

具有集成平衡-非平衡变压器的 LMH9135 3.2GHz 至 4.2GHz 差分至单端放大器

1 特性

- 单通道、窄带差分输入至单端输出射频增益块放大器
- 支持 3.2GHz 至 4.2GHz 1dB BW 典型值
- 在整个频带内具有 18dB 的典型增益
- 3.8dB 噪声系数
- 31.5dBm OIP3
- 18dBm 输出 P1dB
- +3.3V 单电源供电，具有 395mW 功耗
- 工作温度高达 105°C T_C

2 应用

- 适用于 GPS DAC 的差分 DAC 输出驱动器
- 差分至单端放转换
- 平衡-非平衡变压器替代产品
- 小型蜂窝或 m-MIMO 基站
- 5G 有源天线系统 (AAS)
- 无线蜂窝基站

3 说明

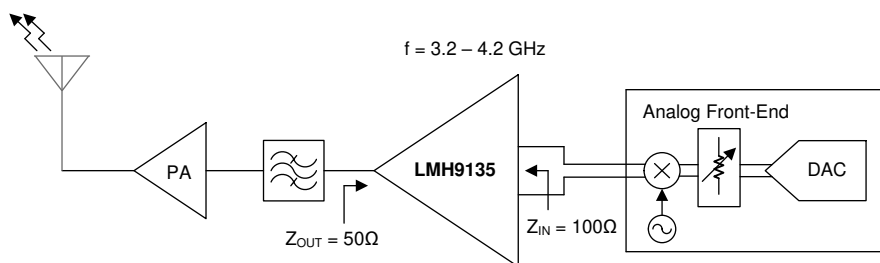
LMH9135 是一款高性能、单通道、差分输入至单端输出传输射频增益块放大器，支持 3.2GHz 至 4.2GHz 频段。该器件可满足下一代 5G 有源天线系统 (AAS) 或小型蜂窝应用的要求，同时可驱动功率放大器 (PA) 输入端。该射频放大器可提供 18dB 的典型增益，并具有 +31.5dBm 输出 IP3 的出色线性性能，同时在整个 1dB 带宽内保持小于 4dB 的噪声系数。该器件内部匹配 100 Ω 差分输入阻抗，可轻松与输入端的射频采样或零中频模拟前端 (AFE) 相连。此外，该器件内部匹配 50 Ω 单端输出阻抗，可与后置放大器、表面声波 (SAW) 滤波器或功率放大器 (PA) 轻松交互。

该器件使用 3.3V 单电源供电，其典型有功功率约为 395mW，因此适用于高密度 5G 大规模 (MIMO) 应用。此外，该器件采用节省空间的 2mm x 2mm、12 引脚 QFN 封装。该器件的额定工作温度高达 105°C，可提供稳健的系统设计。该器件具有符合 JEDEC 标准的 1.8V 断电引脚，可为该器件快速断电和上电，适用于时分双工 (TDD) 系统。

器件信息 (1)

器件型号	封装	封装尺寸 (标称值)
LMH9135	WQFN (12)	2.00mm x 2.00mm

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。



LMH9135 : 差分至单端放大器



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4 Revision History

注：以前版本的页码可能与当前版本的页码不同

DATE	REVISION	NOTES
August 2020	*	Initial Release

5 Pin Configuration and Functions

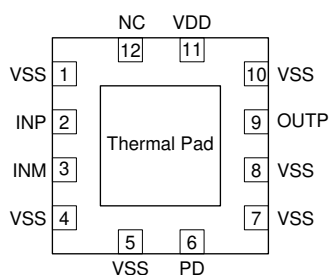


图 5-1. RRL Package 12-Pin WQFN Top View

Pin Functions

PIN		I/O	DESCRIPTION
NO.	NAME		
1	VSS	Power	Ground
2	INP	Input	RF differential positive input into amplifier
3	INM	Power	RF differential negative input into amplifier
4	VSS	Power	Ground
5	VSS	Power	Ground
6	PD	Input	Power down connection. PD = 0 V = normal operation; PD = 1.8 V = power off mode
7	VSS	Power	Ground
8	VSS	Output	Ground
9	OUTP	Output	RF single-ended output from amplifier
10	VSS	Power	Ground
11	VDD	Power	Positive supply voltage (3.3 V)
12	NC	—	Do not connect this pin
Thermal Pad		—	Connect the thermal pad to Ground

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Supply voltage	VDD	– 0.3	3.6	V
RF Pins	INP, INM, OUTP	– 0.3	VDD	V
Digital Input PIN	PD	– 0.3	VDD	V
Continuous wave (CW) input	T = 25 °C		18	dBm
T _J	Junction temperature		150	°C
T _{stg}	Storage temperature	– 65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±1000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±500	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
VDD	Supply voltage	3.15	3.3	3.45	V
T _C	Case (bottom) temperature	– 40		105	°C
T _J	Junction temperature	– 40		125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		DEVICE	UNIT
		PKG DES (PKG FAM)	
		PINS	
R _{θJA}	Junction-to-ambient thermal resistance	74.8	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	72.4	°C/W
R _{θJB}	Junction-to-board thermal resistance	37.1	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	3.2	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	37.1	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	14.2	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Electrical Characteristics

$T_A = +25^\circ\text{C}$, $V_{DD} = 3.3\text{V}$, Frequency (f_{in}) = 3.5 GHz, Differential Input Impedance (Z_{IN}) = $100\ \Omega$, Output Load (Z_{LOAD}) = $50\ \Omega$ (unless otherwise noted)

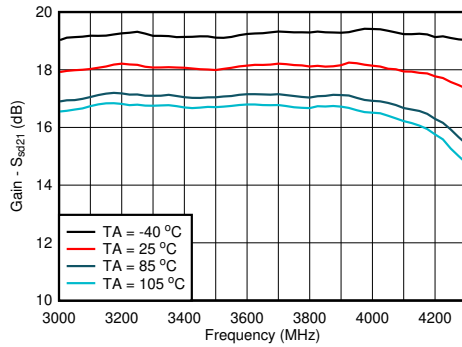
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
RF PERFORMANCE - LMH9135						
F_{RF}	RF frequency range		3200		4200	MHz
BW_{1dB}	1dB Bandwidth			1000		MHz
S_{21}	Gain			18		dB
NF	Noise Figure	$R_S = 100\ \Omega$ differential		3.8		dB
OP1dB	Output P1dB	$Z_{LOAD} = 50\ \Omega$		18		dBm
OIP3	Output IP3	$f_{in} = 3.5\text{ GHz} \pm 5\text{ MHz Spacing}$, $P_{OUT}/TONE = 2\text{ dBm}$		31.5		dBm
	Differential Input Gain Imbalance			± 0.5		dB
	Differential Input Phase Imbalance			± 4		degree
S11	Input return loss ⁽¹⁾	$f_{in} = 3.3 - 3.8\text{ GHz}$		-10		dB
		$f_{in} = 3.3 - 4.2\text{ GHz}$		-10		dB
S22	Output return loss ⁽¹⁾	$f_{in} = 3.3 - 3.8\text{ GHz}$		-10		dB
		$f_{in} = 3.3 - 4.2\text{ GHz}$		-10		dB
S12	Reverse isolation			35		dB
CMRR	Common Mode Rejection Ratio ⁽²⁾			30		dB
Switching and Digital input characteristics						
t_{ON}	Turn-ON time	50% VPD to 90% RF		0.2		μs
t_{OFF}	Tun-OFF time	50% VPD to 10% RF		0.2		μs
V_{IH}	High-Level Input Voltage	PD pin	1.4			V
V_{IL}	Low-Level Input Voltage	PD pin			0.5	V
DC current and Power Consumption						
I_{VDD_ON}	Supply Current			120		mA
I_{VDD_PD}	Power Down Current			10		mA
P_{dis}	Power Dissipation			395		mW

(1) Reference impedance: Input = $100\ \Omega$ differential, Output = $50\ \Omega$ single-ended

(2) CMRR is calculated using $(S_{12} - S_{13})/(S_{12} + S_{13})$ for Transmit (1 is output port, 2 & 3 are differential input ports)

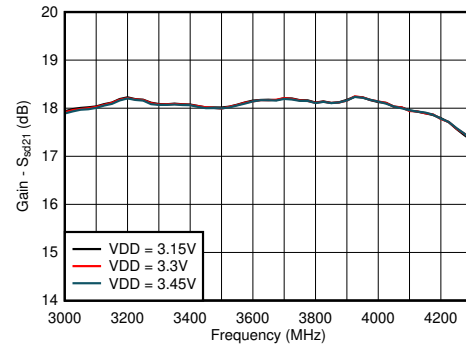
6.6 Typical Characteristics

At $T_A = 25^\circ\text{C}$, $V_{DD} = 3.3\text{ V}$, differential input impedance (Z_{IN}) = $100\ \Omega$, single-ended output impedance (Z_{LOAD}) = $50\ \Omega$ (unless otherwise noted).



$P_{OUT} = 2\text{ dBm}$

图 6-1. Gain vs Frequency and Temperature



$P_{OUT} = 2\text{ dBm}$

图 6-2. Gain vs Frequency and Supply Voltage

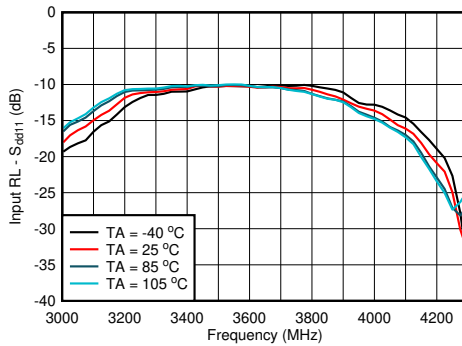


图 6-3. Input Return Loss vs Frequency

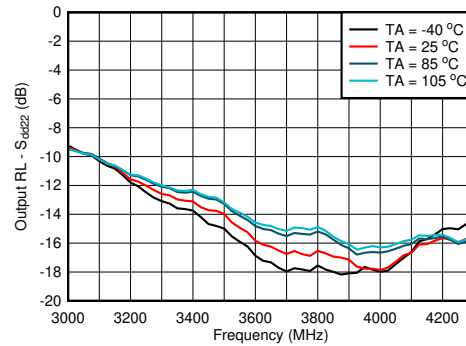


图 6-4. Output Return Loss vs Frequency

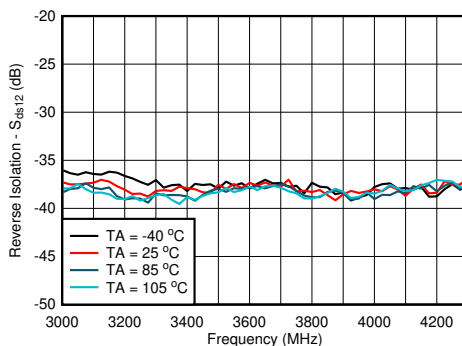
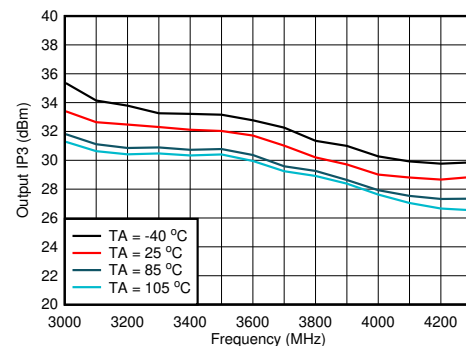
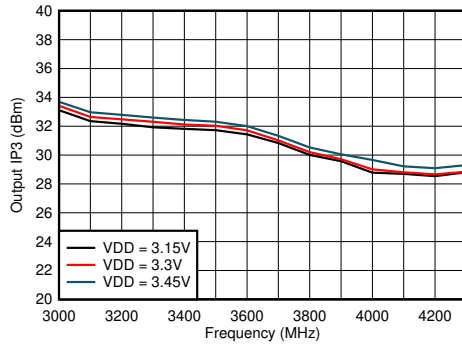


图 6-5. Reverse Isolation vs Frequency



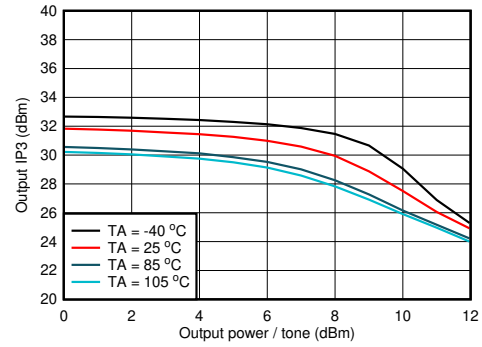
$P_{OUT/TONE} = 2\text{ dBm}$, 10-MHz tone spacing

图 6-6. Output IP3 vs Frequency and Temperature



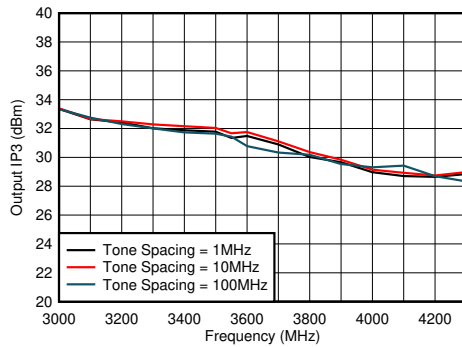
$P_{OUT/TONE} = 2 \text{ dBm}$, 10-MHz tone spacing

图 6-7. Output IP3 vs Frequency and Supply Voltage



$f = 3.5 \text{ GHz}$, 10-MHz tone spacing

图 6-8. Output IP3 vs Output Power per Tone



$P_{OUT/TONE} = 2 \text{ dBm}$

图 6-9. Output IP3 vs Frequency and Tone Spacing

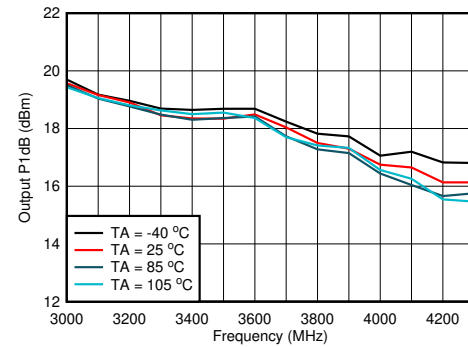


图 6-10. Output P1dB vs Frequency and Temperature

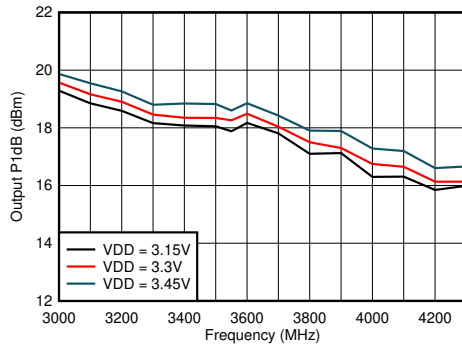
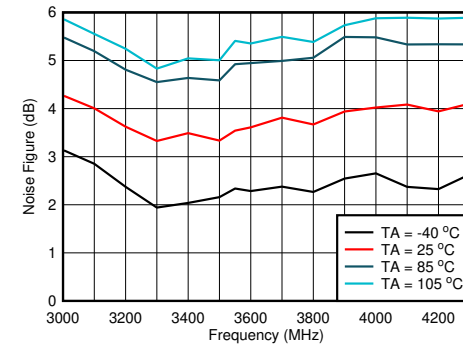


图 6-11. Output P1dB vs Frequency and Supply Voltage



$Z_{SOURCE} = 100\text{-}\Omega$ differential

图 6-12. Noise Figure vs Frequency and Temperature

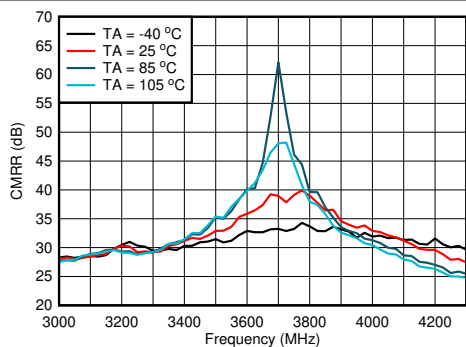


图 6-13. Common-mode Rejection Ratio vs Frequency

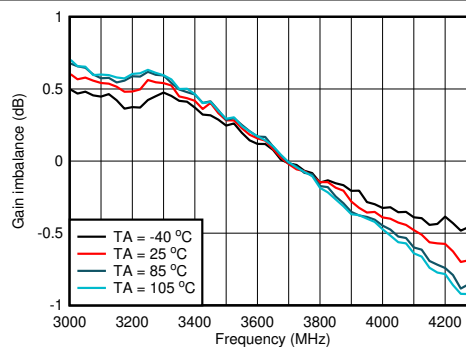


图 6-14. Gain Imbalance vs Frequency and Temperature

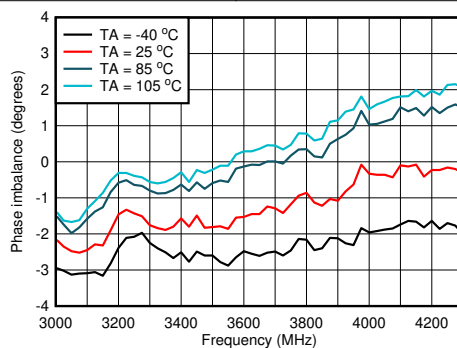


图 6-15. Phase Imbalance vs Frequency and Temperature

7 Detailed Description

7.1 Overview

The LMH9135 device is a differential input to single-ended output narrow-band RF amplifier used in transmitter applications. The device provides 18 dB fixed power gain with excellent linearity and noise performance across the frequency band 3.2 – 4.2 GHz. The device is internally matched for 100- Ω impedance at the differential input and 50- Ω impedance at the single-ended output, as shown in 图 7-1.

LMH9135 have on-chip active bias circuitry to maintain device performance over a wide temperature and supply voltage range. The included power down function allows the amplifier to shut down saving power when the amplifier is not needed. Fast shut down and start up enable the amplifier to be used in a host of time division duplex applications.

Operating on a single 3.3 V supply and 120 mA of typical supply current, the devices are available in a 2 mm x 2 mm 12-pin QFN package.

7.2 Functional Block Diagram

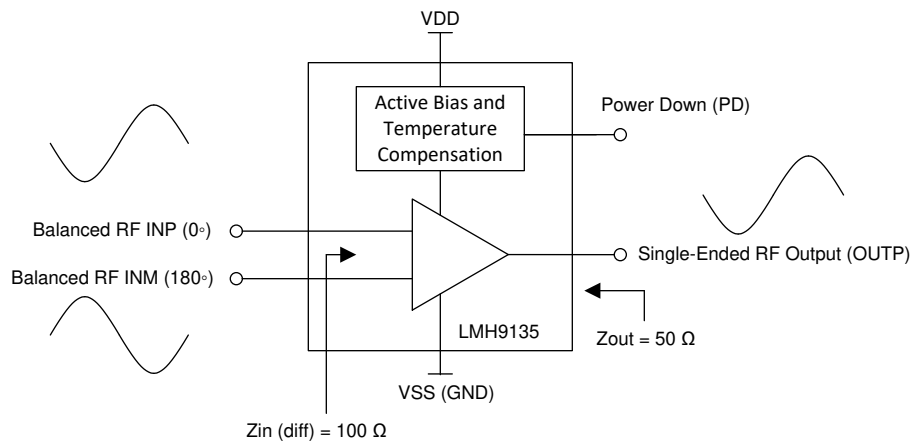


图 7-1. Functional Block Diagram

7.3 Feature Description

The LMH9135 device is a differential input to single-ended output RF amplifier for narrow band active balun implementation. The device integrates the functionality of a single-ended RF amplifier and passive balun in traditional transmitter applications achieving small form factor with comparable linearity and noise performance, as shown in 图 7-2.

The active balun implementation coupled with higher operating temperature of 105°C allows for more robust receiver system implementation compared to passive balun that is prone to reliability failures at high temperatures. The robust operation is achieved by the on-chip active bias circuitry which maintains device performance over a wide temperature and supply voltage range.

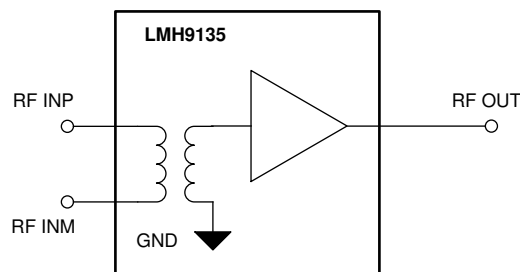


图 7-2. Differential Input to Single-Ended Output, Active Balun Implementation

7.4 Device Functional Modes

LMH9135 features a PD pin which should be connected to GND for normal operation. To power down the device, connect the PD pin to a logic high voltage of 1.8 V.

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

LMH9135 is a differential to single-ended RF gain block amplifier, which works as an active balun in the transmit path of a 3.3-GHz to 3.8-GHz 5G, TDD m-MIMO or small cell base station. The device replaces the traditional passive balun and single-ended RF amplifier offering a smaller footprint solution to the customer. TI recommends following good RF layout and grounding techniques to maximize the device performance.

8.2 Typical Application

LMH9135 is typically used in a four transmit and four receive (4T/4R) array of active antenna system for 5G, TDD, wireless base station applications. Such a system is shown in [Figure 8-1](#), where the LMH9135 is used in the transmit path as an active balun that converts differential DAC output from Tx AFE to single-ended signal. Also shown in the figure is the application of LMH9235 chip, which is the counter-part of LMH9135 in the Receive path.

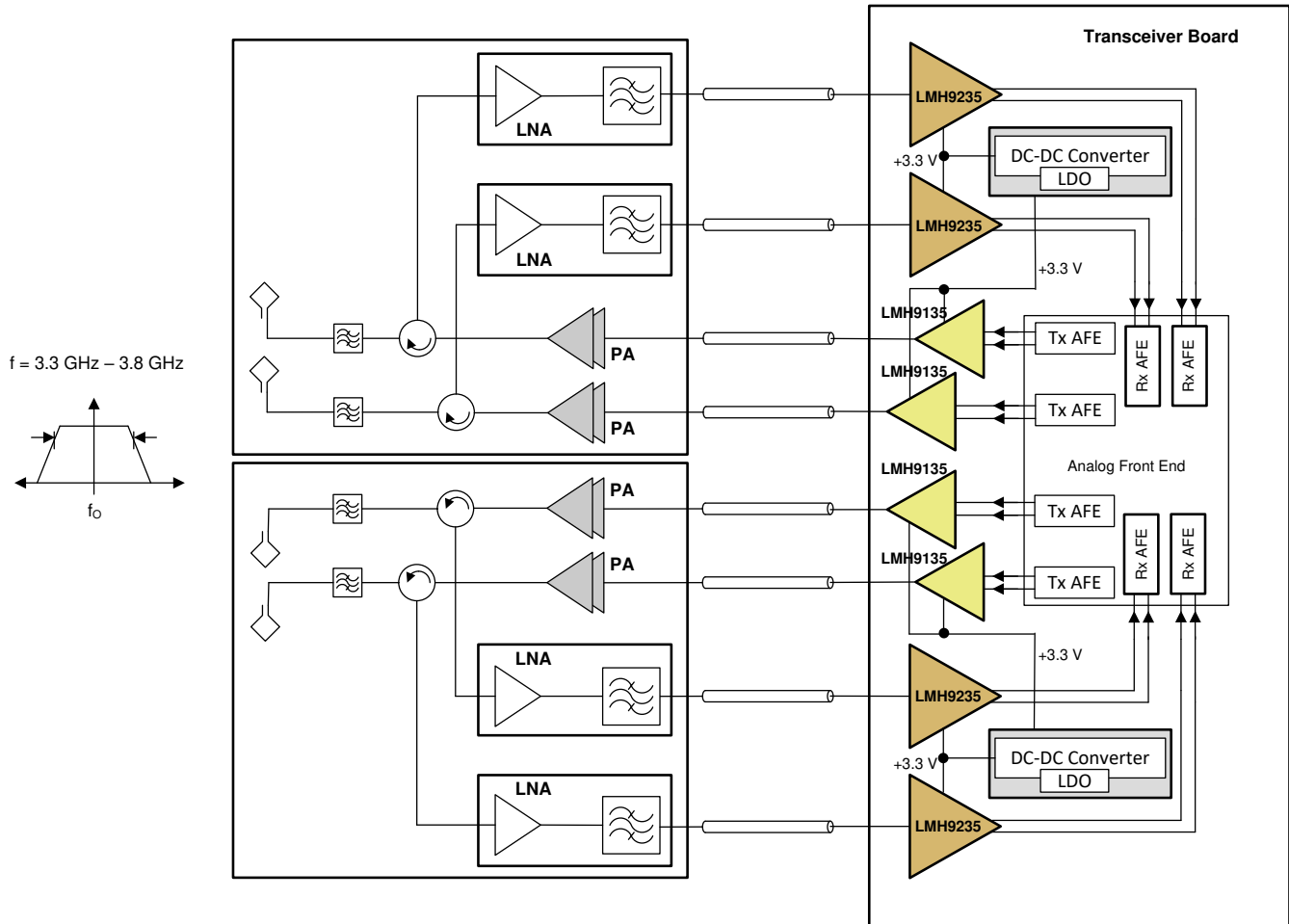


图 8-1. LMH9135 in a 4T/4R 5G Active Antenna System

The 4T/4R system can be scaled to 16T/16R, 64T/64R, or higher antenna arrays that result in proportional scaling of the overall system power dissipation. As a result of the proportional scaling factor for multiple channels in a system, the individual device power consumption must be reduced to dissipate less overall heat in the system. Operating on a single 3.3-V supply, the LMH9135 consumes only about 400 mW and therefore provides power saving to the customer. Multiple LMH9135 devices can be powered from a single DC/DC converter or a low-dropout regulator (LDO) operating on a 3.3-V supply. A DC/DC converter provides the most power efficient way of generating the 3.3-V supply. However, care must be taken when using the DC/DC converter to minimize the switching noise using inductor chokes and adequate isolation must be provided between the analog and digital power domains.

8.2.1 Design Requirements

Input of LMH9135 is matched to $100\ \Omega$ and therefore can be directly driven by a DAC that has $100\ \Omega$ source impedance without any external matching network. If a DAC with different impedance is used, then it should be appropriately matched to get the best RF performance.

The example in 图 8-2 shows how LMH9135 can be matched to a DAC that has $200\text{-}\Omega$ differential termination.

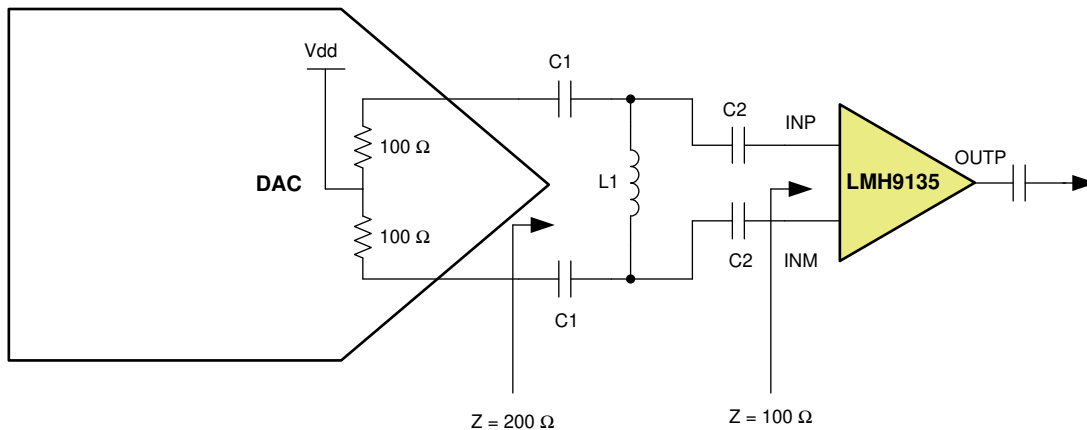


图 8-2. LMH9135 Driven by a DAC with 200-Ω Termination

8.2.2 Detailed Design Procedure

A simple differential LC network is used here as the matching network. In 图 8-2, shunt inductor L1 and series capacitors C2 form the matching network. The series capacitors C1 act as the DC-blocking capacitors. 表 8-1 shows the matching network component values.

表 8-1. Matching Network Component Values for 200-Ω Termination

COMPONENT	VALUE
C1	5.6 pF
L1	6.8 nH
C2	0.8 pF

8.2.3 Application Curves

The graphs given below show the gain, input return loss, and output return loss of the design with different DAC terminations.

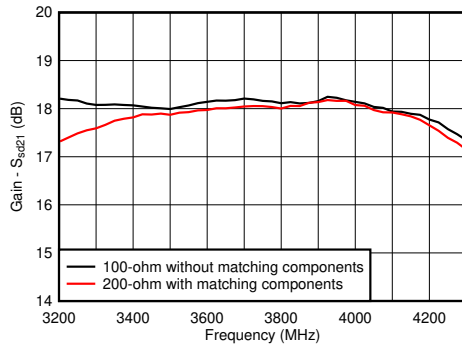


图 8-3. Gain vs Frequency for Different DAC Terminations

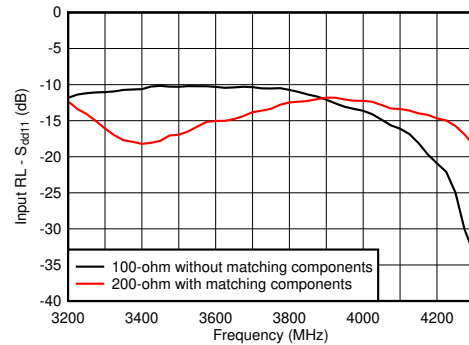


图 8-4. Input Return Loss vs Frequency for Different DAC Terminations

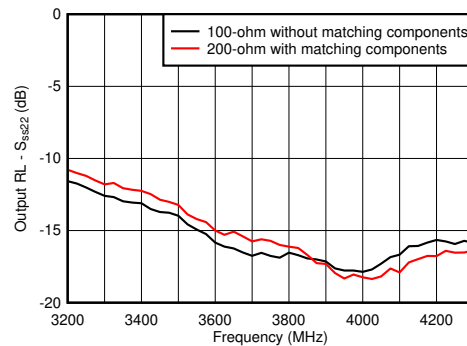


图 8-5. Output Return Loss vs Frequency for Different DAC Terminations

9 Power Supply Recommendations

The LMH9135 device operates on a common nominal 3.3 V supply voltage. It is recommended to isolate the supply voltage through decoupling capacitors placed close to the device. Select capacitors with self resonant frequency above the application frequency. When multiple capacitors are used in parallel to create a broadband decoupling network, place the capacitor with the higher self-resonant frequency closer to the device.

10 Layout

10.1 Layout Guidelines

When designing with an RF amplifier operating in the frequency range 3 GHz to 4.2 GHz with relatively high gain, certain board layout precautions must be taken to ensure stability and optimum performance. TI recommends that the LMH9135 board be multi-layered to improve thermal performance, grounding, and power-supply decoupling. 图 10-1 shows a good layout example. In this figure, only the top signal layer is shown.

- Excellent electrical connection from the thermal pad to the board ground is essential. Use the recommended footprint, solder the pad to the board, and do not include a solder mask under the pad.
- Connect the pad ground to the device terminal ground on the top board layer.
- Ensure that ground planes on the top and any internal layers are well stitched with vias.
- Design the two input and one output RF traces for 50- Ω impedance. TI recommends grounded coplanar waveguide (GCPW) type transmission lines for the RF traces. Use a PCB trace width calculator tool to design the transmission lines.
- Avoid routing clocks and digital control lines near RF signal lines.
- Do not route RF or DC signal lines over noisy power planes.
- Place supply decoupling close to the device.
- The differential output traces must be symmetrical in order to achieve the best differential balance and linearity performance.

See the [LMH9135 Evaluation Module user's guide](#) for more details on board layout and design.

10.2 Layout Example

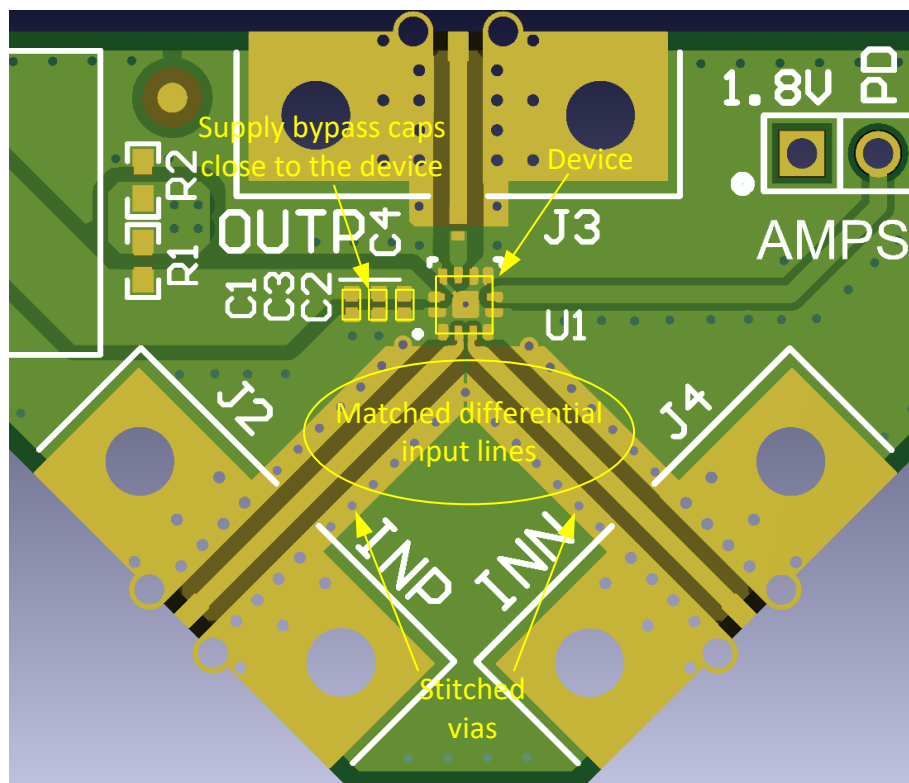


图 10-1. Layout Showing Matched Differential Traces and Supply Decoupling

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, [LMH9135 Evaluation Module User's Guide](#)
- Texas Instruments, [LMH9135 S-parameter Models](#)
- Texas Instruments, [LMH9135RRLEVM EU Declaration of Conformity \(DoC\)](#)

11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 Support Resources

TI E2E™ [support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

11.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

所有商标均为其各自所有者的财产。

11.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
LMH9135IRRLR	Active	Production	WQFN (RRL) 12	3000 LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 105	35AO
LMH9135IRRLR.B	Active	Production	WQFN (RRL) 12	3000 LARGE T&R	Yes	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 105	35AO

- ⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).
- ⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.
- ⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.
- ⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.
- ⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.
- ⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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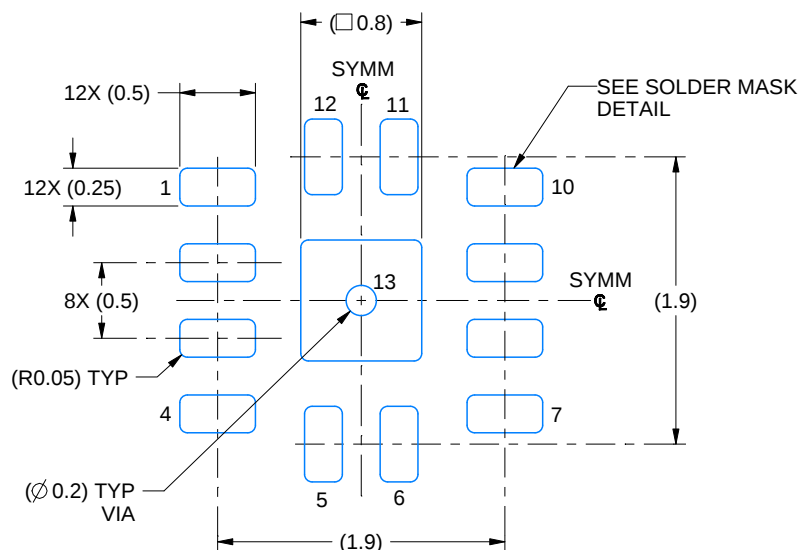


1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

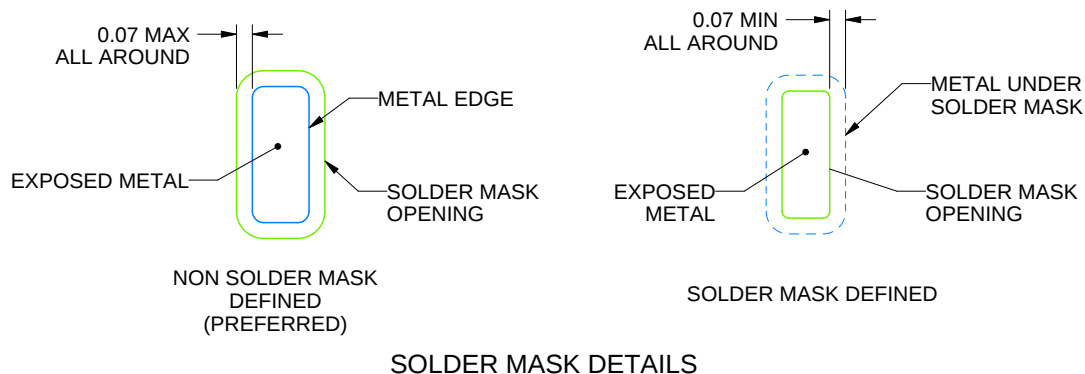
RRL0012A

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE: 20X



4224942/A 04/2019

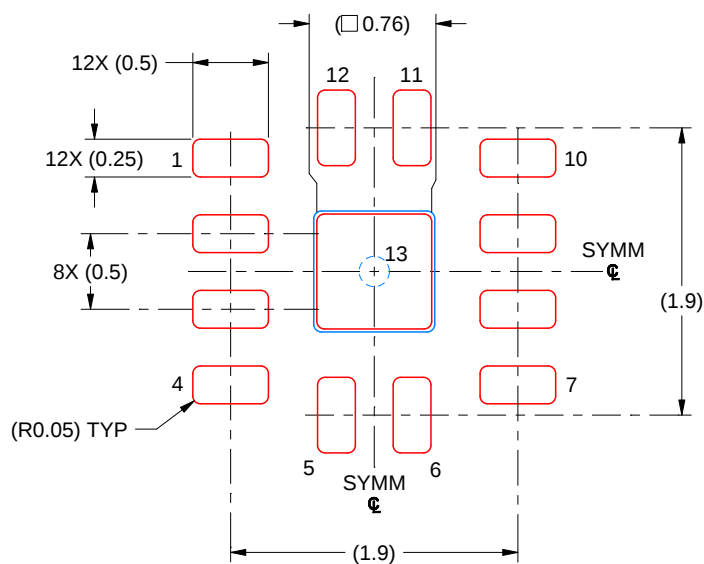
NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

RRL0012A

WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



SOLDER PASTE EXAMPLE
BASED ON 0.125 MM THICK STENCIL
SCALE: 20X

EXPOSED PAD 13
90% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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