

针对 2.4GHz 射频 (RF) 应用的片上系统

特性

- 射频 (RF) 部分
 - 单片 **2.4GHz RF** 收发器和微控制器 (MCU)
 - 支持 **250kbps, 500kbps, 1Mbps** 和 **2Mbps** 数据速率
 - 出色的链路预算, 无需外部前端即可实现长距离通信
 - 高达 **4dBm** 的可编程输出功率
 - 出色的接收器灵敏度 (**2Mbps** 时为 **-90Bm**, **256kbps** 时为 **-9dBm**)
 - 适用于符合世界范围内的射频标准的系统: **ETSI EN 300 328** 和 **EN 300 440 2** 类 (欧洲), **FCC CFR47 15** 部分 (美国), 和 **ARIB STD-T66** (日本)
 - 精确的接收到的信号强度指示器 (RSSI) 功能
- 布局
 - 极少的外部组件
 - 适用于单层印刷电路板 (PCB) 应用的引脚引线
 - 提供参考设计
 - **48 引脚 7mm x 7mm** 四方扁平无引线 (QFN) (**31 个通用 I/O 引脚**) 封装
- 低功耗
 - 激活模式 **RX** 最佳性能: **20.8mA**
 - 激活模式 **TX (0dBm)**: **26.3mA**
 - 功率模式 **1** (**5μs** 唤醒): **235μA**
 - 功率模式 **2** (睡眠定时器打开): **0.9μA**
 - 功率模式 **3** (外部中断): **0.4μA**
 - 宽电源电压范围 (**2V** 至 **3.6V**)
 - 在所有功率模式下的完全 **RAM** 和寄存器保持
- 微控制器
 - 具有代码预取功能的高性能和低功耗 **8051** 微控制器内核
 - **32KB** 闪存程序内存
 - **1KB SRAM**
 - 支持硬件调试
 - 扩展基带自动化, 包括自动确认和地址解码
- 外设
 - 可访问所有内存区域和外设的两个通道 **DMA**
 - 通用定时器 (**1 个 16 位, 2 个 8 位**)
 - 无线电定时器, **40 位**
 - 红外 (IR) 生成电路
 - 几个振荡器:
 - **32MHz** 晶体振荡器 (XOSC)
 - **16MHz RC** 振荡器 (RCOSC)
 - **32kHz XOSC**
 - **32MHz RCOSC**
 - 具有捕捉功能的 **32kHz** 睡眠定时器
 - 高级加密标准 (AES) 安全协处理器
 - 通用异步收发器 (UART) / 串行外设接口 (SPI) / ² 串行接口
 - **31 个通用 I/O 引脚** (**3 x 20mA** 驱动强度, 剩余的引脚有 **4mA** 的驱动强度)
 - 安全装置定时器
 - 真随机数生成器
 - 模数转换器 (ADC) 和模拟比较器

应用范围

- 私有的 **2.4 GHz** 系统
- 人机接口器件 (键盘、鼠标)
- 消费类电子产品



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



这些装置包含有限的内置 ESD 保护。

存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

说明

CC2545 是一款经优化的针对数据速率高达 2Mbps 且使用低物料清单成本制造的片上系统 (SoC) 解决方案。

CC2545 将处于领先地位的 RF 收发器的出色性能与一个单周期 8051 兼容 CPU，32KB 系统内可编程闪存存储器，高达 1KB RAM，31 个通用 I/O 引脚与很多其它功能强大的特性组合在一起。CC2545 有高效的功率模式（RAM 和寄存器保持电流低于 1μA），这使得它非常适合应用于要求超低功耗的低占空比系统。运行模式间较短的转换时间进一步确保了低能耗。

CC2545 与 CC2541/CC2543/CC2544 兼容。它采用带有 SPI/UART/I2C 接口的 7mm x 7mm QFN48 封装。供货时，CC2545 带有由德州仪器 (TI) 提供的完整参考设计。

此器件针对无线消费类产品和人机接口器件 (HID) 应用。CC2545 专为诸如无线键盘等的外设器件而设计。

方框图请见图 7。

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Supply voltage VDD	All supply pins must have the same voltage	−0.3	3.9	V
Voltage on any digital pin		−0.3	VDD+0.3 <= 3.9	V
Input RF level			10	dBm
Storage temperature range		−40	125	°C
ESD ⁽²⁾	All pads, according to human-body model, JEDEC STD 22, method A114 (HBM)		2000	V
	According to charged-device model, JEDEC STD 22, method C101 (CDM)		750	V

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) CAUTION: ESD sensitive device. Precaution should be used when handling the device in order to prevent permanent damage.

RECOMMENDED OPERATING CONDITIONS

PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
Operating ambient temperature range, T _A		−40	85	°C
Operating supply voltage VDD	All supply pins must have same voltage	2.0	3.6	V

ELECTRICAL CHARACTERISTICS

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$ and $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	2 Mbps, GFSK, 320-kHz deviation				
I_{core} – Core current consumption	RX mode, no peripherals active, low MCU activity		20.8		mA
	TX mode, 0-dBm output power, no peripherals active, low MCU activity		26.3		mA
	TX mode, 5-dBm output power, no peripherals active, low MCU activity		30.2		mA
	Active mode, 16-MHz RCOSC, Low MCU activity		3		mA
	Active mode, 32-MHz clock frequency, low MCU activity		6		mA
	Power mode 0, CPU clock halted, all peripherals on, no clock division, 32-MHz crystal selected		4.5		mA
	Power mode 0, CPU clock halted, all peripherals on, clock division at max (Limits max speed in peripherals except radio), 32-MHz crystal selected		3		mA
	Power mode 1. Digital regulator ON; 16-MHz RCOSC and 32-MHz crystal oscillator OFF; 32.753-kHz RCOSC, POR, BOD, and sleep timer active; RAM and register retention		235		μA
	Power mode 2. Digital regulator OFF, 16 MHz RCOSC and 32 MHz crystal oscillator OFF; 32.753 kHz RCOSC, POR and sleep timer active; RAM and register retention		0.9		μA
	Power mode 3. Digital regulator OFF, no clocks, POR active; RAM and register retention		0.4		μA
I_{peri} – Peripheral current consumption (Adds to core current I_{core} for each peripheral unit activated)	Timer 1 (16-bit). Timer running, 32-MHz XOSC used		90		μA
	Radio timer(40 bit). Timer running, 32-MHz XOSC used		90		μA
	Timer 3 (8-bit). Timer running, 32-MHz XOSC used		60		μA
	Timer 4 (8-bit). Timer running, 32-MHz XOSC used		70		μA
	Sleep timer. Including 32.753-kHz RCOSC		0.6		μA

GENERAL CHARACTERISTICS

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$ and $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
WAKE-UP AND TIMING					
Power mode 1 → Active	Digital regulator on, 16-MHz RCOSC and 32-MHz crystal oscillator off. Start-up of 16-MHz RCOSC.		5		μs
Power mode 2 or 3 → Active	Digital regulator off, 16 MHz RCOSC and 32 MHz crystal oscillator off. Start-up of regulator and 16 MHz RCOSC.		130		μs
Active → TX or RX	Crystal ESR = 16 Ω . Initially running on 16-MHz RCOSC, with 32-MHz XOSC OFF.		500		μs
	With 32-MHz XOSC initially ON.		180		μs
RX/TX turnaround	RCOSC, with 32MHz XOSC OFF.		130		μs
RADIO PART					
RF frequency range	Programmable in 1-MHz steps	2379		2496	MHz
Data rates and modulation formats	2 Mbps, GFSK 320-kHz deviation 2-Mbps, GFSK 500 kHz deviation 1-Mbps, GFSK 250 kHz deviation 1-Mbps, GFSK 160 kHz deviation 500 kbps, MSK 250 kbps, MSK 250 kbps, GSK 160 kHz deviation				

RF RECEIVE SECTION

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3\text{ V}$, and $f_c = 2440\text{ MHz}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
2 Mbps, GFSK, 320-kHz DEVIATION, 0.1% BER					
Receiver sensitivity			–86		dBm
Saturation			–8		dBm
Co-channel rejection	Wanted signal at –67 dBm		–13		dB
In-band blocking rejection	± 2 -MHz offset, wanted signal at –67 dBm		–1		dB
	± 4 -MHz offset, wanted signal at –67 dBm		34		
	$> \pm 6$ -MHz offset, wanted signal at –67 dBm		38		
Out-of-band blocking rejection	1-MHz resolution. Wanted signal at –67 dBm, $f < 2\text{ GHz}$ Two exception frequencies with poorer performance		–32		dBm
	1-MHz resolution. Wanted signal at –67 dBm, $2\text{ GHz} > f < 3\text{ GHz}$ Two exception frequencies with poorer performance		–38		
	1-MHz resolution. Wanted signal at –67 dBm, $f > 3\text{ GHz}$ Two exception frequencies with poorer performance		–12		
Intermodulation	Wanted signal at –64 dBm, 1 st interferer is CW, 2 nd interferer is GFSK-modulated signal. Offsets of interferers are: 6 and 12 MHz 8 and 16 MHz 10 and 20 MHz		–43		dBm
Frequency error tolerance ⁽¹⁾	Including both initial tolerance and drift. Sensitivity better than –70 dBm. 250 byte payload.	–300		300	kHz
Symbol rate error tolerance ⁽²⁾	Sensitivity better than –70 dBm. 250 byte payload.	–120		120	ppm
2 Mbps, GFSK, 500 kHz DEVIATION, 0.1% BER					
Receiver sensitivity			–90		dBm
Saturation			–3		dBm
Co-channel rejection	Wanted signal at –67 dBm		–10		dB
In-band blocking rejection	± 2 MHz offset, wanted signal at –67 dBm		–3		dB
	± 4 MHz offset, wanted signal at –67 dBm		36		dB
	$> \pm 6$ MHz offset, wanted signal at –67 dBm		44		dB
Frequency error tolerance ⁽¹⁾	Including both initial tolerance and drift. Sensitivity better than –70 dBm. 250 byte payload.	–300		300	kHz
Symbol rate error tolerance ⁽²⁾	Sensitivity better than –70 dBm. 250 byte payload.	–120		120	ppm
1 Mbps, GFSK, 250 kHz DEVIATION, 0.1% BER					
Receiver sensitivity			–94		dBm
Saturation			6		dBm
Co-channel rejection	Wanted signal at –67 dBm		–7		dB
In-band blocking rejection	± 1 MHz offset, wanted signal –67 dBm		0		dB
	± 2 MHz offset, wanted signal –67 dBm		30		
	± 3 MHz offset, wanted signal –67 dBm		34		
	$> \pm 5$ MHz offset, wanted signal –67 dBm		38		
Frequency error tolerance	Including both initial tolerance and drift. Sensitivity better than –70 dBm. 250 byte payload.	–250		250	kHz
Symbol rate error tolerance	Sensitivity better than –70 dBm. 250 byte payload.	–80		80	ppm

(1) Difference between center frequency of the received RF signal and local oscillator frequency

(2) Difference between incoming symbol rate and the internally generated symbol rate

RF RECEIVE SECTION (接下页)

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3\text{ V}$, and $f_c = 2440\text{ MHz}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
1 Mbps, GFSK, 160 kHz DEVIATION, 0.1% BER					
Receiver sensitivity			–91		dBm
Saturation			6		dBm
Co-channel rejection	Wanted signal at –67 dBm		–8		dB
In band blocking rejection	±1 MHz offset, wanted signal at –67 dBm		2		dB
	±2 MHz offset, wanted signal at –67 dBm		28		
	±3 MHz offset, wanted signal at –67 dBm		33		
	>±5 MHz offset, wanted signal at –67 dBm		36		
Frequency error tolerance	Including both initial tolerance and drift, Sensitivity better than –67 dBm	–250		250	kHz
Symbol rate error tolerance	Maximum packet length	–80		80	ppm
500 kbps, MSK, 0.1% BER					
Receiver sensitivity			–98		dBm
Saturation			6		dBm
Co-channel rejection	Wanted signal at –67 dBm		–5		dB
In band blocking rejection	±1 MHz offset, wanted signal at –67 dBm		21		dB
	±2 MHz offset, wanted signal at –67 dBm		32		
	>±2 MHz offset, wanted signal at –67 dBm		33		
Frequency error tolerance	Including both initial tolerance and drift, Sensitivity better than –67dBm	–150		150	kHz
Symbol rate error tolerance	Maximum packet length	–60		60	ppm
250 kbps, GFSK, 160 kHz DEVIATION , 0.1% BER					
Receiver sensitivity			–98		dBm
Saturation			6		dBm
Co-channel rejection	Wanted signal at –67 dBm		–2		dB
In-band blocking rejection	±1 MHz offset, wanted signal at –67 dBm		22		dB
	±2 MHz offset, wanted signal at –67 dBm		32		
	>±2 MHz offset, wanted signal at –67 dBm		32		
Frequency error tolerance	Including both initial tolerance and drift, Sensitivity better than –67 dBm	–150		150	kHz
Symbol rate error tolerance	Maximum packet length	–60		60	ppm
250 kbps, MSK, 0.1% BER					
Receiver sensitivity			–98		dBm
Saturation			6		dBm
Co-channel rejection	Wanted signal at –67 dBm		–5		dB
In-band blocking rejection	±1 MHz offset, wanted signal at –67 dBm		21		dB
	±2 MHz offset, wanted signal at –67 dBm		32		
	>2 MHz offset, wanted signal at –67 dBm		33		
Frequency error tolerance	Including both initial tolerance and drift, Sensitivity better than –67 dBm	–150		150	kHz
Symbol rate error tolerance	Maximum packet length	–60		60	ppm
ALL RATES/FORMATS					
Spurious emission in RX. Conducted measurement	$f < 1\text{ GHz}$		–67		dBm
Spurious emission in RX. Conducted measurement	$f > 1\text{ GHz}$		–60		dBm

RF TRANSMIT SECTION

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, and $f_C = 2440\text{ MHz}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output power, maximum setting	Delivered to a single-ended 50- Ω load through a balun using maximum recommended output power setting.		5		dBm
Output power, minimum setting	Delivered to a single-ended 50- Ω load through a balun using minimum recommended output power setting.		-20		dBm
Programmable output power range	Delivered to a single-ended 50- Ω load through a balun.		25		dB
Spurious emission in TX. Conducted measurement	$f < 1\text{ GHz}$		-46		dBm
	$f > 1\text{ GHz}$		-44		dBm
	Suitable for Systems Targeting Compliance With Worldwide Radio Frequency Regulations: ETSI EN 300 328 and EN 300 440 Class 2 (Europe), FCC CFR47 Part 15 (US), and ARIB STD-T66 (Japan)				

Use a simple LC filter (1.6nH and 1.8pF in parallel to ground) to pass ETSI conducted requirements below 1GHz in restricted bands. For radiated measurements low antenna gain for these frequencies (depending on antenna design) can achieve the same attenuation of these low frequency components (see EM reference design).

32-MHz CRYSTAL OSCILLATOR

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Crystal frequency			32		MHz
Crystal frequency accuracy requirement	250 kbps and 500 kbps data rates 1 Mbps data rate 2 Mbps data rate	-30 -40 -60		30 40 60	ppm
Equivalent series resistance		6		60	Ω
Crystal shunt capacitance		1		7	pF
Crystal load capacitance		10		16	pF
Start-up time			0.25		ms
Power-down guard time	The crystal oscillator must be in power down for a guard time before it is used again. This requirement is valid for all modes of operation. The need for power-down guard time can vary with crystal type and load.	3			ms

32.768-kHz CRYSTAL OSCILLATOR

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Crystal frequency			32.768		kHz
Crystal frequency accuracy requirement ⁽¹⁾		-100		+100	ppm
Equivalent series resistance			40	130	Ω
Crystal shunt capacitance			0.9	2	pF
Crystal load capacitance			12	16	pF
Start-up time			0.4		s

- (1) Crystal frequency accuracy requirement is highly dependent on application. Higher accuracy enables more accurate duty-cycling which in turn will reduce current consumption. The chip can handle much less accurate crystals.

32-kHz RC OSCILLATOR

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Calibrated frequency			32.753		kHz
Frequency accuracy after calibration			$\pm 0.2\%$		
Temperature coefficient			0.4		%/ $^\circ\text{C}$
Supply-voltage coefficient			3		%/V
Calibration time			2		ms

16-MHz RC OSCILLATOR

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Calibrated frequency			16		MHz
Uncalibrated frequency accuracy			$\pm 18\%$		
Frequency accuracy after calibration			$\pm 0.6\%$		
Start-up time			10		μs
Initial calibration time			50		μs

RSSI CHARACTERISTICS

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3\text{ V}$, unless otherwise noted.

2Mbps, GFSK, 320-kHz Deviation, 0.1% BER and 2 Mbps, GFSK, 500-kHz Deviation, 0.1% BER				
RSSI range ⁽¹⁾	Reduced gain by AC algorithm	64		dB
	High gain by AGC algorithm	64		
RSSI offset ⁽¹⁾	Reduced gain by AGC algorithm	79		dBm
	High gain by AGC algorithm	99		
Absolute uncalibrated accuracy ⁽¹⁾		± 3		dB
Step size (LSB value)		1		dB
All Other Rates/Formats				
RSSI range ⁽¹⁾		64		dB
RSSI offset ⁽¹⁾		99		dBm
Absolute uncalibrated accuracy		± 3		dB
Step size (LSB value)		1		dB

(1) Assuming CC2545 EM reference design. Other RF designs give an offset from the reported value.

FREQUENCY SYNTHESIZER CHARACTERISTICS

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Phase noise, unmodulated carrier	At $\pm 1\text{ MHz}$ from carrier		-112		dBc/Hz
	At $\pm 3\text{ MHz}$ from carrier		-119		
	At $\pm 5\text{ MHz}$ from carrier		-122		

ANALOG TEMPERATURE SENSOR

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$ unless otherwise noted

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output	Measured using integrated ADC, internal band-gap voltage reference, and maximum resolution		1480		12-bit
Temperature coefficient			4.5		/ 0.1°C
Voltage coefficient			1		/ 0.1V
Initial accuracy without calibration			± 10		$^\circ\text{C}$
Accuracy using 1-point calibration			± 5		$^\circ\text{C}$
Current consumption when enabled			0.5		mA

COMPARATOR CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_{DD} = 3\text{ V}$. All measurement results are obtained using the CC2545 reference designs, post-calibration.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Common-mode maximum voltage			VDD		V
Common-mode minimum voltage			-0.3		
Input offset voltage			1		mV
Offset vs temperature			16		$\mu\text{V}/^\circ\text{C}$
Offset vs operating voltage			4		mV/V
Supply current			230		nA
Hysteresis			0.15		mV

ADC CHARACTERISTICS

 $T_A = 25^\circ\text{C}$ and $V_{DD} = 3\text{ V}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input voltage		VDD is voltage on AVDD5 pin	0		VDD	V
External reference voltage		VDD is voltage on AVDD5 pin	0		VDD	V
External reference voltage differential		VDD is voltage on AVDD5 pin	0		VDD	V
Input resistance, signal		Simulated using 4-MHz clock speed		197		kΩ
Full-scale signal ⁽¹⁾		Peak-to-peak, defines 0 dBFS		2.97		V
ENOB ⁽¹⁾	Effective number of bits	Single-ended input, 7-bit setting		5.7		bits
		Single-ended input, 9-bit setting		7.5		
		Single-ended input, 10-bit setting		9.3		
		Single-ended input, 12-bit setting		10.3		
		Differential input, 7-bit setting		6.5		
		Differential input, 9-bit setting		8.3		
		Differential input, 10-bit setting		10		
		Differential input, 12-bit setting		11.5		
		10-bit setting, clocked by RCOSC		9.7		
		12-bit setting, clocked by RCOSC		10.9		
Useful power bandwidth		7-bit setting, both single and differential		0–20		kHz
THD	Total harmonic distortion	Single ended input, 12-bit setting, –6 dBFS ⁽¹⁾		–75.2		dB
		Differential input, 12-bit setting, –6 dBFS ⁽¹⁾		–86.6		
	Signal to nonharmonic ratio	Single-ended input, 12-bit setting ⁽¹⁾		70.2		dB
		Differential input, 12-bit setting ⁽¹⁾		79.3		
		Single-ended input, 12-bit setting, –6 dBFS ⁽¹⁾		78.8		
		Differential input, 12-bit setting, –6 dBFS ⁽¹⁾		88.9		
CMRR	Common-mode rejection ratio	Differential input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
Crosstalk		Single ended input, 12-bit setting, 1-kHz sine (0 dBFS), limited by ADC resolution		>84		dB
Offset		Midscale		–3		mV
Gain error				0.68%		
DNL	Differential nonlinearity	12-bit setting, mean ⁽¹⁾		0.05		LSB
		12-bit setting, maximum ⁽¹⁾		0.9		
INL	Integral nonlinearity	12-bit setting, mean ⁽¹⁾		4.6		LSB
		12-bit setting, maximum ⁽¹⁾		13.3		
		12-bit setting, mean, clocked by RCOSC		10		
		12-bit setting, max, clocked by RCOSC		29		
SINAD (–THD+N)	Signal-to-noise-and-distortion	Single ended input, 7-bit setting ⁽¹⁾		35.4		dB
		Single ended input, 9-bit setting ⁽¹⁾		46.8		
		Single ended input, 10-bit setting ⁽¹⁾		57.5		
		Single ended input, 12-bit setting ⁽¹⁾		66.6		
		Differential input, 7-bit setting ⁽¹⁾		40.7		
		Differential input, 9-bit setting ⁽¹⁾		51.6		
		Differential input, 10-bit setting ⁽¹⁾		61.8		
	Conversion time	7-bit setting		20		μs
		9-bit setting		36		
		10-bit setting		68		
		12-bit setting		132		

(1) Measured with 300-Hz sine-wave input and VDD as reference.

ADC CHARACTERISTICS (接下页)

$T_A = 25^\circ\text{C}$ and $V_{DD} = 3\text{ V}$

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Power consumption			1.2		mA
Internal reference VDD coefficient			4		mV/V
Internal reference temperature coefficient			0.4		mV/10°C
Internal reference voltage			1.15		V

DC CHARACTERISTICS

Measured on Texas Instruments CC2545EM reference design with $T_A = 25^\circ\text{C}$, $V_{DD} = 3.0\text{ V}$, unless otherwise noted.⁽¹⁾

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Logic-0 input voltage				0.5	V
Logic-1 input voltage		2.5			V
Logic-0 input current		–50		50	nA
Logic-1 input current		–50		50	nA
I/O pin pullup and pulldown resistors			20		kΩ
Logic-0 output voltage 4-mA pins	Output load 4 mA			0.5	V
Logic-1 output voltage 4-mA pins	Output load 4 mA	2.4			V
Logic-0 output voltage 20-mA pins	Output load 20 mA			0.5	V
Logic-1 output voltage, 20-mA pins	Output load 20 mA	2.4			V

(1) Note that only two of the three 20-mA pins can drive in the same direction at the same time, and toggle at the same time.

CONTROL INPUT AC CHARACTERISTICS

$T_A = -40^\circ\text{C}$ to 85°C , $V_{DD} = 2\text{ V}$ to 3.6 V .

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
System clock, f_{SYSCLK} $t_{\text{SYSCLK}} = 1/f_{\text{SYSCLK}}$	The undivided system clock is 32 MHz when crystal oscillator is used. The undivided system clock is 16 MHz when calibrated 16-MHz RC oscillator is used.	16		32	MHz
RESET_N low duration	See item 1, 图 1. This is the shortest pulse that is recognized as a complete reset pin request. Note that shorter pulses may be recognized but do not lead to complete reset of all modules within the chip.	1			μs
Interrupt pulse duration	See item 2, 图 1. This is the shortest pulse that is recognized as an interrupt request.	20			ns

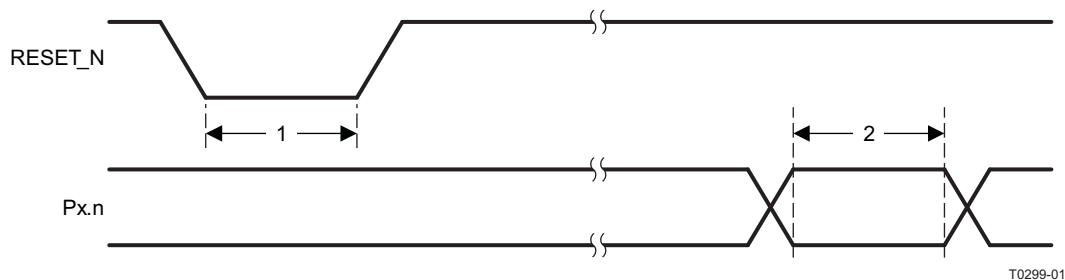


图 1. Control Input AC Characteristics

SPI AC CHARACTERISTICS

$T_A = -40^{\circ}\text{C}$ to 85°C , $V_{DD} = 2\text{ V}$ to 3.6 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_1 SCK period	Master, RX and TX	250			ns
	Slave, RX and TX	250			
SCK duty cycle	Master		50%		
t_2 SSN low to SCK, 图 2 and 图 3	Master	63			ns
	Slave	63			
t_3 SCK to SSN high	Master	63			ns
	Slave	63			
t_4 MOSI early out	Master, load = 10 pF			7	ns
t_5 MOSI late out	Master, load = 10 pF			10	ns
t_6 MISO setup	Master	90			ns
t_7 MISO hold	Master	10			ns
SCK duty cycle	Slave		50%		ns
t_{10} MOSI setup	Slave	35			ns
t_{11} MOSI hold	Slave	10			ns
t_8 MISO early out	Slave, load = 10 pF			0	ns
t_9 MISO late out	Slave, load = 10 pF			95	ns
Operating frequency	Master, TX only			8	MHz
	Master, RX and TX			4	
	Slave, RX only			8	
	Slave, RX and TX			4	

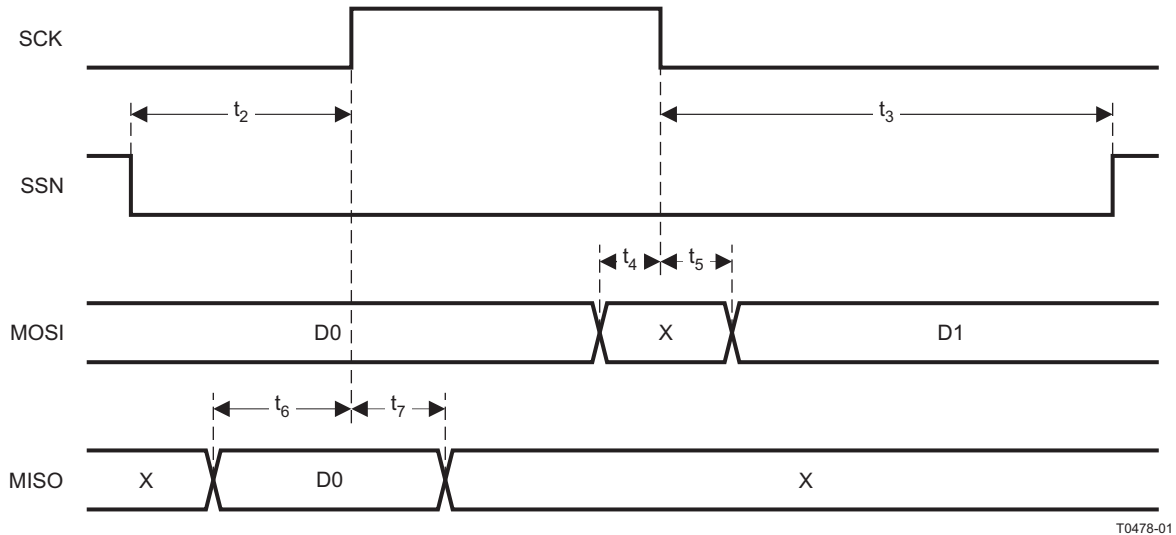


图 2. SPI Master AC Characteristics

T0478-01

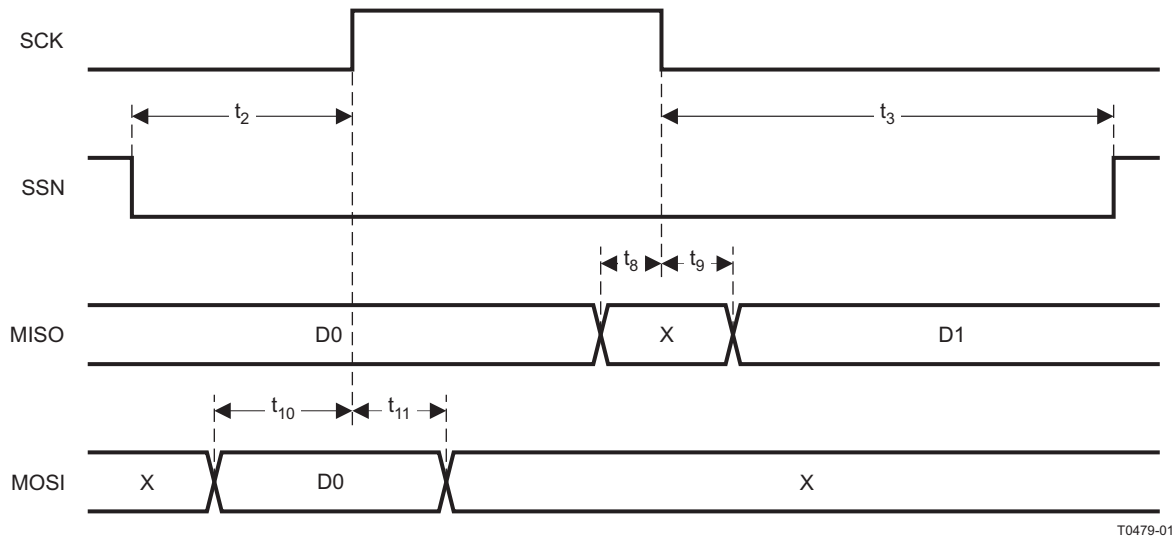


图 3. SPI Slave AC Characteristics

DEBUG INTERFACE AC CHARACTERISTICS

$T_A = -40^{\circ}\text{C}$ to 85°C , $V_{DD} = 2\text{ V}$ to 3.6 V

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$f_{\text{clk_dbg}}$	Debug clock frequency (see 图 4)					12	MHz
t_1	Allowed high pulse on clock (see 图 4)			35			ns
t_2	Allowed low pulse on clock (see 图 4)			35			ns
t_3	EXT_RESET_N low to first falling edge on debug clock (see 图 5)			167			ns
t_4	Falling edge on clock to EXT_RESET_N high (see 图 5)			83			ns
t_5	EXT_RESET_N high to first debug command (see 图 5)			83			ns
t_6	Debug data setup (see 图 6)			2			ns
t_7	Debug data hold (see 图 6)			4			ns
t_8	Clock-to-data delay (see 图 6)	Load = 10 pF				30	ns

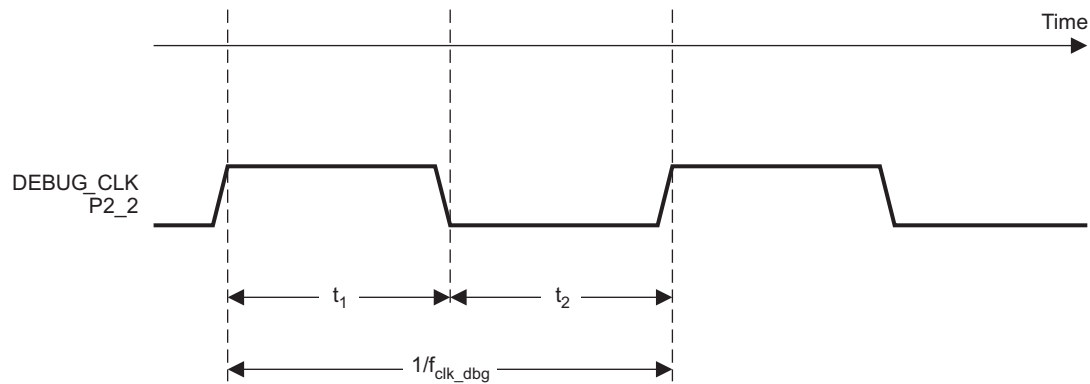
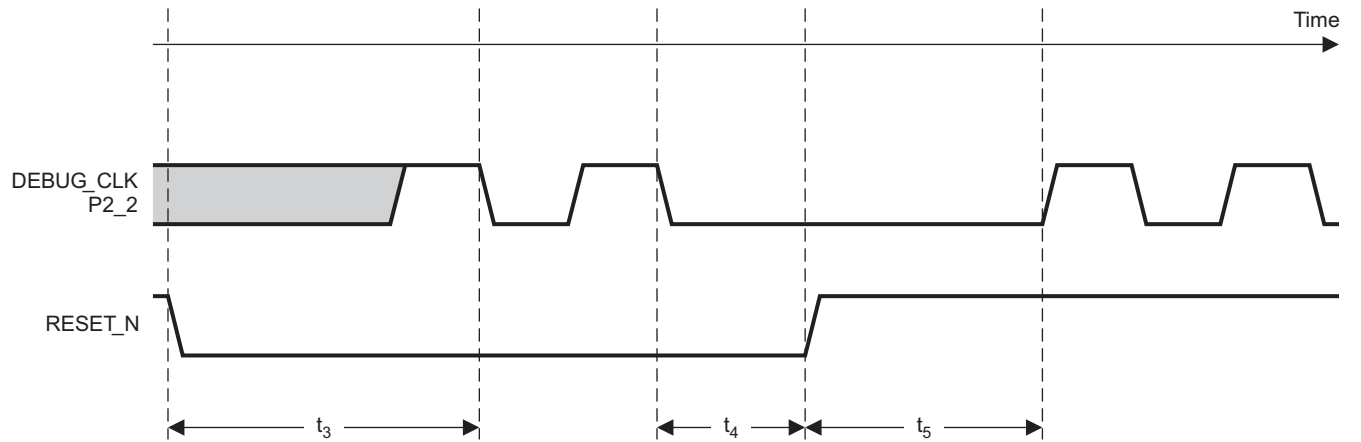
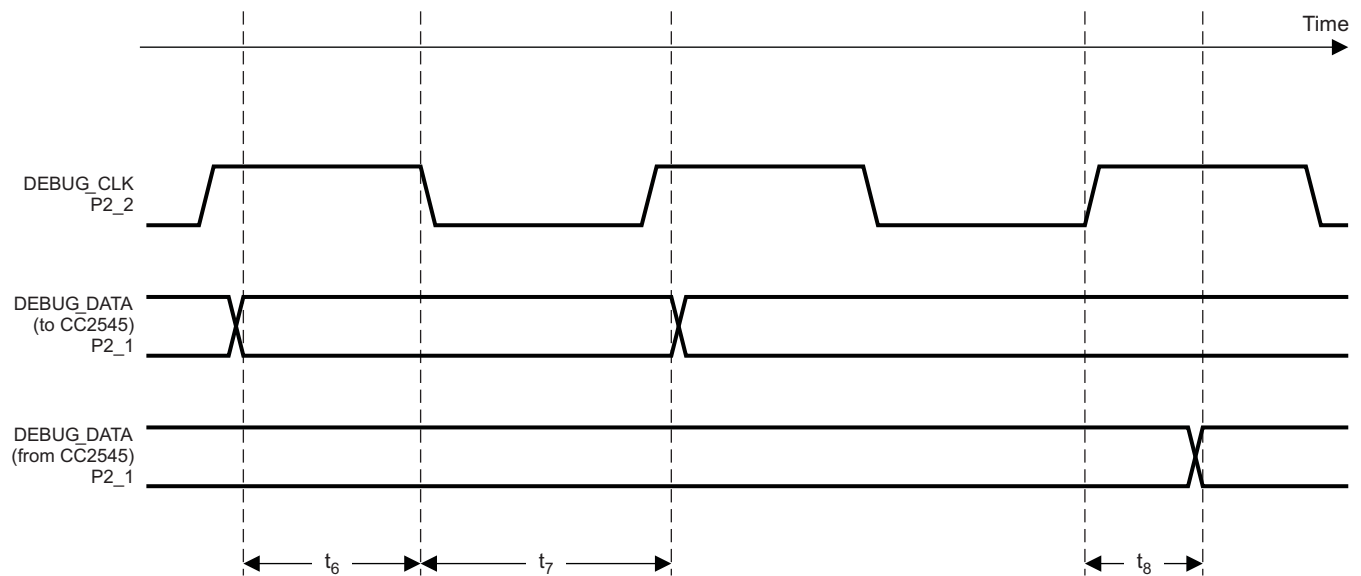


图 4. Debug Clock – Basic Timing



T0437-01

图 5. Debug Enable Timing



T0438-03

图 6. Data Setup and Hold Timing

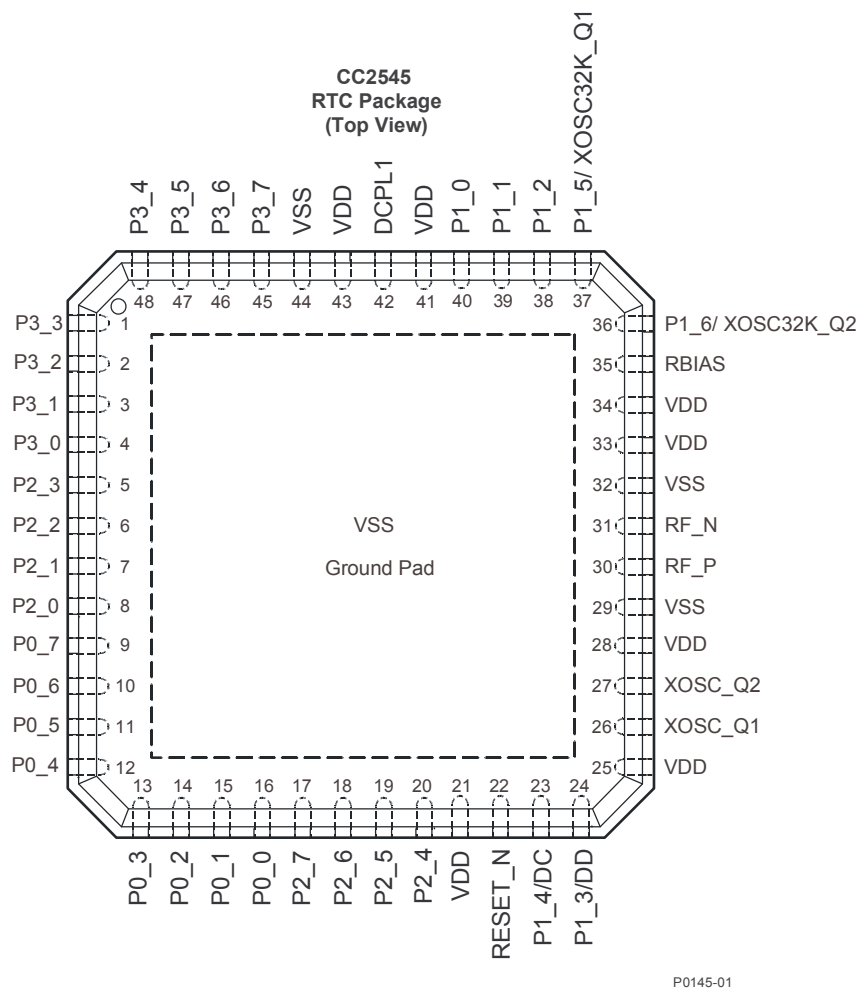
TIMER INPUTS AC CHARACTERISTICS

$T_A = -40^{\circ}\text{C}$ to 85°C , $V_{DD} = 2\text{ V}$ to 3.6 V

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input capture pulse duration	Synchronizers determine the shortest input pulse that can be recognized. The synchronizers operate at the current system clock rate (16 MHz or 32 MHz).	1.5			t_{SYSCLK}

DEVICE INFORMATION

PIN DESCRIPTIONS



- (1) NOTE: The exposed ground pad must be connected to a solid ground plane; this is the main ground connection for the chip.

表 1. Pin Description Table

NAME	PIN	PIN TYPE	DESCRIPTION
P3_3	1	Digital I/O	Port 3.3
P3_2	2	Digital I/O	Port 3.2
P3_1	3	Digital I/O	Port 3.1
P3_0	4	Digital I/O	Port 3.0
P2_3	5	Digital I/O	Port 2.3
P2_2	6	Digital I/O	Port 2.2
P2_1	7	Digital I/O	Port 2.1
P2_0	8	Digital I/O	Port 2.0
P0_7	9	Digital I/O	Port 0.7
P0_6	10	Digital I/O	Port 0.6
P0_5	11	Digital I/O	Port 0.5
P0_4	12	Digital I/O	Port 0.4
P0_3	13	Digital I/O	Port 0.3

表 1. Pin Description Table (接下页)

NAME	PIN	PIN TYPE	DESCRIPTION
P0_2	14	Digital I/O	Port 0.2
P0_1	15	Digital I/O	Port 0.1
P0_0	16	Digital I/O	Port 0.0
P2_7	17	Digital I/O	Port 2.7
P2_6	18	Digital I/O	Port 2.6
P2_5	19	Digital I/O	Port 2.5
P2_4	20	Digital I/O	Port 2.4
VDD	21	Power (analog)	2-V-3.6V analog power-supply connection
RESET_N	22	Digital input	Reset, active-low
P1_4/DC	23	Digital I/O / Debug	Port 1.4/Debug
P1_3/DD	24	Digital I/O / Debug	Port 1.3/Debug
VDD	25	Power (analog)	2-V-3.6V analog power-supply connection
XOSC_Q1	26	Analog I/O	32-MHz crystal oscillator pin 1 or external-clock input
XOSC_Q2	27	Analog I/O	32-MHz crystal oscillator pin 2
VDD	28	Power (analog)	2-V-3.6V analog power-supply connection
VSS	29	Unused pin	Connect to ground
RF_P	30	RF I/O	Positive RF input signal to LNA during RX Positive RF output signal from PA during TX
RF_N	31	RF I/O	Negative RF input signal to LNA during RX Negative RF output signal from PA during TX
VSS	32	Unused pin	Connect to ground
VDD	33	Power (analog)	2-V-3.6V analog power-supply connection
VDD	34	Power (analog)	2-V-3.6V analog power-supply connection
RBIAS	35	Analog I/O	External precision bias resistor for reference current
P1_6/ XOSC32K_ Q2	36	Digital I/O / Analog I/O	Port 1.6/32.768-kHz XOSC
P1_5/ XOSC32k_ Q1	37	Digital I/O / Analog I/O	Port 1.5/32.768-kHz XOSC
P1_2	38	Digital I/O	Port 1.2, 20mA
P1_1	39	Digital I/O	Port 1.1, 20mA
P1_0	40	Digital I/O	Port 1.0, 20mA
VDD	41	Power (analog)	2-V-3.6V analog power-supply connection
DCPL1	42	Power (digital)	1.8-V digital power-supply decoupling. Do not use for supplying external circuits.
VDD	43	Power (analog)	2-V-3.6V analog power-supply connection
VSS	44	Unused pin	Connect to ground
P3_7	45	Digital I/O	Port 3.7
P3_6	46	Digital I/O	Port 3.6
P3_5	47	Digital I/O	Port 3.5
P3_4	48	Digital I/O	Port 3.4
VSS	Ground Pad	Ground	Must be connected to solid ground as this is the main ground connection for the chip. See Pinout Diagram .

BLOCK DIAGRAM

A block diagram of the CC2545 is shown in 图 7. The modules can be roughly divided into one of three categories: CPU-related modules; modules related to power, test, and clock distribution; and radio-related modules. In the following subsections, a short description of each module is given. See CC2543/44/45 User's Guide (SWRU283) for more details.

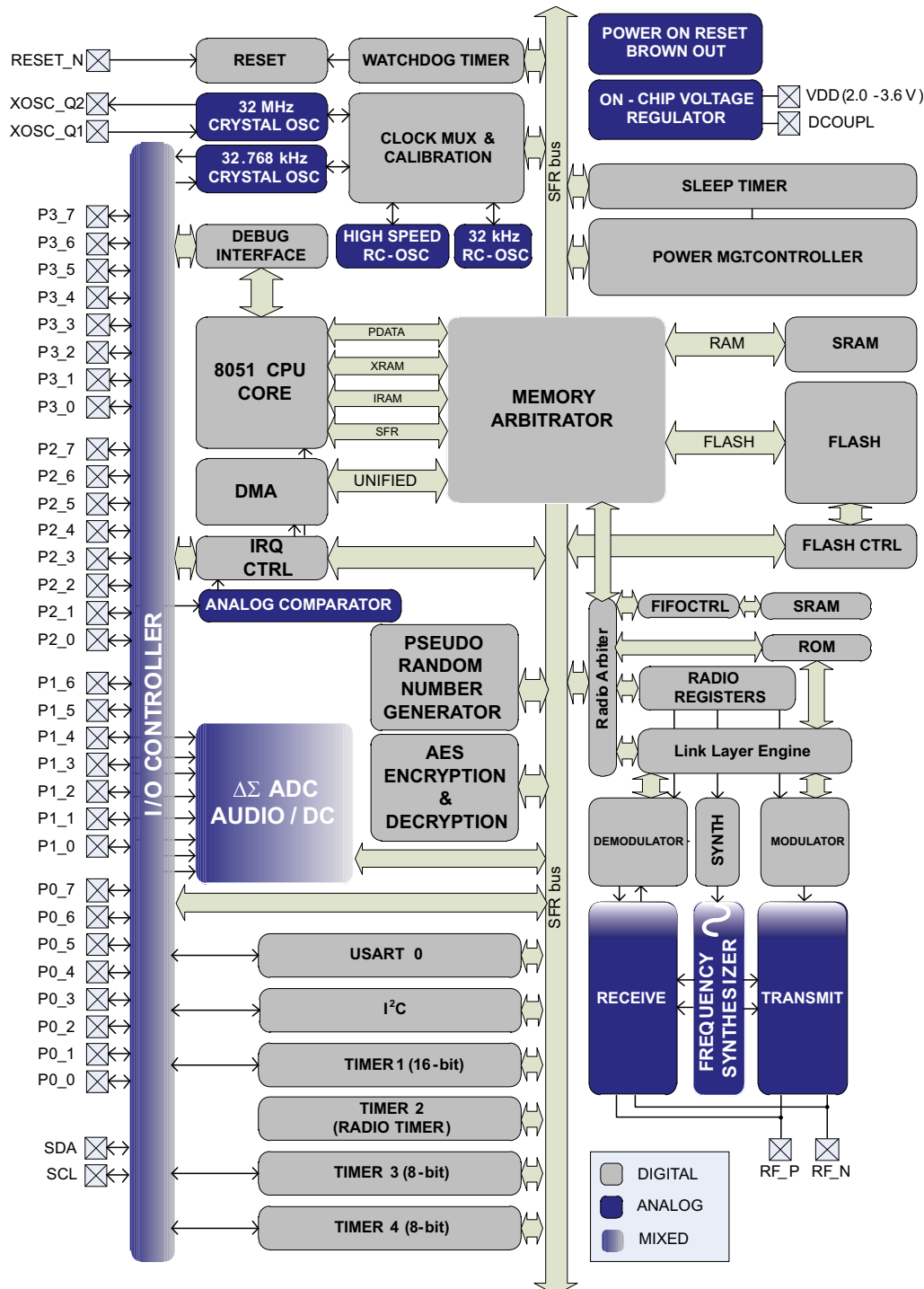


图 7. CC2545 Block Diagram

BLOCK DESCRIPTIONS

CPU and Memory

The **8051 CPU core** is a single-cycle 8051-compatible core. It has three different memory access busses (SFR, DATA, and CODE/XDATA), a debug interface, and an 15-input extended interrupt unit.

The **memory arbiter** is at the heart of the system, as it connects the CPU and DMA controller with the physical memories and all peripherals through the SFR bus. The memory arbiter has four memory-access points, access of which can map to one of three physical memories: an SRAM, flash memory, and XREG/SFR registers. It is responsible for performing arbitration and sequencing between simultaneous memory accesses to the same physical memory.

The **SFR bus** is drawn conceptually in [图 7](#) as a common bus that connects all hardware peripherals to the memory arbiter. The SFR bus in the block diagram also provides access to the radio registers in the radio register bank, even though these are indeed mapped into XDATA memory space.

The **1-KB SRAM** maps to the DATA memory space and to parts of the XDATA memory spaces.

The **18-KB/32-KB flash block** provides in-circuit programmable non-volatile program memory for the device, and maps into the CODE and XDATA memory spaces.

Peripherals

Writing to the flash block is performed through a **flash controller** that allows page-wise erasure and 4-bytewise programming. See User Guide for details on the flash controller.

A versatile two-channel **DMA controller** is available in the system, accesses memory using the XDATA memory space, and thus has access to all physical memories. Each channel (trigger, priority, transfer mode, addressing mode, source and destination pointers, and transfer count) is configured with DMA descriptors that can be located anywhere in memory. Many of the hardware peripherals (AES core, flash controller, USART, timers, etc.) can be used with the DMA controller for efficient operation by performing data transfers between a single SFR or XREG address and flash/SRAM.

The **interrupt controller** services a total of 17 interrupt sources, divided into six interrupt groups, each of which is associated with one of four interrupt priorities. Any interrupt service request is serviced also when the device is in idle mode by going back to active mode. Some interrupts can also wake up the device from sleep mode (when in sleep mode, the device is in low-power mode PM1, PM2 or PM3).

The **debug interface** implements a proprietary two-wire serial interface that is used for in-circuit debugging. Through this debug interface, it is possible to perform an erasure of the entire flash memory, control which oscillators are enabled, stop and start execution of the user program, execute supplied instructions on the 8051 core, set code breakpoints, and single-step through instructions in the code. Using these techniques, it is possible to perform in-circuit debugging and external flash programming elegantly.

The **I/O controller** is responsible for all general-purpose I/O pins. The CPU can configure whether peripheral modules control certain pins or whether they are under software control, and if so, whether each pin is configured as an input or output and if a pullup or pulldown resistor in the pad is connected. Each peripheral that connects to the I/O pins can choose between several different I/O pin locations to ensure flexibility in various applications.

The **sleep timer** is an ultralow-power timer that can use either an external 32.768-kHz XOSC or an internal 32.753-kHz RC oscillator. The sleep timer runs continuously in all operating modes. Typical applications of this timer are as a real-time counter or as a wake-up timer to get out of power modes 1 or 2.

A built-in **watchdog timer** allows the CC2545 to reset itself if the firmware hangs. When enabled by software, the watchdog timer must be cleared periodically; otherwise, it resets the device when it times out.

Timer 1 is a 16-bit timer with timer/counter/PWM functionality. It has a programmable prescaler, a 16-bit period value, and five individually programmable counter/capture channels, each with a 16-bit compare value. Each of the counter/capture channels can be used as a PWM output or to capture the timing of edges on input signals. It can also be configured in IR generation mode, where it counts timer 3 periods and the output is ANDed with the output of timer 3 to generate modulated consumer IR signals with minimal CPU interaction.

Timer 2 is a 40-bit timer used by the Radio. It has a 16-bit counter with a configurable timer period and a 24-bit overflow counter that can be used to keep track of the number of periods that have transpired. A 40-bit capture register is also used to record the exact time at which a start-of-frame delimiter is received/transmitted or the exact time at which a packet ends. There are two 16-bit timer-compare registers and two 24-bit overflow-compare registers that can be used to give exact timing for start of RX or TX to the radio or general interrupts.

Timer 3 and timer 4 are 8-bit timers with timer/counter/PWM functionality. They have a programmable prescaler, an 8-bit period value, and one programmable counter channel with an 8-bit compare value. Each of the counter channels can be used as PWM output.

USART 0 is configurable as either an SPI master/slave or a UART. It provides double buffering on both RX and TX and hardware flow control and is thus well suited to high-throughput full-duplex applications. The USART has its own high-precision baud-rate generator, thus leaving the ordinary timers free for other uses. When configured as SPI slaves, the USART samples the input signal using SCK directly instead of using some oversampling scheme, and are thus well-suited for high data rates.

I²C module provides a digital peripheral connection with two pins and supports both master and slave operation.

The **AES encryption/decryption core** allows the user to encrypt and decrypt data using the AES algorithm with 128-bit keys. The AES core also supports ECB, CBC, CFB, OFB, CTR, and CBC-MAC, as well as hardware support for CCM.

The ultralow power **analog comparator** enables applications to wake up from PM2 or PM3 based on an analog signal. Both inputs are brought out to pins; the reference voltage must be provided externally. The comparator output is mapped into the digital I/O port and can be treated by the MCU as a regular digital input.

TYPICAL CHARACTERISTICS

All curves are for measurements performed at 2Mbps, GFSK, 320-kHz deviation.

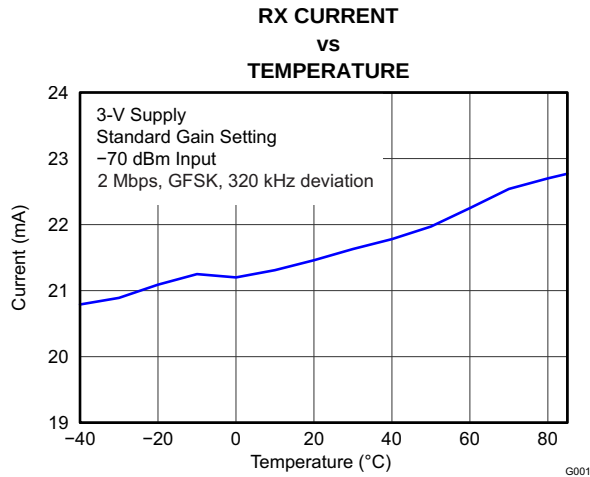


图 8.

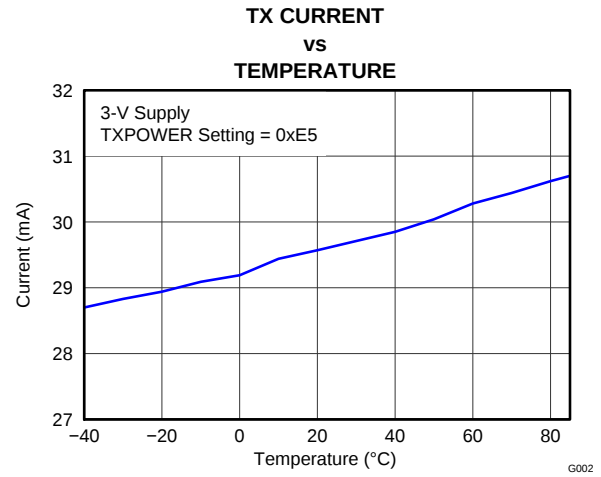


图 9.

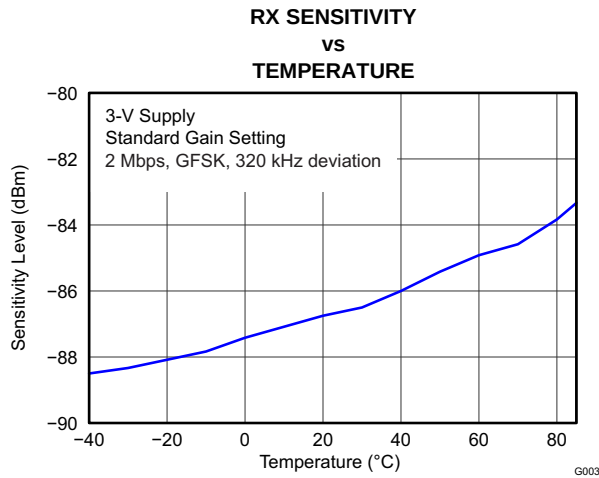


图 10.

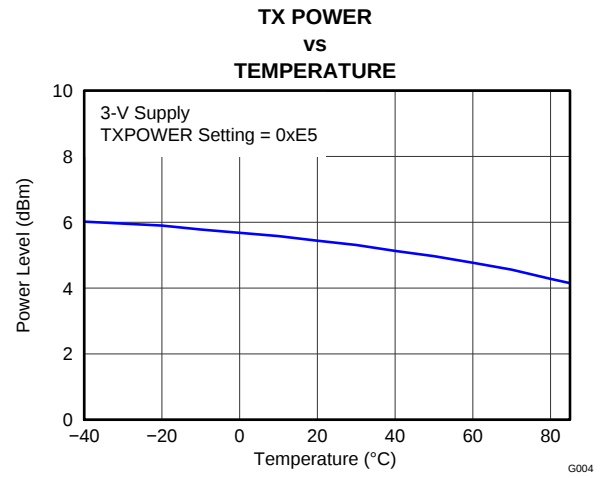


图 11.

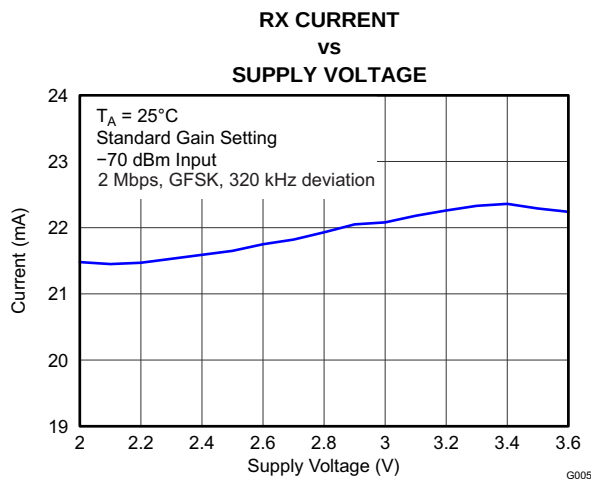


图 12.

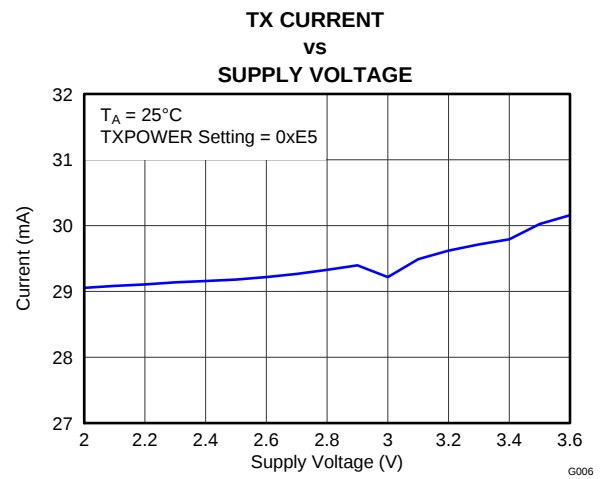


图 13.

TYPICAL CHARACTERISTICS (接下页)

All curves are for measurements performed at 2Mbps, GFSK, 320-kHz deviation.

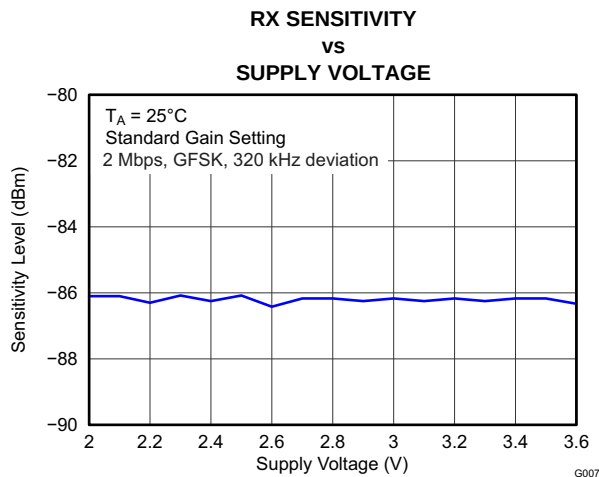


图 14.

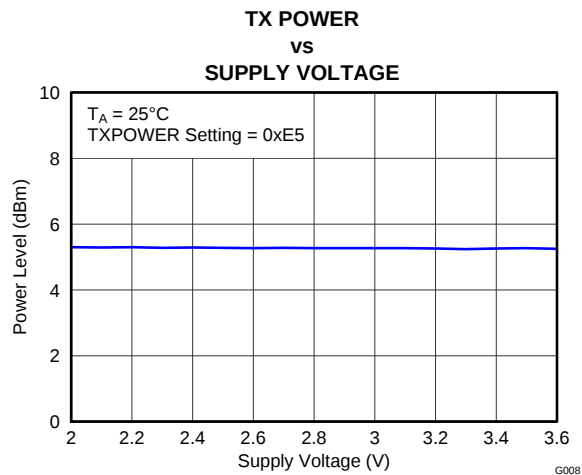


图 15.

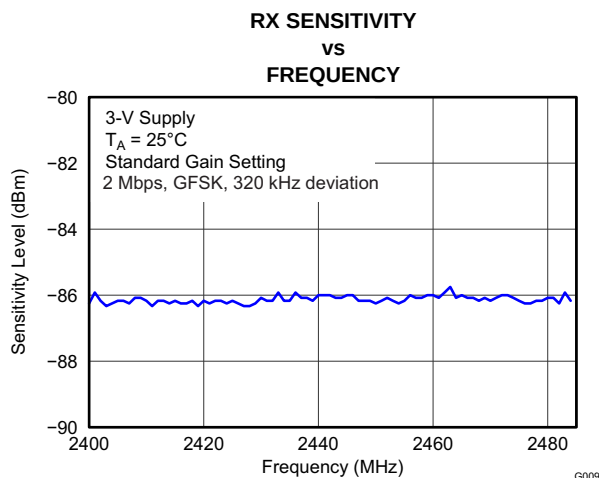


图 16.

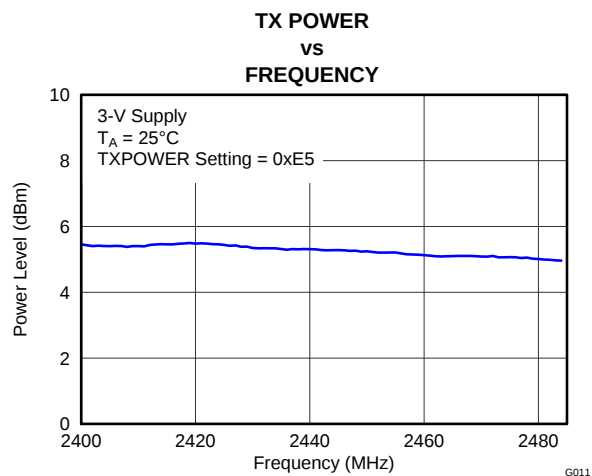


图 17.

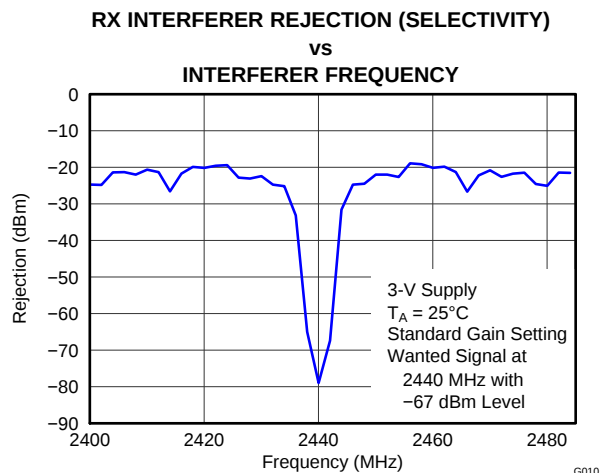


图 18.

TYPICAL CHARACTERISTICS (接下页)
表 2. Recommended Output Power Settings⁽¹⁾

TXPOWER Register Setting	Typical Output Power (dBm)
0xE5	5
0xD5	4
0xC5	3
0xB5	2
0xA5	0
0x95	-2
0x85	-3
0x75	-4
0x65	-6
0x55	-8
0x45	-11
0x35	-13
0x25	-15
0x15	-17
0x05	-20

(1) Measured on Texas Instruments CC2545 EM reference design with TA = 25°C, VDD = 3 V, and fc = 2440 MHz. See [SWRU283](#) for recommended register settings.

APPLICATION INFORMATION

APPLICATION INFORMATION

Few external components are required for the operation of the CC2545. A typical application circuit is shown in 图 19. For suggestions of component values other than those listed in 表 3, see reference design CC2545EM. The performance stated in this data sheet is only valid for the CC2545EM reference design. To obtain similar performance, the reference design should be copied as closely as possible.

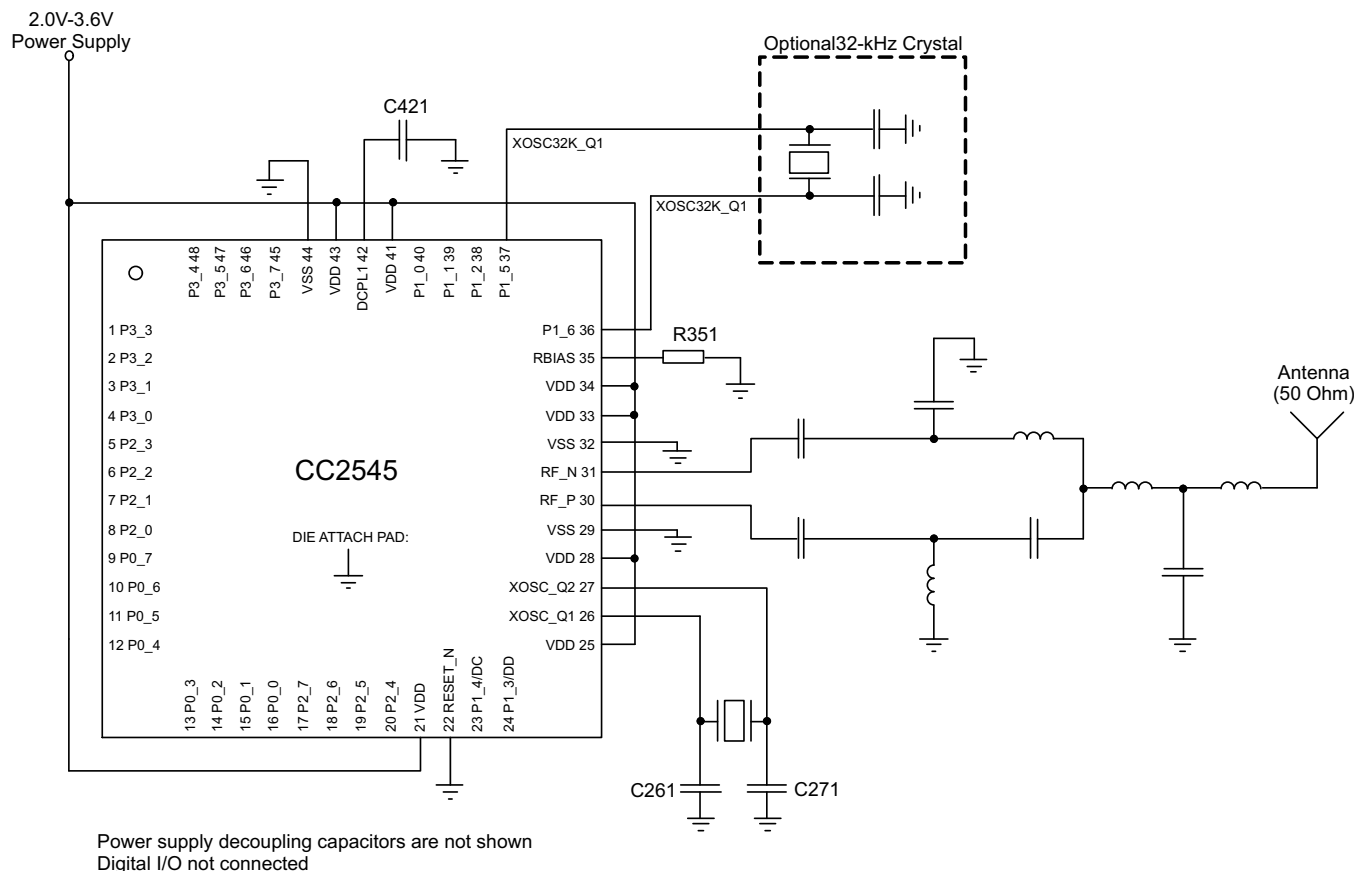


图 19. CC2545 Application Circuit

表 3. Overview of External Components (Excluding Balun, Crystal and Supply Decoupling Capacitors)

Component	Description	Value
C421	Decoupling capacitor for the internal 1.8V digital voltage regulator	1 μ F
R351	Precision resistor $\pm 1\%$, used for internal biasing	56 k Ω

Input/Output Matching

When using an unbalanced antenna such as a monopole, a balun should be used to optimize performance. The balun can be implemented using low-cost discrete inductors and capacitors. See reference design, CC2545EM, for recommended balun.

Crystal

An external 32-MHz crystal with two loading capacitors is used for the 32-MHz crystal oscillator. The load capacitance seen by the 32-MHz crystal is given by:

$$C_L = \frac{1}{\frac{1}{C_{261}} + \frac{1}{C_{271}}} + C_{\text{parasitic}} \quad (1)$$

A series resistor may be used to comply with ESR requirement.

On-Chip 1.8-V Voltage Regulator Decoupling

The 1.8-V on-chip voltage regulator supplies the 1.8-V digital logic. This regulator requires a decoupling capacitor (C421) for stable operation.

Power-Supply Decoupling and Filtering

Proper power-supply decoupling must be used for optimum performance. The placement and size of the decoupling capacitors and the power supply filtering are very important to achieve the best performance in an application. TI provides a compact reference design that should be followed very closely.

REVISION HISTORY

Changes from Original (June 2012) to Revision A	Page
• Deleted 产品预览标题	1
• Changed the Temperature coefficient Unit value From: mV/°C To: / 0.1°C	8
• Changed 图 19	22

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
CC2545RGZR	Active	Production	VQFN (RGZ) 48	2500 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	CC2545
CC2545RGZR.A	Active	Production	VQFN (RGZ) 48	2500 LARGE T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	CC2545
CC2545RGZT	Active	Production	VQFN (RGZ) 48	250 SMALL T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	CC2545
CC2545RGZT.A	Active	Production	VQFN (RGZ) 48	250 SMALL T&R	Yes	NIPDAU	Level-3-260C-168 HR	-40 to 85	CC2545

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

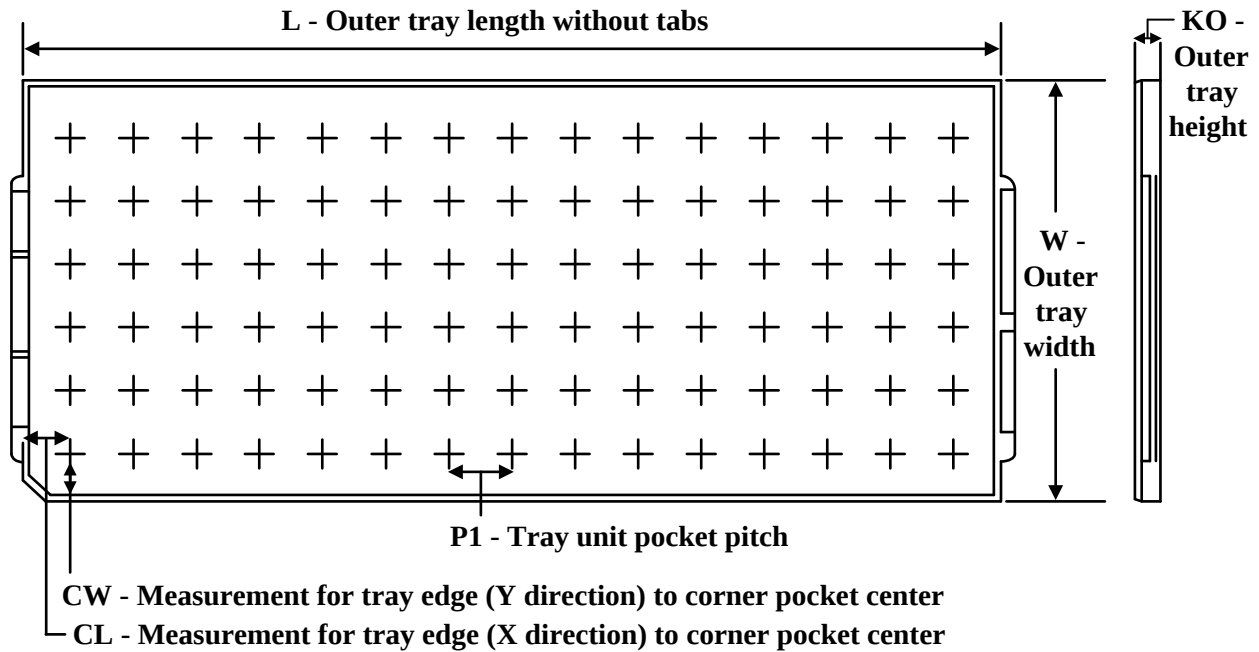
⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

TRAY



Chamfer on Tray corner indicates Pin 1 orientation of packed units.

*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	Unit array matrix	Max temperature (°C)	L (mm)	W (mm)	K0 (μm)	P1 (mm)	CL (mm)	CW (mm)
CC2545RGZR	RGZ	VQFN	48	2500	26 x 10	150	315	135.9	7620	11.8	10	10.35
CC2545RGZR.A	RGZ	VQFN	48	2500	26 x 10	150	315	135.9	7620	11.8	10	10.35
CC2545RGZT	RGZ	VQFN	48	250	26 x 10	150	315	135.9	7620	11.8	10	10.35
CC2545RGZT.A	RGZ	VQFN	48	250	26 x 10	150	315	135.9	7620	11.8	10	10.35

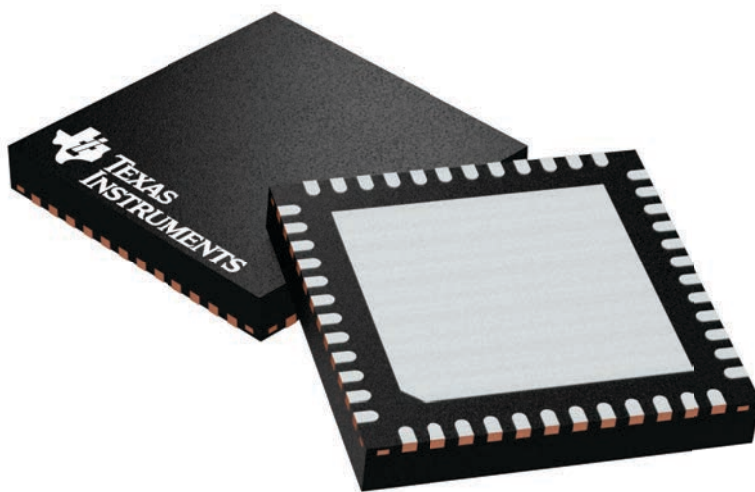
GENERIC PACKAGE VIEW

RGZ 48

VQFN - 1 mm max height

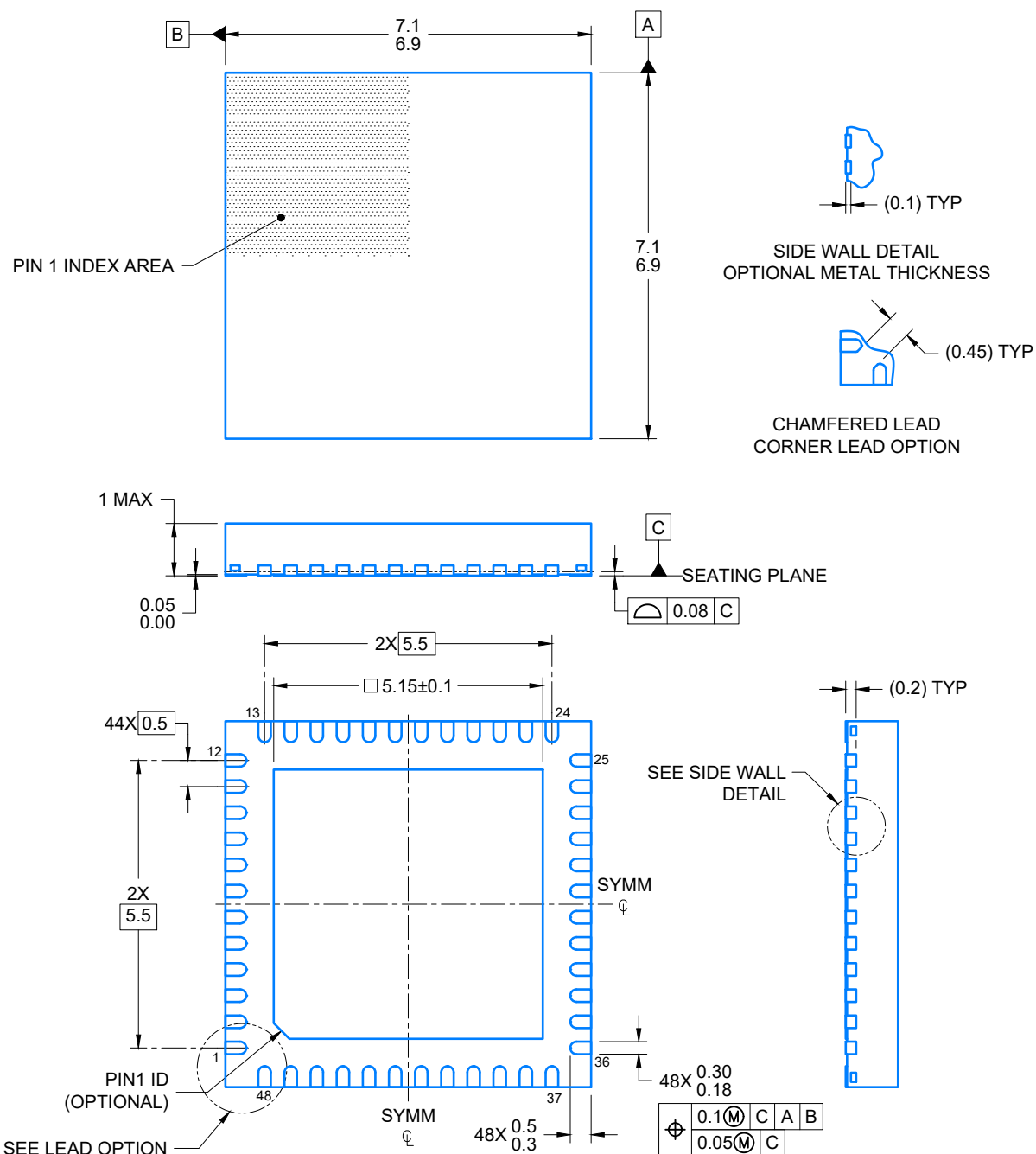
7 x 7, 0.5 mm pitch

PLASTIC QUADFLAT PACK- NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

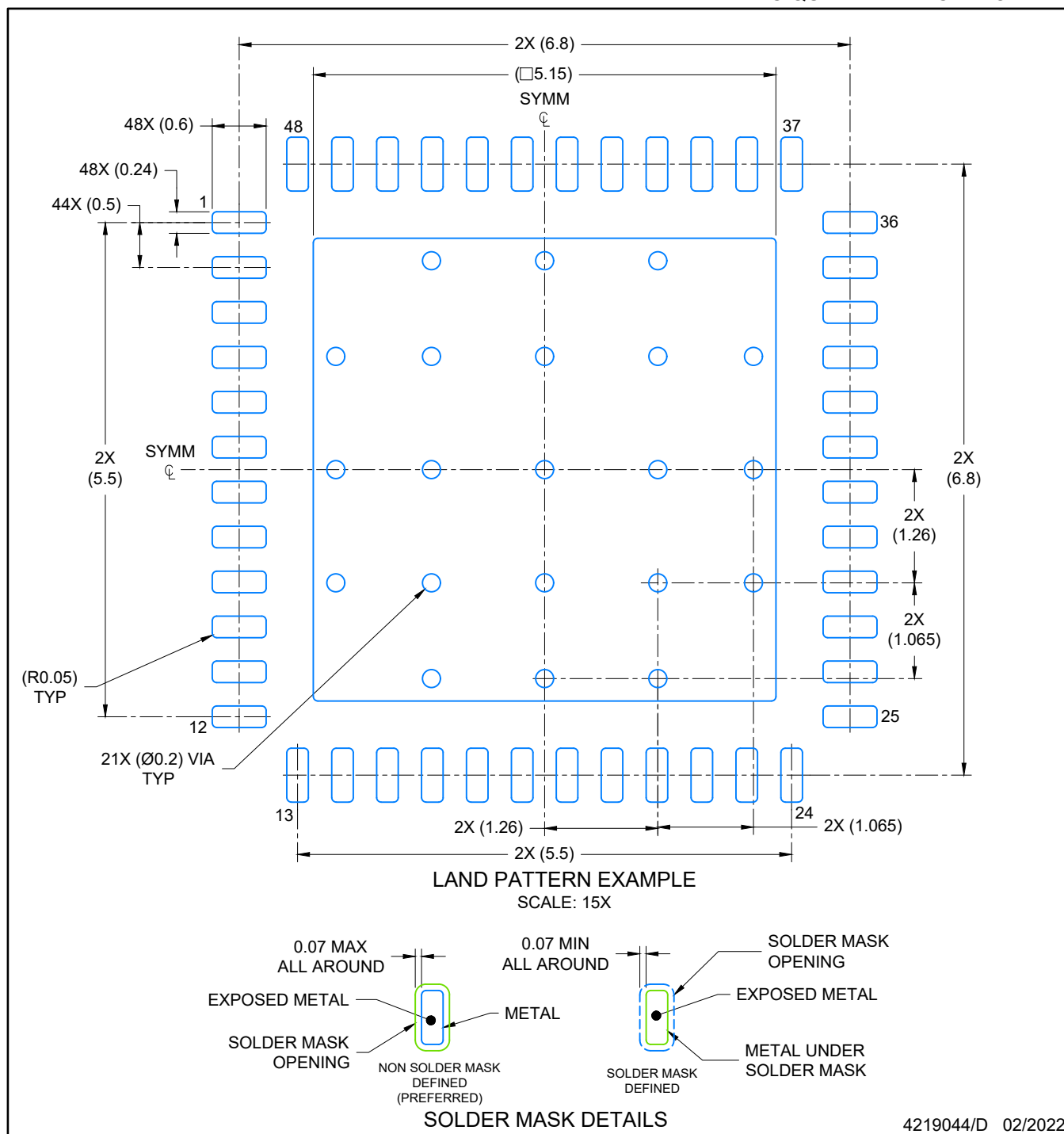
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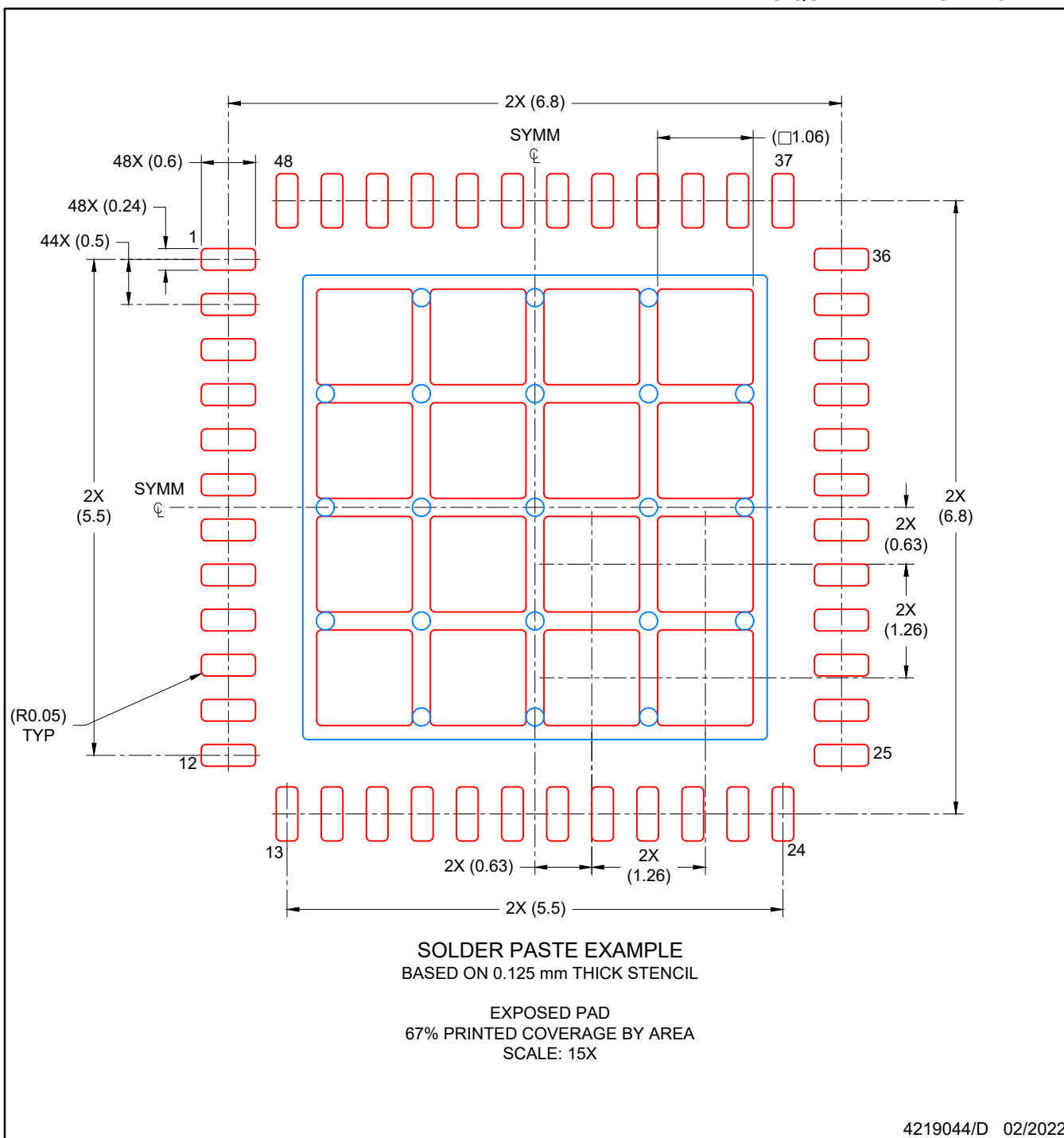
NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/sluea271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

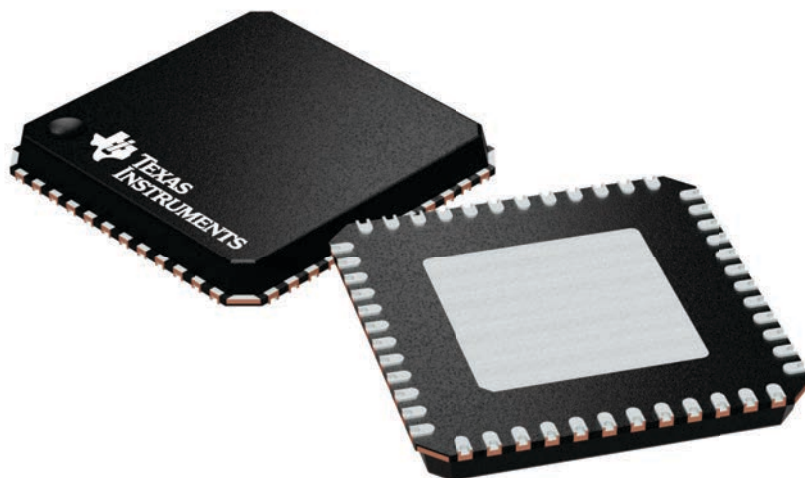
GENERIC PACKAGE VIEW

RTC 48

VQFNP - 0.9 mm max height

7 x 7, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

4224601/A

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