normal operation; PD = 1.8 V, power off mode.
7	VSS	Power	Analog ground
8	OUTM	Output	RF single-ended output negative
9	OUTP	Output	RF single-ended output positive
10	VSS	Power	Analog ground
11	VDD	Power	Positive supply voltage (3.3 V)
12	NC	—	Do not connect this pin
Thermal Pad		—	Connect the thermal pad to ground (VSS).

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Supply voltage	VDD	−0.3	3.6	V
RF pins	INP, OUTP, OUTM	−0.3	VDD	V
Continuous wave (CW) input	$f_{in} = 2.6$ GHz at INP		25	dBm
Digital input pin	PD	−0.3	VDD	V
Junction temperature	T_J		150	°C
Storage temperature	T_{stg}	−65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±1000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±500	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
VDD	Supply voltage	3.15	3.3	3.45	V
T_A	Ambient temperature	−40		105	°C
T_J	Junction temperature	−40		125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		LMH9226	UNIT
		RRL PKG	
		12-PIN WQFN	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	74.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	72.4	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	37.1	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	3.2	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	37.1	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	14.2	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Electrical Characteristics

$T_A = +25^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$, center frequency (f_{in}) = 2.6 GHz, single-ended input impedance (Z_{in}) = 50 Ω , differential output impedance (Z_{LOAD}) = 50 Ω , $P_{OUT(TOTAL)} = 8\text{ dBm}$ into $Z_{LOAD} = 50\text{ }\Omega$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
RF PERFORMANCE						
f_{RF}	RF frequency range		2300		2900	MHz
BW_{1dB}	1-dB bandwidth	Center Frequency (f_{in}) = 2.6 GHz		400		MHz
S_{21}	Gain	$f_{in} = 2.6\text{ GHz}$		17		dB
NF	Noise figure	$f_{in} = 2.6\text{ GHz}$, $R_S = 50\text{ }\Omega$		3		dB
OIP1	Output P1dB	$f_{in} = 2.6\text{ GHz}$, $R_{LOAD} = 50\text{ }\Omega$		17.5		dBm
OIP3	Output IP3	$f_{in} = 2.6\text{ GHz} \pm 10\text{ MHz spacing}$, $P_{OUT/TONE} = 2\text{ dBm}$		35		dBm
	Differential output gain imbalance ⁽¹⁾			0.5		dB
	Differential output phase imbalance ⁽¹⁾			4		degree
S_{11}	Input return loss	$f_{in} = 2.6\text{ GHz}$, $BW = 400\text{ MHz}$		-11		dB
Z_{in}	Single ended input reference impedance			50		Ω
S_{22}	Differential output return loss	$f_{in} = 2.6\text{ GHz}$, $BW = 400\text{ MHz}$		-12		dB
Z_{LOAD}	Differential output reference impedance			50		Ω
S_{12}	Reverse isolation	$f_{in} = 2.6\text{ GHz}$		-35		dB
CMRR	Common-mode rejection ratio ⁽²⁾			27		dB
SWITCHING AND DIGITAL INPUT CHARACTERISTICS						
t_{ON}	Turn-on time	PD pin = 1.8 V to 0 V, $f_{in} = 2.6\text{ GHz}$		0.5		μs
t_{OFF}	Turn-off time	PD pin = 0 V to 1.8 V, $f_{in} = 2.6\text{ GHz}$		0.2		μs
V_{IH}	High-level input voltage ⁽³⁾	At the PD pin	1.4			V
V_{IL}	Low-level input voltage ⁽³⁾	At the PD pin			0.5	V
I_{IH}	High-level input current ⁽³⁾	At the PD pin		28	60	μA
I_{IL}	Low-level input current ⁽³⁾	At the PD pin		10	30	μA
DC CURRENT AND POWER CONSUMPTION						
I_{VDD_ON}	Supply current ⁽³⁾	PD pin = 0 V		84	100	mA
I_{VDD_PD}	Power-down current ⁽³⁾	PD pin = 1.8 V			10	mA
P_{dis}	Power dissipation	$V_{DD} = 3.3\text{ V}$		275		mW

(1) Measured at $f_{in} = 2.6\text{ GHz}$, over the BW_{1dB}

(2) CMRR is calculated using $(S_{21}-S_{31})/(S_{21}+S_{31})$ for receive (1 is input port, 2 and 3 are differential output ports)

(3) 100% tested at $T_A = 25^\circ\text{C}$

6.6 Typical Characteristics

at $T_A = 25^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$, center frequency (f_{IN}) = 2.6 GHz, single-ended input impedance (Z_{IN}) = $50\ \Omega$, differential output impedance (Z_{LOAD}) = $50\ \Omega$, and $P_{OUT(TOTAL)} = 8\text{ dBm}$ into $Z_{LOAD} = 50\ \Omega$ (unless otherwise noted)

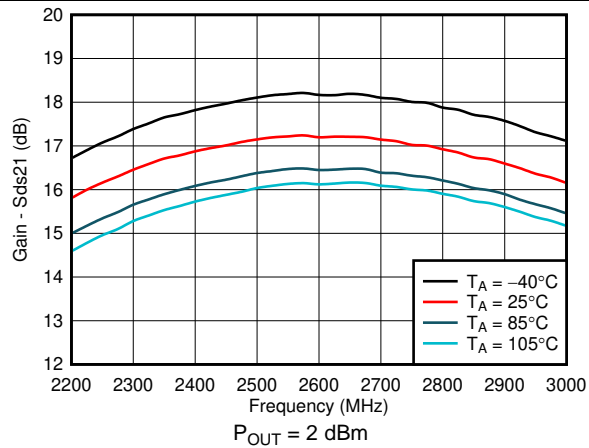


图 1. Gain vs Frequency and Temperature

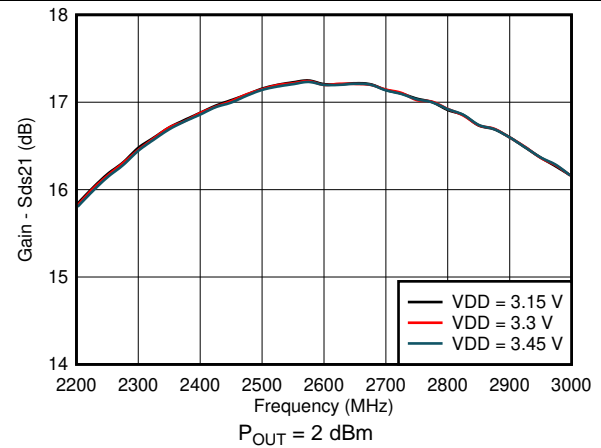


图 2. Gain vs Frequency and Supply Voltage

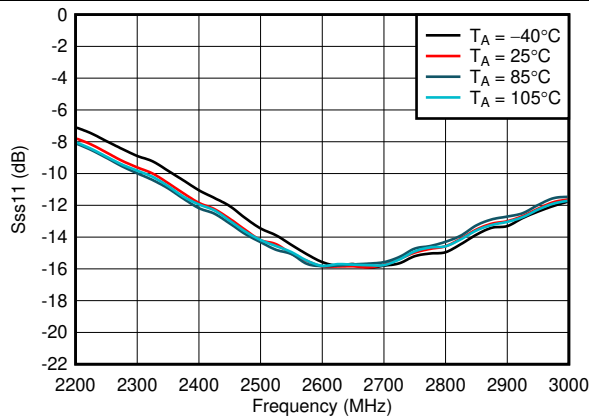


图 3. Input Return Loss vs Frequency

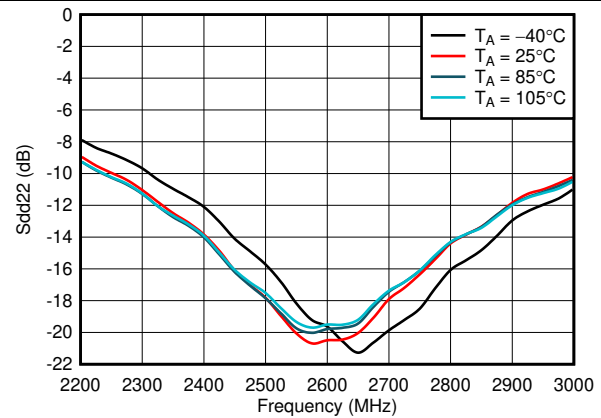


图 4. Output Return Loss vs Frequency

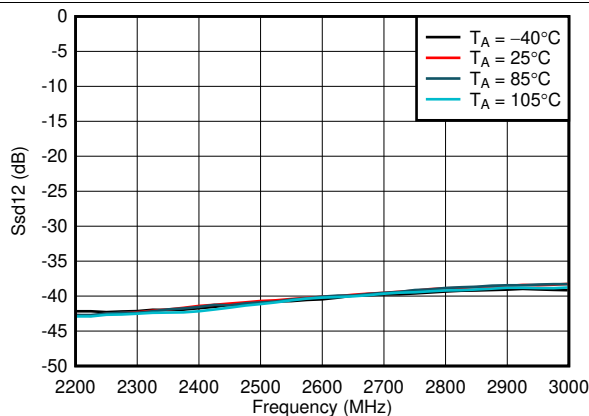


图 5. Reverse Isolation vs Frequency

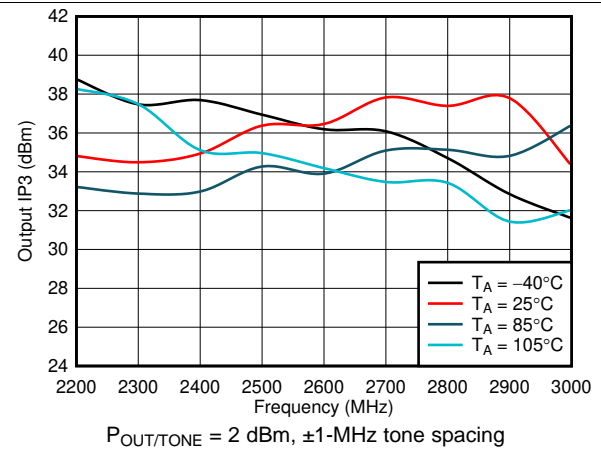


图 6. Output IP3 vs Frequency and Temperature

Typical Characteristics (接下页)

at $T_A = 25^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$, center frequency (f_{IN}) = 2.6 GHz, single-ended input impedance (Z_{IN}) = 50 Ω , differential output impedance (Z_{LOAD}) = 50 Ω , and $P_{OUT(TOTAL)} = 8\text{ dBm}$ into $Z_{LOAD} = 50\text{ }\Omega$ (unless otherwise noted)

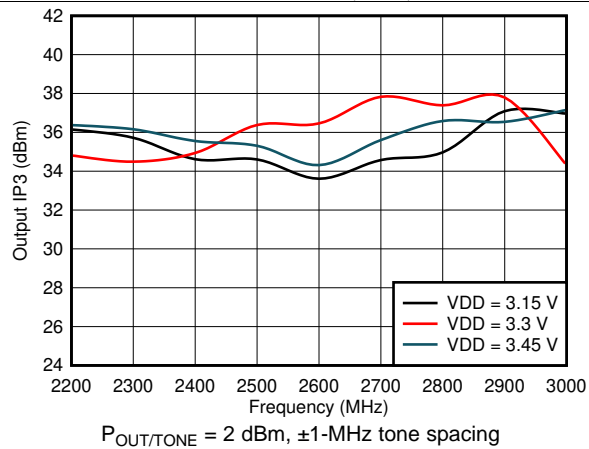


图 7. Output IP3 vs Frequency and Supply Voltage

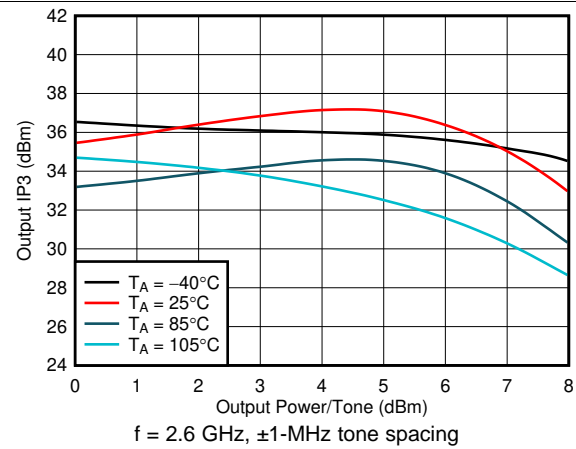


图 8. Output IP3 vs Output Power per Tone

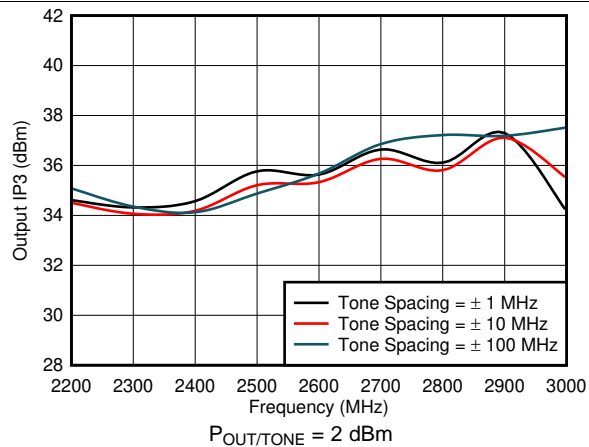


图 9. Output IP3 vs Frequency and Tone Spacing

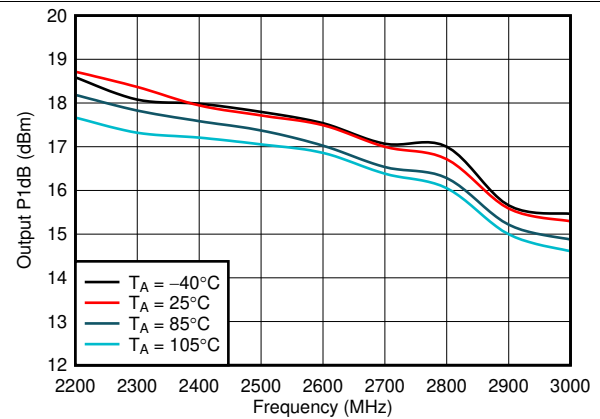


图 10. Output P1dB vs Frequency and Temperature

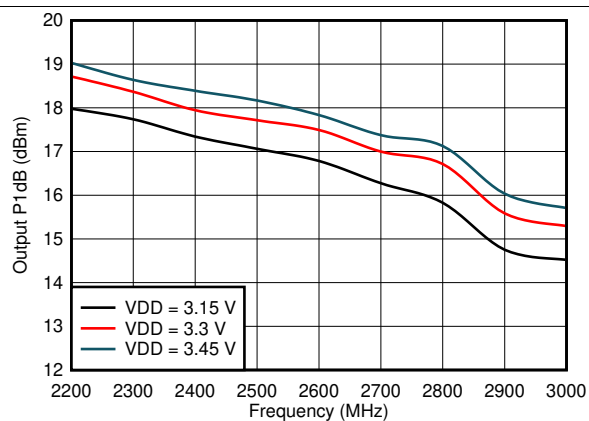


图 11. Output P1dB vs Frequency and Supply Voltage

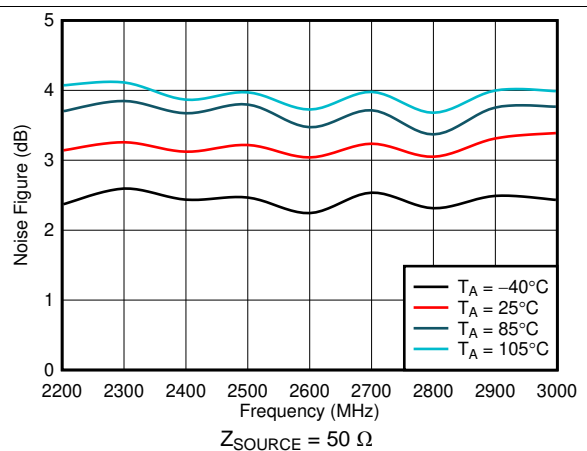


图 12. Noise Figure vs Frequency and Temperature

Typical Characteristics (接下页)

at $T_A = 25^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$, center frequency (f_{IN}) = 2.6 GHz, single-ended input impedance (Z_{IN}) = $50\ \Omega$, differential output impedance (Z_{LOAD}) = $50\ \Omega$, and $P_{OUT(TOTAL)} = 8\text{ dBm}$ into $Z_{LOAD} = 50\ \Omega$ (unless otherwise noted)

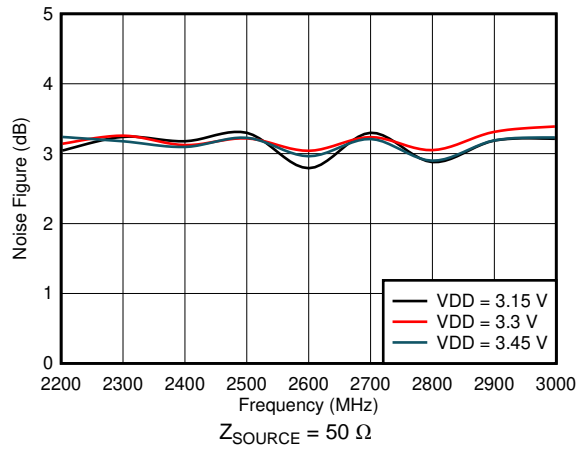


图 13. Noise Figure vs Frequency and Supply Voltage

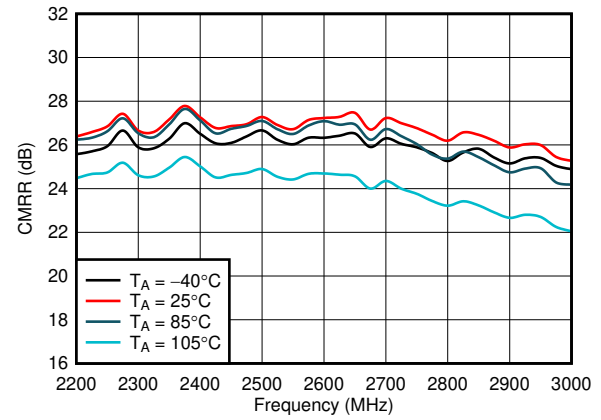


图 14. CMRR vs Frequency

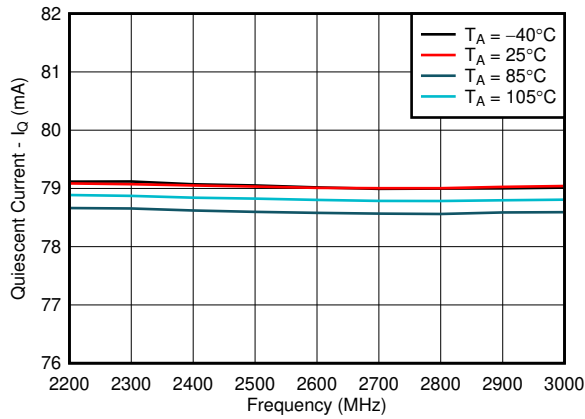


图 15. Quiescent Current vs Frequency and Temperature

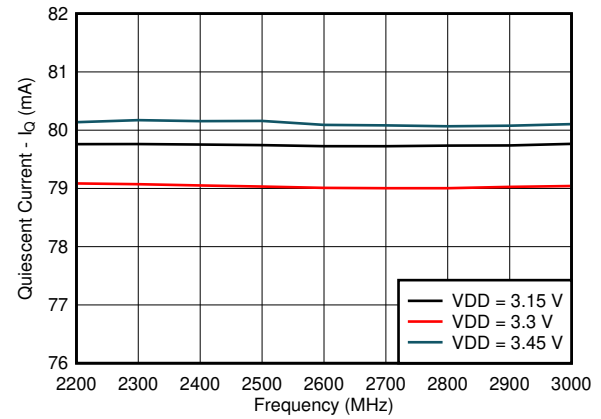


图 16. Quiescent Current vs Frequency and Supply Voltage

7 Detailed Description

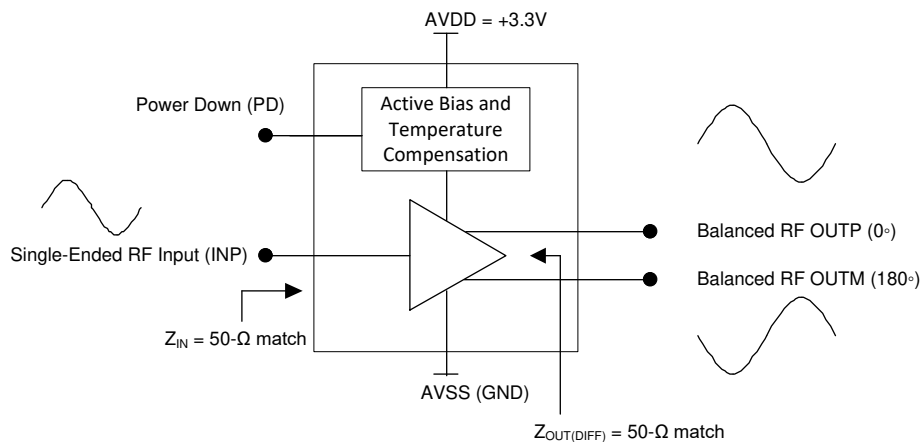
7.1 Overview

The LMH9226 is single-ended, 50-Ω input to differential 50-Ω or 100-Ω output RF gain block amplifier used in 2.3-GHz to 2.9-GHz, frequency-band, 5G, m-MIMO TDD receiver applications. The device provides a 17-dB fixed power gain with excellent linearity and noise performance across 400 MHz of the 1-dB bandwidth at the 2.6-GHz center frequency. The device is internally matched for a 50-Ω input impedance at 2.6 GHz. The device differential output can be matched to the 50-Ω impedance without external matching circuitry, or to the 100-Ω impedance with external matching circuitry (see the [Application and Implementation](#) section for details). The device is typically used in the final stage of a receive signal chain to drive the differential input of an analog-to-differential converter (ADC), while providing additional gain to a low-noise amplifier (LNA) to increase dynamic range and the required single-ended to differential conversion.

The LMH9226 has an on-chip active bias circuitry to maintain device performance over a wide temperature and supply voltage range. The included power-down function allows the amplifier to shut down and save power when the amplifier is not needed. Fast shut-down and start-up enable the amplifier to be used in a host of time division duplex (TDD) applications.

Operating on a single 3.3-V supply and consuming 84 mA of typical supply current, the device is available in a 2-mm × 2-mm, 12-pin WQFN package.

7.2 Functional Block Diagram



7.3 Feature Description

As shown in [Figure 17](#), the LMH9226 integrates the functionality of a single-ended RF amplifier and passive balun in a traditional receive application, achieving a small form factor with good linearity and noise performance. The active balun implementation, along with a higher operating temperature of 105°C, allows for a more robust receiver system implementation compared to a passive balun that is prone to reliability failures at high temperatures. The high-temperature operation is achieved by the on-chip, active bias circuitry that maintains device performance over a wide temperature and supply voltage range.

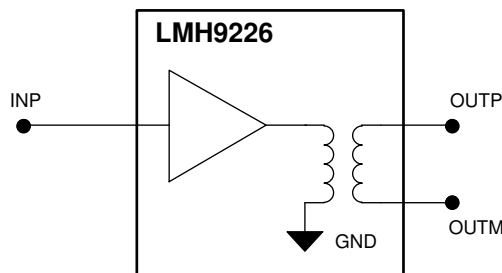


图 17. Single-Ended Input to Differential Output, Active Balun Implementation

7.4 Device Functional Modes

The LMH9226 features a PD pin that must be connected to GND for normal operation. For power-down mode, connect the PD pin to a logic high voltage of 1.8 V.

8 Application and Implementation

注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LMH9226 is a single-ended, 50-Ω input to differential 50-Ω or 100-Ω output RF gain block amplifier, used in the receive path of a 2.6-GHz center frequency, 5G, TDD m-MIMO or small cell base station. The device replaces the traditional single-ended RF amplifier and passive balun offering a smaller footprint solution to the customer. TI recommends following good RF layout and grounding techniques to maximize the device performance.

8.2 Typical Application

The LMH9226 is typically used in a four transmit and four receive (4T/4R) array of active antenna system for 5G, TDD, wireless base station applications. Such a system is shown in 图 18, where the LMH9226 is used in the receive path as the final stage differential driver to an ADC input. TI typically recommends reducing the trace distance between the LMH9226 output and the ADC input to minimize amplitude and phase imbalance during the single-to-differential conversion.

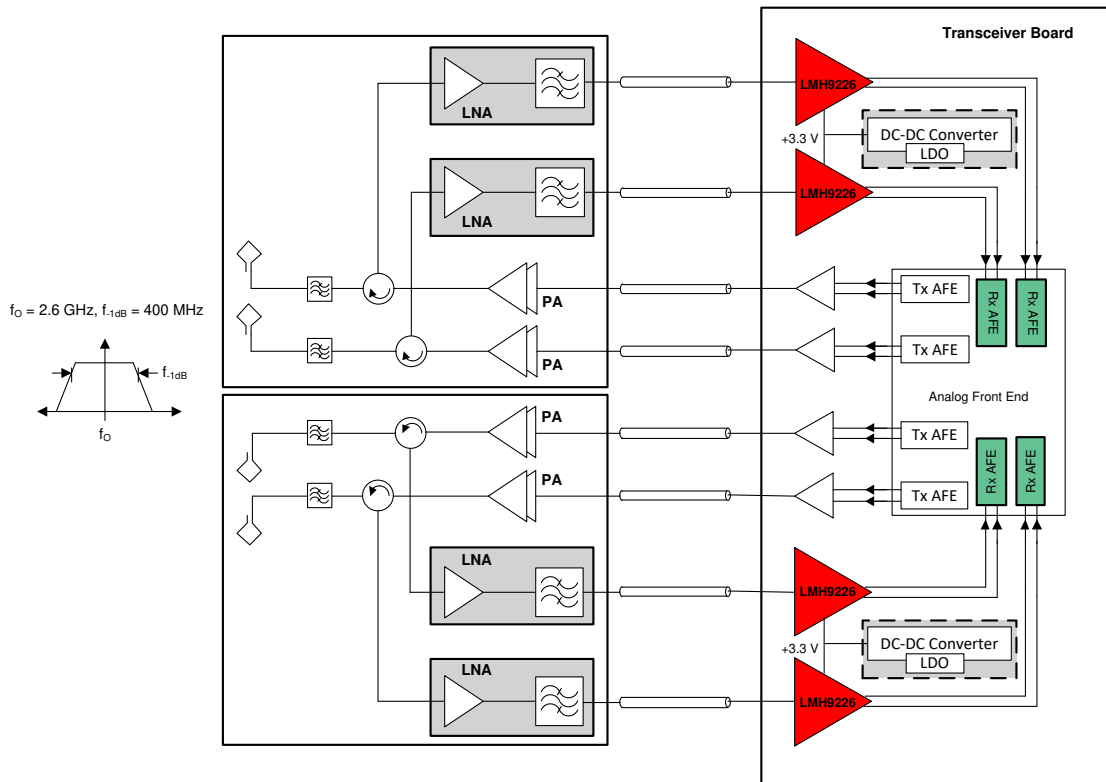


图 18. LMH9226 in a 4T/4R 5G Active Antenna System

Typical Application (接下页)

The 4T/4R system is easily scaled to 16T/16R, 64T/64R, or higher antenna arrays that result in proportional scaling of the overall system power dissipation. As a result of the proportional scaling factor for multiple channels in a system, the individual device power consumption must be reduced to dissipate less overall heat in the system. Operating on a single 3.3-V supply, the LMH9226 consumes only 275 mW and therefore provides power saving to the customer. Multiple LMH9226 devices can be powered from a single DC/DC converter or a low-dropout regulator (LDO) operating on a 3.3-V supply. A DC/DC converter provides the most power efficient way of generating the 3.3-V supply. However, care must be taken when using the DC/DC converter to minimize the switching noise using inductor chokes and adequate isolation must be provided between the analog and digital supplies.

8.2.1 Design Requirements

表 1 shows example design requirements for an RF amplifier in a typical 5G, active antenna TDD system. The LMH9226 meets these requirements.

表 1. Design Parameters

DESIGN PARAMETERS	EXAMPLE VALUE
Frequency range and 1-dB BW	2300 MHz to 2900 MHz with 400 MHz of 1-dB BW
Configuration	Single-ended 50-Ω input to differential 50-Ω output
Power gain	> 15 dB
Output IP3 at $P_{OUT/TONE} = 2$ dBm	> 32 dBm
Noise figure at $Z_{in} = 50 \Omega$	< 4 dB
Output P1dB	< 17 dBm
Power consumption	< 350 mW
Turn-on time	< 1 μ s
Package size	2 mm × 2 mm ²

8.2.2 Detailed Design Procedure

The LMH9226 is a single-to-differential RF gain block amplifier for a 2.6-GHz center frequency application with 400 MHz of the 1-dB bandwidth. 图 19 shows a single receive channel consisting of a low-noise amplifier (LNA) that sits close to the antenna and drives the signal into a single-ended, 50-Ω coaxial cable that then connects to a transceiver board. The LMH9226 that sits at the transceiver board input converts this single-ended signal received from the coax cable into a differential signal, thereby offering low noise and distortion performance while interfacing with the receiver analog front-end (AFE). The LMH9226 input impedance must be matched to 50 Ω to prevent any signal reflections resulting from the coax cable. The device differential output interfaces directly with the differential input of an AFE. The output matching is optimized for a 50-Ω output at the 2.6-GHz center frequency with 400 MHz of the 1-dB bandwidth. The AFE input impedance must be matched to 50 Ω at 2.6 GHz as well to prevent any ripple in the frequency response.

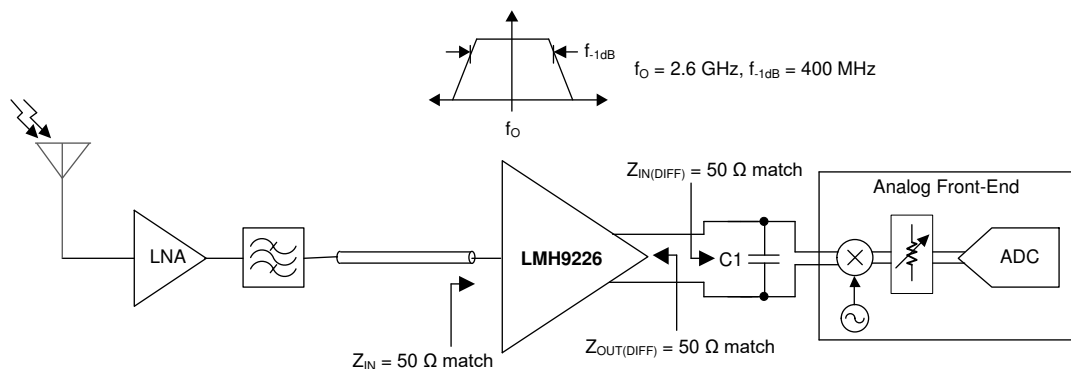


图 19. LMH9226 in a Receive Application Driving an AFE ($Z_{OUT(DIFF)} = 50 \Omega$)

For interfacing with a 100- Ω differential input AFE, as shown in 图 20, an external matching circuitry is needed close to the LMH9226 output. 表 2 lists example recommended component values when transforming the LMH9226 output impedance from 50 Ω to 100 Ω . The component values must be tweaked on the board, depending on the trace length between the matching circuitry and the AFE input to maintain 400 MHz of the 1-dB BW at the 2.6-GHz center frequency. LC component values must be selected with $Q(\min) > 30$ that have a self resonant frequency (SRF) sufficiently higher than the desired frequency of operation. 图 21 and 图 22 provide a comparison of device performance when interfacing with a 50- Ω output matching as compared to a 100- Ω output matching. As depicted in 图 21, the forward path gain (S_{DS21}) is slightly lower for the 100- Ω differential output impedance because of the extra loss in the external matching circuitry.

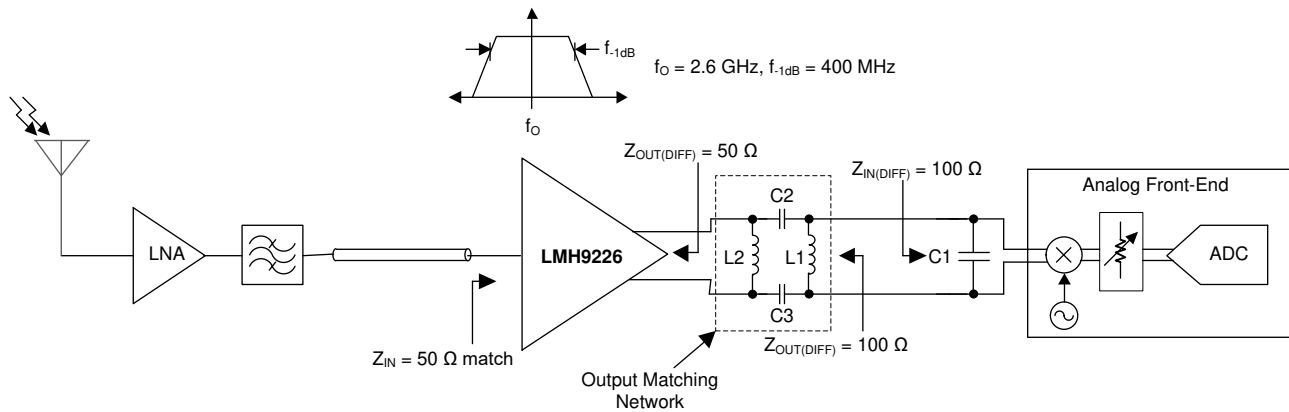


图 20. LMH9226 in a Receive Application Driving an AFE ($Z_{OUT(DIFF)} = 100 \Omega$)

表 2. Output Matching Network Component Values

COMPONENT	VALUE
C2, C3	2.2 pF
L1	6.2 nH
L2	Do not install (DNI)

Following the recommended RF layout with good quality RF components and local DC bypass capacitors ensures optimal performance is achieved. TI provides various support materials including S-parameter and ADS models to allow the design to be optimized to the application-specific performance needs.

8.2.3 Application Curves

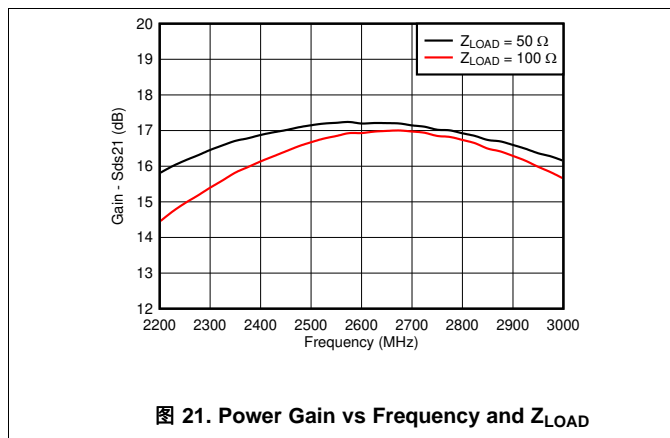


图 21. Power Gain vs Frequency and Z_{LOAD}

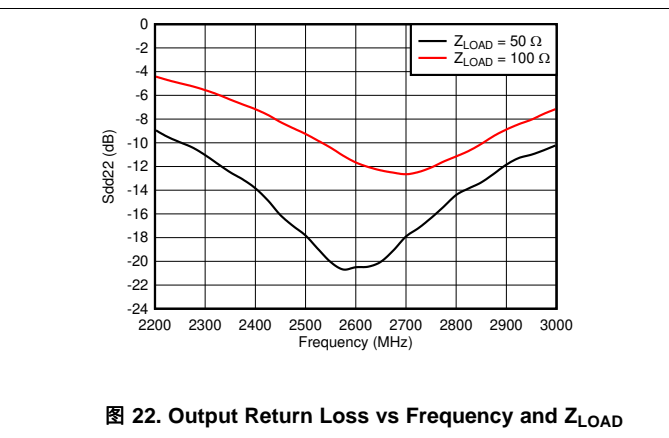


图 22. Output Return Loss vs Frequency and Z_{LOAD}

9 Power Supply Recommendations

The LMH9226 operates on a common nominal 3.3-V supply voltage. The supply voltage is recommended to be isolated through the decoupling capacitors placed close to the device. Select capacitors with a self-resonant frequency near the application frequency. When multiple capacitors are used in parallel to create a broadband decoupling network, place the capacitor with the higher self-resonant frequency closer to the device.

The LMH9226 can be powered from a DC/DC converter or an LDO operating on a 3.3-V supply. A DC/DC converter provides the most power efficient way of generating the 3.3-V supply. However, care must be taken when using the DC/DC converter to minimize the switching noise from inductor chokes and adequate isolation must be provided between the analog and digital supplies.

10 Layout

10.1 Layout Guidelines

When dealing with an RF amplifier with relatively high gain and a center frequency of 2.6 GHz, certain board layout precautions must be taken to ensure stability and optimum performance. TI recommends that the LMH9226 board be multi-layered to improve thermal performance, grounding, and power-supply decoupling. [Figure 23](#) shows a good layout example. In [Figure 23](#), only the top signal layer and its adjacent ground reference plane are shown.

- Excellent electrical connection from the thermal pad to the board ground is essential. Use the recommended footprint, solder the pad to the board, and do not include a solder mask under the pad.
- Connect the pad ground to the device terminal ground on the top board layer.
- Verify that the return DC and RF current path have a low impedance ground plane directly under the package and that the RF signal traces into and out of the amplifier.
- Ensure that ground planes on the top and any internal layers are well stitched with vias.
- Do not route RF signal lines over breaks in the reference ground plane.
- Avoid routing clocks and digital control lines near RF signal lines.
- Do not route RF or DC signal lines over noisy power planes. Ground is the best reference, although clean power planes can serve where necessary.
- Place supply decoupling close to the device.
- The differential output traces must be symmetrical in order to achieve the best linearity performance.

A board layout software package can simplify the trace thickness design to maintain impedances for controlled impedance signals. To isolate the affect of board parasitic on frequency response, TI recommends placing the external output matching resistors close to the amplifier output pins. See the [LMH9226 Evaluation Module user guide](#) for more details on board layout and design.

10.2 Layout Example

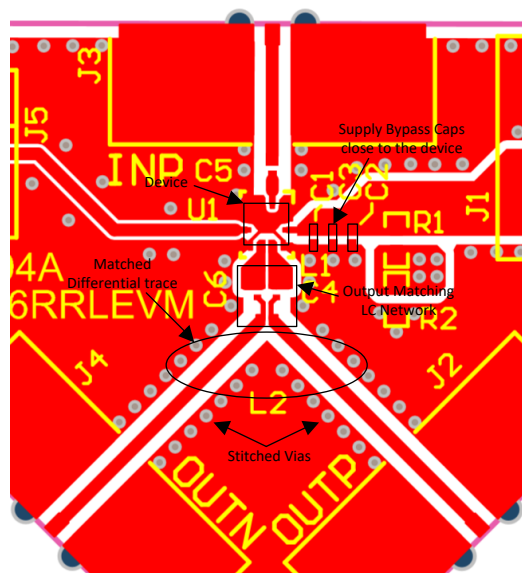


图 23. Supply Bypass and Output Matching

11 器件和文档支持

11.1 文档支持

11.1.1 相关文档

请参阅如下相关文档：

德州仪器 (TI)，《LMH9226 评估模块》 用户指南

11.2 接收文档更新通知

要接收文档更新通知，请导航至 ti.com.cn 上的器件产品文件夹。单击右上角的通知我进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

11.3 社区资源

TI E2E™ support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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11.5 静电放电警告



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ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

11.6 Glossary

[SLYZ022](#) — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LMH9226IRRLR	Active	Production	WQFN (RRL) 12	3000 LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 105	22GO
LMH9226IRRLR.B	Active	Production	WQFN (RRL) 12	3000 LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 105	22GO

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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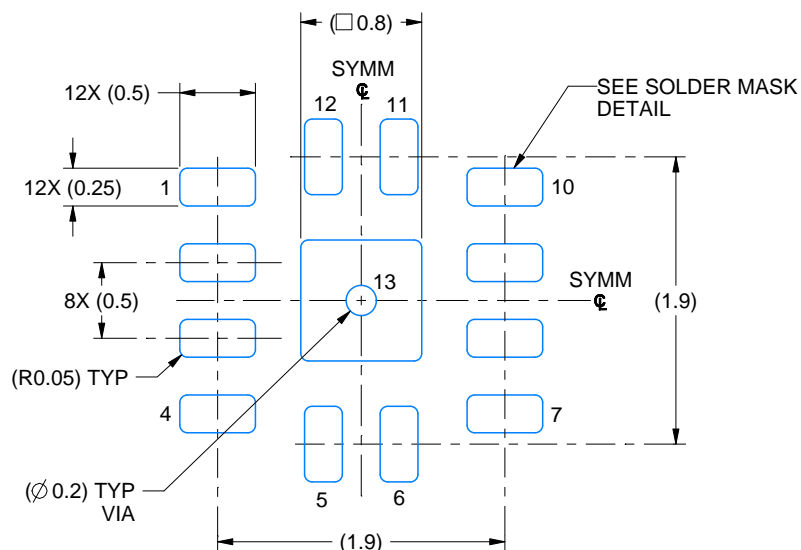
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

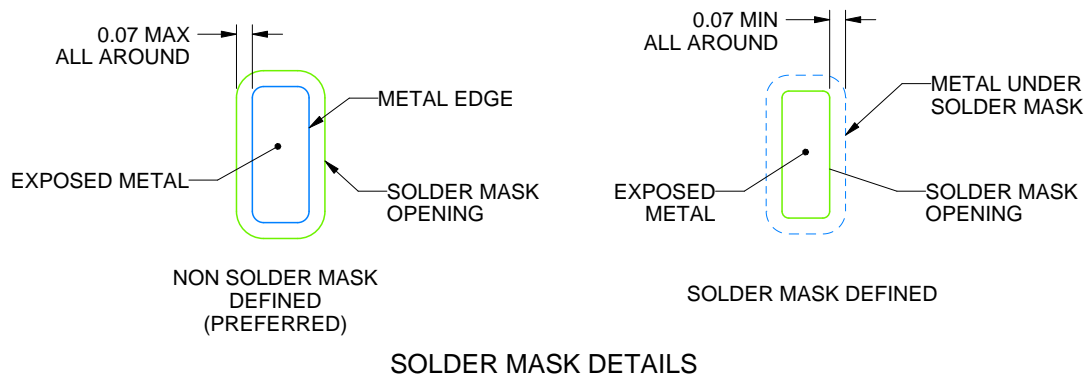
RRL0012A

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 20X



4224942/A 04/2019

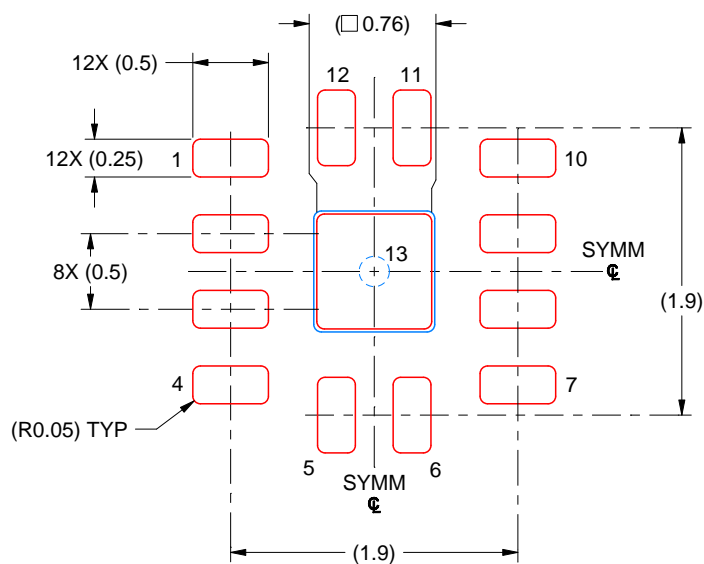
NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

RRL0012A

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 MM THICK STENCIL
SCALE: 20X

EXPOSED PAD 13
90% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

4224942/A 04/2019

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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