









BQ34Z100-R2

ZHCSRC1 - DECEMBER 2022

BQ34Z100-R2 采用 Impedance Track™ 技术的宽量程电量监测计

1 特性

- 支持锂离子、磷酸铁锂、PbA、镍氢和镍镉化学物
- 对电压为 3V 至

16.7 KV 的电池使用已获得专利的 Impedance Track™ 技术估算容量

- 老化补偿
- 自放电补偿
- 支持的电池容量高达 7000Ah,并且提供标准配置
- 支持的充电和放电电流高达 8160 A,并且提供标准 配置选项
- 外部负温度系数 (NTC) 热敏电阻支持
- 支持与主机系统的两线制 I2C 和 HDQ 单线制通信 接口
- SHA-1/HMAC 认证
- 一个或者四个 LED 直接显示控制
- 五个 LED 和通过端口扩展器的更多显示
- 节能模式(典型电池组运行范围条件)
 - 正常工作: < 145µA 平均电流

- 睡眠: < 84µA 平均电流 - 全睡眠: < 30µA 平均电流

• 封装: 14 引脚 TSSOP

2 应用

- 轻型电动车辆
- 医疗仪器
- 移动无线电
- 电动工具
- 不间断电源 (UPS)

3 说明

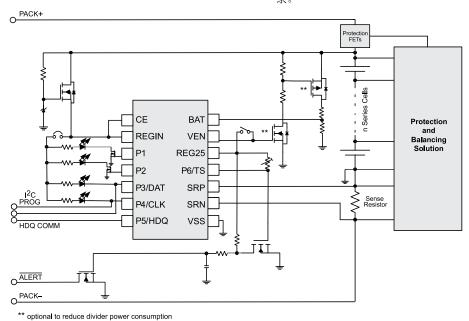
BQ34Z100-R2 器件是适用于锂离子、铅酸、镍氢和镍 镉电池的 Impedance Track™ 电量监测计,并且独立于 电池串联配置工作。通过外部电压转换电路可轻松支持 3V 至 16.7KV 的电池,此电路可通过自动控制来降低 系统功耗。

BQ34Z100-R2 器件提供多个接口选项,其中包括一个 I²C 外设接口、一个 HDQ 外设接口、一个或四个直接 LED 接口以及一个 ALERT 输出引脚。此外, BQ34Z100-R2 还支持外部端口扩展器,连接四个以上 的 LED。

器件信息

器件型号 ⁽¹⁾	封装	封装尺寸(标称值)		
BQ34Z100-R2	TSSOP (14)	5.00mm × 4.40mm		

如需了解所有可用封装,请参阅数据表末尾的可订购产品附



简化原理图



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4 Revision History 注:以前版本的页码可能与当前版本的页码不同

DATE	REVISION	NOTES
December 2022	*	Initial Release



5 Pin Configuration and Functions

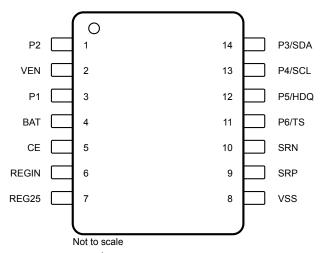


表 5-1. Pin Functions

F	PIN	T\0=	DECORPTION
NAME	NUMBER	TYPE	DESCRIPTION
P2	1	0	LED 2 or Not Used (connect to VSS)
VEN	2	0	Active High Voltage Translation Enable. This signal is optionally used to switch the input voltage divider on/off to reduce the power consumption (typ 45 μ A) of the divider network. If not used, then this pin can be left floating or tied to VSS.
P1	3	0	LED 1 or Not Used (connect to VSS). This pin is also used to drive an LED for single-LED mode. Use a small signal N-FET (Q1) in series with the LED as shown on 图 8-4.
BAT	4	I	Translated Battery Voltage Input
CE	5	I	Chip Enable. Internal LDO is disconnected from REGIN when driven low.
REGIN	6	Р	Internal integrated LDO input. Decouple with a 0.1-µF ceramic capacitor to VSS.
REG25	7	Р	2.5-V output voltage of the internal integrated LDO. Decouple with 1-µF ceramic capacitor to VSS.
VSS	8	Р	Device ground
SRP	9	I	Analog input pin connected to the internal coulomb-counter peripheral for integrating a small voltage between SRP and SRN where SRP is nearest the BAT - connection.
SRN	10	I	Analog input pin connected to the internal coulomb-counter peripheral for integrating a small voltage between SRP and SRN where SRN is nearest the PACK - connection.
P6/TS	11	I	Pack thermistor voltage sense (use a 103AT-type thermistor)
P5/HDQ	12	I/O	Open-drain HDQ Serial communication line (target). If not used, then this pin can be left floating or tied to VSS.
P4/SCL	13	I	Target I^2C serial communication clock input. Use with a 10-k Ω pullup resistor (typical). This pin is also used for LED 4 in the four-LED mode. If not used, then this pin can be left floating or tied to VSS.
P3/SDA	14	I/O	Open-drain target I ² C serial communication data line. Use with a 10-k Ω pullup resistor (typical). This pin is also used for LED 3 in the four-LED mode. If not used, then this pin can be left floating or tied to VSS.



6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1)

		MIN	MAX	UNIT
V _{REGIN}	Regulator Input Range	- 0.3	5.5	V
V _{CC}	Supply Voltage Range	- 0.3	2.75	V
V _{IOD}	Open-drain I/O pins (SDA, SCL, HDQ, VEN)	- 0.3	5.5	V
V_{BAT}	Bat Input pin	- 0.3	5.5	V
VI	Input Voltage range to all other pins (P1, P2, SRP, SRN)	- 0.3	VCC + 0.3	V
ESD	Human-body model (HBM), BAT pin		1.5	kV
E3D	Human-body model (HBM), all other pins		2	kV
T _A	Operating free-air temperature range	- 40	85	°C
T _F	Functional temperature range	- 40	100	°C
т	Storage temperature range	- 65	150	°C
T _{STG}	Lead temperature (soldering, 10 s)	- 40	100	°C

⁽¹⁾ Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	_:	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	\/
		Charged device model (CDM), per ANSI/ESDA/JEDEC JS-002 ⁽²⁾	±500	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

 T_A = -40°C to 85°C; Typical Values at T_A = 25°C C_{LDO25} = 1.0 μ F, and V_{REGIN} = 3.6 V (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V	Cupply Voltage	No operating restrictions	2.7		4.5	V
V_{REGIN}	Supply Voltage	No FLASH writes	2.45		2.7	V
C _{REGIN}	External input capacitor for internal LDO between REGIN and VSS	Nominal capacitor values specified. Recommend a 10% ceramic X5R type		0.1		μF
C _{LDO25}	External output capacitor for internal LDO between VCC and VSS	capacitor located close to the device.	0.47	1		μF
I _{cc}	NORMAL operating-mode current	Gas Gauge in NORMAL mode, I _{LOAD} > Sleep Current		145		μ А
I _{SLP}	SLEEP operating-mode current	Gas Gauge in SLEEP mode, I _{LOAD} < <i>Sleep Current</i>		84		μ A
I _{SLP+}	FULLSLEEP operating-mode current	Gas Gauge in FULL SLEEP mode, I _{LOAD} < <i>Sleep Current</i>		30		μ A
V _{OL}	Output voltage, low (SCL, SDA, HDQ, VEN)	I _{OL} = 3 mA			0.4	V
V _{OH(PP)}	Output voltage, high	I _{OH} = -1 mA	V _{CC} - 0.5			V
V _{OH(OD)}	Output voltage, high (SDA, SCL, HDQ, VEN)	External pull-up resistor connected to V _{CC}	V _{CC} - 0.5			V
V _{IL}	Input voltage, low		- 0.3		0.6	V

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6.3 Recommended Operating Conditions (continued)

 T_A = -40°C to 85°C; Typical Values at T_A = 25°C C_{LDO25} = 1.0 μ F, and V_{REGIN} = 3.6 V (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V _{IH(OD)}	Input voltage, high (SDA, SCL, HDQ)		1.2		6	V
V _{A1}	Input voltage range (TS)	VSS	3 - 0.05		1	V
V _{A2}	Input voltage range (BAT)	VSS	- 0.125		5	V
V _{A3}	Input voltage range (SRP, SRN)	VSS	- 0.125		0.125	V
I _{LKG}	Input leakage current (I/O pins)				0.3	μА
t _{PUCD}	Power-up communication delay			250		ms

6.4 Thermal Information

		BQ34Z100-R2	
	THERMAL METRIC(1)	TSSOP (PW)	UNIT
		14 PINS	
R ₀ JA, High K	Junction-to-ambient thermal resistance	103.8	
R _{θ JC(top)}	Junction-to-case(top) thermal resistance	31.9	
R ₀ JB	Junction-to-board thermal resistance	46.6	°C/M/
ψJT	Junction-to-top characterization parameter	2.0	°C/W
ψ ЈВ	Junction-to-board characterization parameter	45.9	
R _{θ JC(bottom)}	Junction-to-case(bottom) thermal resistance	N/A	

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics Application Report, SPRA953.

6.5 Electrical Characteristics: Power-On Reset

 $T_A = -40$ °C to 85°C; Typical Values at TA = 25°C and $V_{REGIN} = 3.6$ V (unless otherwise noted)

PARAMETER	र	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IT+}	Positive-going battery voltage input at REG25		2.05	2.20	2.31	V
V _{HYS}	Power-on reset hysteresis		45	115	185	mV

6.6 Electrical Characteristics: LDO Regulator

 $T_A = 25$ °C, $C_{LDO25} = 1.0 \mu F$, $V_{REGIN} = 3.6 \text{ V}$ (unless otherwise noted)⁽¹⁾

PARAMETER	1		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V	Regulator output	$I_{OUT} \leqslant V_{REGIN} \leqslant 4.5 \text{ V},$	T _A = - 40°C to 85°C	2.3	2.5	2.7	V
V _{REG25}	REG25 voltage	$2.45~{ m V} \leqslant { m V}_{ m REGIN} < 2.7~{ m V}$ (low battery), ${ m I}_{ m OUT} \leqslant 3~{ m mA}$	T _A = -40°C to 85°C	2.3			V
I _{SHORT} (2)	Short Circuit Current Limit	V _{REG25} = 0 V	$T_A = -40^{\circ}C \text{ to } 85^{\circ}C$			250	mA

⁽¹⁾ LDO output current, I_{OUT}, is the sum of internal and external load currents.

6.7 Electrical Characteristics: Internal Temperature Sensor Characteristics

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARA	METER	TEST CONDITIONS	 \	MAX	
G _{TEMP}	Temperature sensor voltage gain		- 2		mV/°C

⁽²⁾ Specified by design. Not production tested.



6.8 Electrical Characteristics: Low-Frequency Oscillator

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f _(LOSC)	Operating frequency		•	32.768		kHz
		TA = 0°C to 60°C	- 1.5%	0.25%	1.5%	
$f_{(LEIO)}$	Frequency error ⁽¹⁾ (2)	TA = - 20°C to 70°C	- 2.5%	0.25%	2.5%	
		$TA = -40^{\circ}C \text{ to } 85^{\circ}C$	- 4%	0.25%	4%	
t _(LSXO)	Start-up time ⁽³⁾			500		μ S

- (1) The frequency drift is included and measured from the trimmed frequency at VCC = 2.5 V, T_A = 25°C.
- (2) The frequency error is measured from 32.768 kHz.
- (3) The startup time is defined as the time it takes for the oscillator output frequency to be ±3%.

6.9 Electrical Characteristics: High-Frequency Oscillator

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f _(OSC)	Operating frequency			8.389		MHz
		T _A = 0°C to 60°C	- 2%	0.38%	2%	
f _(EIO)	Frequency error ⁽¹⁾ (2)	$T_A = -20^{\circ}C \text{ to } 70^{\circ}C$	- 3%	0.38%	3%	
		T _A = -40°C to 85°C	- 4.5%	0.38%	4.5%	
t _(SXO)	Start-up time ⁽²⁾			2.5	5	ms

- (1) The frequency error is measured from 2.097 MHz.
- (2) The startup time is defined as the time it takes for the oscillator output frequency to be ±3%.

6.10 Electrical Characteristics: Integrating ADC (Coulomb Counter) Characteristics

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER	R	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _(SR)	Input voltage range, $V_{(SRN)}$ and $V_{(SRP)}$	$V_{(SR)} = V_{(SRN)} - V_{(SRP)}$	- 0.125		0.125	V
+	Conversion time	Single conversion		1		s
^t sr_conv	Resolution		14		15	bits
V _{OS(SR)}	Input offset			10		μV
I _{NL}	Integral nonlinearity error			±0.007%	±0.034%	FSR ⁽²⁾
Z _{IN(SR)}	Effective input resistance ⁽¹⁾		2.5			ΜΩ
I _{lkg(SR)}	Input leakage current ⁽¹⁾				0.3	μΑ

- (1) Specified by design. Not tested in production.
- (2) Full-scale reference

6.11 Electrical Characteristics: ADC (Temperature and Cell Measurement) Characteristics

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IN(ADC)}	Input voltage range		0.05		1	V
t _{ADC_CONV}	Conversion time				125	ms
	Resolution		14		15	bits
V _{OS(ADC)}	Input offset			1		mV
Z _{ADC1}	Effective input resistance (TS) ⁽¹⁾		8			ΜΩ
Z _{ADC2}	Effective input resistance (BAT) ⁽¹⁾	BQ34Z100-R2 not measuring cell voltage	8			ΜΩ
	Effective input resistance (BAT)	BQ34Z100-R2 measuring cell voltage		100		ΚΩ

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6.11 Electrical Characteristics: ADC (Temperature and Cell Measurement) Characteristics (continued)

 $T_A = -40$ °C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at $T_A = 25$ °C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I _{lkg(ADC)}	Input leakage current ⁽¹⁾				0.3	μA

⁽¹⁾ Specified by design. Not tested in production.

6.12 Electrical Characteristics: Data Flash Memory Characteristics

T_A = -40°C to 85°C, 2.4 V < REG25 < 2.6 V; Typical Values at T_A = 25°C and REG25 = 2.5 V (unless otherwise noted)

PARAMETER	· ·	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{DR}	Data retention ⁽¹⁾		10			Years
	Flash-programming write cycles ⁽¹⁾		20,000			Cycles
t _{WORDPROG}	Word programming time ⁽¹⁾				2	ms
I _{CCPROG}	Flash-write supply current ⁽¹⁾			5	10	mA

⁽¹⁾ Specified by design. Not tested in production.

6.13 Timing Requirements: HDQ Communication

 $T_A = -40^{\circ}\text{C}$ to 85°C, 2.45 V < $V_{REGIN} = V_{BAT} < 5.5$ V; typical values at $T_A = 25^{\circ}\text{C}$ and $V_{REGIN} = V_{BAT} = 3.6$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
t _(CYCH)	Cycle time, host to BQ34Z100-R2		190	'		μs
t _(CYCD)	Cycle time, BQ34Z100-R2 to host		190	205	250	μs
t _(HW1)	Host sends 1 to BQ34Z100-R2		0.5		50	μs
t _(DW1)	BQ34Z100-R2 sends 1 to host		32		50	μs
t _(HW0)	Host sends 0 to BQ34Z100-R2		86		145	μs
t _(DW0)	BQ34Z100-R2 sends 0 to host		80		145	μs
t _(RSPS)	Response time, BQ34Z100-R2 to host		190		950	μs
t _(B)	Break time		190			μs
t _(BR)	Break recovery time		40			μs
t _(RISE)	HDQ line rising time to logic 1 (1.2 V)				950	ns
t _(RST)	HDQ Reset		1.8		2.2	s



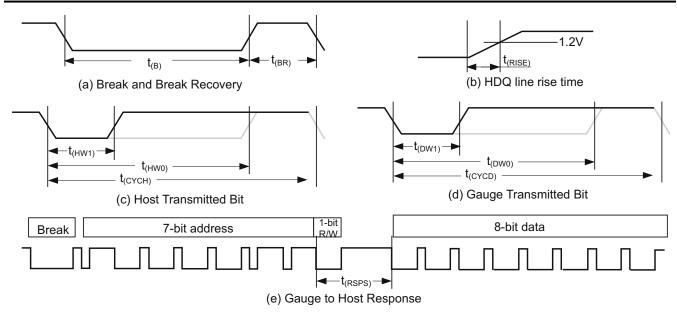


图 6-1. Timing Diagrams

6.14 Timing Requirements: I²C-Compatible Interface

 $T_A = -40$ °C to 85°C, 2.45 V < $V_{REGIN} = V_{BAT} < 5.5$ V; typical values at $T_A = 25$ °C and $V_{REGIN} = V_{BAT} = 3.6$ V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
t _r	SCL/SDA rise time				300	ns
t _f	SCL/SDA fall time				300	ns
t _{w(H)}	SCL pulse width (high)		600			ns
t _{w(L)}	SCL pulse width (low)		1.3			μS
t _{su(STA)}	Setup for repeated start		600			ns
t _{d(STA)}	Start to first falling edge of SCL		600			ns
t _{su(DAT)}	Data setup time		100			ns
t _{h(DAT)}	Data hold time		0			ns
t _{su(STOP)}	Setup time for stop		600			ns
t _{BUF}	Bus free time between stop and start		66			μs
f _{SCL}	Clock frequency			-	400	kHz

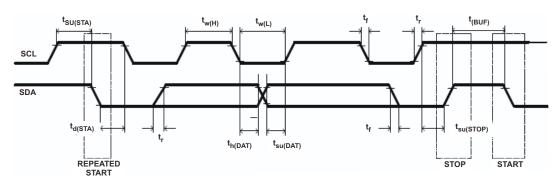
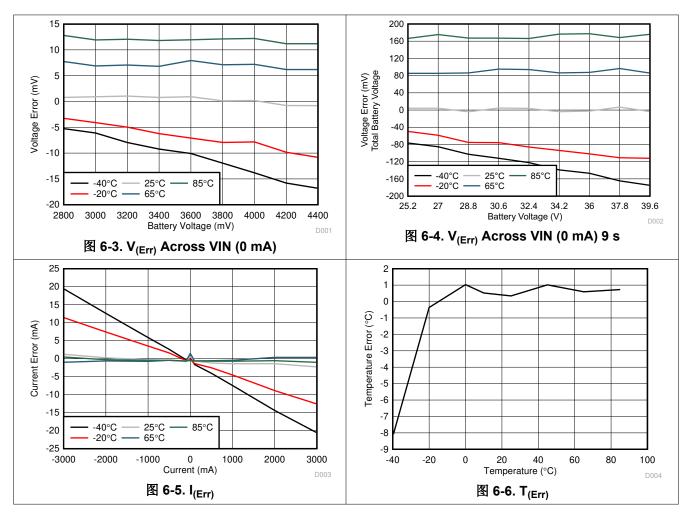


图 6-2. I²C-Compatible Interface Timing Diagrams

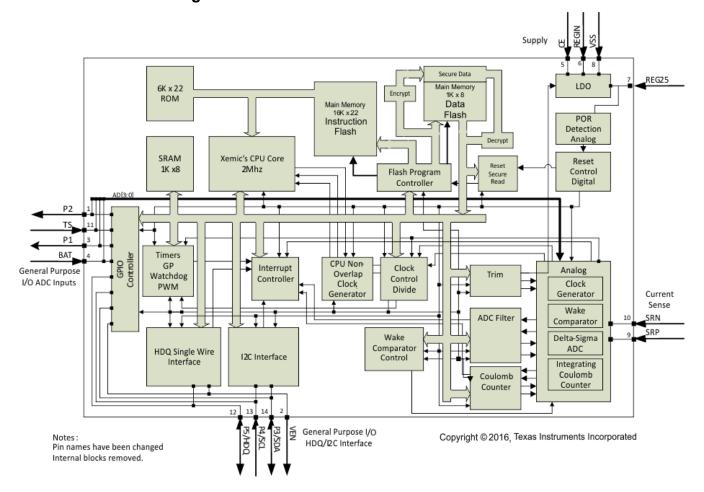


6.15 Typical Characteristics





7 Functional Block Diagram





8 Application and Implementation

备注

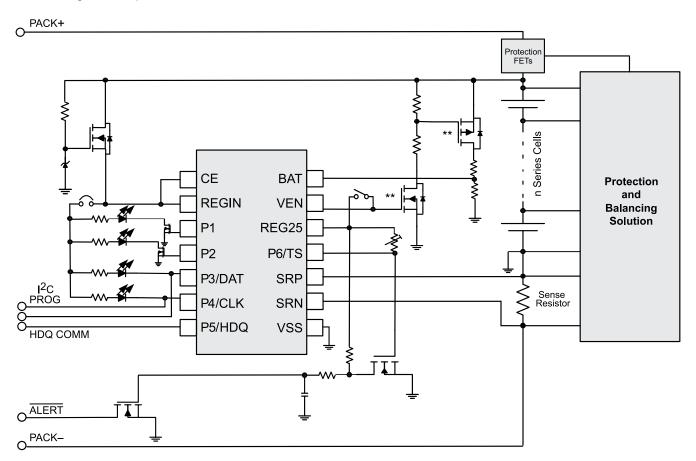
以下应用部分中的信息不属于 TI 元件规格, TI 不担保其准确性和完整性。TI 的客户负责确定元件是否适合其用途,以及验证和测试其设计实现以确认系统功能。

8.1 Application Information

The BQ34Z100-R2 is a flexible gas gauge device with many options. The major configuration choices comprise the battery chemistry, digital interface, and display.

8.2 Typical Applications

8-1 is a simplified diagram of the main features of the BQ34Z100-R2. Specific implementations detailing the main configuration options are shown later in this section.



^{**} optional to reduce divider power consumption

图 8-1. BQ34Z100-R2 Simplified Implementation



The BQ34Z100-R2 can be used to provide a single Li-ion cell gas gauge with a 5-bar LED display.

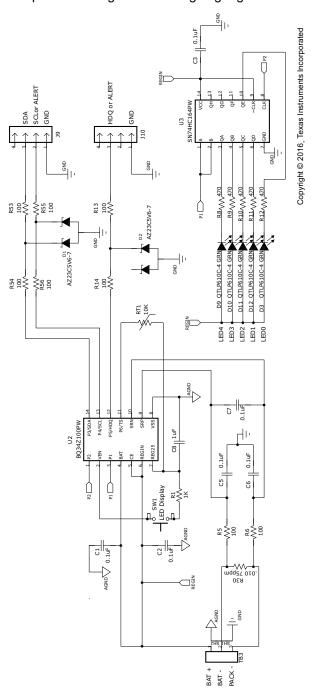


图 8-2. 1-Cell Li-ion and 5-LED Display

The BQ34Z100-R2 can also be used to provide a gas gauge for a multi-cell Li-ion battery with a 5-bar LED display.

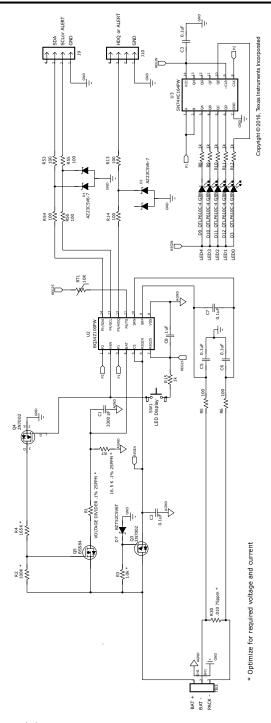


图 8-3. Multi-Cell and 5-LED Display



8-4 shows the BQ34Z100-R2 full features enabled.

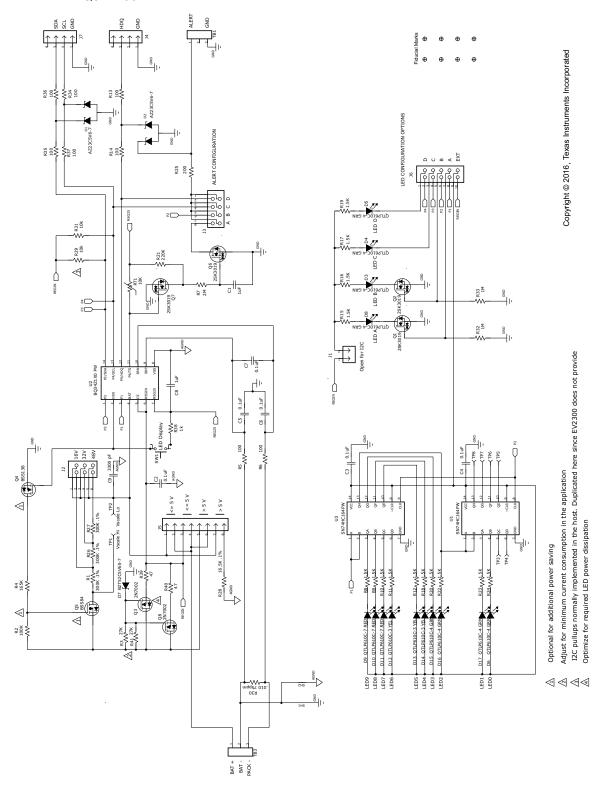


图 8-4. Full-Featured Evaluation Module EVM

8.2.1 Design Requirements

For additional design guidelines, refer to the BQ34Z100 EVM Wide Range Impedance Track Enabled Battery Fuel Gauge User's Guide (SLUU904).

8.2.2 Detailed Design Procedure

8.2.2.1 Step-by-Step Design Procedure

8.2.2.1.1 STEP 1: Review and Modify the Data Flash Configuration Data.

While many of the default parameters in the data flash are suitable for most applications, the following should first be reviewed and modified to match the intended application.

- **Design Capacity**: Enter the value in mAh divided by *CurrScale()* for the battery, even from the "design energy" point of view.
- Design Energy: Enter the value in cWh divided by EnergyScale() .
- Cell Charge Voltage Tx-Ty: Enter the desired cell charge voltage for each JEITA temperature range.

8.2.2.1.2 STEP 2: Review and Modify the Data Flash Configuration Registers.

- **LED_Comm Configuration**: See in the BQ34Z100-R2 Technical Reference Manual to aid in selection of an LED mode. Note that the pin used for the optional Alert signal is dependent upon the LED mode selected.
- Alert Configuration: See the BQ34Z100-R2 Technical Reference Manual to aid in selection of which faults trigger the ALERT pin.
- · Number of Series Cells
- **Pack Configuration**: Ensure that the VOLSEL bit is set for multicell applications and cleared for single-cell applications.

8.2.2.1.3 STEP 3: Design and Configure the Voltage Divider.

If the battery contains more than 1-s cells, a voltage divider network is required. Design the divider network, based on the formula below. The voltage division required is from the highest expected battery voltage, down to approximately 900 mV. For example, using a lower leg resistor of 16.5 K Ω where the highest expected voltage is 32000 mV:

Rseries = 16.5 K Ω (32000 mV - 900 mV)/900 mV = 570.2 K Ω

Based on price and availability, a 600-K resistor or pair of 300-K resistors could be used in the top leg along with a 16.5-K resistor in the bottom leg.

Set the **Voltage Divider** in the Data Flash Calibration section of the Evaluation Software to 32000 mV with VoltScale() =1.

Use the Evaluation Software to calibrate to the applied nominal voltage; for example, 24000 mV. After calibration, a slightly different value appears in the **Voltage Divider** parameter, which can be used as a default value for the project. For the applications with voltage higher than 65535 mV, please refer to the <u>BQ34Z100-R2</u> Technical Reference Manual.

Following the successful voltage calibration, calculate and apply the value to *Flash Update OK Cell Volt* as: *Flash Update OK Cell Volt* = 2800 mV × *Number Of Series Cells* × 5000 / *Voltage Divider* /*VoltScale()* .

8.2.2.1.4 STEP 4: Determine the Sense Resistor Value.

To ensure accurate current measurement, the input voltage generated across the current sense resistor should not exceed +/ - 125 mV. For applications with a very high dynamic range, it is allowable to extend this range to absolute maximum of +/ - 300 mV for overload conditions where a protector device will be taking independent protective action. In such an overloaded state, current reporting and gauging accuracy will not function correctly.

The value of the current sense resistor should be entered into both *CC Gain* and *CC Delta* parameters in the Data Flash Calibration section of the Evaluation Software.

8.2.2.1.5 STEP 5: Review and Modify the Data Flash Gas Gauging Configuration, Data, and State.

- Load Select: See Current Model Used When Load Mode = 0 and Constant-Power Model Used When Load Mode = 1 in the BQ34Z100-R2 Technical Reference Manual.
- Load Mode: See Current Model Used When Load Mode = 0 and Constant-Power Model Used When Load Mode = in the BQ34Z100-R2 Technical Reference Manual.

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- **Cell Terminate Voltage**: This is the theoretical voltage where the system begins to fail. It is defined as a zero state-of-charge. Generally, a more conservative level is used to have some reserve capacity. Note the value is for a single cell only.
- Quit Current: Generally, this should be set to a value slightly above the expected idle current of the system.
- Qmax Cell 0: Start with the C-rate value of your battery.

8.2.2.1.6 STEP 6: Determine and Program the Chemical ID.

Use the BQChem feature in the Evaluation Software to select and program the chemical ID matching your cell. If no match is found, use the procedure defined in Tl's (*Mathcad Chemistry Selection Tool* (SLUC138).

8.2.2.1.7 STEP 7: Calibrate.

Follow the steps on the **Calibration** screen in the Evaluation Software. Achieving the best possible calibration is important before moving on to Step 8. For mass production, calibration is not required for single-cell applications. For multi-cell applications, only voltage calibration is required. Current and temperature may be calibrated to improve gauging accuracy if needed.

8.2.2.1.8 STEP 8: Run an Optimization Cycle.

Refer to the *Preparing Optimized Default Flash Constants for Specific Battery Types Application Report* (SLUA334B).

Product Folder Links: BQ34Z100-R2



9 Power Supply Recommendations

Power supply requirements for the BQ34Z100-R2 are simplified due to the presence of the internal LDO voltage regulation. The REGIN pin accepts any voltage level between 2.7 V and 4.5 V, which is optimum for a single-cell Li-ion application. For higher battery voltage applications, a simple pre-regulator can be provided to power the bq34Z100-R2 and any optional LEDs. Decoupling the REGIN pin should be done with a 0.1-µF 10% ceramic X5R capacitor placed close to the device. While the pre-regulator circuit is not critical, special attention should be paid to its quiescent current and power dissipation. The input voltage should handle the maximum battery stack voltage. The output voltage can be centered within the 2.7-V to 4.5-V range as recommended for the REGIN pin.

For high stack count applications, a commercially available LDO is often the best quality solution, but comes with a cost tradeoff. To lower the BOM cost, the following approaches are recommended.

In

9-1, Q1 is used to drop the battery stack voltage to roughly 4 V to power the BQ34Z100-R2 REGIN pin and also to feed the anode of any LEDs used in the application. To avoid unwanted quiescent current consumption, R1 should be set as high as is practical. It is recommended to use a low-current Zener diode.

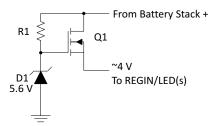


图 9-1. Q1 Dropping Battery Stack Voltage to 4 V

Alternatively, if the range of a high-voltage battery stack can be well defined, a simple source follower based on a resistive divider can be used to lower the BOM cost and the quiescent current. For example:

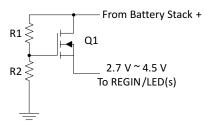


图 9-2. Source Follower on a Resistive Divider

Power dissipation of the linear pre-regulator may become an important design decision when multiple LEDs are employed in the application. For example, the BQ34Z100-R2 EVM uses a pair of FETs in parallel to inexpensively dissipate enough power for 10-LED evaluation.

10 Layout

10.1 Layout Guidelines

10.1.1 Introduction

Attention to layout is critical to the success of any battery management circuit board. The mixture of high-current paths with an ultralow-current microcontroller creates the potential for design issues that are not always trivial to solve. Some of the key areas of concern are described in the following sections, and can help to enable success.

10.1.2 Power Supply Decoupling Capacitor

Power supply decoupling from VCC to ground is important for optimal operation of the gas gauge. To keep the loop area small, place this capacitor next to the IC and use the shortest possible traces. A large loop area renders the capacitor useless and forms a small-loop antenna for noise pickup.

Ideally, the traces on each side of the capacitor should be the same length and run in the same direction to avoid differential noise during ESD. If possible, place a via near the VSS pin to a ground plane layer.

10.1.3 Capacitors

Power supply decoupling for the gas gauges requires a pair of 0.1-µF ceramic capacitors for (BAT) and (VCC) pins. These should be placed reasonably close to the IC without using long traces back to VSS. The LDO voltage regulator, whether external or internal to the main IC, requires a 0.47-µF ceramic capacitor to be placed fairly close to the regulation output pin. This capacitor is for amplifier loop stabilization and as an energy well for the 2.5-V supply.

10.1.4 Communication Line Protection Components

The 5.6-V Zener diodes, used to protect the communication pins of the gas gauge from ESD, should be located as close as possible to the pack connector. The grounded end of these Zener diodes should be returned to the Pack(-) node rather than to the low-current digital ground system. This way, ESD is diverted away from the sensitive electronics as much as possible.

In some applications, it is sometimes necessary to cause transitions on the communication lines to trigger events that manage the gas gauge power modes. An example of one of these transitions is detecting a sustained low logic level on the communication lines to detect that a pack has been removed. Given that most of the gas gauges do not have internal pulldown networks, it is necessary to add a weak pulldown resistor to accomplish this when there's an absence of a strong pullup resistor on the system side. If the weak pulldown resistor is used, it may take less board space to use a small capacitor in parallel instead of the Zener diode to absorb any ESD transients that are received through communication lines.

10.2 Layout Example

10.2.1 Ground System

The gas gauge requires a low-current ground system separate from the high-current PACK(-) path. ESD ground is defined along the high-current path from the PACK(-) terminal to low-side protector FETs (if present) or the sense resistor. It is important that the low-current ground systems only connect to the BAT(-) path at the sense resistor Kelvin pick-off point. It is recommended to use an optional inner layer ground plane for the low-current ground system. In 10-1, the green is an example of using the low-current ground as a shield for the gas gauge circuit. Notice how it is kept separate from the high-current ground, which is shown in red. The high-current path is joined with the low-current path only at one point, shown with the small blue connection between the two planes.



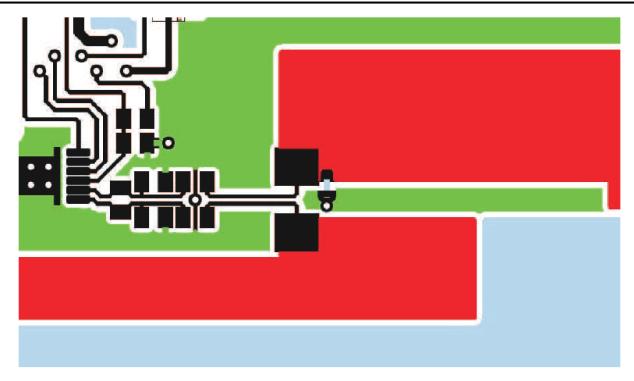


图 10-1. Differential Filter Component with Symmetrical Layout

10.2.2 Kelvin Connections

Kelvin voltage sensing is very important to accurately measure current and cell voltage. Notice how the differential connections at the sense resistor do not add any voltage drop across the copper etch that carries the high current path through the sense resistor. See
10-1 and 10-2.

10.2.3 Board Offset Considerations

Although the most important component for board offset reduction is the decoupling capacitor for V_{CC} , additional benefit is possible by using this recommended pattern for the coulomb counter differential low-pass filter network. Maintain the symmetrical placement pattern shown for optimum current offset performance. Use symmetrical shielded differential traces, if possible, from the sense resistor to the 100- Ω resistors, as shown in \mathbb{Z} 10-2.



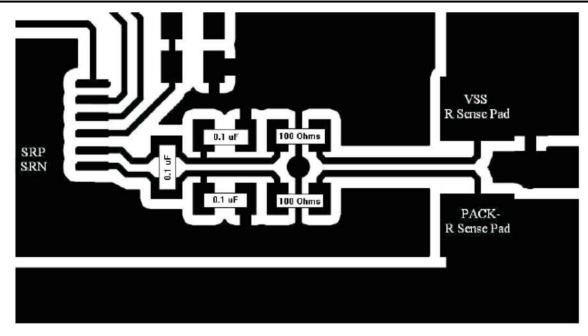


图 10-2. Differential Connection Between SRP and SRN Pins with Sense Resistor

10.2.4 ESD Spark Gap

Protect the communication lines from ESD with a spark gap at the connector. 🖺 10-3 shows the recommended pattern with its 0.2-mm spacing between the points.

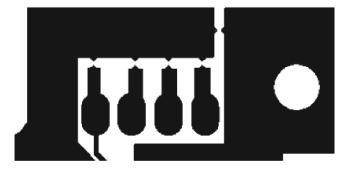


图 10-3. Recommended Spark-Gap Pattern Helps Protect Communication Lines from ESD



11 Device and Documentation Support

11.1 Documentation Support

For related documentation, see the following:

- BQ34Z100-R2 Technical Reference Manual
- BQ34Z100-R2 High Cell Count and High Capacity Applications application report

11.2 接收文档更新通知

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11.6 术语表

TI术语表本术语表列出并解释了术语、首字母缩略词和定义。

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/ Ball material	MSL rating/ Peak reflow	Op temp (°C)	Part marking (6)
BQ34Z100PWR-R2	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	34Z100
BQ34Z100PWR-R2.B	Active	Production	TSSOP (PW) 14	2000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	34Z100

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

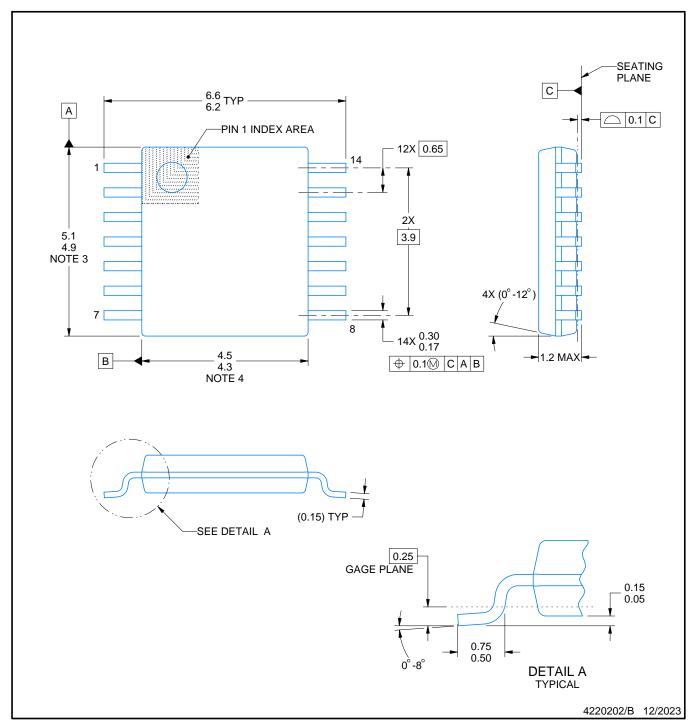
⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



SMALL OUTLINE PACKAGE



NOTES:

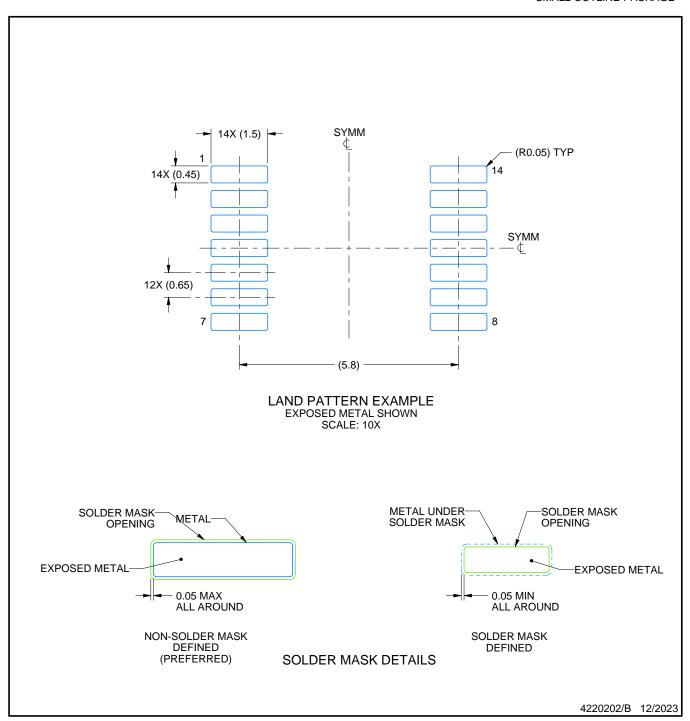
- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

 2. This drawing is subject to change without notice.

 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153.



SMALL OUTLINE PACKAGE



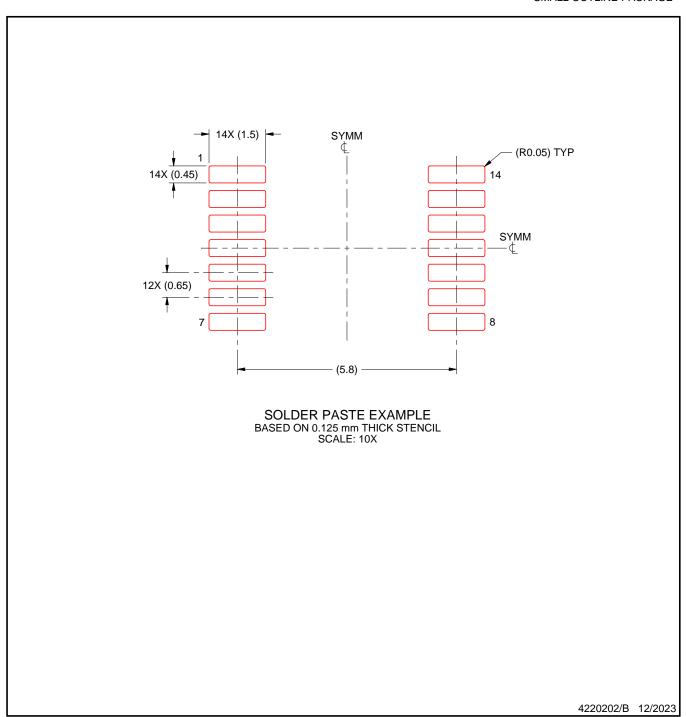
NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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